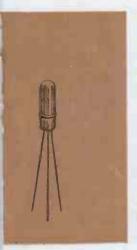
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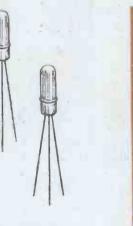
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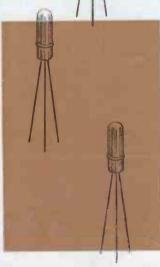
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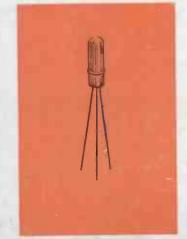




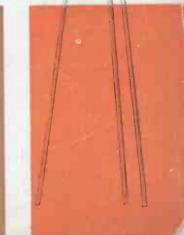












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ELECTRONICS, RADIO; TELEVISION

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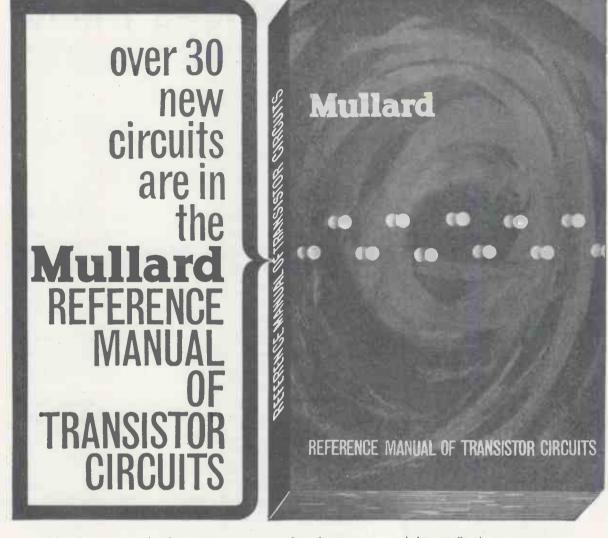
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Wireless World

#### **VOL 67 NO 1 JANUARY 1961**

#### COLOUR BRINKMANSHIP

FOR a long time now colour television in this country has been in the doldrums. Those who want to get under way and who expected a wind from the Television Advisory Committee's Report were disappointed, but they have recently received what in nautical language might be termed a puff from an unexpected quarter.

The T.A.C. Report (May, 1960) said, amongst other things: " . . . we are of the opinion that present technical and economic limitations make it undesirable to introduce a colour television system in the near future," and that "any decision with regard to the introduction of colour must follow a decision on line standards."

The Director-General of the B.B.C. in a speech to the Radio and Television Retailers' Association in London (October 25th, 1960) said: "We in the B.B.C. are ready and very eager to proceed with a small compatible colour service within the framework of our existing programme in Band I on 405lines, without waiting for a decision on whether there is to be a future move to 625-lines in Bands IV and V . . . to wait for colour on 625-lines means that there will be no colour for a national audience for something like another ten years-and that, after the development work the B.B.C. has done during the past six years, would be most regrettable.'

All very confusing and reminiscent of the ancient battle in which "those behind cried 'Forward,' and those before cried 'Back'." At least that is how it must appear to the layman who is confronted with such contradictory statements. But he should understand the feelings of the D.-G. His engineers have produced for him a beautiful toy. No one who has seen recent B.B.C. colour transmissions on a good colour receiver in any of the industry's research laboratories will deny that they are superb. Naturally he wishes to share his pleasure with others; the more the merrier, and the sooner the better. It may seem a small favour to ask of the P.M.G. that he should give permission to take the out-of-hours experimental transmissions which the B.B.C. has been putting out regularly for years, double their duration and include them in regular broadcasting hours: a start would have been made. A start of what? Not of colour television; that was made by the B.B.C. nearly six years ago. The start of a service, and we should be over the brink.

Once a service is started anything may happen. The public may cold-shoulder it on the score of cost, or a few Joneses may set the pace, and industry may be inundated with orders from their neighbours before it is in a position to meet the demand. If it decides to supply the die will have been cast and the pattern will be difficult to change.

Six years ago the Americans took the plunge and

committed themselves to the N.T.S.C. system which, with the shadow-mask three-gun tube and with competent handling, is capable of giving superb results, but which failed to fulfil in the field the high promise of its sponsors. As a result of the American experience significant improvements have been made in receiver stability and reliability, but the three-gun shadow-mask picture tube, to the characteristics of which the N.T.S.C. system is tailored, is costly and likely to remain so. More important, the formulation of colour information in the signal has been compromised in its favour to produce an improved. though still not technically perfect, overall system performance. In so doing, difficulties have been placed in the way of the development of alternative and cheaper display tubes. These difficulties are not insuperable, as the article on a single-gun beamindexing tube, which starts in this issue, will show, but a strong case can be made<sup>1, 2</sup> for complete independence of brightness and colour information to give more freedom for possible future developments. In other respects the N.T.S.C. standards were a masterpiece of ingenuity and the basic concepts leading to compatibility with ordinary black-andwhite television will no doubt be retained in any future standardization.

At the moment the shadow-mask three-gun tube undoubtedly leads the field, but it is being strongly challenged by single-gun tubes3 of various types which are cheaper to produce, and if proved and adopted could help to remove the present obstacles of high first cost and maintenance. In the running are electro-luminescent and eidophor (oil film light gate) methods, and there may be as yet unknown outsiders working their way up towards the leaders.

On the present showing the B.B.C. has undoubtedly backed a winner in the shadow-mask tube, which even removes the lininess from the 405 standard! But it has also made sure of success by giving it the N.T.S.C. track to run on; other promising runners have at times found the going a little hard. The B.B.C.'s proper function is to provide and maintain the track, not the horses, and the choice of going will be the privilege and the duty of the Postmaster General after hearing the advice of the Pilkington Committee. They have called for evidence and it is now up to industry to bring out all those promising ideas on which they have been working for so long, so that their requirements may be taken into account in arriving at colour standards which are fair to all, and will endure.

" "N.T.S.C. Colour Information," by E. L. C. White. Wireless

World, February, 1957. <sup>2</sup> "An Alternative Colour TV System," by E. J. Gargini, <sup>w</sup>ireless World, August, 1957. <sup>a</sup> "Single-Gun v. Three-gun Tubes: Their Influence on Colour Receiver Design," by R. N. Jackson, *Journal of the Television Society*, April-June, 1960.

### **Beam Indexing Tubes**

1.-AN ALTERNATIVE TO THE SHADOW-MASK PRINCIPLE FOR COLOUR TELEVISION DISPLAYS

#### By IAN MACWHIRTER,\* A.M.I.E.E.

The deficiencies of the shadow-mask colour display tube are outlined and proposals are made for the design of a tube which is completely free from these deficiencies. It is shown that the N.T.S.C. type of colour television signal is not suitable for direct application to the new tube and means are described for transforming the N.T.S.C. signal into a suitable form. The problems of synchronizing the colour signal with the instantaneous beam position are discussed and various solutions are suggested.

Comments are made on the advisability of modifying the formulation of the N.T.S.C. type of signal into a form better suited to the new display.

HE Television Advisory Committee has made no specific recommendations to the Postmaster General in connection with colour television standards.<sup>1</sup> Potential manufacturers of colour display tubes now have an opportunity for investigation of the usefulness of display tubes other than the shadowmask tube. It is clear that had the early introduction of a public colour television service been recommended in the United Kingdom, the manufacturers of colour receivers would almost certainly have had to use a three-gun shadow-mask cathode ray tube. This tube, pioneered by R.C.A., has now been engineered to a standard with which all new developments in single unit colour display tubes will be compared. There are, however, operational shortcomings which are well known and can be briefly stated as follows:-

1. Only a small percentage of beam current reaches the phosphor because of shadow-mask trapping. Result: the maximum useful high-light brightness is limited to some 20ft. lamberts.

2. Because the phosphors have unequal efficiencies and because cathode ray tube guns have powerlaw characteristics, it is not easy to match the three guns in operating conditions. Result: the grey scale requires critical adjustment.

3(a). Each of the three beams must be individually aligned to excite the appropriate phosphor dots exclusively and completely. Result: incorrect purity adjustment will cause a colour shade in the picture both in brightness and in hue.

3(b). The three beams must converge on to one trio of phosphor dots. Failure of this requirement will cause colour fringing around the boundaries of picture components quite independently of the three differential driving voltages applied to the tube.

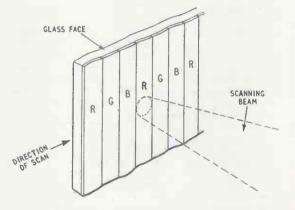


Fig. 1. Enlarged section of screen showing vertical phosphor strips.

4. Presently available tubes have a deflection angle limited to some  $70^{\circ}$ . Result: the tube is significantly more bulky than a  $110^{\circ}$  black and white tube of similar screen area.

A fundamental drawback is that the three-gun shadow-mask tube is difficult and expensive to make. Moreover, the tube must be used with a variety of magnets for convergence, beam-positioning and field purity, all of which help to increase receiver costs.

**Proposed Solution.** All four operational drawbacks may be obviated with the use of a singlegun tube whose beam can scan sequentially a series of tricolour phosphor strips; such strips may be arranged either horizontally or vertically, but this discussion relates primarily to tubes with vertical phosphor strips.<sup>2,3</sup> A steady beam current will

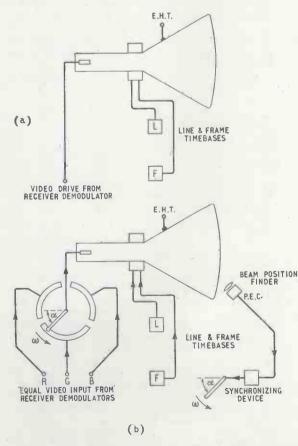
<sup>\*</sup>Associated Electrical Industries, Ltd.

produce equal current excitation of the three phosphor strips during scanning and it is a fundamental requirement of this type of tube that the relative efficiencies of the phosphors should be balanced, so that white light is produced for a steady beam current. Small errors in the white point may conveniently be adjusted by means of external colour-correction filters.

The inevitable decrease in brightness caused by the balancing of the phosphor efficiencies, mainly in the green, can be alleviated if an elliptical spot is used whose minor axis is no greater than the width of a phosphor strip, and whose major axis is such that the scanning lines almost touch, i.e., satisfy the conditions for a "flat field." The brightness of a tube with such a gun can be as good as a modern black and white tube.

Fig. 1 is a drawing of a cutaway portion of the screen of such a tube, but the phosphor strips need not necessarily be contiguous as shown. It is of some importance that should the beam "spread" with increasing video drive, the intended excitation of, say, one primary should not be accompanied by unwanted partial excitation of the other two primaries.

Ideally, there should be a sufficient number of colour phosphor triplets such that the highest frequency component of the video luminance signal would be unable to excite less than one complete triplet of colour phosphors; it is axiomatic, then,



that the average eye would be unable to resolve the individual colour strips at a normal viewing distance. In other words, a normal video signal applied to the tube would appear as a black and white picture, free from purity, convergence and electron-gun differential contrast characteristics. In practice, however, the use of a coarse strip structure appears to simplify some of the circuitry to be described later, and the number of triplets is chosen by a compromise.

In order to show a colour picture, it is necessary that the beam current should be excited with voltages proportional to the colour signal when it is passing over the appropriate colour phosphor. This implies that an indicating or index mechanism is necessary to seek out the instantaneous horizontal position of the beam and to switch on, or "gate," the appropriate colour signal: red, green or blue.

At this stage it is useful to make a comparison between the relative complexities of the following displays, a black and white, a single-gun strip phosphor and a three-gun shadow-mask. Fig. 2 illustrates this comparison. It should be appreciated that the fundamental difference between the threegun shadow-mask display and the single-gun beam indexing display is that the former requires a threestimuli, simultaneous-colour signal, whereas the latter requires a three-stimuli, dot-sequential signal.

The use of a beam position indexing mechanism in a colour receiver creates great problems and some time must now be spent in understanding them.

As an introduction to this, it would be as well to see how a typical band shared colour television signal is matched by the requirements of the threegun shadow-mask display and by a "beam position indexing" display.

A typical colour signal, e.g., the N.T.S.C. signal, consists of two parts: (i) a wide-band black and white, or luminance, signal  $E_{\rm x}'$  and (ii) two bandwidth-limited components proportional to monochromatic colour minus luminance,  $a(E_{\rm B}^{1/\gamma} - E_{\rm x}')$ ,  $b(E_{\rm B}^{1/\gamma} - E_{\rm x}')$  which doubly modulate a carrier in balanced

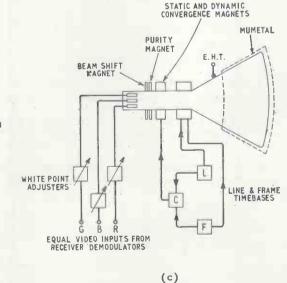


Fig. 2. Basic elements of (a) a black and white display, (b) a single-gun beam sensing display, (c) a shadow-mask display.

modulators, and these constitute the chrominance signal.<sup>4</sup> (The carrier of single frequency is divided into two orthogonal components, i.e., a quarter of a cycle apart, before the modulators). The composition of the luminance signal  $E_{y}'$  is of some impor-tance; starting with three suitable reproducing primaries, the amounts of red, green and blue are proportioned according to their relative contributions to luminance, which give rise to the visual sensation of brightness. This is because the colour difference channels are scaled so that they contribute nothing to brightness; this brightness reproduction is left entirely to the luminance signal. For N.T.S.C.

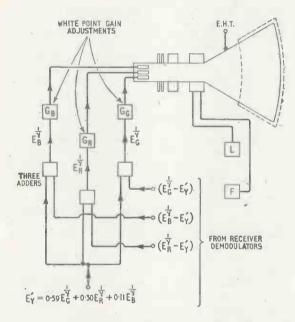


Fig. 3. In a typical display,  $G_{G} = G_{B} = 0.7G_{R}$ , but may vary with different phosphor conversion efficiencies.

primaries and for a normalizing white of 6,500°K,  $E_{x'} = 0.30 \ E_{B}^{1/\nu} + 0.59 E_{0}^{1/\nu} + 0.11 E_{B}^{1/\nu}$ . The colour difference signals  $(E_{B}^{1/\nu} - E_{x})$ ,  $(E_{B}^{1/\nu} - E_{x})$ .

 $E_{y}$  and the derived  $(E_{0}^{1/\gamma} - E_{y})$  when added to the luminance signal  $E_{y}$ , produce colour signals for the display.<sup>4</sup>

Fig. 3 shows a three-gun shadow-mask display and the associated driving voltages derived from a colour signal proportioned to N.T.S.C. specifications.

If colour difference signals only are applied to the display, their relative contributions to luminance will be:

$$\begin{split} \mathbf{Y}_{\mathbf{R}} &= 0.30 \; (\mathbf{E}_{\mathbf{R}}^{1/\nu} - \mathbf{E}_{\mathbf{Y}}') \\ \mathbf{Y}_{\mathbf{0}} &= 0.59 \; (\mathbf{E}_{\mathbf{G}}^{1/\nu} - \mathbf{E}_{\mathbf{Y}}') \\ \mathbf{Y}_{\mathbf{B}} &= 0.11 \; (\mathbf{E}_{\mathbf{B}}^{1/\nu} - \mathbf{E}_{\mathbf{Y}}') \\ \therefore \mathbf{Y}_{total} &= \sum \mathbf{Y}_{\mathbf{EGB}} = 0.30 \; \mathbf{E}_{\mathbf{R}}^{1/\nu} + 0.59 \; \mathbf{E}_{\mathbf{G}}^{1/\nu} + \\ &\quad 0.11 \; \mathbf{E}_{\mathbf{B}}^{1/\nu} - 1.0 \; \mathbf{E}_{\mathbf{Y}}' = 0 \end{split}$$

i.e., the colour difference signals themselves do not

contribute to luminance changes (in a linear system). Therefore, the relative gains in the chrominance channels of the receiver may be proportioned so that an interfering signal present on all three colour difference signals will produce changes only in chromaticity of the display, i.e., constant luminance operation can be achieved. These necessary changes in relative gains are recognized in the formulation of the N.T.S.C. signal.

From the foregoing it will be appreciated that the three-gun shadow-mask tube is well matched to the N.T.S.C. colour signal.

Fig. 4 shows a single-gun beam indexing display and its associated driving voltages from which it will be seen that the N.T.S.C. simultaneous colour signal is transformed into a dot sequential signal by means of gate circuits. Again the display exhibits constant luminance but the three gate circuits must either be very linear in order to preserve the inherent perfect grey scale of the single gun tube or they must gate colour difference signals.5.6

The composition of the luminance signal which results in the smallest voltage swings in the colour difference gates, has equal weightings by red, green and blue.

Thus  $E_{L}' = \frac{1}{3}E_{B}^{1/\gamma} + \frac{1}{3}E_{O}^{1/\gamma} + \frac{1}{3}E_{B}^{1/\gamma}$ . From this it is possible to predict the maximum excursion of the colour difference signals, and how these are matched to the permissible amplitude swing of the video signals in the gates for linear operation. If a saturated primary colour is transmitted, e.g., blue, then  $E_{L}' = \frac{1}{3}$ , since  $E_{R}^{1/\gamma} = E_{0}^{1/\gamma}$ = 0 and  $E_{B}^{1/\gamma} = 1$ ,  $\therefore (E_{B}^{1/\gamma} - E_{L}') = 1 - \frac{1}{3} = 0.67$ . Now let the complementary colour yellow, be transmitted then

$$E_{\rm L}' = \frac{2}{3}, \text{ since } E_{\rm R}^{1/\gamma} = E_{\rm G}^{1/\gamma} = 1 \text{ and } E_{\rm B}^{1/\gamma} = 0 \therefore (E_{\rm R}^{1/\gamma} - E_{\rm L}') = 0 - \frac{2}{3} = -0.67.$$

Thus the peak excursion of the  $(\mathbf{E}_{\mu}^{1/\gamma} - \mathbf{E}_{L})$  signal is from + 0.67 to - 0.67 = 1.33. Because of the composition of the luminance signal, the peak excursion of all three colour difference signals is 33% greater than for single primary signals. This means that the gates must be designed to accept this increase in voltage swing. However, small differential contrast errors will not be objectionable since, for neutrals, the colour difference signals vanish.

A better use of the display tube would be one in which the chrominance signal is applied directly to the tube and so allows the sequential scanning of the colour phosphors to "gate" the chrominance signal at the appropriate time intervals.7 Assuming that the line scanning is linear and that the phosphor strips are of equal width, the separation of vector components of the chrominance signal should be equi-angular rather than that of the N.T.S.C. signal formulations. A suitable chrominance vector signal is shown in Fig. 5, and superimposed is the N.T.S.C. chrominance signal which has been merely redrawn with three vector components.

Two points arise from this new chrominance vector.

1. The angular rotation, i.e., colour switching frequency, is decided by the number of phosphor strips on the screen. This frequency should be a little higher than the sub-carrier (if the visibility of the strip structure is not to be objectionable at a

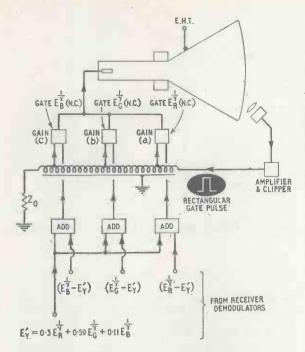


Fig. 4. Essential elements of a single-gun beam indexing display.

normal viewing distance), and it will vary if the line sweep is non-linear, or if the amplitude of the line sweep changes. Means must therefore be provided for synchronizing the new chrominance vector to the colour switching rate.

2. The vectors are shown as having equal amplitudes and the reason for this may not be immediately obvious. It will be recalled that in the design of a three-gun shadow-mask receiver, the gains in the colour channels (either in the three demodulators or in subsequent amplifiers) are such as to achieve constant luminance and the relative amplitudes of the transmitted colour vectors match the receiver requirements. Now the three " demodulators" of the single-gun display are made by the screen structure itself and although the relative "gains," or conversion efficiencies, of the three phosphors may be adjusted so that an interfering signal in the band shared channel will appear with minimum luminance disturbance, the colour of the screen will no longer be white. Because the screen must show white when an unmodulated beam scans it, it is not possible to change the "gains" of the single-gun "demodulator?." Therefore an equiamplitude chrominance vector is necessary.

In order to complement this equi-angular, equi amplitude chrominance signal, a luminance component of the form  $E_{L'} = \frac{1}{3}E_{R}^{1/\nu} + \frac{1}{3}E_{G}^{1/\nu} + \frac{1}{3}E_{B}^{1/\nu}$ is required to provide the residual drive signals,  $E_{R}^{1/\nu}$ ,  $E_{G}^{1/\nu}$ ,  $E_{B}^{1/\nu}$ . (This composition also happens to satisfy the requirement giving the smallest voltage swings in the colour-difference gate circuits.)

It should now be clear that, unlike the three-gun shadow-mask display, the single-gun beam indexing tube when screened with phosphors of balanced efficiency is not at all well matched for *direct* use with the N.T.S.C. formulated signals. Even if the widths and spacings of the three phosphor strips

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were made to match the non equi-angular spacings of the N.T.S.C. chrominance vectors, both constant luminance operation would remain unachievable and there would be incorrect colour rendering.

It is possible to transform the N.T.S.C. luminance and chrominance components into a form suitable for a single-gun display and also permit constant luminance operation to be achieved.<sup>8</sup> (In practice, however, certain simplifications which do not give an exact transformation may be justified on the grounds of economy, even though the resulting colour reproduction will not be exact.)

Now the N.T.S.C. luminance signal is weighted according to the relative luminosities of the specified primaries, i.e.,

$$E_{y'} = 0.30E_{p}^{1/\gamma} + 0.59E_{q}^{1/\gamma} + 0.11E_{p}^{1/\gamma}$$

but for the single-gun beam indicating tube, the required luminance signal  $E_{L}$ ' is given by

$$E_{L}' = \frac{1}{3}E_{R}^{1/\gamma} + \frac{1}{3}E_{A}^{1/\gamma} + \frac{1}{3}E_{R}^{1/\gamma}$$

The difference  $(E_{\rm L}'-E_{\rm Y}')$  vanishes for a neutral colour, i.e.,  $E_{\rm L}'=E_{\rm Y}'$ 

$$\therefore (\mathbf{E_{L}'} - \mathbf{E_{Y}'}) = \frac{1}{3}(\mathbf{E_{R}^{1/\gamma}} - \mathbf{E_{Y}'}) + \frac{1}{3}(\mathbf{E_{P}^{1/\gamma}} - \mathbf{E_{Y}'}) + \frac{1}{3}(\mathbf{E_{P}^{1/\gamma}} - \mathbf{E_{Y}'})$$

From the N.T.S.C. colour signal formulation,  $(E_{\alpha}^{1/\gamma} - E_{\gamma}') = -0.51 (E_{\gamma}^{1/\gamma} - E_{\gamma}') -$ 

$$0.19 (E_{\rm L}^{1/\gamma} - E_{\rm Y}) = 0.33 (E_{\rm B}^{1/\gamma} - E_{\rm Y}) + 0.33 (E_{\rm B}^{1/\gamma} - E_{\rm Y}) - 0.51 (E_{\rm B}^{1/\gamma} - E_{\rm Y}) - 0.19 (E_{\rm B}^{1/\gamma} - E_{\rm Y}) = 0.16 (E_{\rm B}^{1/\gamma} - E_{\rm Y}) + 0.26 (E_{\rm B}^{1/\gamma} - E_{\rm Y}) : E_{\rm Y} = E_{\rm Y}' + 0.16 (E_{\rm B}^{1/\gamma} - E_{\rm Y}) + 0.26 (E_{\rm B}^{1/\gamma} - E_{\rm Y})$$

 $\therefore E_{L} = E_{Y} + 0.10 (E_{R} - E_{Y}) + 0.20 (E_{B} - E_{Y})$ Rewriting this in terms of N.T.S.C. colour difference signal weightings,

$$E_{L}' = E_{Y}' + 0.18 \frac{(E_{R}^{1/\gamma} - E_{Y}')}{1.14} + 0.58 \frac{(E_{B}^{1/\gamma} - E_{Y}')}{2.03}$$

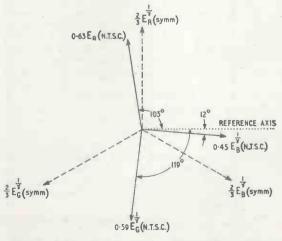


Fig. 5. Equi-angle and N.T.S.C. chrominance vectors.

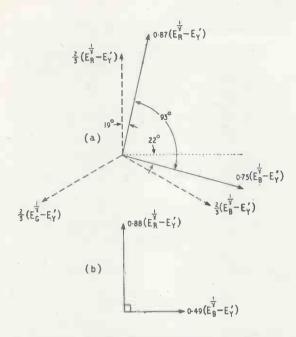


Fig. 6(a). Transformation of equi-angle vectors to twocolour difference vectors, (b) Orthogonal N.T.S.C. chrominance vectors.

i.e., the correcting signal  $(E_{L}' - E_{Y}')$  which must be added to  $(E_{Y}')$  can be derived by demodulating the N.T.S.C. chrominance signal with a gain of  $\sqrt{0.18^2 + 0.53^2} = 0.56$  and at an angle of  $\tan \frac{^{-1}0.18}{0.53} \simeq 19^\circ$  i.e.,  $0.56 \angle 19^\circ$ .

It is of interest to note that the luminance correcting signal is derived directly from the two colour difference chrominance components which have been band limited. These components, i.e.  $(E_L' - E_Y')$  in no way impair luminance detail since, during an achromatic transition in picture information,  $(E_L' - E_Y') = 0$  and the transition is described by  $(E_Y')$  only. The N.T.S.C. chrominance signal can be trans-

The N.T.S.C. chrominance signal can be transformed into an equi-angle, equi-amplitude signal and Fig. 6 (a) shows the redrawing of the required symmetrical vectors into two colour difference vectors. In order to facilitate this transformation, a signal  $(-\frac{3}{2}E_{y})$  has been added to each vector. These should be compared with the available

vectors. In order to facilitate this transformation, a signal  $(-{}_{3}^{2}E_{y}')$  has been added to each vector. These should be compared with the available N.T.S.C. colour difference vectors in Fig. 7, from which it will be seen that whilst the  $(E_{B}^{1/\gamma} - E_{y}')$ vector is greatly increased in amplitude, the  $(E_{B}^{1/\gamma} - E_{y}')$ vector is slightly reduced. In addition, the new vectors are no longer orthogonal, but are separated by an angle of 93°.

This vector transformation is conveniently accomplished with an amplifier whose gain is varied with the second harmonic of sub-carrier frequency,  $(\omega_c)$ . Let the gain  $G = a\cos(2\omega_c t + \theta)$  and the input signal be  $E_c = b\cos(\omega_c t + \phi)$ . Then output =  $GE_c = ab\cos(2\omega_c t + \theta)\cos(\omega_c t + \phi)$ .

$$= \frac{ab}{2} \left\{ \cos \left( 3\omega_{e}t + \theta + \phi \right) + \cos \left( \omega_{c}t + \theta - \phi \right) \right\}$$
  
Filtering out all components above  $(\omega_{c})$ 

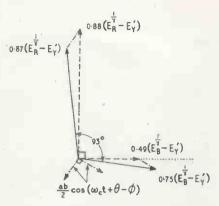


Fig. 7. Transformation of N.T.S.C. colour difference vectors.

output  $= \frac{ab}{2}\cos(\omega_c t + \theta - \phi)$ . Clearly, this

includes the input signal with a reversed phase sequence  $(-\phi)$ , and this may be added in a suitable proportion to the unmodified N.T.S.C. chrominance signal, see Fig. 7. It is suggested that this proportion and the phase  $(\theta)$  of the frequency-doubled subcarrier be adjusted when observing the output signal on a calibrated vectorscope. A suitable test signal consists of colour bars containing primaries, complementaries and a neutral. In practice it has been found possible to set the vector angles to within 2 or 3 degrees of their correct value.

It will be appreciated that a beam indexing display when used with a N.T.S.C. type of signal needs to have rather complex signal translation apparatus in order to give correct colour rendition. There is, however, a further colour error for which correction should be provided. Fig. 8(a) shows a diagrammatic section of a typical beam indexing tube screen and a chrominance signal corresponding to a saturated blue field (the luminance value of the blue field is not considered here). Clearly, there will be considerable excitation of the adjoining red and green phosphors, and this will cause desaturation of the intended pure blue light. Although this defect does not cause serious distortion of pastel colours, the loss of saturation on highly coloured objects is rather objectionable and correction for this can take one of three forms.

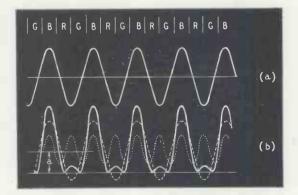


Fig. 8. (a) De-saturation of pure blue field. (b) Effect of adding 2nd harmonic term.

(i) The phosphor strips may be spaced with inert bands between adjacent colours, so that the conduction angle of the phosphors will be reduced.

(ii) The conduction angle can be reduced electrically by deducting from the luminance signal a signal proportional to chrominance carrier amplitude.<sup>9</sup> This idea is based on the fact that whilst the phase of the chrominance signal is proportional to saturation. An envelope detector will generate a signal proportional to saturation only, and this can be used to control the luminance amplitude. The inevitable decrease in brightness can be corrected by an increase in the gain of the chrominance channel.

(iii) The most satisfactory method for saturation correction requires the use of a second harmonic chrominance component which is added to the original chrominance signal<sup>10</sup>. In Fig. 8(b) is shown the effect of this where the added signal is of the form  $a + b\cos(2\omega_{\mathbf{R}}t + \phi)$  where  $(\omega_{\mathbf{R}})$  is the colour switching rate.

#### (To be concluded)

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<sup>6</sup> K. G. Freeman, "A Gating Circuit for Single-gun Colour Television Tubes," Brit. I.R.E. Television Convention 1959, pre-print page 5. Also in Journal Brit. I.R.E., Nov. 1959, Vol. 19, No. 11, p. 667 et seq.

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### **Silicon Carbide Transistor**

#### NEW HIGH-TEMPERATURE DEVICE

**DESEARCH** scientists of the Westinghouse Electric Corporation in the United States have recently developed a transistor capable of operating above 650 degrees Fahrenheit, a temperature higher than the melting point of lead. The new transistor is the first to be successfully made from silicon carbide, a hard crystalline material which, in impure form, is used as an abrasive in grinding wheels.

The high-temperature capabilities of the new transistor mark it as a significant advancement in the technology of these semiconductor devices. Presentday transistors, manufactured almost exclusively from germanium and silicon, can operate at temperatures no higher than about 200 degrees F (ger-manium) or 400 degrees F (silicon). Germanium and silicon transistors, however, cannot always meet the high-temperature requirements of today's existing and planned aircraft and space vehicles. Such applications, therefore, have furnished strong motivation for the development of higher-temperature transistors. Because of its great chemical stability and desirable electrical properties which it retains at elevated temperatures, silicon carbide is one of the most promising transistor materials for extremely high-temperature applications. Laboratory tests show that the new silicon carbide transistor still amplifies at 670 degrees F, and with further development, an upper operating temperature of more than 925 degrees F should be achievable.

The new device is actually a "unipolar" or "fieldeffect" transistor, which differs in operating prin-

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ciple from those usually made from germanium and silicon. Such conventional transistors regulate the flow of an electric current through them by the injection of electric charge carriers across a junction built into the semiconductor material. The unipolar transistor, on the other hand, acts more like a valve which opens and closes to regulate the electron flow.

By MICHAEL LORANT

The new transistors are made from exceptionallypure crystals about two-thousandths of an inch thick. The necessary junction is built into the material by exposing it to vaporized aluminium at the white-hot temperature of 3,900 degrees F. The aluminium atoms diffuse into the silicon carbide crystal, changing its electrical behaviour from socalled n-type material to p-type. The junction is formed where the two types meet, and the process is controlled to an accuracy of a few millionths of an inch.

Then, to establish the input and output terminals of the transistor, the wafer is etched at two points in such a way that the silicon carbide is eaten away until the junction within the body of the crystal is reached. Electrical connections at these two points and to the body of the wafer complete the transistor.

A typical finished transistor is about 80-thousandths of an inch long and 40-thousandths of an inch wide, and the "working" area of the crystal surface is smaller than the head of a pin. Electrical measurements on the finished transistors show them to give a power gain of about 60 at room temperatures.

# Electronic Telephone Exchanges

#### RECENT DEVELOPMENTS

DISCUSSED AT THE I.E.E.

T

HE papers presented at the Institution of Electrical Engineers' Conference on Electronic Telephone Exchanges on November 22 and 23 were contributed by experts from Europe, Asia and America and covered a very wide field in considerable detail. They will be reprinted in full in 1961 in Part B of the Proceedings, and the following notes are intended only as an introduction for those who may wish to study the subject further.

#### Time Division

In Great Britain, feasibility studies of the design of a completely electronic exchange have been carried out. The studies, undertaken independently by the Post Office and five principal manufacturers of telephone equipment, and later by the Joint Electronic Research Committee, have resulted in the experimental exchange shortly to be installed at Highgate Wood, in London.

In order to provide consistency of service to subscribers during the experimental period, a complete electromechanical exchange will be installed. If, at any time, the electronic exchange is removed from service to allow modification, the mechanical system will function in its place, and in fact will remain permanently in service when the experiment reaches its conclusion.

The electronic system employs the time-division multiplex method of switching, each unidirectional "highway" carrying 100 channels. Each connection requires two antiphase channels,  $\frac{1}{2}$  cycle (50µsec) apart, and is established by switching line and interhighway diode gates by means of channel pulses, which are obtained from a channel store consisting of a circulating delay-line of the magnetostriction variety.

The central control equipment of the exchange is also, in the main, composed of magnetostriction delay-lines, which employ channel pulses to set up and supervise the progress of calls.

The speech-path channels are amplitude-modulated, the slight loss encountered during transmission through gates being made good by amplification in the demodulator.

Permanent control information, such as directory

When the design of the experimental exchange commenced, it was thought that transistors would restrict the transmission performance of the system and thermionic valves and semi-conductor diodes were chosen. It is reasonable to suppose that advances in semiconductor research have brought an exclusively solid-state exchange within the bounds of possibility.

#### **Space Division**

An American experimental exchange, the Morris Electronic Central Office, is described in a paper by Keister, Ketchledge and Lovell. This exchange, designed by Bell Telephone Laboratories, employs a space-division system using gas-filled diodes as switching elements. The diodes have a negativeimpedance characteristic which offsets loss in transformers and other elements in the transmission path.

Permanent control information is stored in a flying-spot photographic store, while transient information is held on barrier-grid tubes.

The more important sections of the equipment are duplicated, while fault detection and, in certain cases, correction are automatic.

Novel facilities provided by the exchange include the possibility of obtaining frequently called numbers by dialling two digits only, and of connecting extension telephones, via the exchange, in an intercommunication circuit.

Switching Elements used in modern space-division systems fall broadly into two classes, electronic and electromechanical. Representative of the electromechanical switch are the reed relay, and a development known as the "ferreed."\* The reed relay consists of contacts mounted on two metallic strips or reeds, the whole being mounted in a sealed glass tube. Surrounding the tube is a solenoid carrying the control current, which magnetizes the reeds and brings them into contact. Long life is a feature of this type of switch. In the ferreed, the reeds form part of a magnetic circuit, in part of which is ferrite material. The remanence of the ferrite is capable of maintaining closure of the contacts in the absence of controlling current.

Electronic switches include the neon-filled diode and the p-n-p-n junction diode. Both of these devices have the slight negative slope to their characteristics referred to earlier. The junction diode requires a much lower operating voltage than the neon diode, heat-dissipation problems with both these devices being negligible.

Exchanges employing time-division and spacedivision multiplexing are under active development in many parts of the world, and it may reasonably be assumed that one or more systems, both economically and technically competitive with the electromagnetic exchange, will have proved themselves within a few years.

\* The Ferreed-A New Switching Device. A. Feiner et al. B.S.T.J. Vol. XXXIX. No. 1.

numbers, and transient information such as state of line, is stored on a magnetic drum. The drum is a magnetically coated cylinder with reading and writing heads. In the case of the permanent information, the writing head is normally isolated from its amplifier, to prevent accidental mutilation of the information. The type of permanent store employed lends itself to modification as a service expands.

### PHYSICAL SOCIETY EXHIBITION

MANUFACTURERS AND RESEARCH ESTABLISHMENTS EXHIBITING

THE 45th show of scientific instruments and apparatus which has become known as the Physical Society Exhibition will open at the Royal Horticultural Society's Halls, Westminster, on January 16th for five days. This year, however, the exhibition assumes a new official title, the Annual Exhibition of the Institute of Physics and the Physical Society, because of the recent amalgamation of the two organizations.

On a very large proportion of the 150 stands there will be equipment of interest to radio and electronic engineers and, as in the past, the emphasis will be on new developments in instruments and research techniques as well as on standard instruments and equipment.

From 10.30 to 2.0 on the opening day admission is limited to members of the society and the Press. On succeeding days the exhibition will be open to ticket holders

A.E.I. Admiralty Research Estabs. Advance Components Airmec Archæology and the History of Art, Research Lab. Avo

Baird & Tatlock Baker, C., Instruments Baldwin Industrial Controls Barr & Stroud Beck, R. & J. Bellingham & Stanley Birmingham University Divisit Long & Steal Research British Iron & Steel Research Assoc

C.N.S. Instruments \* Cambridge Instrument Co. Casella, C. F., & Co. Cawkell Cooke, Troughton & Simms Cossor Instruments

D.S.I.R. Dawe Instruments Decca Radar Distillers Co. McInnes Dobbie McInnes Doran Instrument Co. Dynatron Radio

E.M.I. Electronics Edwards High Vacuum Ekco Electronics **Electro** Methods Electronic Instruments Electronic Tubes Electrothermal Engineering Elliott Brothers English Electric Valve Co. Ericsson Telephones Evans Electroselenium

Ferranti Flann Microwave Insts. Fleming, J. & R. Fleming Radio Furzehill Laboratories

G.E.C Gallenkamp, A., & Co. General Radiological Griffin & George Guy's Medical School

Hilger & Watts

International Computers & Tabulators Isotope Developments

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Joyce, Loebl & Co.

Kelvin & Hughes Labgear Lintronic Locarte Co. Lucas, Joseph

M-O Valve Co. Marconi Instruments Marconi's W/T Co. Marshall of Cambridge Megatron Mervyn Instruments Metals Research Ministry of Aviation Morgan Crucible Co. Muirhead & Co. Mullard

N.G.N. Electrical N.P.L. & Post Office Engg. Nagard Nash and Thompson National Res. Dev. Corp. New Electronic Products Newport Instruments Northern Polytechnic Nuclear Enterprises (G.B.)

Oertling, L. Oliver & Boyd Optica United Kingdom Optical Works Ottway, W., & Co. Panax Equipment Paton Hawksley Electronics Perkin-Elmer Planer, G. V. Plessey Co. Post Office Engineering Dept. Prior, W. R., & Co. Pullin, R. B., & Co. Pye, W. G., & Co.

Racal Instruments Rank Cintel Reading University Royal College of Surgeons Royal Meteorological Society **Royston Instruments** 

S.T.C. Sanders, W. H. (Electronics) Sangamo Weston Science Museum Servomex Controls Shackman, D., & Sons Singer Instruments Co. Solartron Electronic Group Solus-Schall Stanton Instruments

at 10 a.m. It will close at 7.0 on the 16th, 18th, and 19th, at 9.0 on the 17th and 1.0 on the last day. Tickets are obtainable from exhibitors or from 47 Belgrave Square, London, S.W.1.

Wireless World, together with our associate journal Electronic Technology, are among the publications which have taken space in the exhibition.

The following demonstration-lectures will be given at 5.45 on the middle three days: "Hydrodynamic Research" by F. S. Burt of the Admiralty Research Laboratory (17th); "The Physics of the Oceans" by Dr. G. E. R. Deacon of the National Institute of Oceano G. E. R. Deacon of the National Institute of Oceano-graphy (18th); and "Some Physical Problems in Travel-ling at Supersonic Speed" by Dr. F. P. Bowden of Cambridge University (19th). The Acoustics Group is holding a symposium on traffic noise at 2.0 on the 18th.

Stanley, W. F., & Co. Sunbury Glass Works

Taylor, Taylor & Hobson Techne (Cambridge) Telcon Metals Texas Instruments Thermal Syndicate Thompson, J. Langham Thom Electrical Industries Tinsley, H., & Co. Tintometer Towers, J. W., & Co.

Townson & Mercer 20th Century Electronics

U.K.A.E.A. Ultrasonoscope Co. Unicam Instruments

Venner Electronics Vinten, W.

Watson, W., & Sons Wayne Kerr Laboratories Wray (Optical Works)

#### Industrial Groups—IV

WITH the growing practice of diversification in industry new names are coming into the field of radio and electronics as companies or groups are merged with concerns A case in point is Metal Industries, Ltd. The M.I. group which now comprises 38 companies in the U.K. and overseas and employs over 12,000 came into our field with the acquisition in 1959 of Avo (formed in 1923) and its subsidiary Taylor Electrical Instruments (1938). The group had been on the fringe of our field since its acquisition of several electrical concerns including Brookhirst Switchgear and Igranic Electric, now known as Brookhirst Igranic.

The group recently doubled its size with the acquisi-tion of Lancashire Dynamo Holdings. Among the Lancashire Dynamo companies is Lancashire Dynamo Electronic Products and also the International Rectifier Company (Great Britain) which was formed in May, 1959, jointly by L.D. and the International Rectifier Corporation of Los Angeles.

The Metal Industries group now includes:-

Avo Brookhirst Igranic Cable Jointers Cox & Danks Crypto Crypto Crypton Equipment Dynamo & Motor Repairs Farmer Brothers (Shifnal) Fawcett Preston & Company Foster Electrical Supplies Foster Transformers Hughes Bolckow J. G. Statter & Company John Allan & Company John Allan & Company (Glen-park) Lancashire Dynamo & Crypto Lancashire Dynamo Electronic Products Lancashire Dynamo Nevelin Malcolm & Allan (London) Metal Industries (Salvage)

Minerva Mouldings

New Eagle Foundry Company Shipbreaking Industries Taylor Electrical Instruments Towler Brothers (Patents) Overseas Companies Brookhirst Igranic S.A. (South Africa) Fawcett Preston (Europe) S.A. (France) Lancashire Dynamo Central Africa (Rhodesia) Lancashire Dynamo South Africa M.I. Australia Metal Industries Europe S.A. (Belgium) Olaer France S.A. Associated Companies Bepco Canada International Rectifier Company (G.B.) oc. Representation d'Appareils

Soc. Representation Mecaniques (France)

### **Scientific Radio Conference**

#### U.R.S.I. AND INTERNATIONAL RADIO MEASUREMENTS

By R. L. SMITH-ROSE, C.B.E., D.Sc., F.C.G.I., M.I.E.E.

**D** URING, September, the International Scientific Radio Union held its XIIIth General Assembly at University College, London, when over six hundred scientists and engineers from twenty-four countries discussed various fields of scientific research ranging from precision laboratory measurements to the propagation of radio waves and communications in space. In his opening address, Dr. L. V. Berkner, the president since 1957, briefly reviewed the work of the Union over the past three years. It was during this period that the outstanding programme of the International Geophysical Year (1957/58) had been completed according to the carefully arranged plan, and this was followed by a year of International Geophysical Co-operation in 1959.

Since its formal organization in 1921 as an international scientific union, U.R.S.I. (the initials of its French title "Union Radio Scientific Internationale") has taken a keen interest in the characteristics of the atmosphere which determine the transmission of radio waves, and of the influence outside the atmosphere which determine these characteristics. It was thus perhaps a natural consequence that as the president remarked, U.R.S.I. was the first union to express its confidence in the scientific benefits that would result from artificial satellites in orbit around the earth. It is significant to note that the spectacular developments in the launching and use of such satellites and other vehicles in outer space have all taken place in the past three years. To deal with the scientific aspects of this work, an Space International Committee on Research (C.O.S.P.A.R.) was formed with U.R.S.I. as a charter member and having direct representation.

During the past triennium the allocation and protection of frequency channels for the use of radio astronomers has been actively discussed. With the coming of artificial earth satellites and their associated radio transmissions, this question has become of much greater importance. Although some frequency assignments were made for research purposes at the Geneva conference of the International Telecommunication Union at the end of 1959, it has been considered necessary to explore the future needs much more thoroughly; and at the recent U.R.S.I. General Assembly, representatives of U.R.S.I. and C.O.S.P.A.R., together with members of the International Astronomical Union, agreed to set up an inter-union committee for this purpose.

#### Scientific Programme of U.R.S.I.

The work of the International Scientific Radio Union is divided among seven commissions, dealing with individual portions of the field as indicated by their titles. These, together with the names of the chairmen appointed in September for the current three-year period (1960-63), are as follows:---

I Radio measurements and Dr. U. Adelsberger (Gerstandards many)

II Radio and troposphere

III Ionospheric radio IV Radio noise of terrestrial

origin

V Radio astronomy VI Radio waves and circuits

VI Radio waves, and circ VII Radio-electronics Prof. A. C. B. Lovell (U.K.) Dr. J. Loeb (France) Prof. W. G. Shepherd (U.S.A.)

Prof. R. A. Helliwell (U.S.A.)

many) P. Voge (France) J. A. Ratcliffe (U.K.)

At the General Assembly in London, the programme was arranged so that all these commissions met concurrently to discuss subjects which had been selected. A few joint sessions were also held to deal with subjects of common interest to two or more commissions. Outstanding among these was a morning session for the whole assembly allotted to the radio aspects of space research. Several papers described experiments made with rockets and satellites for studying the ionosphere and outer atmosphere as well as cosmic noise and solar ionizing radiations. Other papers stimulated a discussion on the scientific and technical aspects of communications by means of active and passive satellite relay systems. One of these papers outlined a proposal to discharge from a satellite large quantities of small metallic filaments into an orbital belt of microwave-resonant dipoles, which would reflect radio waves for long-distance communications purposes. This proposal was viewed with some misgivings by both optical and radio astronomers, since the existence of a permanent reflecting or scattering belt might seriously impair future scientific research in the astronomical field.

At the closing plenary session, the individual chairmen reported on the activities of their commissions and working parties; and put forward resolutions and recommendations for adoption by the Union. The individual papers presented at the meetings of all commissions, together with reports of the scientific discussions, will be published as a series of U.R.S.I. monographs in the near future. In the meantime, a brief review of some of the items discussed will be given in this and succeeding contributions.

It was at the closing session also, that the following officers were elected for the next triennium.

President:	Dr. R. L. Smith-Rose (U.K.)
Vice-Presidents:	B. Decaux (France)
	Professor I. Koga (Japan)
	A. Prochorov (U.S.S.R.)
	Professor G. A. Woonton (Canada)
	Professor Ch. Manneback (Belgium)
The Headquarters	of the Union are at 7 Place Emile Danco.
Brussels 18, Be	elgium, and the secretary-general is

Brussels 18, Belgium, and the secretary-general is E. Herbays.

These, together with the commission chairmen listed above, will prepare for and organize the next General

Assembly to be held in Tokyo in 1963, when the Union will be celebrating its Jubilee. It grew out of the International Provisional Commission of Scientific Wireless Telegraphy formed in 1913.

From the earliest days of U.R.S.I., one of its most important activities has been to establish agreement throughout the world on the accuracy of measurement of such quantities as frequency, power and field-strength, which form the basis of all international research in the radio field.

As a result of intensive research and development over the past quarter of a century in several national laboratories, the frequency of an alternating current can now be measured to an accuracy which surpasses that of any other type of physical measurement. After passing through phases when a temperaturecontrolled tuning-fork and, later, a specially cut quartz crystal, formed the essential element in a standard frequency source, one of the resonant modes of a cæsium atom has now proved to be the most accurate and stable basis of reference.<sup>1</sup> The stability of radio-frequency oscillators controlled by a cæsium atomic standard has been shown to be of the order of one or two parts in 10<sup>10</sup>.

Direct comparisons have been made in the same laboratory between the standard developed at the National Physical Laboratory in this country and the corresponding U.S.A. standard. The results<sup>2</sup> showed that the standards of the two countries agreed to within about 2 parts in  $10^{10}$ , and that part of this discrepancy could be attributed to the different electronic arrangements used to drive the cæsium oscillators.

In order to state the actual frequency of such an oscillator, it is necessary to specify what standard of time is to be used. It is convenient for many purposes to adopt a uniform scale of time; but for precise scientific work, it is now internationally agreed that ephemeris time, which is based on the revolution of the earth round the sun, should be used. As a result of a joint programme of measurements carried out between the National Physical Laboratory in this country and the United States Naval Observatory, Washington,<sup>3</sup> the mean value of the frequency of the cæsium oscillator was found to be 9 192 631 770 cycles per second, with a probable error of  $\pm 20$  c/s (Ephemeris Time). At the recent U.R.S.I. General Assembly, it was recommended that the Bureau International de l'Heure should keep under review the appropriate astronomical observations and announce a nominal value of this frequency for use during the following year.

In order to make such a standard available to users in various laboratories, programmes of international standard frequency transmissions have been established in several countries on various frequencies between 2.5 and 25Mc/s. In this country, the transmissions are emitted by a G.P.O. station at Rugby under the call sign MSF; and they are now supplemented on an experimental basis with a much lower frequency (16kc/s) from Rugby (GBR). These signals are measured at the N.P.L. by reference to the cæsium standard, and the values of the deviations from the nominal frequency are published monthly in *Electronic Technology*.

At a meeting of Commission I of the General Assembly of U.R.S.I., a review of the present position of "Standard Frequencies and Time Signal Transmissions" was introduced by U. Adelsberger (Germany), who gave particular attention to the problem of modulating or interrupting the continuous wave emissions in order to produce accurate time signal ticks. The Commission recommended that the day-to-day phase stability of the very lowfrequency transmissions should be measured in various countries in order to assess their usefulness for high precision time synchronization and frequency comparison purposes.

Another radio measurement of reasonably high accuracy is that of the velocity of electromagnetic waves in a vacuum. At the XIIth General Assembly at Boulder, U.S.A., in 1957, a value of 299,792.5  $\pm 0.4$  km/s was adopted; and this has been found to give consistent results in the measurement of geodetic distances by electronic means. At the last Assembly it was recommended that agreed formulæ for the refractive index of the atmosphere should be adopted for both light and radio frequencies, so that the value of the velocity can be corrected for the appropriate ambient conditions, which prevail in practical use.

Considerable attention has been given by U.R.S.I. at previous meetings to the specification of field strength measuring techniques and their inter-national comparison. As the frequencies in use have increased into the microwave region, it has become customary to specify a received signal in terms of power flux rather than field strength; and a recommendation was made in 1957 that the national laboratories should compare their standards of power measurement in the neighbourhood of 3,000 and 10,000Mc/s. On this occasion, a paper by J. A. Lane (U.K.), presented the results of measurements made at the Radio Research Station, Slough, at a frequency of 9.375 Gc/s (wavelength 3.2cm) on wire and film bolometers submitted by the National Bureau of Standards, Boulder, U.S.A., and by the University of Tokyo, Japan. The measurements were made in several ways, including one using a constantflow water calorimeter in which the microwave power was dissipated.<sup>4</sup> The general conclusion from these comparisons was that the techniques used in the three participating countries agree within the limits of their estimated accuracy, about 1.5 or 2.0%.

In view of the general trend of radio research and application to increasingly higher frequencies, the U.R.S.I. General Assembly recommended that this work should be extended to 140 Gc/s; and that similar experimental comparisons should be made on other electrical parameters, such as voltage, attenuation, impedance and voltage standing-wave ratios in waveguides. Such measurements are clearly fundamental to the design of terminal radio equipment as well as to the study of radio wave propagation.

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(b) J. A. Lane. "Transverse Film Bolometers for

(b) J. A. Lane. "Transverse Film Bolometers for the Measurement of Power in Rectangular Waveguides." *Proc. I.E.E.*, 1958, Vol. 105, Part B, p. 77.

#### Audio Festival

PLANS are well in hand for the International Audio Festival and Fair being organized by a committee of audio manufacturers for April 6th to 9th at the Russell Hotel, London, W.C.1. Over 50 British manufacturers are subscribers to this international show at which quality of reproduction will be the keynote.

The following have been elected to form the council: —H. J. Leak chairman, R. Merrick (Ferrograph) vice-chairman, R. Arbib (Multimusic), D. Chave (Lowther), Major J. F. E. Clarke (Clarke & Smith), H. Farquharson (Armstrong), H. S. Futter (Gramophone Co.), L. B. Livingstone (Tannoy), J. Maurice (Lustraphone), J. Rogers (Rogers Developments), L. Stone (E.A.R.), L. Smith (Wharfedale), J. Swift (Goodmans), P. Walker (Acoustical), and L. Young (Vitavox). G. A. Briggs (Wharfedale) and J. Maurice have been appointed Trustees.

In addition to the usual stands and individual demonstration rooms for the 60 or more exhibitors, plans are being made for special lecture sessions covering reproduction from disc, tape and radio. Admission to the exhibition will be by ticket obtainable free from exhibitors and audio dealers.

Further details are obtainable from C. Rex-Hassan, 42 Manchester Street, London, W.1.

#### **Television** Afloat



The new 42,500-ton P. & O. luxury liner Oriana, now on her maiden voyage, has a television installation which provides not only a closed-circuit service but also for the reception of 405-, 525- and 625-line transmissions. The incoming television signals are, where necessary, converted to 525 lines for display on Ekco receivers and, moreover, where alternative programmes are available the viewer can make his own choice. Initially 60 receivers have been installed in public rooms and first-class cabins but provision is made for nearly 400. The television system was installed by Marconi's who also provided the ship's communications equipment, navigational aids and sound reproducing equipment incorporating 1,600 loudspeakers.

#### **Radio** Research

MANY interesting aspects of the research programme initiated by the Radio Research Board and carried out by the Radio Research Station at Slough are given in "Radio Research 1959"\* which contains the report of the Board and of the Director of Radio Research. As might be expected the advantages to research in radio propagation afforded by rockets and satellites is given more than a passing reference. O. W. Humphreys, chairman of the Board, stresses that in order to obtain the full effect from the opportunities offered, the Radio Research Station "should devise its own experiments for flying in rockets and satellites." The proportion of the station's effort devoted to the general field of space research is being increased to 50%.

The present members of the Board are: --Prof. H. E. M. Barlow (University College, London), Prof. W. J. G. Beynon (University College of Wales), Dr. R. L. F. Boyd (University College, London), Prof. E. C. Cherry (Imperial College, London), O. W. Humphreys (G. E. C. Research Labs.), Dr. F. E. Jones (Mullard), K. I. Jones (Ferguson), C. J. V. Lawson (Cable & Wireless), F. C. McLean (B.B.C.), A. W. Montgomery (S.T.C.), Capt. J. S. Raven (Admiralty) and Capt. F. J. Wylie (Radio Advisory Service).

The 22-page report of Dr. R. L. Smith-Rose is of particular interest as it is his last as Director of Radio Research; he is succeeded by J. A. Ratcliffe. \*H.M.S.O. 3s.

College of Aeronautics.—For the past year or so the College of Aeronautics, Cranfield, Bletchley, Bucks., has been pursuing a policy of diversification of teaching activities, and as a result a number of new advanced courses have been developed in subjects which although relevant to aeronautics are of equal interest to many other branches of engineering. It has therefore been decided to change the titles of some of the departments thereby giving a better indication of the scope of their activities. Under this re-arrangement the Department of Aircraft Electrical Engineering becomes the Department of Electrical and Control Engineering. Professor G. A. Whitfield has been head of the department since its formation in 1955. In addition to providing specialized courses for aeronautical engineers as part of the two-year diploma course the department is now offering advanced courses in industrial control engineering and flight control as well as courses in space technology and guided missiles.

Satellite Tracking.—Two of the chain of 18 stations to be used for tracking the first American manned space satellite—those at Kano (Nigeria) and Zanzibar—are to be staffed by Cable & Wireless men for maintaining point-to-point communications. The stations will be linked by radio to London whence information will be passed to the U.S.A. via the transatlantic cable. W. A. Coslett, who during the war was engineer-in-charge of one of the mobile telecommunication units known as Blue Trains and until his retirement in 1959 at the age of 55 was C. & W. manager in Jamaica, has been appointed engineer-in-charge at Kano. The assistant engineers are J. D. Munday and T. Shepherd. At Zanzibar, D. G. Hicks will be E.-in-C. with P. J. Harlow and D. J. Payne as assistants.

S.O.N.D.E.- The Society of Non-Destructive Examination is to hold the first Croxson Memorial Lecture (in memory of the late Charles Croxson the founder chair-man) on February 17th at 6.15 in the Caxton Hall, Westminster. Dr. L. Mullins will speak on "The Evolution of Non-Destructive Testing". Free tickets— not more than two per applicant—are obtainable from the Hon. Secretary of the Society, D. T. Carter, E.S.A.B. Ltd., Gillingham, Kent.

Computer Consortium.-The reconstituted British Conference on Automation and Computation, of which Sir Walter Puckey is chairman, has set up three panels to cover the following aspects of its interests:-education to cover the following aspects of its interests:—education and training (with Professor G. D. S. MacLellan, of Glasgow University, as chairman); research and develop-ment (J. F. Coales, Cambridge University); and public relations (W. C. F. Hessenberg, British Iron and Steel Research Association). The general title of the con-ference of the B.C.A.C. to be held in Harrogate from June 27th to 30th is "Automation-Men and Money"

Television Society .--- Sir Harold Bishop, Director of Engineering, B.B.C., has accepted the invitation of the Television Society to become president in succession to the late Sir George Barnes. He took office on December 8th for two years. The Society, which was formed in 1927 "for the furtherance of research in television and allied problems" now has a membership of over 1,200.

R.S.G.B. Membership .--- For the fourth successive year after a period of recession the membership of the Radio Society of Great Britain had again increased at the end of June, 1960, when it passed the 10,000 mark. The year's increase recorded in the Society's annual report was 496 bringing the total to 10,036. Of the U.K. total of 8,729 holders of amateur (sound) licences, 6,473 (60%) were members of the Society.

Annual Dinner of the Royal Flying Corps Wireless Operators Old Comrades Association will be held in London on March 18th. Details are obtainable from E. J. F. C. Hogg, 57 Hendham Road, London, S.W.17.

Police Radio-Features of this latest motor-cycle f.m. transmitter-receiver, produced by G.E.C. to a Lancashire Police design, are a handlebar press-totalk switch, loudspeaker .unit on the petrol tank and microphoneearpiece in the helmet.

African Radio Union.-Broadcasting organizations in Morocco, Tunisia, the United Arab Republic, Ghana, Guinea and Libya have set up an international organiza-tion under the name of African Radio Union. Principal aim of the Union is for technical and administrative co-operation between the different national broadcasting organizations in Africa. Abdoulaye Toure, director of the state-owned broadcasting service of Guinea, is the first president.

V.H.F. Sound Broadcasting .--- The sixth edition of the list of European v.h.f. sound broadcasting stations, giving the situation on January 1st, will be published by the European Broadcasting Union early in February. The cost, including five bi-monthly supplements, is 50 Belgian francs. The list can be obtained from the E.B.U. Technical Centre, 32 avenue Albert Lancaster, Brussels 18, Belgium.

Receiving Licences .- During October the number of combined television and sound licences current in the U.K. increased by 82,397 bringing the total to 10,962,867. Sound-only licences totalled 4,226,094 including 459,856 for sets fitted in cars.

Pulse Techniques .- An 11-week laboratory course on pulse techniques begins at the Borough Polytechnic, Borough Road, London, S.E.1, in January. It will be held on Monday afternoons commencing January 9th and will be repeated on Thursday evenings from January 12th. The course will also be repeated on Monday afternoons from April 10th. (Fee £1).

Educational Publications.-The latest pamphlets in the series "Demonstrations and Experiments in Electronics" issued by the Mullard Educational Service give details of an all-transistor one-watt amplifier (No. 13) and an "echo" method of determining the velocity of sound (No. 14). The pamphlets are available free to schools, technical colleges and training establishments from the Mullard Educational Service, Mullard House, Torrington Place, London, W.C.1.

> **Television receiver production** in Hungary for the next five years, has been planned to total 1M. More than half the receivers are scheduled for export.

Moscow TV .- A tower carrying the aerials for five transmitters is a feature of the new television centre being built in Moscow.

Can you Help?-An American reader is anxious to secure a copy of Wireless World for January, 1952, to complete a 15-year file. Offers should be addressed to the Editor.



NEWS

Bradford.-Dr. G. N. Patchett of the Bradford Technical of the Bradford Amateur Radio Society. A fortnight later D. Millard (G3OGV) will deal with amateur receiver align-ment. The club meets at 7.30 at Cambridge House, 66 Little Horton Lane.

Cleckheaton.—"'Scope Interpretation" is the title of the talk to be given by T. C. Isaac of Bradford Technical College at the January 4th meeting of the Spen, Valley Amateur Radio Society. On the 18th a member of staff of Philips Electrical will discuss tape recording. Meetings are held at 7.30 at the Labour Rooms.

Mitcham.—Lifeboat radio equipment is the subject of the talk being given by W. D. Pye to the Mitcham and District Radio Society on January 13th at 8.0 at "The Canons," Madeira Road.

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Personalities

G. A. Marriott, B.A., is retiring in March from the managing directorship of the M-O Valve Company, a wholly owned subsidiary of the G.E.C., and is to be succeeded by J. Bell, B.Sc., F.Inst.P., at present deputy director and manager of G.E.C. Research Laboratories, and a member of the M-O Valve board. Mr. Marriott, who is 68 and a graduate of Caius College, Cambridge, has been with the G.E.C. throughout his professional life. He served on the board of the British Radio Valve Manufacturers' Association (B.V.A.) for nearly 20 years, and is now chairman of V.A.S.C.A., the associated organization which was formed in 1959 to take over from the B.V.A. the responsibilities for semiconductors and industrial valves and tubes. He was president of the Brit.I.R.E. from 1956 to 1958. Mr. Bell, deputy director of the G.E.C. Research Laboratories, Wembley, where he was manager of the telecommunications division from 1953 to 1958, was appointed to the board of the M-O Valve Company a few months ago.



#### G. A. Marriott

Paul Adorian, M.I.E.E., M.Brit.I.R.E., F.C.G.I., managing director of Associated-Rediffusion Ltd., has been elected a Fellow of the Institute of Radio Engineers of America, "for the development of electronic distribution networks used in broadcasting and television." He is the only British recipient among the 76 elected for 1961. Mr. Adorian is on the board of a number of companies within the Rediffusion group and is chairman of Central Rediffusion Services Ltd., Redifon Ltd., and Rediweld Ltd.

E. S. Hall, M.I.E.E., this year's chairman of the Rugby Sub-Centre of the I.E.E., is a director of A.E.I. Sound Equipment, Ltd., and divisional assistant chief engineer of A.E.I. Electronic Apparatus Division, New Parks, Leicester. He joined B.T.H. as a student apprentice in 1924 and on completion of the course in 1929 went to the research laboratory on the development of sound film equipment. In 1946 he was appointed chief assistant to the manager of the B.T.H. Electronic Engineering Department. Mr. Hall has held his present position since the formation of the A.E.I. divisions in 1959.

**R. H: Booth**, at one time chief engineer of the industrial division of E.M.I. Electronics Ltd., which he joined in 1949, and now personal assistant to the company's technical director, has gone to North America for a two-year tour of duty. He will maintain liaison with E.M.I.'s representatives in the U.S.A. and undertake a survey of the American market. **Professor F. C. Williams, O.B.E., D.Sc., D.Phil.,** F.R.S., of Manchester University, has received the American John Scott award in recognition of his work on the development of I.F.F. (Identification, Friend or Foe) during the war. Dr. Williams, who joined Watson-Watt's radar research team at Bawdsey in 1939, was employed throughout the war on radar circuitry. In 1947 he was appointed professor of electrotechnics at Manchester University, where he is now professor of electrical engineering and has been working on the development of digital computers. He is 49. The John Scott award is made by the City Trust of Philadelphia for "developing inventions for the benefit of mankind."

**Professor A. C. B. Lovell,** O.B.E., F.R.S., director of the Nuffield Radio Astronomy Laboratory of the University of Manchester at Jodrell Bank, has been awarded one of the two Royal Medals of the Royal Society for 1960, "for his distinguished contributions to radio astronomy."

**Dr. D. Gabor**, professor of applied electron physics at Imperial College, and **Dr. D. Jones** have been granted £2,000 by the Paul Instrument Fund Committee of the Royal Society for additional equipment for use in connection with the development of an electron interference microscope.

W. E. J. Farvis, B.Sc.(Eng.), M.I.E.E., senior lecturer in applied electricity and head of the post-graduate School of Electronics and Radio at Edinburgh University, has been appointed to the newly established chair of electrical engineering at the university. He has been in the Electrical Engineering Department for the past 12 years, prior to which he was for three years a lecturer at University College, Swansea. During part of the war he was engaged on the development of decimetric radar systems and later headed a radio counter-measures group.

P. L. Stride, formerly manager of Ekco's Malmesbury works, has become manager of the aviation division of Ekco Electronics Ltd., which is now centred at Southend. The division handles the design and technical liaison for ground and airborne equipment for the aviation industry. P. J. Harvey, formerly chief electronics engineer, becomes manager of the nucleonics and industrial division which is responsible for research and development covering nucleonic and physical instrumentation and control. E. B. Thompson continues handling the commercial activities of both divisions as sales manager.



**S. R. Mullard,** M.B.E., recently celebrated his 40th anniversary as a director of the Mullard company. He is seen here with (left) Sir Arthur Vere Harvey, C.B.E., M.P., who has been with the company 25 years and (right) ' S. S. Eriks, O.B.E., managing director of the company.

J. Bell

Sir Leslie Gamage has retired from the chairmanship of the General Electric Company and is succeeded by Arnold Lindley, who joined the company as an engineering apprentice in 1918. In 1958 he was made an assistant managing director in charge of the company's heavy engineering group, and for the past 18 months has been vice-chairman and a managing director.

E. Green, M.Sc., M.I.E.E., formerly head of the transmitter advanced development group of Marconi's, and for the past six years consultant engineer to the company, has retired. He has been with Marconi's since graduating at Manchester University in 1913. In his early days with the company Mr. Green was engaged in the installation of marine radio equipment and in 1915 he was seconded to the R.N.V.R. and was sent to Hong Kong to assist in the installation and operation of a 25-kW spark set and a Poulsen arc transmitter for the Royal Navy. In the years following World War I, he was assistant to C. S. Franklin during the whole of the development work on the short-wave beam system. His textbook "Amplitude-Frequency Characteristics of Ladder Networks" is a standard source of reference.



E. Green

L. C. Jesty

L. C. Jesty, B.Sc., M.I.E.E., has been appointed manager of Sylvania-Thorn Colour Television Laboratories in succession to B. C. Fleming-Williams, B.Sc., A.M.I.E.E., who has resigned. Mr. Jesty joined the Laboratories in 1957, and has been mainly responsible for the research and development programme on colour television and cathode-ray tubes. For seven years prior to joining the Laboratories he was with Marconi's, where he led the television research group and was closely associated with the development of the Anglicized version of the American N.T.S.C. colour system. From 1927 to 1946 he was at the G.E.C. Research Laboratories, and for three years at the Cinema-Television Laboratories. J. K. Oxenham, M.A., who has been in charge of the circuit work on colour and closedcircuit television at the Sylvania-Thorn Laboratories, has been appointed deputy manager.

T. E. Greenfield has been appointed sales development manager of the industrial process control division of Gresham Automation Ltd. After war service in the Merchant Navy he joined J. Langham Thompson Ltd. in 1946, and three years later went to the General Electric Co. at Stanmore, where he was concerned with flight instrumentation trials. In 1953 he returned to J. Langham Thompson, and since 1957 he has been with English Electric Aviation Ltd. He was appointed head of quality control for the production of guided weapons at their Stevenage works, and was also test project engineer for the Thunderbird II.

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The Rietzke Award being received from the donor Eugene Rietzke, founder and president of Capitol Radio and Engineering Institute, of Washington, by Sgt. J. H. O. Willacy "the airman showing the greatest promise in the field of electronics" at the R.A.F. Bomber Command Station at Cottesmore, near Oakham, Rutland. On the left is Air Commodore J. R. Morgan, Command Education Officer, and on the right Group Captain A. D. Mitchell, station commander. Some 70 airmen on the station took the electronics course of C.R.E.I.

J. D. MacEwan, B.Sc., A.M.I.E.E., A.M.Brit.I.R.E., has been appointed by the B.B.C. engineer-in-charge, television, Birmingham, in succession to H. G. Whiting, who recently became Regional Engineer, Midland Region. Mr. MacEwan joined the Operations and Maintenance Department of the B.B.C. in 1947. Since 1956 he has been senior lecturer (technical operations) at the Corporation's Engineering Training Department.

G. W. Short, the first part of whose article on the bootstrap follower is on page 21, was for two years on the editorial staff of our sister journal *Electronic Technology* until 1958, when he joined the B.B.C. He served in Royal Signals from 1944 to 1947. He then went to Oxford University, where he graduated in 1951 and the following year received his M.A. From 1953 to 1956 he was assistant press officer at Mullard's.

Horace Freeman, who, as already announced, retired recently from active business life after spending nearly 40 years in radio and electrical advertising, has been elected an honorary vice-president of the Radio Society of Great Britain. It was appropriate that the certificate was presented to him at the opening of the Radio Hobbies Exhibition in the Royal Horticultural Society's Old Hall, for it was there that he organized the first all-British wireless exhibition in 1922.

**B.** M. Lee, son of E. M. Lee, director and general manager of Belling & Lee, Ltd., has been appointed manager of the company's industrial group and to the board of executive directors.

#### OBITUARY

Dr. W. R. G. Baker, vice-president of the General Electric Company of America when he retired in 1958, has died at the age of 67. He joined the company in 1917. Since his retirement he had been vice-president of Syracuse University. Dr. Baker's best-known contribution to the American radio industry is probably the direction of its two National Television System Committees, the first to establish standards for the American monochrome service and the second, appointed in 1950, to set up standards for what is now known as the N.T.S.C. colour system.

### **News from Industry**

Telemeter "Pay-TV."—A new company, British Telemeter Home Viewing, has acquired from International Telemeter Co. the British rights of the "Pay-TV" system developed by its associates, Paramount Pictures Corp. The system, which is being used in a pilot scheme in Toronto, provides for the reception of a television programme specially transmitted by wire or over the air which is un-scrambled by a coin-in-the-slot unit fitted to a standard receiver. The subscribers to the new company include British Lion Films and the Granada Group.

Decca.—E. R. Lewis, the chairman of the Decca Record Company, announced at the extraordinary general meeting on November 10th that the year's results for the group were "by far the best in the company's history." The balance from the trading account was  $\pounds 3,714,547$ —an increase of some  $\pounds 400,000$ on the previous year. The net balance carried forward was  $\pounds 1,821,984$ , a 28% increase on the previous year.

Plessey's consolidated profit and loss account for the year ended last June shows a trading profit of over  $\pounds 4M$ . After allowing for taxation and making various deductions the balance carried forward into the current year was  $\pounds 2,240,044$  compared with  $\pounds 1,536,042$  the previous year.

**M.S.S. Recording Company** have arranged with E.M.I. to manufacture and distribute the discs and disc cutting apparatus previously undertaken by M.S.S. who are concentrating on the production of their Master-tape and Data Tape.

Non-destructive Testing.—Solus-Schall, Ltd., of County Building, Honeypot Lane, Stanmore, Middx., have produced an information booklet reviewing the sitemethods of non-destructive testing provided by the company. Copies of the booklet are available free of charge.

The Gramophone Company are developing a combined four-track tape recorder and 45/33 r.p.m. record reproducer in which the tape and record are driven (simultaneously if required) by means of the same motor.

Aero Electronics Ltd., of Gatwick Airport, have been appointed the exclusive export agents and sole distributors to U.K. aircraft manufacturers of the products of Communications (Air) Ltd., of Bagshot, Surrey. Their products include a V.O.R./I.L.S. simulator and airborne V.O.R./I.L.S. equipment.

Beme Telecommunications Ltd., of 24, Upper Brook Street, London, W.1, a member of the Derritron Group of companies, has announced two new marine radio products—a depth indicator and a f.m./a.m. receiver. The transistor depth indicator "Diver 60" has a range of 0 to 360 feet.

Southern Instruments Ltd., who in association with the Drayton Regulator and Instrument Company of West Drayton, some months ago formed Drayton-Southern Ltd. (specializing in instrumentation and control systems), have more recently formed Storno-Southern Ltd. in association with the Danish Storno Radio Co. Storno-Southern is to manufacture in this country under licence v.h.f. communications equipment designed by the Copenhagen company. Another company, Southern Analytical Ltd., has been formed to take over the analytical instrument work of the parent company. Ocean Weather Ships.—Four transmitters, eight receivers and two direction-finders are being supplied by Marconi's for each of two frigates Amberley Castle and Pevensey Castle which are being converted for use as ocean weather ships. The company is also modernizing Admiralty ranging and height-finding radar in the vessels. H.M.S. Amberley Castle has been renamed O.W.S. Weather Adviser.

Raytheon Co., of Waltham, Mass., have supplied two air traffic control radar systems for Switzerland's major air routes. The radars will be connected by microwave links to the Geneva-Cointrin and Zurich-Kloten airports which they will serve. The Geneva installation will be located on the 5,500-foot La Dole, 16 miles away, and the Zurich equipment on a mountain some seven miles away.

Central Electronics Inc., the recently formed whollyowned French subsidiary of Zenith Radio Corporation, are building a factory in Paris for the production of components and receivers.

J. W. Maunder, U.K. agent for Shure audio components, now has an office at 22 Orchard Street, London, W.1 (Tel.: Hunter 4116).

#### NEW FACTORIES

Associated Transistors, Ltd., which is jointly operated by A.T.E., English Electric and Ericsson Telephones, has acquired an 18-acre site at Basingstoke, Hants, on which a transistor factory covering about 120,000 sq ft is to be built.

**G.E.C.**—A mill at Reddish, near Manchester, has been acquired by the G.E.C. Semiconductor Division. It will be known as Broadstone Works and has a working space of 600,000 sq ft. Manufacture of semiconductors is expected to start in the early part of this year.

Nottingham Electronic Valve Co., Ltd., manufacturers of Nev cathode-ray tube pumping equipment, have moved their offices and works from Netherfield to Main Street, East Bridgford, Notts. (Tel.: East Bridgford 276).

Elremco.—An extension to their factory at Bush Fair, Harlow, Essex, has increased the production area of the Electrical Remote Control Co. to 14,000 sq ft.

Marconi's Aeronaut'cal Division is being regrouped in new premises at Basildon, Essex, where the bulk of the company's aeronautical equipment is manufactured.

L. E. Simmonds Ltd., relay manufacturers of 5 Byron Road, Harrow, Middx., have opened a new factory at Thetford, Norfolk. The factory gives an additional 6,000 sq ft to the company's present manufacturing capacity at Harrow, which will remain the headquarters.

Alma Components Ltd., manufacturers of precision wire-wound resistors, have recently opened a new factory in Diss, Norfolk (Tel.: Diss 2288). The London works have been closed but an office is being maintained at the old address: 551 Holloway Road, London, N.19 (Tel.: Archway 0014).

Sifam.—A new factory with a total floor area of 21,000 sq ft and accommodation for 260 employees has been opened in Torquay by Sifam Electrical Instrument Co.

#### By F. BUTLER,

O.B.E., B.Sc., M.I.E.E., M.Brit.I.R.E.

### TRANSISTOR INVERTER

#### FREQUENCY STABILIZED CIRCUIT SUITABLE FOR RUNNING A TAPE RECORDER

IGH-POWER transistor inverters commonly use two or a multiple of two transistors in a special type of push-pull oscillator circuit which is designed to give a square-wave output. High-efficiency operation is secured because the transistors alternate between two conditions in each of which the internal energy dissipation is very low. In one condition the transistor is cut off and there is no energy loss. In the other state the transistor is bottomed and although it then carries a large current the voltage drop across it is quite small and the energy loss is again very low.

If a sinusoidal output waveform is required the maximum efficiency is lower and cannot possibly exceed 78.5 per cent. In a practical case the efficiency would be nearer 60 per cent compared with about 85 per cent for the square-wave case.

Frequency Stability of Self-excited Inverters.— Push-pull square-wave oscillators are of two distinct types. In one of these the rectangular waveform results from magnetic saturation effects in the core of the oscillator transformer. High-efficiency

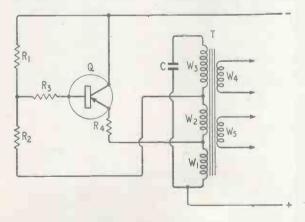


Fig. 1. Driver oscillator circuit for frequency-stabilized inverter.

operation therefore calls for the use of low-loss core materials. Suitable magnetic alloys are expensive but economy in the core size can be achieved by operation at a relatively high frequency, usually between 400 and 1,200 c/s. Within this range of frequencies the sum of all the losses (eddy current, hysteresis, dielectric and copper winding resistance) reaches an acceptable minimum value.

The second type of self-excited square-wave generator also employs a ferro-magnetic core but operates at flux densities which are always below the

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saturation level. Losses are thus reduced and it becomes possible to make use of conventional core materials. Amplitude limiting of the square-wave output is in this case caused by cut-off or saturation (bottoming) in the transistors. These must be accurately matched if it is important to generate a square-wave output with a mark-to-space ratio of unity.

Oscillators which make use of core-saturation effects generate a frequency which is proportional to the battery supply voltage. A constant output frequency thus calls for a regulated power supply. The output frequency is not markedly dependent on the connected load. A valuable feature is that the circuit is self protecting against overloads.

By contrast, inverters which operate with unsaturated transformers deliver an output frequency which is dependent on the supply voltage and on the connected load. A heavy overload will normally result in the destruction of the transistors.

**Constant-frequency Inverters.**—When used for d.c. to d.c. conversion there is no particular need to specify precisely the frequency of the oscillator. It is convenient to make this fairly high in order to simplify the output filtering arrangements, and a value in the region of 1000 c/s is entirely suitable.

There are cases in practice where there is no such latitude regarding the choice of output frequency. This is certainly true if the equipment to be operated contains squirrel-cage or synchronous motors which are required to be run at a fixed speed. A taperecorder capstan drive motor comes into this category and for use in vehicles, in aircraft, or on board ship it must be supplied from a battery-operated constantfrequency a.c. source.

To operate mobile equipment a 28-V d.c. supply is normally available. Experience shows that it is possible to design an inverter for operation from such a supply which will give a continuous output of 100 W at any fixed frequency between 40 and 60 c/s.

In principle, a constant output frequency could be derived by amplifying the output of a low-power oscillator of stable frequency. If a sinusoidal voltage waveform is required with a substantial power output, a practical scheme would be to drive a Class-B amplifier from the oscillator, using intermediate amplifiers as necessary. Such an assembly would be expensive and inefficient. At best, the output stage would have an efficiency of under 70 per cent. The amplifier would have a high output impedance so that the voltage regulation would be extremely poor. This could be corrected by the use of negative feedback which in turn would call for higher gain in the amplifier, requiring at least one extra stage. At the expense of further circuit complication it might be possible to devise some form of automatic gain control to give a constant output voltage.

The situation is much more favourable if a squarewave output is acceptable. It is not considered very good practice to run large a.c. motors under these conditions. Heating is increased by the harmonic currents and the motor may tend to "cog" round at a small fraction of its fundamental synchronous speed. Nevertheless, many low-power motors appear to perform satisfactorily with a square-wave input.

In respect of its output impedance, the squarewave inverter is effectively an electronic switch, periodically reversing the supply battery connections to the primary of the inverter transformer. The output impedance at the transformer secondary terminals includes the secondary winding resistance plus the total primary circuit resistance referred to the secondary side, i.e. multiplied by the square of the transformer turns ratio. It also contains a term sufficient to account for all the other circuit losses. The voltage regulation of an inverter is normally about 20 per cent between no load and full load. During the conduction phase there is also a voltage drop across each transistor of about 0.75 V. Assuming that a 28-V battery supply is used, this is responsible for a further reduction in efficiency of 2.7 per cent.

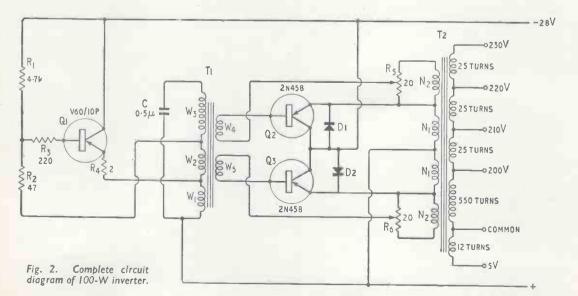
As regards the design of a constant-frequency oscillator to drive the square-wave amplifier there are two possibilities to be considered, remembering that the output stage calls for a substantial drive power. One is to use a high-power tuned-circuit oscillator and the other is to use a low-power version, say of the Wien bridge type, followed by an inter-The requirements for mediate power amplifier. high oscillator stability and high output power are incompatible, but the amplifier drive power requirements can be reduced by making this stage regenerative. Sufficient positive feedback may be applied from the output stage to cause incipient self-oscillation and if the regeneration control is adjustable it may be set at such a level that only moderate power demands are made on the stable-frequency driving

source. This technique is used in the actual equipment to be described later.

A circuit diagram of the 50 c/s oscillator is shown in Fig. 1. It is of a conventional feedback type, the arrangement being such that the transistor collector terminal may be connected directly to the negative supply lead which is at earth potential. The chassis can thus be employed as a convenient heat sink. The iron-cored transformer T carries five windings. Of these, W4 and W5 are the output secondaries used to couple the oscillator stage to the output amplifier. The primary windings W1,  $W_2$  and  $W_3$  are connected series aiding, the total inductance being such as to resonate with C at a frequency of 50 c/s. A convenient value of C is about  $0.5 \ \mu F$  which calls for a very large number of turns on  $W_3$ . The windings  $W_1$  and  $W_2$  are calculated to provide the requisite feedback voltage for reliable oscillation and to match the input and output impedance of the power transistor Q. The resistances  $R_1$ ,  $R_2$  and  $R_4$  set the base bias of Q, while  $R_3$ is sufficiently large to ensure a constant-current drive to the transistor base circuit. The transformer employs a C-core of grain-orientated silicon steel and has a small airgap to avoid magnetic saturation due to the d.c. component of the collector current.

**Practical Inverter Circuit.**—Fig. 2 shows the complete circuit diagram of a 100-W inverter. A power transistor Q1 associated with the transformer T1 constitutes the master oscillator, operating at 50 c/s. Apart from the specification of the transformer, to be given later, this stage calls for no further comment.

The output transistors Q2 and Q3 feed power to a centre-tapped winding on the transformer T2, each half-primary having  $N_1$  turns. Two additional windings, each with  $N_2$  turns, serve to provide positive feedback to the bases of Q2 and Q3, in phase with the oscillator drive from the secondaries  $W_4$  and  $W_5$  of the transformer T1. The magnitude of the feedback is controllable by the slider settings on the two variable resistors  $R_5$  and  $R_6$ . If excessive regeneration is employed the master oscillator may lose control and the output stage will run as a self-



excited inverter. This condition must be avoided but a moderate amount of feedback makes the amplifier stage easier to drive. The two resistances also serve another useful purpose in that they can be used to equalize the drive to the output transistors if these have widely-differing characteristics. Heavyduty wire-wound potentiometers should be chosen for this purpose. Except for  $R_4$  the remaining resistors can be of 1-W rating and 5 per cent tolerance;  $R_4$  should be a 5-W wire-wound component.

The diodes D1 and D2 call for comment. They must be in circuit if the inverter is used to supply a highly inductive load. Under these circumstances the output transistor load line degenerates into a distorted ellipse and there is a reverse flow of current which is in fact carried by the diodes. They are unnecessary if the load is resistive or if power-factor correction is applied. They can be of the germanium or silicon junction type, rated to carry a maximum current of 5 A and with an inverse voltage rating of 100 V.

The transformer specifications are as follows:---

#### OSCILLATOR TRANSFORMER T<sub>1</sub>

Core: Double-C pattern, centre limb lin × lin (1 sq in cross-sectional area)

Airgap: 0.031in (1/32-in paxolin spacer)

Windings (all enamel and rayon covered):---

$W_1$ :	160	turns,	22	s.w.g.		
W2:	70	22	24			
W 3:	4330	22	34	33		
W4:	15	S	10	D!	C1	
W .5:	15	) ce	10	,, Bi	mar	woi
0						

**OUTPUT TRANSFORMER, T2** 

Core: Double-C pattern, centre limb  $1\frac{1}{2}$ in  $\times 1\frac{1}{2}$ in (2.25 sq in cross-sectional area)

und

Airgap: None (butt joint)

Windings (all enamel covered):-

N<sub>1</sub>: 56+56 turns, 14 s.w.g. Bifilar wound

N<sub>2</sub>: 8+ 8 " 16 " Biniar wound

Secondary windings as shown in Fig. 2, 22 s.w.g. In both cases the heavy gauge wire is wound on first (nearest to the core).

The two transformers are bolted to a rectangular sole plate of  $\frac{1}{2}$ -in thick aluminium which constitutes the heat sink for all three transistors. These are in direct contact with the plate which is connected to the negative terminal of the supply battery. It may be earthed if required. The circuit wiring is straightforward and the layout of components is in no way critical.

Alternative types of transistors suitable for use in the circuit are readily available. They include the following:—

MANUFACTURER	Q1	Q2 and Q3
Ediswan-Mazda G.E.C	{XC142 XC155 GET 9	XC155 XC156 GET573
Mullard	{OC28 OC29 (V60/10P	OC28 OC29
Newmarket	V60/20P V60/30P	
Texas Instruments	{2N457 2N458	2N457 2N458

**Operation with Inductive Loads.**—The dynamic characteristic of a power amplifier is almost a straight

line for a purely resistive load. As previously mentioned it becomes a distorted ellipse if the load is reactive. The normal square-wave output from the inverter can under some circumstances be transformed to that shown in Fig. 3. This waveform was actually observed when the capstan drive motor of a high-grade tape recorder formed the inverter load. The motor called for a nominal input of 70 VA at 230 V and although its operation appeared to be entirely normal an attempt was made to correct the waveform by adjustment of the load power factor. It was found that a shunt capacitance of  $1\mu$ F removed the effect but introduced some ringing on the leading edge of the square wave. This in turn was eliminated by connecting a resistance of  $400 \Omega$  in series with the capacitor.

**Power Output and Efficiency.**—The inverter shown in Fig. 2 is easily capable of delivering 100 W into a resistive load. It can be switched on in the

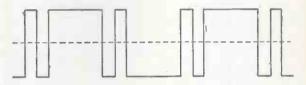


Fig. 3. Inverter output waveform with inductive load.

battery circuit with a 100 W lamp as the connected load and will quickly bring the lamp up to full brilliance in spite of the very low initial resistance of the cold filament. At full load the transistors remain cool and are clearly being run well below their rated maximum dissipation.

Measurements show that the unit draws 4.8 A from a 28 V battery when delivering 220 V, 0.46 A. An output of 101 W is thus obtained when the input is 134 W, corresponding to an efficiency of almost exactly 75 per cent. Some caution must be used when making measurements with rectifiertype a.c. meters. These in fact measure mean values but are calibrated to indicate the r.m.s. value of a sinusoidal voltage or current. They read about 10 per cent high when used to measure square waveforms.

The no-load current is 0.4 A at 28 V and the voltage regulation is about 10 per cent between one quarter load and full load output. Load changes have a negligible effect on the generated frequency, while halving the battery supply voltage changes the frequency by about 1 c/s. Finally the unit forms a suitable driver stage for a very-high-power amplifier. Using four Texas Instruments Type 2N 514B transistors in parallel push-pull an output in excess of 1 kW could easily be obtained.

**Tape-Recorder Wow and Flutter.**—There is no obvious reason why the performance of a recorder in respect of wow and flutter should be worse when the capstan drive motor is supplied from an inverter than when it is run from the mains. Elastic couplings, eccentric capstans and stretched tape are the usual causes of these types of distortion and are common to both cases.

Careful observations have shown that there is still another source of irregularity. This is due to

a periodic fluctuation of motor torque which appears to be a function of the number of stator slots or rotor bars. It is sufficient to cause a slight frequency modulation of any recorded tone unless the torque variations are smoothed out by a mechanical lowpass filter.

Actual measurements on a particular recorder (E.M.I. Type TR 52) give a total figure of 0.2 per cent wow and flutter from all causes when the machine is operating from the mains. The figure is exactly the same when running from the inverter provided that the battery gives a full 28 V output. When the voltage is reduced to 24 the figure rises to 0.3 per cent.

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Α

"DO-IT-YOURSELF"

METHOD

By H. DAGNALL, M.A.



Typical nameplates as they appear on plastic-faced hardboard.

### PANEL NAMEPLATES

THE use of transfers does much to improve the appearance of home-built equipment and helps to give it a professional finish. But, excellent as they are for many purposes, ready-made transfers do not entirely solve the problem of panel lettering for the amateur constructor. He is soon likely to discover that the exact wording required does not exist on any of the sheets available, and he is then forced either to abandon his chosen name for that switch or socket or to make up the word from single letters. If he attempts the latter alternative, the difficulty he has in aligning them will remain for ever distressingly visible—at least in the writer's case it does. Another difficulty likely to be encountered is that transfers cannot be applied to a crackle surface, nor can they easily be curved, compressed or expanded to fit a particular requirement.

The object of this note is to describe a method by which neat panel nameplates can be made in any desired size and wording. Lettering is done with the aid of a "Uno" stencil in black Indian ink on the reverse side of a strip of Perspex. Since the letters are on the back of the strip when it is affixed to the panel they are fully protected from damage and, being in optical contact with the Perspex, they appear intensely black.

Perspex having a thickness of 1/24 in is the most suitable kind to use. It should be cut into strips  $\frac{1}{4}$  in wide and about 12 in long.

A jig to hold the Perspex and stencil in alignment while lettering is essential for neat work, but it need be nothing more elaborate than a board about 12 in  $\times$ 4 in on which are nailed two pieces of hardboard of the same thickness as the Perspex. These should be 4 in apart so that the Perspex strip is a push fit in the gap; if the edges of the hardboard are purposely left rather rough, the strip will be gripped quite securely between them. A rail, parallel to the strip, against which the top edge of the stencil moves, completes the jig and ensures that the letters are aligned along the centre of the strip.

"Uno?" stencil UC.1½ (UF.1½ for figures) and pen No. O are suitable for the smallest (4-in) nameplates, but of course other sizes of strip and stencil can be used as required.

Writing on the Perspex is not difficult providing that it has been cleaned with carbon tetrachloride or other degreasing agent and the pen is kept clean to ensure free flow of ink. As the lettering is done on the rear surface of the Perspex the stencil must be reversed and writing done from right to left. To obviate the need for accurate centring of words on the strip, several legends are written on one strip and separated later.

The nameplates are fixed to the panel by means of countersunk 10 BA or 12 BA bolts. The heads of the bolts can afterwards be enamelled to match the panel.

On a light-coloured panel the words show up clearly but on a dark surface they should be backed with a light-coloured paper or paint; this also permits the use of colour coding for distinguishing different channels. Since the nameplates are transparent they are particularly suitable for mounting on light-coloured wood veneers or plastic-faced hardboard.

A circular dial plate can be lettered in a similar manner using a simple jig; circular plates can also be attached with 12 BA bolts to the brass bush of a knobto form a skirt.

Do not be put off because a required symbol does not appear on the stencil, for with a little ingenuity two or more characters or parts of characters can be combined to form the wanted symbol, for example parts of "O" and "L" to produce a neat capital omega symbol.

### The Bootstrap Follower

ITS USE IN AUDIO AMPLIFIERS

By G. W. SHORT

HE title of this article might well have been "When Is A Cathode-Follower Not A Cathode-Follower?" However, I have a feeling that the Editor would have found that a wee bit too long. Moreover, using this as an opening gambit enables me to have the best of both worlds, since readers will have guessed that the answer to the question is "When it's a bootstrap follower," whatever that may be.

The object of the exercise is to investigate the behaviour at audio frequencies of the "cathode-follower" circuit of Fig. 1(a). This circuit is arrived at when it is desired to operate a cathode-follower with a cathode resistance much larger than is required for obtaining "cathode bias." A logical way out of the difficulty is then to use a normal-sized cathode-bias resistance in series with the desired high resistance, and to return the grid to the appropriate point, as in Fig. 1(a), so that only the voltage drop across the smaller resistance  $R_{k1}$  is operative as grid bias. In many practical circuits  $C_k$  is omitted, and the output is taken directly from the cathode.

The difference between this type of circuit and a true cathode-follower (Fig. 5(a)) is apparently trivial. It is merely a question of convenience in biasing, or so I thought for a long time. My first misgivings came when I encountered Jeffery's phasesplitter.1 In this circuit use is made of the fact that, in the Fig. 1(a) type of circuit, the signal source " sees " not  $r_g$ , but a resistance many times greater. Thus the source delivers an increased voltage to the valve, since less is dropped across its internal resistance  $r_s$ . If  $r_s$  is large, the increase may be substantial. (In Jeffery's circuit,  $r_{s}$  is the anode resistance  $r_{\rm a}$  of a pentode valve, and runs to several megohms.) This "impedance multiplication" effect is one which does not make itself very obvious in a true cathode-follower, where all it does is to reduce the input capacitance a little, without altering the input resistance.

It came as something of a surprise to have it pointed out years later, by "Cathode Ray,"<sup>2</sup> that the Fig. 1(a) circuit differs from a true cathodefollower in another important respect; it has less negative feedback. This is because the proportion of the output voltage which is fed back negatively to the grid is reduced by the source resistance  $r_s$ , because r and  $R_g$  act as a potential divider. The amount actually fed back; i.e., the amount developed between grid and cathode, is therefore  $R_g/(R_g+r_s)$ times the actual output voltage. Thus the fraction fed back approaches unity as  $r_g$  approaches zero,

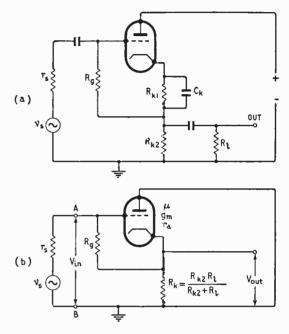


Fig. I (a) "Cathode-follower" circuit. (b) An a.c. equivalent circuit.

and it approaches zero as the signal-source resistance becomes infinite.

One thing which puzzled me at the time was this: if the feedback fraction is reduced, then the gain of the triode ought to be increased. Yet nobody seemed to have got out more voltage than he put in. Odd. But there are so many more interesting things to do than get out pen and paper and analyse circuits. Easier things, too, as far as I'm concerned. It wasn't until the circuit appeared again recently in *Wireless World*<sup>3</sup> that I finally got around to it, and discovered that, unlike the true cathode-follower which it so closely resembles, this circuit has an input impedance which depends on the load impedance, and an output impedance which depends on the signal-source impedance. The gain, oddly enough, is the same as that of a cathode-follower proper, though this depends on how it is defined.

For the purposes of analysis, Fig. 1(a) can be simplified to Fig. 1(b) which shows only those parts which are relevant to an a.c. signal.  $R_k$  is now the effective value of  $R_{k2}$  and the external load  $R_l$  in parallel.

Input Resistance.-This is the resistance "seen"

<sup>&</sup>lt;sup>1</sup> "Push-Pull Phase-Splitter," by E. Jeffery, Wireless World, August 1947, p. 274. <sup>2</sup> "Cathode Followers, With Particular Reference to Grid Bias Arrangements," by "Cathode Ray," Wireless World, June 1955, p. 292.

<sup>\* &</sup>quot;Economical High-Gain A.F. Amplification," by A. R. Bailey, Wireless World, January 1960, p. 25.

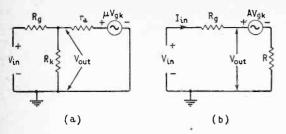


Fig. 2. Equivalent circuits of Fig. (1); (a) with normal equivalent valve circuit, (b) with valve and load combined.

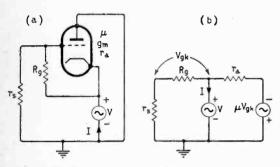


Fig. 3 (a) Circuit for calculating output impedance. (b) Equivalent circuit.

by the signal source. It is the resistance measured between points A and B with the source disconnected. If a voltage  $V_{in}$  is applied between A and B, a current will flow in  $R_g$ . The voltage drop across  $R_g$  is the true input to the valve; i.e., the voltage between grid and cathode  $(V_{ga})$ . The voltage across  $R_k$  is the output, and is  $AV_{gk}$ , where A = $\mu R_k/(r_a + R_k)$ . The situation is shown in the equivalent circuit of Fig. 2(a). It is simpler, instead of analysing this circuit as it stands, to convert it into the straightforward series circuit of Fig. 2(b) by using the Thévenin equivalent for the valve and its load. This is a generator with an open-circuit output  $AV_{gk}$  and an internal resistance R = $R_k r_a/(R_k + r_a)$ . The current supplied by  $V_{in}$  is  $I_{in} = (V_{in} - AV_{gk})/(R_g + R)$ where  $V_{gk} = I_{in}R_g$ . These expressions yield  $V_{in}/I_{in} = R_{in} = R + R_g(A + 1)$ . The term (A + 1)

where  $V_{gk} = I_{in}R_g$ . These expressions yield  $V_{in}/I_{in} = R_{in} = R + R_g(A + 1)$ . The term (A + 1) represents the impedance multiplication effect. The grid-cathode resistance appears to the source to be much larger than the actual physical value.

**Gain.**—It is clear from Fig. 1(b) that the voltage across  $R_s$  cannot exceed  $V_{in}$ . If it did, current would flow into the source instead of out of it. Thus  $V_{out}$  can approach  $V_{in}$ , but not exceed it, and the maximum possible value for the gain is 1, as in a normal cathode-follower. Here  $V_{in}$  is regarded as the input voltage. If  $v_s$  is considered to be the input voltage, then the gain is smaller and falls to zero as  $r_s$  becomes infinite.

From Fig 2(b),

$$V_{out} = AV_{gk}$$
  
= AI<sub>1</sub>,R<sub>g</sub>  
= AV<sub>in</sub>R<sub>g</sub>/[R + R<sub>g</sub>(1 + A)]  
V<sub>out</sub>/V<sub>in</sub> = AR<sub>g</sub>/[R + R<sub>g</sub>(1 + A)] = A'

In most practical circuits, R is  $100k\Omega$  or less,

and  $R_s(1 + A)$  is  $10M\Omega$  or more, so that  $A' \approx A/(1 + A)$ , which is the same as the "gain" of a cathode-follower.

**Output Resistance.**—This is the resistance seen by the load  $R_k$  in Fig. 1(b). To compute it we replace  $R_k$  by a generator of e.m.f. V, and let  $V_s$  be zero. The situation is then as shown in Fig. 3. Only that part of V which is developed across  $R_r$ acts as an input voltage  $(V_{gk})$  to the valve. We have.

I = 
$$(V + \mu V_{gk})/r_a + V/(R_g + r_s)$$
  
and  $(V_{rk} = VR_s/(R_s + r_s))$ 

The second term in the first equation merely represents a current through  $R_{\rm g}$  and  $r_{\rm e}$  which would flow even if the valve were not there. If we ignore it, we obtain,

$$r_{\text{out}} = \frac{V}{I} = \frac{r_{\text{a}}}{1 + \frac{\mu R_{\text{g}}}{R_{\text{g}} + r_{\text{s}}}}$$

When r is zero; i.e., when the amplifier is driven from a constant-voltage source, the expression for  $r_{out}$  reduces to  $r_a/(1 + \mu)$ , which is exactly what one would expect from a cathode-follower. But

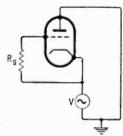


Fig. 4. When  $r_s$  is infinite, it can be removed without altering the output impedance.

when  $r_{a}$  is infinite; i.e., when the amplifier is driven from a constant-current source,  $r_{out} = r_{a}$ , which is what would be seen by the load if the circuit were a straightforward amplifier with no feedback. This is what it is, under these conditions, as far as output resistance is concerned. We have made r infinite, which means that it can be removed without altering the output resistance. The circuit then becomes that of Fig. 4. Here, V is applied directly between anode and cathode, and there is no input between grid and cathode, so that the resistance seen by V is  $r_{a}$ .

At this point it is instructive to see what sort of values one gets in practice. Using a valve such as the ECC81 (12AT7) one might choose working conditions such that  $\mu = 60$ ,  $r_a = 17k\Omega$ , A = 34,  $R_k = 22k\Omega$ ,  $R_g = 1M\Omega$ ,  $r_s = 100k\Omega$ . With these values, the input resistance is  $35M\Omega$ , and the output resistance is about  $300\Omega$  (compared with  $275\Omega$  for a true cathode-follower using the same valve). So for most practical purposes the circuit can be regarded as an ordinary cathode-follower. In some circumstances, it may be positively useful to increase the output impedance; for example, one might want to match a line impedance. On the other hand, there is no difficulty in raising the output resistance of a true cathode-follower: all one needs to do is to connect a suitable resistor in series with the live output terminal.

Hybrid Circuit.—The foregoing analysis shows (Continued on page 23) that the behaviour of the circuit is conditioned largely by the impedances to which its input and output terminals are connected. The impedance of the source of input voltage is usually the most important, since it governs the output impedance. When  $r_s = 0$ , the circuit becomes a true cathode-follower (shown in idealized form in Fig. 5(a)). This has a gain of A/(A + 1) where A is the gain without feedback, in other words the gain defined as  $V_{out}/V_{gk}$ . The output resistance of the valve is then low, being  $r_a/(\mu + 1) \approx 1/g_m$ . (In practical circuits, the output is usually shunted by something, so that the net output resistance is less than this.)

When  $r_s$  is infinite; i.e., when the triode is driven from a constant-current source, the valve presents the usual non-feedback output resistance  $r_a$ . Under these conditions the circuit becomes as shown in Fig. 5 (b); it has no feedback, and the input voltage is  $IR_g$ . This voltage is developed directly between grid and cathode, as in an ordinary triode amplifier. The difference is that the output is taken from the cathode, the anode being earthy. Now, a triode amplifier with a floating input voltage and cathode output is a bootstrap amplifier, and this is what our circuit becomes under these conditions.

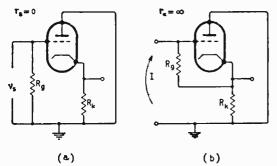


Fig. 5 (a) True cathode-follower. (b) "Bootstrap follower" with current drive. This has the same characteristics as a bootstrap amplifier with voltage drive.

Practical signal sources have neither zero nor infinite resistance, so the circuit behaves in practice like a hybrid between a cathode-follower and a bootstrap amplifier. It seems logical to call it a bootstrap follower. No originality is claimed for this title: it is so obvious that somebody must have used it before now.

In a slightly modified form (Fig. 6) this circuit forms part of the phase-splitter described in this journal by Jeffery<sup>1</sup>. Advantage is taken of the impedance multiplication effect to present the pentode with a high effective load, thereby increasing its gain to an appreciable fraction of the pentode  $\mu$ , instead of the miserable fragment of it which is all one usually gets. The great advantage of making use of impedance multiplication is that the anode load resistor can have the usual sort of resistance and the pentode can be operated with the usual anode current (say 1mA) yet a large gain can still be obtained. Moreover, because the pentode is operating under normal conditions, as opposed to "starvation" conditions, it can deliver a large output voltage. For example, one might have an effective anode load of  $3M\Omega$ , even though the actual load resistance is only  $100k\Omega$ . With an anode current

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of 1mA, the voltage drop across the load is only 100V. With a high tension voltage of 250V, the remaining 150V is available for the valve, which might well deliver a peak voltage of something approaching 100V before "bottoming" begins: this is more than the triode could handle. The gain of the pentode might be 1,000 or more.

While one might get a high gain by operating the pentode under starvation conditions with a physical anode load resistance of a few megohms, it is unlikely that the gain would be as high or that the available output voltage would be as large as the values obtained using the bootstrap follower.

These facts have led to the use of a single-ended combination of a pentode and a bootstrap follower as a means of obtaining high overall gain from a pentode and a triode. An equivalent circuit for this combination, neglecting direct voltages, is given in Fig. 7. This is the same as Fig. 1 (b), except that  $r_s$  is now the anode resistance of the pentode  $(r_{ap})$  and  $v_s$  is  $\mu_p V$ , where  $\mu_p$  is the pentode  $\mu$ , which may be several thousand.

The maximum gain of the combination cannot exceed the pentode amplification factor. This is much less than the maximum possible gain of the

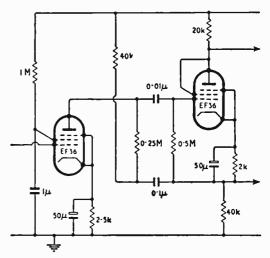
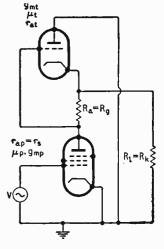


Fig. 6. Jeffery's phase splitter.

Fig. 7. Single-ended pentode-b.f. triode combination. Simplified circuit neglecting direct voltages.



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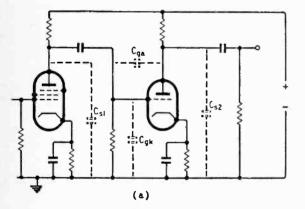
two valves in cascade, which is the product of the two amplification factors. In practice, however, a cascaded circuit would have an overall gain much less than this product. All the same, the cascade amplifier might be expected to produce more gain than the bootstrap follower with pentode drive.

However, one requires gain over a band of frequencies, and to compare the two circuits on this basis it is useful to calculate the products of gain and bandwidth. This is done in the Appendix. In terms of the stray capacitances of Fig. 8, the ratio of the gain-bandwidth product of the cascade amplifier to that of the pentode-bootstrap follower combination is:

$$\frac{\mathrm{GB}_{1}}{\mathrm{GB}_{2}} = \begin{bmatrix} (c_{\mathtt{s}1} + c_{\mathtt{g}\mathtt{s}}) (\mathrm{A}_{2} + 1) + c_{\mathtt{g}\mathtt{k}} \\ c_{\mathtt{g}\mathtt{s}} (\mathrm{A}_{1} + 1) + c_{\mathtt{s}1} + c_{\mathtt{g}\mathtt{k}} \end{bmatrix} \quad \begin{pmatrix} \mathrm{A}_{1} \\ \mathrm{A}_{2} \end{pmatrix}$$

where A1 is the gain of the triode in the cascade circuit and A2 is the gain  $(V_{out}/V_{gk})$  in the bootstrap follower circuit. In the numerator, the stray capacitance  $c_{s13}$  which contains the output capacitance of the pentode in Fig. 8, is multiplied by (A2 + 1), but it is not multiplied by anything in the denominator. In practice,  $c_{s1}$  usually exceeds  $c_{s23}$  so the effect of this difference can be very large.

The effect of  $c_{s1}$  in the cascade circuit is to add to  $c_{gk}$ . This is comparatively harmless, since the important capacitance is the Miller capacitance



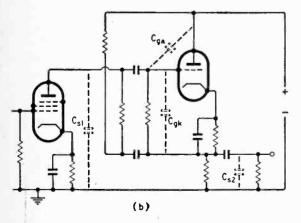


Fig. 8 (a) Cascade amplifier and (b) bootstrap follower ccmbination showing stray capacitances

 $c_{s*}$  (A<sub>1</sub> + 1). But in the bootstrap follower amplifier,  $c_{s_1}$  adds to  $c_{g*}$ , and the result is disastrous. In the bootstrap follower amplifier, the impedance seen by the preceding stage is the normal sort of triode input impedance, including the part due to Miller effect, multiplied by (A<sub>2</sub> + 1). The effect of adding  $c_{s_1}$  to  $c_{g*}$  is the same as would be the effect of adding it in the cascade circuit, only in the cascade circuit it is not added.

Before leaving the subject of gain and bandwidth, it should be mentioned that, with the cascade circuit, one has the opportunity of putting the triode in the first stage and the pentode in the second. With some values of signal-source impedance and load capacitance this might produce a marked improvement. In the bootstrap follower combination there is no point in putting the triode in the first stage since its amplification cannot exceed about 100 even with the best high- $\mu$  values. We shall continue to consider circuits in which the pentode is always in the first stage, partly to preserve a sound basis for comparison, and partly because the best audio pentodes are rather better than the best triodes in regard to hum. In a high-gain amplifier it would be logical to use a pentode such as the EF86 in the first stage.

#### Appendix

#### Gain-Bandwidth Products

To find these, we calculate the pentode gain and bandwidth for each circuit and multiply their product by the triode gain. (The effect of the triode bandwidth is discussed in the article). To find the pentode gain, we need to know the load resistance and capacitance. In the cascade amplifier, the load resistance is just R<sub>s</sub> in parallel with the triode grid resistance R<sub>g</sub>. The load capacitance, however, contains a portion due to Miller effect. This portion is here taken to be  $c_{ga}$  (A + 1), where  $c_{ga}$  is the inter-electrode capaci-tance (grid-anode) of the triode, plus any stray capaci-tance between grid and anode. This is not strictly correct, because the Miller capacitance is not a pure capacitance. A small resistance appears in series with it, but at audio frequencies, with  $c_{p}$  of a few pF, this resistance is negligible, and for all practical purposes the Miller capacitance is a pure capacitance. Similarly, in computing the anode load of the pentode when the triode is a bootstrap follower, a resistance which appears in series with  $R_{s}$  (A + 1) is neglected. (See under "Input Resistance"). Again, this is justified in practical circuits, where  $R_g (A + 1)$  is nearly always much larger.

By neglecting these unimportant resistances, we get the equivalent circuits shown in Fig. A, which have the advantage of being readily comparable. In each case, the pentode is shown by the current generator  $g_m V$  and the anode resistance  $r_s$ . The circuits are simple parallel RC circuits. The pentode gain is  $g_m R$ , where R is the net resistance, and the bandwidth is  $1/2\pi RC$ , where C is the total capacitance.

R and C are different for the two circuits. If we call the values in the cascade circuit  $R_o$  and  $C_c$ , and the values in the bootstrap follower amplifier  $R_b$  and  $C_b$ , we have for the pentode in the cascade circuit, a gain-bandwidth product

 $g_m R_c \times 1/2\pi R_c C_c = g_m/2\pi C_c$ 

The overall gain-bandwidth product is  $A_1$  times this, where  $A_1$  is the triode gain in the cascade circuit,

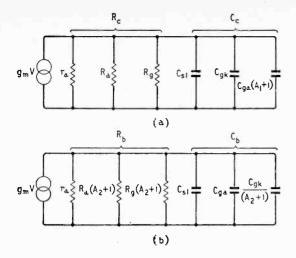


Fig. A (a) Pentode operating conditions in the cascade amplifier. (b) Pentode operating conditions in the bootstrap follower combination.

so we can put, for the cascade amplifier,  $GB_1 = A_1 g_m/2\pi C_e$ 

For the pentode in the bootstrap follower amplifier, the gain  $isg_m R_b$  and the bandwidth is  $1/2\pi R_0 C_b$ . The gain-bandwidth product for the pentode is thus  $g_m/2\pi C_b$ . To obtain the overall gain-bandwidth product for the bootstrap follower amplifier, this must be multiplied, not by  $A_2$ , the gain  $V_{out}/V_{gk}$  of the bootstrap follower triode, but by  $A_2/(A_2 + 1)$ . This is because the output of the pentode is applied between the grid of the triode and "earth," like  $V_{in}$  in Fig. 1(b). The "gain" of the triode is then  $A_2/(A_2 + 1)$ , as shown in the article. So the overall gain-bandwidth product for the bootstrap follower amplifier can be written

$$GB_{2} = \frac{g_{m}}{2\pi C_{b}} \times \frac{A_{2}}{A_{2} + 1} = \frac{(A_{2})g_{m}}{2\pi C_{b}(A_{2} + 1)}$$

The ratio of the gain-bandwidths is

$$\frac{\text{GB}_{1}}{\text{GB}_{2}} = \frac{\text{A}_{1}\text{g}_{m}}{2\pi\text{C}_{c}} \times \frac{2\pi\text{C}_{L}(\text{A}_{2}+1)}{\text{A}_{2}\text{g}_{m}} = \frac{\text{C}_{v}\text{A}_{1}(\text{A}_{2}+1)}{\text{C}_{c}(\text{A}_{2})}$$

If  $A_1 = A_2 = A$ , this approximates to

$$\frac{\mathrm{AC}_{\mathrm{b}}}{\mathrm{C}_{\mathrm{c}}} = \frac{\mathrm{A}(c_{\mathrm{s1}} + c_{\mathrm{ga}}) + c_{\mathrm{gk}}}{\mathrm{A}c_{\mathrm{ga}} + c_{\mathrm{s1}} + c_{\mathrm{gk}}}$$

In practical circuits,  $c_{s1}$  is the largest of the strays, so the cascade amplifier is much superior. However,  $A_1$  is not likely to be the same as  $A_2$ . For greater precision, we must write the full formula,

$$\frac{\text{GB}_{1}}{\text{GB}_{2}} = \frac{\text{A}_{1}(\text{A}_{2}+1)\left[c_{s1}+c_{g3}+\frac{c_{g4}}{\text{A}_{2}+1}\right]}{\text{A}_{2}\left[c_{s1}+c_{g4}+c_{g3}\left(\text{A}_{1}+1\right)\right]}$$
$$= \frac{(c_{s1}+c_{g3})\left(\text{A}_{2}+1\right)+c_{g4}}{c_{g3}\left(\text{A}_{1}+1\right)+c_{g1}+c_{g4}} \times \frac{\text{A}_{1}}{\text{A}_{2}}$$

To get an idea of what this means in practice, we substitute the following typical values:

 $c_{\rm gl} = 10 {\rm pF}$ ,  $c_{\rm ga} = 2 {\rm pF}$ ,  $c_{\rm gk} = 5 {\rm pF}$ ,  $A_1 = 30$ ,  $A_2 = 25$ This produces a ratio of just under 5. For equal bandwidths, the gain of the bootstrap follower, amplifier is nearly 14 dB below that of the cascade amplifier.

(To be concluded)

#### **Electronic Pointer Generator**

ALTHOUGH a conventional lecture pointer can usually be used with large-scale diagrams for illustrating television talks, there are often occasions when this is impracticable, for example, in describing surgical operations or in microscopy. To meet these and other



requirements, Pye, Ltd., have developed an electronic pointer generator by means of which an arrow can be superimposed on the picture and moved to any desired point by a remote-control "joystick," which carries a push-button for switching on or off. An auxiliary switch reverses the direction of the pointer. The "joystick" movement is resolved into settings

The "joystick" movement is resolved into settings of horizontal and vertical potentiometers which provide d.c. shifts for the arrow blanking pulses in the line and field periods. The arrow formation is generated by mixing sawtooth waveforms in diode coincidence circuits and can be adjusted in size. It is made more clearly visible by filling with black and white vertical striations which are generated by bursts from a triggered Hartley oscillator.

The equipment is designed to operate on 405-, 525or 625-line systems and is available for rack mounting. (Type 2443) or in portable form (Type 2444).

#### "Permeability Tuners for Television"—a correction

In the formula for the conditions for balance (right-hand col., p. 476 of the October 1960 issue) it is regretted that a term was omitted from the denominator of the left-hand expression. The equation should read :—

$$\frac{C_{gk}C_{c}}{C_{c}C_{d}+C_{g}C_{d}+C_{g}C_{c}+C_{gk}C_{c}+C_{gk}C_{d}}=\frac{C_{ak}}{C_{out}+C_{ak}}$$

WIRELESS WORLD, JANUARY 1961

# RADIO HOBBIES EXHIBITION

AMATEUR TRANSMISSION, "HI-FI"

AND MUSIC-MAKING ON SHOW



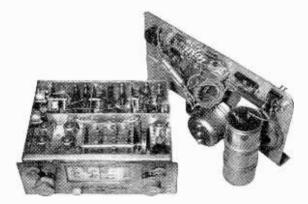
E. St. B. Sydenham's silver plaque winning wavemeter.

**N** the competition organized by the R.S.G.B. for home-constructed apparatus a heterodyne wavemeter, made by E. St. B. Sydenham, G3LOK, was the Silver Plaque winner. R. H. Hammans' (G21G) transistor communications receiver, covering 1 to 30Mc/s with a.m., s.s.b. and c.w. facilities, was judged the best entry from the outside-London area: this set is only 41 in by 6in by 51in. The best club entry was from Aquila Radio Club, who submitted a six-band transistor car receiver covering 1.8 to 30Mc/s, made by C. J. Salvage, G3HRO. This receiver is split into two units, as is fairly common practice, with the output stage on the loudspeaker panel. The receiver proper has three i.f. stages, an r.f. stage, and b.f.o. for c.w. reception. The band switch on the front panel takes the form of a sliding control and this is coupled to a rotary switch at the rear of the set by a mechanical linkage.

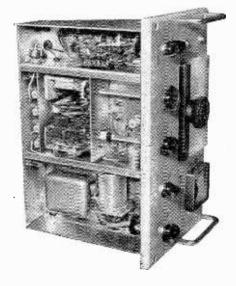
The interest shown by R.S.G.B. members in mobile operation extends to the motor cycle—exemplified by s.s.b. equipment shown by V. Page, G3IVP—and even the humble "push-bike": R. G. Scutt's (G3IBI) 2-metre bicycle transmitter-receiver does not depend on foot power, but uses a miniature 24-V accumulator.

Another exhibit by E. St. B. Sydenham showed how members of the Royal Amateur Emergency Reserve may be alerted automatically. A super-regenerative 2-metre receiver fitted with an r.f. stage to eliminate the characteristic radiation, covers, due to its wide bandwidth, the whole 144Mc/s band. The output from this feeds, through a tuned a.f. amplifier, a pendulum relay. This relay will respond only when fed with pulses at its resonant frequency, and these pulses must be composed of a.f. to which the amplifier responds. Thus only a transmission satisfying these requirements broadcast at any frequency in the 144Mc/s band will energize the pendulum and ring the alarm bell. In the event of a mains failure, the receiver is automatically switched to battery operation.

Five companies producing equipment for the amateur were seen in the "commercial" section for the first time. Electroniques (Felixstowe) (who also produce temperature-compensated tuning coils and coil packs) were showing a complete "top-band" (160m) transmitter called the "Pathfinder," This uses germanium



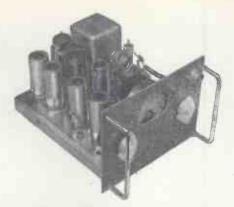
Aquila Radio Club extry—transistor mobile communications receiver made by G3HRO. Band selection is achieved by the ratary switch at the rear, operated by the slide-bar below the dial.



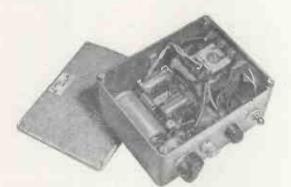
Underside of "Pathfinder" 160m transmitter by Electroniques (Felixstowe). Four-stage TV interference filter is adjustable and is seen in back compartment of centre row.



K. W. Electronics "One Sixty" top-band transmitter.



TW2 10-watt 144Mc/s transmitter by Withers.



Interior of "Tiger Talking Box" transistor modulatordrive amplifier.

diodes as a speech clipper, which enables a high level of modulation to be achieved without danger of overmodulation. A four-stage low-pass output filter (to prevent television interference) is fitted in a screened compartment and the aerial current is monitored permanently by a plug-in thermocouple feeding the frontpanel meter. Another new 'top-band" transmitter on show was the K.W. "One-Sixty," which uses multiplication from the v.f.o. frequency of about 900kc/s. Like the "Pathfinder" this transmitter is complete in a small case, and can run at over the 10 watts input permitted. K.W. Electronics (who incidentally are importing Hammarlund receivers) won the award for commercial equipment this year, with their Viceroy s.s.b. exciter unit. Another new exhibitor was Tiger Radio, who were showing a very wide range of equipment. Items that particularly caught our eye were a 750W p.e.p. linear r.f. amplifier for 10 to 80 metres and the "Tiger Talking Box." The tetrode output stage (QY3-65) of the amplifier has its screen grid supplied from a cathode follower whose control grid is normally at chassis potential. When drive is applied a diode rectifies it providing a positive potential for the cathode-follower grid, consequently lifting the screen potential of the OY3-65 and allowing it to amplify. Complete absence of p.a. anode current (and output) is thus assured until drive is applied. Power supplies for the unit are derived by a chain of silicon rectifiers, whose good regulation is an advantage, and a bias control allows operation in any condition between Classes AB1 and C.

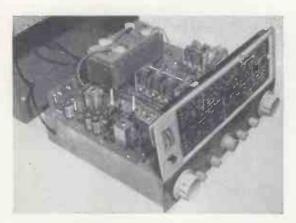
The "Tiger Talking Box" is a small transistor a.f. amplifier which derives its power supply from the 6.3V heater line. It is designed to be used as a microphone stand and it is capable of driving fully a pair of KT88 or 807 valves as the final stage of the modulation amplifier. The sensitivity is 10mV at an impedance of  $750k\Omega$ and the 150-V grid-to-grid output is achieved by two OC74s feeding two transformers.

Equipment made for the amateur for use in the 144Mc/s band is a comparative rarity; but this is a speciality of Withers Electronics. This company were showing for the first time a 10-watt transmitter, receiver converter and aerial, all suitable for mobile or fixed-station working. The TW2 transmitter is crystal controlled, tripling and doubling from 24Mc/s, whilst a clamp valve operates on the QQV03-10 output stage. The aerial is a  $\lambda/2$  dipole, bent into a circle to reduce the directional characteristics and fed with a  $\gamma$ -match section which allows an 80- $\Omega$  feeder to be used although the basic aerial impedance has been changed by the "rolling up" process.

Another unusual form of aerial—called the "birdcage" —was shown by Minimitter: this is designed to act as a directional high-gain aerial for 14Mc/s. Its mode of operation can best be visualized by considering two V-dipoles, stacked, with reverse-V reflectors behind them. Little loss of performance of a dipole is incurred by bending away at right-angles to the main axis half of each of the "rods." The lower dipole and reflector have their ends "bent" upwards to meet and join, as the stacked arrays are  $\lambda/4$  apart vertically, the ends of the upper array so forming the coupling between the two arrays. A gain of 10dB is claimed for the "birdcage" which, Minimitter point out, is roughly equivalent to a five-element Yagi with a boom length of nearly 60ft. The "diameter" of the birdcage aerial is under 20ft.

Sound Vision Services—the fourth of the newcomers —were showing telescopic masts embodying an ingenious principle. The lowest extending section is raised by winding up on a ratchet barrel a wire rope joined to its base and passing over a pulley at the top of the fixed section; but the upper sections have the guy-wires attached to their bases in a similar manner. Thus having made fast the guys, winding up the lowest extending section raises the upper sections. For nontelescopic rotating masts a rotating guy-ring using nylon bearing inserts which do not need greasing should eliminate the necessity for climbing the mast which always seems to be necessary during the worst blizzard of the winter.

James Scott, importers of Hallicrafters receivers, have added Dow-Key accessories to their range. Items on



Heathkit "Mohican" transistor communications receiver using ceramic i.f. transformers (arrowed).

show included a.c.-energized coaxial relays for aerial switching and automatic TR switches which operate by allowing the transmitter output to build up a high bias on the grid of a valve connected as an amplifier in the aerial lead to the receiver. On "receive," the bias decays, allowing the valve to give a slight gain.

Judging by the R.S.G.B. exhibits, transistors have made an enormous impact on the amateur during the last two or three years, particularly for use in receivers; but this year sees the first introduction of a commercial communications receiver using them. The Heathkit "Mohican" was certainly the centre of attraction among the kits on Daystrom's stand: this set covers 550kc/s to 32Mc/s in five bands, using OC171s for r.f., mixer and local oscillator stages. The 455kc/s i.f. is amplified by three type OC45, and to achieve a bandwidth of 3kc/s piezoelectric ceramic i.f. "transformers" are used. Two of these are of ring and dot construction and approximate to double-tuned transformers, whilst two more behave as single-tuned "acceptor" circuits in the emitter leads of the second and third i.f. stages. Control of the b.f.o. is exercised by a voltage-variable capacitance diode and the whole receiver with its 400-mW output stage takes only 35mA from the 12-V battery much less than the dial lamps, for which a separate switch is provided!

Minimitter were showing a transistor converter covering the amateur bands up to 14Mc/s. This has an i.f. of 600kc/s, to feed into an ordinary receiver and the consumption is 1mA (one) from the internal 4.5-V battery.

Not all the exhibition was devoted to amateur radio high fidelity enthusiasts would have found something to interest them from say, Jason, Heathkit or Aveley, who were showing American Dynakit designs and new output and mains transformers using toroidal windings. For mains transformers this offers the valuable advantage that the hum field is greatly reduced.

For those who want to make their own music the British Recording Club were demonstrating "electronic music" with the aid of a tape reproducer, and showing some of the equipment used. For instance, a ring modulator can be used to produce the very thing that most of us spend time getting rid of—intermodulation which has many musical possibilities, or another undesirable—valve noise—can be turned into a "new sound" by passing it through filters.

Finally, for those who want to make the means of making music, Jennings, our fifth newcomer, were showing a basic electric guitar kit. This consists of guitar body parts, strings, etc., and a magnetic pickup unit placed under the strings, which are of steel. Thus an output is produced when a string is plucked and, after passing through a volume and a "top-cut" control (which, as it is connected in an inductive circuit, has a greater variety of effect than is usual) mounted on the instrument this can be fed into an amplifier. Jennings claim that the output is comparatively high and, as most modern radio receivers are fairly sensitive, the gramophone amplifier section of a radio-gram or radio can be used.

#### NEW DATA RECORDING EQUIPMENT

#### UNUSUAL TAPE TRANSPORT MECHANISM

A NOVEL design of magnetic-tape transport mechanism is a feature of new digital data recording equipment manufactured by the Computer Department of Redifon Limited. This mechanism moves the tape past the recording and reading heads in small separate steps, instead of continuously as in conventional systems. The new mechanism permits controlled variations to be made independently in the speed of recording and read-out data: controlled variations can also be made in the time delay produced by the passage of data from the recording to the reading heads.

The tape (35mm wide) accommodates up to sixteen recording tracks side by side. The drive system operates in such a way that the tape is quickly started, advanced at a uniform speed during most of the step, and then quickly stopped. The interval of uniform forward motion in each step is made to coincide with a recording or read-out period.

To get the variation in tape transit time between the recording and read-out heads two sprocket drive wheels are used; one wheel is near the recording heads and the other near the read-out heads. If both sprocket wheels drive the tape at the same speed, then the length of tape between the wheels remains the same, and the time delay between the two heads is constant. If, however, the "read-out" sprocket wheel is made to move at a different speed from the "recording" sprocket wheel, then the length of tape between the two wheels alters, and the time delay between the recording and read-out heads varies accordingly. For driving the two sprocket wheels at independent speeds two separate stepping motors are used. These are energized by pulses from variable-frequency pulse generators—the recording or read-out rate varying according to the pulse

A great advantage of this type of data recording system is that, since the tape velocity is fast during each step, cumbersome high-frequency carrier or flux-sensitive head recording techniques can be avoided, and simple saturation recording by d.c. pulses is effective.

### **RING AERIALS**

#### INSTALLATION AT MOTALA LONG-WAVE BROADCASTING STATION

**T** is unusual nowadays to read of major developments in long-wave broadcasting, and this makes the recently published description of the station under construction at Motala, Sweden<sup>1</sup>, all the more interesting.

The station works on a wavelength of 1,571m (191kc/s), and at present the transmitter power of 150kW is radiated from a single relatively low aerial; this arrangement is similar to the B.B.C. long-wave station at Droitwich. The useful range of such a station may be limited by interference (either manmade or from stations working on nearby channels), by fading, or both. Interference can be minimized by increasing the transmitter power, and it is proposed to increase the power of Motala fourfold, i.e., to 600kW. However, this by itself would not be worth while unless at the same time the aerial is made "anti-fading," i.e., unless the upwards radiation towards the ionosphere is reduced. The usual way of achieving this on medium wavelengths is to use a mast between 0.5 and 0.6 wavelength high, but this is impracticable for long-wave broadcastingthe mast would have to be about 3,000ft high. However, substantially the same performance can be achieved by using, instead of a single tall mast radiator, a number of low aerials in the form of one or more concentric rings, and the interest of the Motala project centres round the decision to use a system of this kind.

Let us consider first a ring containing an infinite number of aerials; from considerations of symmetry we can see that two arrangements will give uniform radiation in all horizontal directions—the usual requirement for a broadcasting station. The aerials may be driven with equal co-phased currents (inphase ring), or with the currents of the same amplitude but with the phase progressing uniformly round the ring, the total phase shift being an integral multiple of  $2\pi$  radians (progressive-phase ring).

In order to achieve anti-fading characteristics the in-phase ring must be associated with a central aerial,

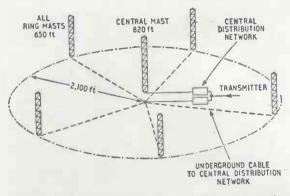


Fig. 1. Schematic diagram of the Motala ring aerial system.

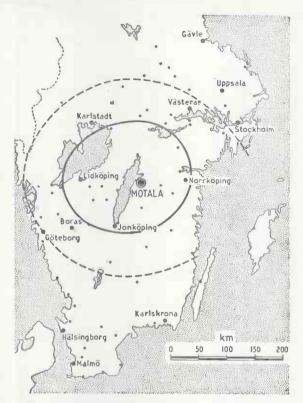


Fig. 2. Fading limits of the old (solid line) aerial system at Motala together with estimated limits of the new ring aerial. This map is reproduced from Magnusson and Stranden's article in the E.B.U. Review.

carrying a current opposite in phase to that of the ring (the central aerial can be regarded as a ring of zero radius). By changing the ratio of the current in the central aerial to that in the ring, the radiation at any specified angle to the vertical can be reduced to zero; in other words, we can achieve a vertical radiation pattern which is very like that for a single aerial of the optimum height. The progressive-phase ring does not require a central aerial, but the vertical radiation pattern is similar to that for a half-wavelength aerial, so that the anti-fading characteristics are less favourable than for the in-phase ring.

The above discussion has been in terms of a ring containing an infinite number of aerials. The effect of using a finite number is to introduce serrations in the horizontal radiation pattern, and a sufficient number of ring aerials must be used to reduce these serrations to an acceptable value. For the in-phase ring it is advantageous to use an odd number.<sup>2</sup>

The Motala ring aerial, shown schematically in Fig. 1, will employ a central mast 820ft high,

\* B.B.C. Research Department.

and an in-phase ring comprising five masts, 650ft high, fed by underground cables from a central point. The ring radius will be 2,100ft and the earth system will occupy an area of 725 acres. For a system adjusted to give zero radiation at an angle of 45° to the vertical (i.e., having a vertical radiation pattern similar to that for a single mast 0.59 wavelength high) the current in each mast is expected to be approximately 200A and the voltage at the base to be approximately 30kV r.m.s. The initial complication of a ring aerial is therefore very high in respect of the number of masts, the area of the site required, and the equipment for energizing aerials. It is, howev r, the only practicable way of achieving a worth-while increase in the coverage of long wavelengths. The estimated fading limit for the new Motala station, shown in Fig. 2, is almost double that of the present installation.

The use of ring aerials to achieve anti-fading characteristics is by no means new. An in-phase ring was first used (but only experimentaly) in Germany in 1931.<sup>3</sup> In 1939 the B.B.C. also tested an in-phase ring aerial on a wavelength of 342m at Brookman's Park; here the object was to increase the range of the Regional medium-wave service without infringing a severe Air Ministry restriction on the height of masts permitted at this station. However, when this restriction was lifted after the Second World War the B.B.C.'s aim was achieved more economically in the case of this medium wavelength station by erecting a single high mast, which is still in service. The progressive-phase ring aerial was first proposed in 1936.<sup>4</sup> One was put into service in 1939 at Allouis, France, working on a wavelength of 1,648m, and designed for a power of 900kW.<sup>5</sup> The station was destroyed during the Second World War and the aerial system was not rebuilt in ring form.

Although much thought has been given over the past thirty years to the application of ring aerials to long- and medium-wave broadcasting, as far as the author is aware the Motala station, when completed, will be the only such system in service. Broadcasting engineers will await with interest to hear how the performance of the station compares with the expectations of the designers.

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<sup>1</sup> Magnusson, E., and Stranden, F: "Planning the New Motala Long-Wave Broadcasting Station," *E.B.U. Review*, Part A.—Technical, No. 61, June, 1960, p. 107. <sup>2</sup> Page, H.: "Ring-Aerial Systems — Minimum Number of Radiators Required," *Wireless Engineer* 

Number of Kadiators Required, Wireless Engineer (now Electronic Technology), October, 1948, p. 308. <sup>3</sup> Harbich, H., and Hahnemann, W.: "Vorläufiger Bericht über Versuche zur Bekämpfung der Schwunderscheinungen im Rundfunk mit Antennengebilden Ublicher Höhe ( $\lambda/4$ ) and Grosserer Horizontalausdehnung," Elektrotechnische Zeitschrift, 17th December, 1931, p. 1545.

1931, p. 1545. <sup>4</sup> Chireix, H.: "Antennes à Rayonnement zénithal réduit," L'Onde Electrique, July, 1936, p. 440. <sup>5</sup> Adem, M.: "Le Nouveau Poste National de la

<sup>6</sup> Adem, M.: "Le Nouveau Poste National de la Radiodiffusion Française à Allouis (Cher)," Le Génie Civil, 11th November, 1939.

#### **BOOKS RECEIVED**

Television Antenna Handbook by Jack Darr. Practical handbook for the service technician covering the principles, choice and installation of all types of television aerials (v.h.f. and u.h.f.). Profusely illustrated by examples of American commercial practice and including a chapter on roof techniques and safety precautions. Pp. 248; Figs. 260. Howard W. Sams & Co., Inc., 2201 East 46th Street, Indianapolis, Indiana, U.S.A. Price \$3.95, U.S.A. Obtainable in U.K. through R. S. R. Hutchison, 60 Arno Vale Road, Woodthorpe, Nottingham.

Antennes voor FM, KG en TV by Ing. H. J. A. Smit and A. J. Dirksen. Practical handbook (in Dutch) of design and installation of f.m., short-wave and television aerials systems. Calculation of element length and spacing, impedance matching; mechanical problems of aerial erection, and a chapter on communal aerial systems. Pp. 191; Figs. 239. De Muiderkring N.V., Bussum, Netherlands. Price Fl. 5,90.

Electron'sche Muz'ekinstrumenten by H. Meiyer Jr. and W. Heggie. Circuits in theory and practice, with descriptions (in Dutch) of the electrical and mechanical details of some representative designs. Pp. 168; Figs. 153. De Muiderkring N.V., Bussum, Netherlands. Price Fl. 7,50.

Grundzüge der Electroakustik by F. A. Fischer, Dr. Phil. Second edition, revised and enlarged, of an authoritative treatise on the theoretical foundations of electroacoustics. Provides a succinct mathematical treatment of transducers both as emitters and receivers of acoustic energy. Pp. 210; Figs. 141. Fachverlag Schiele & Schön, Markgrafenstrasse 11, Berlin, S.W. 61. Price, DM 24. Handbuch des Rundfunk-und Fernseh-Grosshandels 1960/61. Illustrated guide with specifications and prices of current West German radio and television receivers, car radios, record players and tape recorders. Pp. 315, Verlag für Radio-Foto-Kinotechnik G.m.b.H., Berlin-Borsigwalde. Price DM 4,80.

Glossary of Terms Used in Telecommunications (includ ng Radio) and Electronics. British Standard 204: 1960. Third revision of this standard, including the five supplements to the previous edition. Based on current usage, but with guidance in the choice of preferred terms. Covers general electrical terms, telecommunication components and circuits, radio terminal equipment, propagation and media, classification of radio waves and transmissions services, e.g., telegraphy, telephony, broadcasting, radar and navigational aids, and inductive co-ordination (design of systems to minimize interference from power supplies). Pp. 351. British Standards Institution, 2, Park Street, London, W.1. Price 35s.

Rad'o Engineering Formulæ and Calcu'ations by W. E. Pannett, A.M.I.E.E.' Collection of useful formulæ covering a wide field in radio transmission and reception, and dealing with problems in the every-day practice of design, installation and operation of radio stations. Pp. 200; Figs. 165. George New 23, Ltd., Southampton Street, London, W.C.2. Price 17s 6d.

The Story of the Ionosphere by J. A. Harrison, M.A., M.Ed., Ph.D. Elementary exposition of radio propagation from Hertz to the first artificial satellite with sidelights on the development of radio technique, radar and radio-astronomy. Pp. 103; Figs. 123. Hulton Educational Publications, Ltd., 161/166, Fleet Street, London, E.C.2. Price 10s 6d.

## TRANSISTOR NOISE

#### By "CATHODE RAY"

**UVERYTHING** must have a beginning, and incredible though it may seem there are people for whom this will be their very first Wireless World. They must be warned that in order to avoid vain repetition such as the heathen use I am assuming readers' knowledge of my remarks—or their equivalent (if any)—on valve noise in the last issue, and indeed of those on k in the issue before that. It being unrealistic to suppose that such knowledge will return instantly and fully to the minds of even those who read the said remarks, I will recapitulate.

Quite apart from any man-made interference, all amplification is limited by random (or "white") noise caused by electricity and matter not being continuous but made of particles. That is why very weak signals are heard against a background of escaping gas, or seen on a background of animated graininess. There are two main sources of such noise. One of them is the continuous agitation of electrons in matter, caused by heat. (You may say it is a form of heat.) The constant that connects the electrical noise power with the absolute temperature is the k mentioned above, equal to  $1.38 \times 10^{-23}$ joules per degree. The maximum noise power of this self-generated kind that can come from any bit of circuit is

#### *k*TB .. ..

where T is the absolute temperature (beginning at  $-273^{\circ}$ C) and B is the frequency bandwidth in cycles per second. In practice this power usually

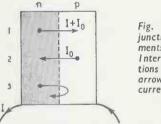


Fig. 1 Diagram of a p-n junction, showing the movements of typical electrons. Internal arrows show directions of electrons; external arrows, positive direction of current.

.. (1)

yields a few microvolts, and the only source of it worth bothering about is the input circuit of the amplifier. The maximum electrical power of any kind always results when a source works into a resistance equal to its own (say R), and from that it follows that the equivalent noise e.m.f.  $E_N$  is given by

$$\mathbf{E}_{\mathrm{N}}^{2} = 4\mathbf{R}\mathbf{k}\mathbf{T}\mathbf{B} \qquad \dots \qquad (2)$$

That is Johnson or circuit noise.

The other main kind is shot noise, caused where electrons stream from one electrode to another under the influence of an electric field, as in a valve. The individual electrons do not follow one another at exactly equal intervals of time, but randomly, and it is the resulting irregularities that constitute shot noise. The basic equation is  $L^2 = 2eB$  (3)

$$I_N^2 = 2eIB \qquad \dots \qquad (3)$$

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where  $I_N$  is the noise current, e is the charge on an electron, and I is the total current.

In any valve as normally used, the anode current I is not limited by the temperature of the cathode but by the crowd of electrons hanging about just outside it (the space charge) and these have a considerable smoothing effect on the noise, reducing it to perhaps only a tenth of the amount given by the above equation. On the other hand, whenever the current divides up—say between an anode and a screen grid—a further random element is introduced which is not smoothed. This contribution, which in a pentode usually exceeds the former kind of shot noise, is called partition noise.

Now we are about ready to start on transistors. But to prevent disappointment I had better make clear that transistor theory is almost always more complicated than valve theory, and the noise aspect is no exception. In fact, until fairly recently I wouldn't have presumed to expound it at all. A paper by an (perhaps *the*) outstanding authority on the subject, van der Ziel, published in 1955,<sup>1</sup> had a most discouraging appearance. But, as so often mercifully happens, theory which at first looks beyond the reach of any but Nobel prizewinners is eventually found to be capable of being explained to school children. By 1958,2 van der Ziel (with a collaborator, Becking) had been having second thoughts to such good effect that they produced what has been described as a more rigorous proof in about half the number of pages, and in spite of that condensation most of it is intelligible even to me. What follows is based on it.

The original transistors, of the point-contact variety, were excessively noisy, and their workings mysterious withal, so it is fortunate that they soon retired in favour of junction types, which are the only ones to be considered now. A feature of the treatment by van der Ziel and Becking that particularly appealed to me was the type of diagram shown here as Fig. 1, in which the various possible ways in which particles could move were considered in This diagram, which applies to junction turn. diodes, shows electrons only, and divides them into three classes. Most of the free electrons in the diode crystal are those belonging to atoms of the " donor ' impurity put there to make one end of the crystal These diffuse around, and of those that n type. cross the frontier into the p-type zone some of them -the majority, if the p end if positive or forward-biased-are gone for good. They form Class 1. Others, in their aimless wanderings, find themselves back again in the n zone; they are Class 3. There are also a few electrons liberated by heat throughout the crystal, regardless of the type of impurity present. Those in the n zone need not be separately considered, because all that cross the frontier can be

<sup>&</sup>lt;sup>1</sup> Proc. I.R.E., Nov. 1955, p. 1639.

Proc. I.R.E., March 1958, p. 589.

included in Classes 1 or 3. Those in the p zone that cross over to n form Class 2.

Note that this classification is the same whatever external e.m.f. may be applied, but of course the numbers of electrons in the three classes are greatly affected thereby. The external current is denoted by I. If  $I_o$  stands for the reverse current due to Class 2 (remember, electrons being negative move oppositely to the conventional direction of current) Class 1 must add up to  $I+I_o$ . Class 3, unable to make up its own mind, obviously adds nothing to the external current.

The "emission" of electrons in Classes 1 and 2 from their zones, and their journeys to the opposite zones, are analogous to the crossing of electrons from cathode to anode in a thermionic diode—except for the absence of space charge, which is one respect at least in which semiconductor electronics is simpler than valves. (The electron charges are neutralized by the equal positive charges of their parent atoms.) So the shot-effect formula in its simpler form without space-charge smoothing (eq. 3) applies:

$$i_1^2 = 2e (\mathbf{I} + \mathbf{I}_o) \mathbf{B} \dots$$
 (4)

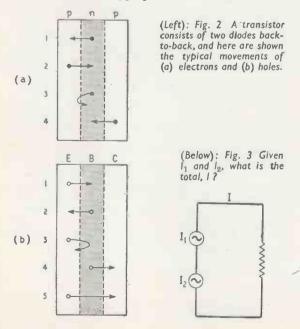
and 
$$i_2^2 = 2e I_0 B$$
 ..... (5)

where  $i_1$  and  $i_2$  are the r.m.s. noise currents due to all the electrons in Classes 1 and 2 respectively. These formulae are for current-squared, partly because they are simpler that way, and partly because random noise powers (proportional to current-squared or voltage-squared) can be added together simply, whereas currents and voltages can't if one wants a correct answer.

The movements of Class 3 are random, due entirely to thermal agitation, and so their contribution to noise is given by an adaptation of eq. 2. That was in terms of voltage, but "Ohm's law" enables us to adapt it:

$$u_{3}^{2} = \frac{E_{N}^{2}}{R^{2}} = \frac{4kTB}{R} = 4kTGB$$

where G is the appropriate conductance. The



question of what is appropriate here is a little tricky. It turns out that G is equal to  $G_a - e(I+I_o)/kT$ , where  $G_a$  is the conductance of the junction to a.c. So

$$i_{3}^{2} = 4kT\left(G_{a} - \frac{e(I + I_{o})}{kT}\right)B.....$$
 (6)

The total noise-current-squared (say  $i^2$ ) is  $i_1^2 + i_2^2 + i_3^2$  and substituting their values from (4), (5) and (6) we get

$$i^{2} = i_{1}^{2} + i_{2}^{2} + i_{3}^{2} = 2e (I + I_{o}) B + 2eI_{o}B + 4kTG_{a}B - 4e(I + I_{o})B = 4kTG_{a}B - 2eIB$$
 (7)

 $I_o$  very conveniently disappears, leaving us with  $i^2$  in terms of the external current I.

This calculation may seem a bit dodgy in places, but it is confirmed by practical measurements, which is a comfort.

So far our currents have been composed exclusively of electrons. In an actual junction diode each class of electrons has its counterpart in a class of holes moving in the opposite direction. The current I includes both, and I don't think it would have occurred to me to doubt that the noise currents are in the same proportion therein as they are in the electron currents just calculated. But just to make sure, van der Ziel gives formal proofs that what holds for holes alone does also for electrons alone and for both combined.

Skipping that, we pass on to transistor triodes. Unlike a triode valve, a transistor consists of a pair of diodes back to back. So it can be tackled as an extension of what we have just done. Fig. 2(a) shows how the electrons move in a p-n-p transistor. The first three classes correspond to those in Fig. 1, with the base as the n zone and the emitter as the p. One might expect the same three to be duplicated in the base-collector junction. But, unlike the baseemitter junction, under working conditions it is always biased in the "reverse" direction, preventing electrons from flowing towards the collector. So Classes 1 and 3 are absent.

Fig. 2(b) shows hole movement. The first four classes are the same as for electrons in reverse, but there is one extra class which is in fact the most important of all, comprising the holes that pass right through and on into the external circuit.

If equation (1) is true for diodes in general it should be true of the two transistor diodes in particular. Let us apply it first to the emitter diode, in which the current and conductance can be called  $I_e$  and  $G_e$  respectively to distinguish them, and of course  $i_e$  is the corresponding noise current. Substituting these in (7) we get

$$i_e^2 = 4kTG_eB - 2eI_eB$$
 ..... (8)

The formula for the collector diode is the same except for c instead of e, but we can save ourselves the trouble of writing the first term, because we have already decided that the collector diode is reverse-biased and so its conductance is negligible. And because  $I_c$  in this case flows from n to p instead of p to n (which we have taken as the positive direction without actually saying so) it is negative. So

 $i_c^2 = 2eI_cB$  ..... (9)

The noise current  $i_e$  can be regarded as coming from a current generator in parallel with the input junction of a noise-free transistor, and  $i_e$  as coming in parallel with the output junction. But that is not quite all.

In Fig. 3, I<sub>1</sub> and I<sub>2</sub> are currents from two a.c.

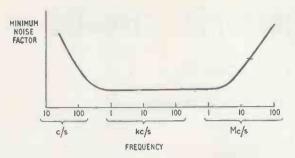


Fig. 4 Typical noise/frequency characteristic for a transistor.

generators working simultaneously. How much is the total current, I? The answer is that it depends on whether the two generators are synchronous or not. If not, then the rule we have been following is correct:  $I^2 = I_1^2 + I_2^2$ . But if they have the same frequency, then  $I^2 = (I_1 + I_2)^2$ . For example, if  $I_1$  happened to be equal to  $-I_2$ , I would clearly be 0. But if they were unequal in frequency, or random, then it wouldn't even be possible to say  $I_1 = -I_2$ , because there couldn't be a constant 180° phase difference between them.

The second method of addition applies even when the currents have no definite frequency-as with noise currents-so long as they are synchronous. This is so with the straight-through current of Class 5, but as that is not the whole current the total result is somewhere between the two extremes. In technical language, the noise currents  $i_e$  and  $i_c$ are *partially correlated*. Note that the difference between  $i_e^2 + i_c^2$  and  $(i_e + i_c)^2$  is  $2i_e i_e$ . We might think that even a partially correlated current would therefore be larger than a totally uncorrelated one, but in this case the correlated parts of  $i_e$  and  $i_e$  have opposite signs, so the total noise is reduced by the correlation. Because  $i_5$  is the main current, correlation is nearly complete, at least at low frequencies.

To calculate the total noise, one has to fit the noise generators given by eqs. 8 and 9 into a transistor equivalent circuit, and, since one is very much interested in how it depends on frequency, this equivalent circuit should include capacitors and anything else that may be necessary to simulate highfrequency effects in the transistor. The calculations then become no easy task. They have been performed-and, what's more, checked with measurements on actual transistors-by J. M. Stewart.<sup>3</sup> The badness of noise is best expressed as noise factor or figure, which is a measure of the extent to which it affects weak signals and so limits effective amplification. Stewart used a common-base circuit for his analysis, but found that the noise factor was essentially the same for the other two configurations. He circumvented another complication—that the noise factor depends on the external impedances connected to the transistor-by assuming that they would be adjusted to give the lowest and therefore best noise factor.

An interesting point is that the forking of emitter current inte collector current and base current causes partition noise, analogues to that in multielectrode valves; but because the base current is usually a very small fraction of emitter current the effect is relatively small.

<sup>3</sup> Proc. I.E.E., Part B Supplement No. 17, May 1959, p. 1056.

A more serious thing is that there is also Johnson noise due to the resistances of the various parts. The most important is that due to what is usually denoted by  $r_{bb}$ —the resistance between the base terminal and the active part of the base.

Such calculations reveal a frequency characteristic in which noise is flat or "white" over a wide middle range, but slopes upward at each end somewhat as in Fig. 4. Why?

At the low-frequency end the additional noise is more or less inversely proportional to frequency, so it is often called 1/f noise. Like the rather similar " flicker" noise in valves, it seems to lack a comprehensive explanation, but for our purpose it may be sufficient to note that it has been traced to surface leakage and similar imperfections of manufacture. While at one time it seriously affected the a.f. band, it has been pushed progressively down towards the sub-audio frequencies. Even those strange types who work among such things as servo-mechanisms and bio-electricity, to whom 20 c/s is an ultra-high frequency, are helped by this, because the lower the frequency at which the 1/f noise starts the less there is of it at any given frequency.

At the other end the rise is due chiefly to the various effects that cause transistor performance in general to fall off. Even "white" noise sources, if inserted in networks that include capacitors, give a net output that varies with frequency, Correlation becomes less, for one thing. It must be remembered that noise factor is worsened by anything that reduces signals more than noise. Against this gloomy thought there is the information that over the valuable middle frequencies the noise factor for junction transistors can be lower even than that for valves. But don't assume that this is necessarily so with the transistors you buy!

### Brit.I.R.E. Premiums

THE first recipients of the newly established Associated-Rediffusion premium of the British Institution of Kacio Redinusion premium of the British Institution of Kadio Engineers are K. G. Freeman (Mullard) and D. C. Brothers (B.B.C.) who share the £50 prize. Their papers "A Gating Circuit for Single-gun Colour Tele-vision Tubes" (Freeman) and "The Testing and Operation of  $4\frac{1}{2}$ -in Image Orthicon Tubes" (Brothers) were read at the Institution's 1959 Convention. The Usersite Userstein (20 cm) cross to B. P.

The Heinrich Hertz premium (20 gn.) goes to P. B. Helsdon (Marconi's) for his Convention paper "Tran-sistors in Video Equipment." K. Burrows, of Imperial College, receives the new A. F. Bulgin premium (15 gn.) for his paper "A Rocket Borne Magnetometer."

Four co-authors who are with the National Research Council of Canada receive the Brabazon award (1) for for "A Low-drain Distress Beacon for a Crash Position Indicator." T. C. R. S. Fowler, of Bristol Aircraft, is awarded the Leslie McMichael premium (10 gn.) for Six-channel High-Frequency Telemetry System.

The new Charles Babbage award for an outstanding The new Unarles Babbage award for an outstanding paper on electronic aspects of computers (15 gn.) is shared by Dr. T. B. Tomlinson (Southern Instruments) whose paper was "Switching Circuits Using Bi-direc-tional Non-linear Impedances" and Dr. M. Prutton (I.C.T.) for "Ferro-electrics and Computer Storage." I. J. P. James' paper "A Vidicon Camera for In-dustrial Colour Television" has been awarded the 10 gn. Marconi award. He is with E.M.I. All three projnents of the graduateship asymptotic

All three recipients of the graduateship examination prizes are from overseas—C. S. Sujan (India), O. Smikt (Israel) and W. W. Cridland (Canada).

# **Elements of Electronic Circuits**

### 21.-Differentiation and Integration

By J. M. PETERS, B.Sc. (Eng.), A.M.I.E.E., A.M.Brit.I.R.E.

**LARLY** sections in this series have illustrated how a simple series C-R combination can "differentiate" or "integrate" a rectangular input waveform with an accuracy depending on the relation between time constant employed and the length of the waveform. These circuits, with the aid of feedback amplifier, can be used to produce waveforms approximately proportional to the derivative or the integral of the input waveform. If certain precautions are taken it is possible for high accuracies to be achieved.

Most of the commonly-used methods are based on the fact that

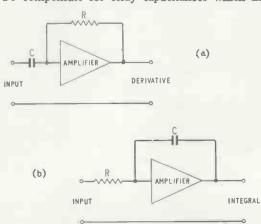
$$v_{\rm c} = \int i_{\rm c} \, \mathrm{d}t$$
 or  $i_{\rm c} = \mathrm{d}v_{\rm c}/\mathrm{d}t$ 

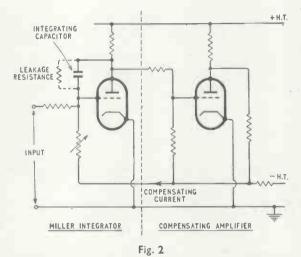
where  $v_{e}$  is the voltage developed across the capacitor and  $i_{e}$  is the current taken by it.

L-R differentiating networks may also be encountered but the resistance of the inductor introduces errors. However, in some applications these errors can be tolerated.

Figs. 1 (a) and (b) illustrate an improvement on the simple C-R circuit for differentiating and integrating respectively. It can be shown mathematically that if the gain of the amplifier without feedback (A) is constant; then the time constant of the differentiating circuit of Fig. 1(a) is effectively divided by (A+1). This is accomplished without the corresponding decrease in amplitude of the output (which would have occurred in the ordinary way with the simple C-R circuit). The gain of the amplifier compensates for this and an improvement in accuracy of waveforms due to the decreased time constant results. A similar treatment for the integrating circuit of Fig. 1 (b) leads to the result that the use of the feedback amplifier effectively multiplies the time constant by the factor (A+1), resulting again in a more accurate integrated waveform.

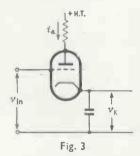
To compensate for stray capacitances which in-





troduce time delays in the feedback loops of differentiators, and leakage resistance and dielectric absorption in the integrating capacitor of integrators, it is usual to introduce compensating circuits. Fig. 2 shows an integrating circuit of the Miller type (see

description of the Miller time base in No. 14, June 1960) followed by a compensating amplifier. The object of this amplifier is to provide a current which always equal and is opposite to the leakage current through the integrating capacitor and the variable resistor is set so that balance of charge and discharge is achieved. One of the aims in both feed-



back differentiating and integrating circuits is to keep the voltage variations at the first grid to a minimum. The closer one can get to **this** ideal the nearer will the output be to either the derivative or the integral of **the** input. A high loop gain is essential and several stages of amplification are often used to achieve this.

Mention must be made of current, as opposed to voltage, feedback for differentiation or integration. Referring to Fig. 3, it will be seen that this circuit acts as a differentiator.

$$v_{\mathbf{K}} \propto \int i_{\mathbf{K}} = i_{\mathbf{a}}$$

and due to feedback  $v_{\text{K}} = v_{in}$ . The anode current therefore varies as the derivative of the input voltage.

Before leaving this section it should be noted that a number of mechanical and electro-mechanical devices are available for carrying out these operations, but their response time is much longer than that of the circuits considered here.

### LETTERS TO THE EDITOR

The Editor does not necessarily endorse the opinions expressed by his correspondents

### **Stereo Broadcasts**

I HAVE listened with interest and enjoyment to the B.B.C.'s series of experimental stereo broadcasts.

Presumably the original intention of these tests was to assess the relative merits of stereo and mono reproduction. This task has recently been made more difficult by the use of a TV sound a.m. transmitter for the righthand channel. This inevitably produces a marked difference in the signal-to-noise ratio of the two channels. Similar performance from each channel can only be provided by using two similar transmitters or, preferably, one transmitter using a multiplex system.

In the meantime the case for stereo has been amply demonstrated by the gramophone record industry. After a somewhat shaky start they are issuing and selling vast numbers of stereo discs to a public which is obviously no longer in any doubt.

Now that the burden of proof has been lifted from the B.B.C., surely they could demonstrate a singletransmitter method of stereo broadcasting. The Mullard system (with its crosstalk 45dB down) seems to have much to recommend it, including the facility of radiating two separate programmes from a single transmitter.

Now seems the ideal time to start regular stereo transmissions from a single transmitter in each B.B.C. region. If this is not possible surely we could have an initial pilot scheme operating from Wrotham only, as Saffron Walden, M. S. GOTCH.

Essex

### **Television Standards**

MR. BANTHORPE (Oct. 1960 issue) advocates a change of field frequency from 50 c/s to 60 c/s for the purpose of reducing flicker. The price to be paid would include an increase of bandwidth and a loss of the mains-hold feature as already mentioned by Mr. Ban-thorpe. But American experience shows that telecine and telerecording process become more involved and give inferior performances under those conditions. Once a field frequency has been reached which is sufficiently high to enable rapid movements to be depicted satisfacfrom flicker by selection of afterglow time.

Mr. Heffernan, in the same issue, claims that "a real advantage . . . is gained in getting one's peak aerial power in the blacks and the all-important synchronizing pulses." Has Mr. Heffernan forgotten that an unmodulated carrier contains no intelligence and that a voltage change from 0-3 produces just the same signal as a change from 7-10?

Mr. Charles Rogers (Nov. 1960 issue) writes of the bandwidth needed for equal horizontal and vertical definition. Concentrating first on the word "needed," is this not putting the cart before the horse? A suitable number of lines is incidental to the bandwidth available. It is the bandwidth which is the raw material and which costs money, lines do not. There is no single correct relationship, and discussion\* has shown that the country's experts disagree strongly amongst themselves as to what is the best relationship, their opinions covering wide ranges.

(\* See, for example, "Relation Between Picture Size, Viewing Distance and Picture Quality," L. C. Jesty, Proc.I.E.E., Part B, No. 23, Vol. 105, Sept. 1958.)

As for equal horizontal and vertical definitions, I suggest that there is no such condition. If there are l lines in a picture, then the number of different vertical heights which images may have is l. But images may be shown with an infinite variety of horizontal lengths. In the vertical plane an image may take up one of l positions. In the horizontal plane it may take up one of an infinite number of positions.

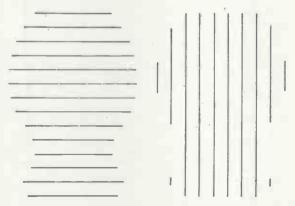
It has been known for at least a decade that the conventional relationship, usually expressed in cart-before-horse fashion as

### $f_{\rm max} = A l^2 f_{\rm p}/2$

gives, when transposed, a number of lines which is too small. The Television Advisory Committee's recommendation for increasing vertical definition to a greater extent than horizontal definition merely recognizes this fact.

I believe that the logical approach to this problem is to determine the available bandwidth, and with a multi-standard camera channel, limited to this band-width, to select that number of lines which gives the most pleasing results. This, of course, has been done, but it is surprising how old ideas stick. This is prob-ably because the subject makes a neat (but very misleading) examination question.

I believe that the fallacy of the conventional equation can be demonstrated in the following manner :---



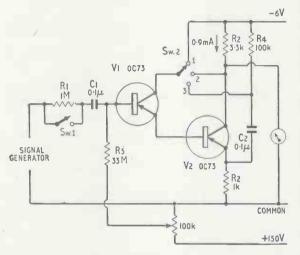
I reason that if this equation were valid, then the definition of an image would be governed by the sum total of the line lengths that lay within the boundaries of the image of an object, irrespective of whether they coincided with the major or the minor axis of an elongated image. The above sketches are of the same object drawn in these two different ways. R. C. WHITEHEAD.

Sutton, Surrey.

### Transistorized Wein Bridge Oscillator

THE use of the super-alpha pair as the input stage in the article in your August 1960 issue has already provoked considerable comment, but one feature of the arrangement has been misrepresented by Mr. F. Butler and overlooked by subsequent correspondents. On page 388 he says that the super-alpha stage is an emitter follower directly coupled to a common-emitter stage, the collector of the emitter follower being taken to the collector of the common

emitter stage and not the supply rail. He then says: "The principal effect of this change is to place in series with the collector circuit of the first transistor the whole output voltage developed across the load resistance. This voltage is opposite in phase to the amplifier input voltage and constitutes a large series negative feedback signal. The effect of this [feedback] is to cause a further increase in the already high input resistance . . ." Surely the effect of this feedback is to reduce the input



resistance, for, when the collector voltage of an emitter follower is reduced, an increase in the input current is required to maintain the emitter current constant. Thus the stage appears to have a resistor connected between base and emitter, the value using hybrid parameters, being  $(B + 1)/h'_{22}$ . By using the super-alpha stage this resistor is connected between input and output of the pair and thus appears as if it were across the input k + 1fimes smaller, k being the stage voltage gain. At high frequencies the collector to base capacitance shunts the input still further, an effect which is k + 1 times greater with the super-alpha stage than with the conventional arrangement.

Using the circuit shown a set of measurements was made with the same two transistors throughout. With Sw2 in position 1 the circuit is that of an emitter follower driving a common-emitter stage, in position 2 the super-alpha stage results, and in position 3 the collector of V1 is connected so that it closely follows the input voltage. The input impedance is measured by inserting  $R_1$  in series with the input and observing the change in the output voltage.

At 100 c/s the change of Sw2 from position 3 to position 2 reduces the input impedance from 1.6 to  $1.3M\Omega$ . This gives  $(B + 1)/h'_{22}$  as  $30M\Omega$  (very approx.) and hence if B = 30 (and experiments show that B is maintained at very low currents)  $1/h'_{22} = 1M\Omega$ , which is much higher than the value at normal currents.

At 3 kc/s changing from position 3 to position 1 reduces the input impedance from 1.4 to  $0.8M\Omega$ . This gives the collector to base capacity of V1 as 5½pF.

At 3 kc/s changing from position 3 to position 2 reduces the input impedance from 1.4 to 0.17 megohms. This gives the collector to base capacity as  $7\frac{1}{2}pF$ , i.e. larger at the lower collector voltage, which seems reasonable though both values of the capacitance are smaller than the value at normal currents.

A table setting out the input impedance of the three circuits at various frequencies is shown below

	Frequency (kc/s)	0.1	1.0	3.0	10	30	Sw 2
impedance <	Conventional circuit Super-alpha Collector " lifted "	1.3	0.6		0.19 0.045 0.5	0.05 0.02 0.07	1 2 3

The technique of lifting the collector of an emitter follower is used in Application Report No. 6 of Texas Instruments. Here a d.c. amplifier is described with an input impedance of  $200M\Omega$ .

J. C. A. TALBOT Barnet, Herts.

The author replies:--Mr. Talbot has drawn attention to what in some circumstances may be a fundamental defect of the superalpha transistor pair. I accept his conclusion that the effect of feedback from the collector load is to reduce the input impedance of the first stage and not to increase it as I stated. Nevertheless, from Mr. Talbot's tabulated figures it is clear that at low frequencies the input impedances of the three circuit arrangements he describes are all of the same order of magnitude but that, partly because of collector-base capacitance, the input impedance of the super-alpha circuit falls off more rapidly with an increase in frequency than is the case with the other two circuits. It turns out, quite fortuitously, that this characteristic is acceptable in the present case because, on any given frequency range, the Wien bridge network has its lowest impedance at the highest frequency. Thus we have in effect an amplifier of variable input impedance driven from a source of variable output impedance and any mis-match is of little practical consequence. Naturally the effect of collector capacitance is much worse in the case of audio-frequency transistors than it is for h.f. or v.h.f. transistors in which this capacitance is very small. If outputs extending up to 100 or 200 kc/s are desired it would be advisable to use transistors with alpha cut-off frequencies in the range 30-100Mc/s.

I have a copy of the excellent Texas Instruments Application Report referred to by Mr. Talbot, but it was issued after my paper was written. It is possible that their techniques for designing transistor circuits of very high input impedance might have applications in the R-C oscillator field but, if carried to extremes, these techniques become too elaborate to incorporate in a simple piece of apparatus.

F. BUTLER.



P.V.C. holders have been produced by Bush Radio to protect the miniature circuit diagrams which they now issue to dealers for the servicing of television and sound receivers.

# MANUFACTURERS' PRODUCTS

### NEW ELECTRONIC EQUIPMENT AND ACCESSORIES

### Automatic Circuit Tester

THE American Lavoie Laboratories' "Robotester" can test to any one of four alternative tolerance limits the resistance and a.c. or d.c. voltage between any two of up to 250 test points at a rate of about 80 tests per minute.



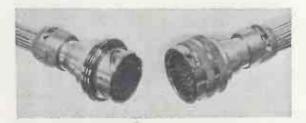
Lavoie Laboratories "Robotester" automatic circuit tester.

The tolerance limit as well as the resistance or voltage to be tested are individually set for each test by means of a punched tape which is fed into the Robotester. If a fault is found the machine automatically stops until switched on again, and at the same time the measurement accuracy can be checked. Alternatively, an additional unit can be attached which automatically records the fault and then restarts the tester. It is claimed that with this instrument on production runs final test times can be shortened by about 80%. The Lavoic Laboratories' Robotester costs £2,850 and is distributed in this country by Metrix Instruments Ltd., of 54 Victoria Road, Surbiton, Surrey.

### Versatile Multi-pin Connectors

A NEW range of aluminium-cased, multi-pin connectors, the Mark 6, is now available from The Plessey Company. These connectors embody many singular features and they also accommodate a considerably greater number of contacts than other Plessey connectors of similar size.

Mark 6 connectors are available in four sizes with a fixed and free unit in each; both units are obtainable



The new versatile Mark 6 Plessey multi-pin connector.

with either plugs or sockets and are easily convertible to coupler units. A silicone-rubber moulding forms the insulator and fully shrouds individual contacts, cable joint and a portion of the cable dielectric.

Insulation resistance between contacts and earth throughout is greater than  $10^{5}M\Omega$  at 500V. The voltage rating is 1kV r.m.s., irrespective of con-

The voltage rating is 1kV r.m.s., irrespective of contact class, at ground level, reducing to 350V at 70,000ft, or under equivalent conditions. Current ratings range from 5A to 12A and the connectors are satisfactory for operation in ambient temperatures between  $-55^{\circ}$ C and + 155°C.

Contacts are silver-plated brass and socket inserts are designed to give equal mating loads irrespective of the length of engagement.

These new connectors can be used in four alternative positions by merely varying the position of the bayonet couplings. This feature prevents mis-coupling to wrong units and also prevents mis-mating. The connector is designed to meet the major performance requirements of British Military Specifications ELI987 and DEF5321, and American Military Specifications MIL-C-5015D and MIL-C-26500 (U.S.A.F.).

A special high quality connector with nickel-plated housings and gold-plated contacts is available to special order. Further details can be obtained from The Plessey Co., Ltd., Ilford, Essex.

### Multi-channel Oscillograph

THE new Savage and Parsons Type RG32-12/15 multichannel oscillograph has a response of up to 1000c/s at a sensitivity of 7.5mA/cm. Eight alternative paper speeds from 5in/min to 150in/sec can be used, and a well-defined trace is obtainable at writing speeds of 12,000in/sec. Up to .200 feet of paper can be recorded at any one time and this can either be run continuously or alternatively automatically stopped after 2, 5, or 10 feet have been used. Identification of each channel every two feet irrespective of the paper



Savage and Parsons Multi-Channel Oscillograph.

speed is provided by a sequential interruption of each trace. The signals are obtained from mirror galvanometers and the traces develoced spontaneously within a few seconds by using ultra-violet sensitive paper. The magnet block carries fifteen galvanometers but some of these will usually be used as time markers or, connected to the mains, as a reference frequency. The cost of this oscillograph is in the region of  $\pounds 1,200$ , the exact value being obtainable on application. It is manufactured by Savage and Parsons Ltd., of Watford, Herts.

### Sub-miniature I.F. Transformers

A TRULY sub-miniature transformer, the "Fidis," which measures only  $\frac{1}{6}$  in in diameter and  $\frac{1}{6}$  in high, is being produced in France primarily for use in pocket-



Orega "Fidis" sub-miniature I.f. transformer compared in size with a cigarette.

sized transistor receivers, but it has, also, applications wherever space is strictly limited. The base soldering pins are positioned so that the transformer can be used in printed circuits.

Despite its diminutive size it is fully screened and the "Q" of the windings is claimed to be of the order of 150.

Further details can be obtained from the French company Orega, 106, rue de la Jarry, Vincennes (Seine) France, a subsidiary of the Compagnie Générale de Télégraphie Sans Fil of Paris.

### Transistor DC/DC Converters

RECENTLY introduced by Aveley Electric is a range of transistor converters supplying h.t. voltages from 6V to 24V batteries. A special feature of the Avel converters as they are called, is the use of toroidally-wound transformers on H.C.R. square-loop type magnetic cores, together with bifilar windings to ensure accurate balance and fast switching of the transistors with over-shoot voltages kept to a minimum.

The converters, which are available with ratings of 5W to 120W, provide normally 300V d.c. output with intermediate tappings of 200V and 250V. Adequate protection is provided to prevent damaging the transis-



Chassis of a typical Avel transistor DC/DC converter made by Aveley Electric.

tors by inadvertently reversing the polarity of the input or by other wrong connections.

Snown in the illustration is one of the basic units employed in the converters, the massive chassis block forming a heat sink for the transistors. This basic unit can be fitted in a variety of housings to meet users' requirements. By duplicating and triplicating the basic units in a single housing higher power, or voltage, outputs and/or multiple voltages are readily obtainable.

The basic units are very compact, the chassis, or heat sink, measuring only  $3\frac{1}{6}in \times 2\frac{1}{4}in$  with height dependent on output rating. In a 45-Watt unit, for example, this is about 3in.

Further details can be obtained from Aveley Electric Ltd., South Ockendon, Essex.

### **Television** Aerial Isolator

TO the Egen range of components has now been added an aerial isolator, Type 364. It isolates the aerial on a.c./d.c. television receivers and is a single compact rugged unit which complies fully with the individual requirements of BS415. Insertion loss is very low and its



Egen television aerial isolator.

electrical specification ensures maximum performance at all frequencies envisaged for domestic receivers.

It is completely co-axial with full screening of the inner conductor. The series inductance of the feedthrough capacitor in the outer conductor is exceptionally low, which is a desirable feature. Feed-through capacitors in both conductors are each 470pF. They are tested to 3,000V d.c.

Provision is made for direct mounting to the receiver chassis or to a separate bracket and the isolator can be supplied with any required length of coaxial cable for connection to the receiver input circuit. The external socket accepts a standard coaxial plug to R.E.C.M.F. specification. The makers are Egen Electric Ltd., Charfleet Industrial Estate, Canvey Island, Essex.

### Transformer Kits

THE introduction of a range of transformer kits comprising a stack of laminations, a bobbin and fixing clamps where appropriate, will come as welcome news to many concerned with the construction of a prototype ironcored transformer, and these also include home constructors.

Filteen sizes and three different lamination materials, Silcor, Radiometal and Mumetal respectively, are available and prices for single kits range from 9s 3d to 15s 6d in Silcor, 9s 3d to 63s 9d in Radiometal and 9s 6d to 93s in Mumetal. Further details and prices for quantities are obtainable from The Belclere Company Ltd., 171 Cowley Road, Oxford.

New Plastic Foil Capacitors.—The following corrections should be made to the report on the Tropyfol capacitors on page 616 of the December issue. The largest capacitance in the 400V range is  $0.47\mu$ F. The 125V type measuring  $4.5 \times 12$ mm is a  $0.01\mu$ F capacitor while the  $13 \times 24$ mm capacitor of  $1\mu$ F is a 125V type.

# **Iron-Cored Coupling Transformers**

With Particular Reference to the Design of Iron-Cored A.F. Components

BY D. SAULL

A TRANSFORMER is like the conjunction in English language—it correctly joins two or more individual circuits together; e.g., a microphone to the input of an amplifier, one stage of an amplifier to the next stage, an amplifier output to a speaker or transmission lines, etc. And for each particular application a transformer must usually be specifically designed. The range of such designs is therefore very large.

The block diagram in Fig. 1 shows the family tree, as it were, illustrating some of the many uses to which transformers may be put in electronic circuitry. If the reader pauses to consider that a complete book could be written on the subject matter of any one application—say pulse transformers—it becomes evident that a single article in a journal could not possibly cover the complete subject of transformers.

The purpose of this article is, therefore, most certainly not a comprehensive study of impedance matching transformers; neither is this article intended to be read by transformer designers. No, its purpose is to aid the engineer, whose mind is centred on the equipment he is designing and from which he does not wish to be deviated, to reach a starting point if he is obliged to get down to designing his own transformer. Alternatively, to assist the engineer to call up his requirements if he is fortunate enough to possess the facility of a transformer design department to supply his needs. For these reasons, the author intends to generalize in his statements to avoid entering into details, which would defeat the purpose of this article.

Before leaving the reference to a transformer design department, it may be worth explaining the value of such a department to a firm employing a fair number of design and development engineers.

It can be shown that the average transformer

design time for a transformer design department, over a period of time, is half to two thirds the average time taken when individual design and development engineers design their own transformers. This is a money-saving consideration for readers in a managerial position in industry to ponder.

The reason for this shorter design time, claimed by transformer design departments, is twofold. First, although the transformer designer is no more skilled than the development engineer, and perhaps even less skilled, he has the advantage of constantly designing transformers and so learns the short cuts; whereas the development engineer is called upon only infrequently to design transformers. Secondly, one design frequently assists another design, and often a transformer is a modification of a previous design; a transformer design department has records of the work carried out but—in the case where separate designs are done by various engineers—Fred doesn't know what George is doing.

As the subject matter of this article is not intended to be "the management of industry," let us press on with transformer design.

A transformer is basically a number of turns of conducting wire wound round a block of iron. If two such windings, consisting of any number of turns, are wound round any block of iron and an a.c. signal is applied to one winding some form of a.c. signal will appear across the second winding. What relation the output signal will bear to the input signal, in this case, is anyone's guess. And, perhaps, here we come very near to the heart of the subject.

As was said earlier, a transformer correctly joins together two or more electric circuits. It may serve to connect a low-output impedance microphone to the high impedance grid circuit of a valve, or the high-output impedance of a valve to the low-input

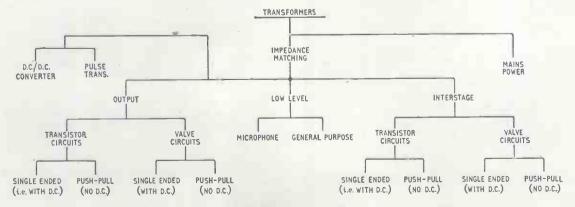


Fig. 1. Transformer family tree showing some of its numerous applications

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impedance of a speaker. Whichever the case, if the transformer is ideally designed, each circuit will be correctly terminated and the respective circuits do not know that the transformer is there at all like the broody hen and the china egg. However, the china egg is cold to sit upon, and the practical transformer is not ideal, having losses, phase shift, etc.

How near to the ideal may the engineer expect to get? The problem is rather like a tug-of-war contest with perfection on one end of the rope and costs on the other. The road to perfection leads to higher grade transformer laminations, larger lamination sizes, and longer development time; factors which may be expressed in terms of  $\mathcal{L}$ , s, and d.

So a compromise must be made; and usually the engineer is obliged to specify the lowest standards that he can tolerate. The maximum insertion loss

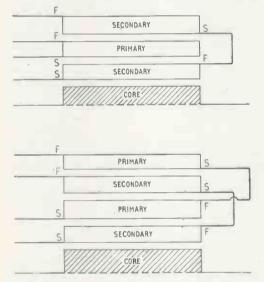


Fig. 2. Two methods of sectionalizing windings

and harmonic distortion, together with the minimum frequency response, that is acceptable to the engineer, must be specified.

A transformer functions electrically the same whatever its application, and its design is a compromise of a number of variables. For one application insertion loss is an important factor, for another it is of little importance; the same applies for winding d.c. resistance, leakage inductance, etc.

Perhaps the easiest method of reaching a compromise between perfection and costs is to examine the merits of the various lines of action that the engineer may take. The author will, therefore, try to take the reader briefly, in turn, through the practical considerations required for the design of each transformer application illustrated in the family tree in Fig. 1. But, before passing on, it might serve well to discuss some of the various grades of transformer irons available, the methods of winding, and three different ways of assembling the laminations.

Winding Methods.—There are two main methods of winding transformers, (a) paper interleaved layer winding on a cheekless former and (b) layer winding without interleaving paper on a former with end cheeks.

In the author's opinion, type (a) is more suitable for the larger variety of transformer, e.g., the output transformer of a high-quality 10-watt a.f. amplifier. The figures for overall space factor and the turns-per-inch-squared quoted in the Tables 1 and 2 of the article "Power Transformer Design," published in the June (1960) issue of WIRELESS WORLD, will hold good for this application. Type (b) should be employed in the smaller variety of transformer for the reason that, when interleaving paper is employed, the thickness of the paper becomes comparable with the diameter of the winding wire used. For this smaller variety, an overall space factor of 30 to 35%, which includes the former, should be assumed.

Flux Gradient .-- The flux density across the window area of the laminations is not constant due to the increasing length of the iron path; if the lines of flux are compared with the rings of water ripple surrounding a stone thrown into a pond, this idea becomes more evident. This flux gradient increases the leakage inductance. To reduce the leakage inductance the primary winding may be split into two or more sections and sandwiched between the secondary windings. The result of so dividing the primary winding into sections unfortunately increases the capacitance between windings. This capacitance may in turn be reduced by separating the primary and secondary windings by several layers of interleaving paper. Further sectionalizing and reversing the direction of the windings, but connecting the windings in the same magnetic sense, results in improved characteristics; Fig. 2 illustrates two possible methods.

A compromise must be arrived at to obtain the required results. For the above considerations, the number of turns on the primary winding should tend towards a minimum, which suggests in order to maintain the required primary inductance, that a core material with as high a permeability as possible—consistent with costs—should be employed. The latter applies particularly to low level transformers.

Grades of Iron.—Commencing from the lowest, in permeability and cost, the following list of core materials will give an approximate guide to the respective applications:

- (i) Silcor 25; 0.014in and 0.020in thickness—1% silicon content.
- (ii) Silcor 107; 0.014in and 0.020in thickness $-3\frac{1}{2}$ % silicon content.
- (iii) "C" Cores; 0.002in to 0.013in thickness—cold rolled grain-oriented silicon steel—may be worked at 30% higher flux densities than (i) or (ii).

These first three grades are suitable for use in output transformers employed with higher-power, good-quality a.f. amplifiers. For this application, it is better to have a large core of cheaper material than a smaller core of the higher, more expensive, grade material such as Radiometal or Mumetal.

"C" cores are particularly adaptable to such uses due to their shorter mean magnetic-path length and higher permeability, and may be conveniently worked at  $5 \times 10^3$  gauss. This material saturates at

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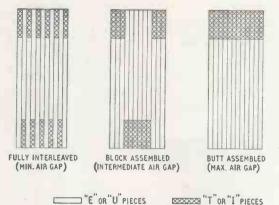


Fig. 3. Methods of assembling core laminations.

a higher value of d.c. component, which is worth bearing in mind when designing transformers carrying d.c. The range of sizes obtainable extends down to small sizes suitable for interstage transformers with a d.c. component.

- (iv) Radiometal lies between silicon steel and Mumetal, and may be worked at flux densities twice that for Mumetal, but does not have as high a permeability as the latter material.
- (v) Mumetal. The use of this material, with its high permeability, makes possible the design of miniature transformers for transistor cir-cuits and transformers of smaller size with high primary inductance and low leakage inductance. The material saturates at low values of d.c. and is, therefore, not suitable for applications where a d.c. current is present. However, where the d.c. is of low value, and the primary inductance is not high, this material may be used, i.e., transistor a.f. interstage transformers.

Average Permeability:

Silicon Steel-350 Radiometal-1,600 Mumetal-15,000

Turns Reduction for a Given Inductance:

Radiometal =2.14 to 1 Silicon Steel  $\frac{\text{Mumetal}}{\text{Silicon Steel}} = 6.8 \text{ to } 1$ Radiometal =3.3 to 1 Mumetal

Methods of Assembling Laminations.-Fig 3 illustrates the three main methods of assembling the laminations to produce various degrees of effective air gap in the iron core.

Low-Level Transformers.—A low-level transformer is one which is used to couple two circuits where the power available from the primary circuit is of a small order, i.e., milliwatts. Such applications might be a microphone transformer, a transformer connecting a 600-ohm signal generator to a 20-ohm load, a moving-coil meter transformer to extend the current measuring range, etc.

The first consideration might be a low insertion

loss, which calls for a low-loss, high-permeability type of lamination-Mumetal or Radiometal-the author's choice would be Mumetal. The second consideration is frequency response; the third consideration, closely related to frequency response, is second harmonic distortion.

Low Frequencies.-The distortion in a transformer at the lowest frequency is dependent upon the maximum operating flux density, and falls off sharply with frequency increase.

The frequency response at the lowest frequency is dependent upon the value of shunt inductance of the primary winding. Fig. 4 illustrates how this fall off at the lower frequencies arises, and the table gives practical values.

High Frequencies.-At high frequencies the leakage inductance and the capacitance of the windings increases the frequency fall-off. A well-designed low-level transformer may have a frequency response of from 20c/s to 100kc/s. The design problems of such a transformer might be listed as follows:

- (a) High primary inductance.
- (b) Low leakage inductance.
- (c) Low flux density-order of 600 gauss.
- (d) Low winding capacitance.
- (e) Low d.c. resistance-dependent upon application.

Factors (a) and (b) plus (d) work in opposition because by increasing the number of turns, to in-

Fig. 4. Equivalent transformer primary circuit.



r = SOURCE IMPEDANCE Z = GENERATOR LOAD (REFLECTED FROM SECONDARY WINDING) X1 = REACTANCE OF PRIMARY WINDING

crease the primary inductance, the leakage inductance increases proportionately (i.e., leakage inductance and primary inductance are both a function of the square of the number of turns). The capacitance of the windings also increases.

Increasing the stack size increases the primary inductance but does not reduce the leakage inductance to the extent that might at first be expected, due to the increased mean turn length of the windings.

The answer lies in using core materials of high

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X <sub>L</sub> equal to	Total % harmonic distortion	Frequency fall-off
6R	0.3	
4R	0.4	
2R	0.7	1dB
R	1.7	3dB

permeability—hence a very good reason for using Mumetal. Radiometal follows some little way behind—if the purse strings are tied.

The tighter the specification, the greater the time taken to produce the completed design. The designer is fortunate indeed if he produces the finished transformer at the first attempt. The procedure is more likely to result in a first-off version to be tested and the shortcomings noted. This is where the transformer design department again has the advantage—the second-off and final design, in this case, will be the modification to a previous near design.

Making a Start.—The required primary inductance should first be calculated from  $\omega L=R$ , 2R, 4R or 6R according to the permissible distortion at the lowest frequency to be reproduced (see table).

The next step is the choice of lamination size, which, unless a direct selection can be made based on previous experience, must satisfy the two following equations:—

(i) 
$$N^2 = \frac{L \times 10^8 l}{3.2 \times A \times \mu}$$

(11)  $B = \frac{1}{28.6 \text{ fNA}}$ 

### Where:

 $\mu =$ Initial permeability.

- L=Henries.
- A = Cross-sectional area of the core in sq in.
- l = Mean magnetic path length in in.
- N = Number of turns.
- f = Frequency (lowest).
- B = Flux density (order of 600 gauss).
- E=Applied maximum voltage when the transformer is in circuit.

Having selected a suitable lamination size, the number of primary turns required should be evaluated from equation (i) and then a check made using equation (ii) to ascertain that the flux density is not in excess of 600 gauss.

The secondary turns are then evaluated from: ----

rns ratio = 
$$\sqrt{L_1/L_2}$$

where  $Z_1$  and  $Z_2$  are the input and output circuit impedances.

When a sample transformer has been wound it should be tested for frequency response. This may be done with the aid of a signal generator and a valve voltmeter. The transformer should be correctly terminated in a resistive load and the voltage appearing across the primary winding at A (see Fig. 5) should be adjusted, with the aid of the valve voltmeter, to be that which will appear across the winding when the transformer is connected in circuit. Fig. 5 illustrates the circuit required for test-

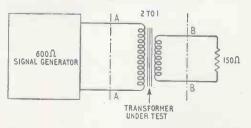


Fig. 5. Arrangement of circuit for testing a transformer

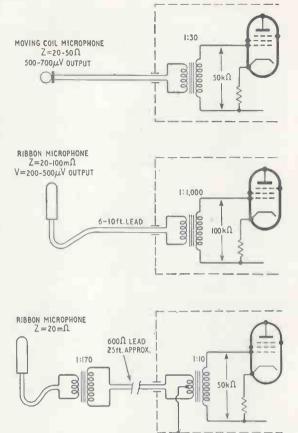


Fig. 6. Matching microphone to input of amplifier

ing such a transformer with a turns ratio of 2 to 1. The valve voltmeter is then connected across the secondary B and the voltage noted. This procedure is repeated in steps throughout the frequency range for which the transformer is designed. A frequency response graph is then plotted and the insertion loss evaluated.

Frequency fall-off at the low-frequency end calls for an increase in primary inductance; frequency fall-off at the high frequency end is due to either leakage inductance or winding capacitance. The latter factors may be identified by connecting a small capacitor across the secondary winding and if the fall-off in frequency response does not increase at this end the trouble is leakage inductance—if it does fall off more sharply, then the trouote is capacitance. The alterations necessary to improve these short-comings have already been discussed earlier in "Flux Gradient."

Microphone Transformers.—The low-impedance ribbon and moving-coil type microphones require a matching transformer to connect the low-impedance of the microphone to the high-impedance grid circuit of the first amplifier valve.

Fig. 6 illustrates these two types of microphone matching. The transformer windings are wound for Radiometal or Mumetal laminations.

To reduce hum pick-up when the transformer is mounted in the amplifier, narrow "U" laminations

are used, the induced hum voltage cancelling out in the coils wound equally on either limb. The secondary winding is usually wound with fine wire order of 50 s.w.g.—the high d.c. resistance being unimportant in this application. The windings of the secondary may conveniently be six separate bobbins, three on either limb of the "U" laminations; the primary consisting of thin copper strip.

Fig. 7 shows a suitable transformer of turns ratio 1 to 1,000; the secondary is made up of six bobbins each of 1,000 turns and the primary consists of 6 turns of thin copper strip.

Fig. 7(b) shows the direction of the winding and the resulting magnetic polarity due to the signal current at any instant. It will be seen that, although the two sets of coils are connected to be additive in inductance, the coils on each limb produce magnetic fields that are in opposite directions in plan view. Hence a magnetic field, caused by hum, would produce a hum voltage equal and opposite in polarity in the windings on either limb. The narrow "U" laminations are used to ensure that both sets of coils are in a field of equal hum flux, which might not be so if the limbs were spaced well apart.

Output Transformers.—This type of transformer covers a very wide range and could not be dealt with fully in an article of this nature. They do, however, fall into two main groups, those for push-pull operation and those for single-ended operation. The former have no d.c. component in their windings whereas the latter may have.

For medium power a.f. amplifiers, ordinary silicon irons function quite well. For higher power a.f. amplifiers, in the order of 20 to 30 watts, it is an advantage to use "C" core laminations.

A good rule-of-thumb guide to lamination size is the weight of the core in lb should be 0.17 times the output watts of the amplifier and the volume in cubic inches should be 0.7 times the output watts of the amplifier. For good fidelity these figures should be doubled.

The maximum flux density of operation should not exceed 5 kilogauss for a reasonable **distortion** factor. The calculations for primary inductance are the same as under "Low-Level Transformers."

When the primary windings carry a d.c. component it may be necessary to increase the effective air gap in the core by block or butt method of assembling the laminations.

The frequency response of output transformers is a similar consideration to the factors effecting interstage transformers.

The permeability of the core will vary with applied signal level—thus the frequency response will normally extend lower as the signal level is increased.

When a transformer is connected to a source of zero impedance there is no voltage drop incurred by the magnetizing current. As in the practical case there is source impedance, a voltage drop will occur due to the magnetizing current; since the magnetizing flux is not by any means a sine-wave for an input sine-wave current, due to the B/H curve, distortion resulting from the out-of-phase magnetizing current occurs.

The ratio of  $\omega L/R$  (where R is the resultant of anode and load resistances in parallel) should be kept as low as possible consistent with other requirements. The d.c. resistance of the primary, and of the reflected secondary, windings may be regarded as part of the source impedance for this purpose. Further to this consideration, the maximum flux density should be restricted to the straight part of the B/H curve. Fig. 8 illustrates these principles.

Interstage Transformers for Class A and Class B Working.—The amplification of the stage at midfrequency is very nearly equal to the product of the

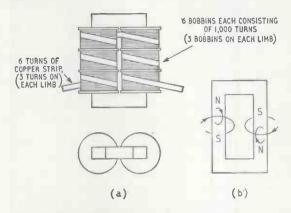


Fig. 7. Hum pick-up by microphone transformer minimized by an "astatic" form of winding using both limbs of a "U"-shaped core. (a) shows practical assembly and (b) the direction of winding

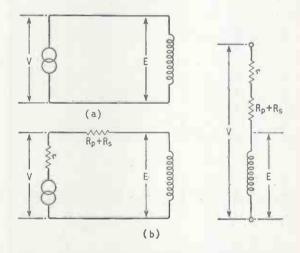


Fig. 8. Output transformer equivalent circuit; (a) no source impedance V = E; (b) with source impedance and resistance of windings. Magnetizing current a function of frequency and load current

amplification factor of the valve and the turns ratio of the transformer.

The gain falls off at low frequencies due to the decrease in the reactance of the transformer primary winding. At frequencies at which  $\omega L = R$  the response will fall off 3dB; at frequencies  $\omega L = 2R$  the response will fall off 1dB.

The leakage inductance and shunt capacitance of the transformer, in conjunction with the anode and the winding resistances, form a low-Q resonant circuit. The gain will fall off sharply above this resonant frequency. This resonant frequency may be varied by altering the value of the leakage inductance or the winding capacitance.

Transistor Output and Interstage Transformers.— These transformers may be wound on very small Mumetal laminations. Because the impedances associated with transistors are small compared with those for valves, primary inductances are of a low order—e.g. 250 to 500 millihenries.

**Pulse Transformers.**—It is not intended to more than mention this type of transformer in passing except to say that, due to the wave shape of the pulses the transformer is required to pass, the harmonic content may reach high frequencies. Trouble is frequently experienced when the leakage inductance resonates with the winding capacitance producing unwanted spikes.

To minimize this effect, a copper screen may be placed between the primary and the secondary windings and the screen connected to an earthy point in the circuit. This screen consists of one turn of 0.002-in copper foil interleaved with paper to prevent a short-circuited turn.

**D.C./D.C.** Transistorized Converters.—Toroidal cores are usually found most efficient for these types of circuits, particularly when the VA output is of a small order. However, transformers utilizing medium grade silicon-iron laminations can be designed for this application which are fully satisfactory for VA ratings in excess of 6VA, working at frequencies up to 1,000c/s.

Ferrite materials are also employed for circuits operating at several kc/s where, due to the low value of smoothing components at this frequency, the overall size of the complete apparatus may be kept small.

The nature of the converter circuit requires that saturation of the core is reached twice in each cycle of operation. The use of silicon-iron laminations therefore restricts the frequency of operation. The iron losses in the core are the chief losses in this type of circuit—hence a minimum VA rating for a high efficiency working is fixed by the relation between total losses and the output VA. Since this type of equipment is mostly used working from storage batteries, efficiencies are important.

Although the subject matter in this article is, of necessity, very briefly dealt with it is hoped that it will serve some use to the engineer, even if only as a pointer to select a starting point from which he may approach his objective, and to the management in the electronic industry in illustrating the advantages of setting aside a department, however small, for the specific purpose of designing transformers.

### HI-FI P.A.

**READERS who** visited the National Radio Show and were impressed with the quality of reproduction on the B.B.C. Gramstand may be interested in the following brief details of the installation:—

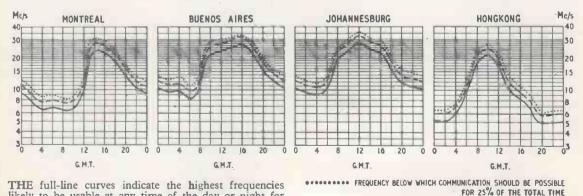
The requirements were for high-quality public address coverage over a large area to feed a standing audience of up to 1,500 at one time. Anything less than high quality would not have served the purpose because the main output from this exhibit was commercial records which nowadays employ such "gimmicks" as closebalance, pre-emphasis, accentuated reverberation and frequency correction which would be lost on the normal type of public address system.

Six high-quality monitoring loudspeakers of a type normally used in control rooms and control cubicles and known as the LSU/10 were sited at strategic points round the audience area. The loudspeakers, mounted on plinths so that the sound output was not baffled by people standing in front of them, were built into the auditorium walls and in pillar casings.

Basically, the construction of an LSU/10 loudspeaker consists of an acoustic cabinet with a vented enclosure containing a large unit with a 15-in cone carrying a 3-in speech coil. The middle and high-frequency unit is mounted concentrically with the large unit and has a light domed aluminium diaphragm, 14-in diameter, attached to a speech coil of the same diameter. There is an additional tweeter unit separate from the main loudspeaker units to extend further the top frequency response. The 10-watt amplifier feeding the speaker units is also contained in the cabinet.

### SHORT-WAVE CONDITIONS

Prediction for January



THE full-line curves indicate the highest frequencies likely to be usable at any time of the day or night for reliable communications over four long-distance paths from this country during January.

Broken-line curves give the highest frequencies that will sustain a partial service throughout the same period.

WIRELESS WORLD, JANUARY 1961

ON ALL UNDISTURBED DAYS

PREDICTED MEDIAN STANDARD MAXIMUM USABLE FREQUENCY

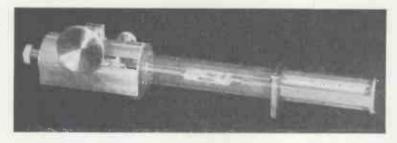
FREQUENCY BELOW WHICH COMMUNICATION SHOULD BE POSSIBLE

# TECHNICAL NOTEBOOK

Adjustable Waveguide Termination recently developed at the U.S. National Bureau of Standards has two moving parts: a short-circuiting plunger and a resistive vane. The double-exposure illustration shows the plunger and vane positioned inside rectangular waveguide. 2

be "epitaxial". Diffused-base transistors have been made at Bell Telephone Laboratories on epitaxial layers of both germanium and silicon and have shown the improvements to be expected from a thinner collector region. For example, the switching time in a typical circuit

continually compared electronically with the unknown voltage so as to fire a stroboscopic lamp whenever the sawtooth and unknown voltages are equal. For a fixed input voltage the stroboscopic lamp will thus be fired at corresponding points on each ramp of the sawtooth waveform, so



Independent mechanical controls are provided for rotating the vane, longitudinally sliding the vane relative to the short circuit, and sliding the entire termination assembly along the waveguide. In operation, the resistive vane reflects some energy back towards the signal source. The re-maining energy is either partially absorbed by the vane or reflected by the short circuit back towards the signal source. The relative phase of the reflections from the vane and short circuit can be altered by varying the distance between them. The amount of energy reflected by the short circuit depends upon the amount left unabsorbed by the vane, and this can be varied by rotating the vane

New Transistor Manufacturing Technique developed by Bell Telephone Laboratories reduces switching times and collector resistances of diffused-base transistors by factors of more than ten. Diffused-base transistors require a collector region of relatively high resistivity in order to attain a low capacitance and high voltage breakdown. A lightly-doped collector region is thus used, and ideally this should only be about 0.1 mil thick. However, if the collector wafer were made as thin as this, it would be extremely difficult, if not impossible, to handle mechanically. Thus a collector thickness about thirty times greater is normally used, although increasing the collector thickness increases both its resistance and also, through carrier-storage effects, the transistor switching time. Thinner collector regions can, however, be made and utilized by growing them on a low-resistivity heavilydoped mechanically supporting wafer. When the thin collector film is a direct extension of a singlecrystal supporting wafer it is said to

was reduced from 200 to 20 musec and the collector series resistance was decreased by a factor of more than ten

Digital Voltmeter using a stroboscopic technique is being imported into this country from America by Scientific Furnishings Ltd. of Poynton, Cheshire. In this voltmeter the 250 indicating numbers are registered on a drum which is rigidly coupled to the spindle of a potentiometer. This potentiometer has a standard voltage developed across it and its spindle is rotated at 1500 r.p.m. so that its wiper generates a sawtooth voltage. This sawtooth voltage is



that the same number on the drum will be illuminated by each flash of the lamp. A slowly-varying input voltage will be indicated by a slowly varying number, and interpolation between two partially-visible numbers will give a measure of a voltage between these two numbers. Alternatively, only the nearest number to the actual voltage can be indicated. This is done by making use of an additional set of pulses produced as the numbers pass a phototransistor.

#### JANUARY MEETINGS

Tickets are required for some meetings; readers are advised, therefore, to communicate with the secretary of the society concerned

### LONDON

4th. Brit.I.R.E.—"Automatic tech-niques in civil air line communications systems" by W. E. Brunt at 6.30 at the London School of Hygiene, Keppel Street, W.C.1. 5th. Institution of Mechanical Engi-neers.—Discussion on "The reliability

of mechanical engineering parts of data processing systems" at 10.30 at the Institution, 1 Birdcage Walk, S.W.I. 5th. I.E.E.—Hunter Memorial Lec-ture on "The application of electronics" of the distribution of the state of the system of the

ture on "The application of electronics to the electnicty supply industry" by Dr. J. S. Forrest at 5.30 at Savoy Place, W.C.2. 9th. I.E.E.—"Recent research in thermionics" by Dr. G. H. Metson at 5.30 at Savoy Place, W.C.2. 10th. I.E.E.—"Precision measure-ment" by G. H. Rayner and A. Felton, with supporting papers at 5.30 at Savoy

With supporting papers, at 5.30 at Savoy Place, W.C.2. 11th. Brit.I.R.E. — "Multi-layer switching devices" by Dr. G. F. Taylor

at 7.15 at the London School of Hygiene, Keppel Street, W.C.1. 12th. Radar and Electronics Asso-ciation.—" Programme circuits on tele-phone plant" by G. Stannard at 7.30 at the Royal Society of Arts, John Adam Street, W.C.2. 13th. Television Society.—" A wide range standards converter" by E. R. Rout and R. F. Vigurs at 7.0 at the Cinematograph Exhibitors' Association, 164 Shaftesbury Avenue, W.C.2. 19th. Brit.I.R.E. — Symposium on "Alpha numeric displays" at 6.30 at the London School of Lygiene, Keppel

the London School of Lygiene, Keppel

the London School of Lygiene, Keppel Street, W.C.1. 20th. Institute of Navigation.— "Presentation of height information in aircraft" by A. Stratton and K. R. Honick at 5.15 at the Royal Geograph-ical Society, I Kensington Gore, S.W.7. 20th. B.S.R.A.—" Modern electro-static microphones" by F. W. O. Bauch at 7.15 at the Royal Society of Arts, John Adam Street, W.C.2.

20th. Junior Institution of Engi-neers.—"Radio investigations of the solar atmosphere" by J. Heywood at 7.0 at Pepys House, 14 Rochester Row, S.W.1.

23rd. Institute of Physics and Physical Society.—" Some new applications of radar" by Dr. E. Eastwood at 6.0

of radar" by Dr. E. Eastwood at 6.0 at 47 Belgrave Square, S.W.1. 24th. I.E.E. — Discussion on "Machine-tool control" opened by Dr. E. H. Frost-Smith at 5.30 at Savoy Place, W.C.2. 25th. I.E.E.—"Generation and am-plification in the millimetre wave field" by W. E. Willshaw at 5.30 at Savoy

by W. E. Willshaw at 5.30 at Savoy Place, W.C.2.

Place, W.C.2. 25th. Brit.I.R.E.—"Noise correla-tion measurements" by K. R. McLach-lan at 6.30 at the London School of Hygiene, Keppel Street, W.C.1. 26th. Television Society.—Fleming Memorial Lecture on "Behind the eye"

by Prof. D. M. MacKay at 7.0 at the Royal Institution, Albemarle Street, W.1.

### BIRMINGHAM

2nd. I.E.E .- Discussion on "Broadening university courses " opened by Dr. H. E. M. Barlow at 6.30 at the James Watt Memorial Institute.

13th. Society of Instrument Techno-logy.—" Controls associated with flying " by Capt. A. M. A. Majendie at 7.0 in the Lecture Theatre, Byng Kendrick Suite, Gosta Green College of Techno-Aston Street.

logy, Aston Street. 18th. Television Society.—"Video recording" by J. Southgate at 7.0 at the New Physics Lecture Theatre, the

University. 23rd. I.E.E. — "Applications of microwaves" by Prof. A. L. Cullen at 6.0 at the James Watt Memorial Institute.

### BRISTOL

20th. Institute of Physics and Phy-sical Society.—" The physicist and the technologist in industry" by G. W. War-ren at 7.0 at the College of Science and Technology.

### CARDIFF

11th. Brit.I.R.E.—"The measure-ment of ionizing radiation" by R. G. Wood at 6.30 at the Welsh College of Advanced Technology.

### CHESTER

26th. Society of Instrument Techno-logy.—" Satellite instrumentation" by Dr. R. C. Jennison at 7.0 in the Lec-ture Theatre, Associated Ethyl Co., Oil Sites Road, Ellesmere Port.

### EDINBURGH

11th. Brit.I.R.E—"A survey of microwave valves" by C. R. Russell at 7.0 at the Department of Natural Philosophy, the University, Drummond Street.

Street. 17th. I.B.E. — "Applications of microwaves" by Prof. A. L. Cullen at 7.0 at the Carlton Hotel, North Bridge. 24th. I.E.E.—"The changing face of electronics" by W. E. J. Farvis et 7.0 at the Carlton Hotel, North Bridge.

### **EVESHAM**

16th. I.E.E.—"Magnetic recording of TV programmes" by H. E. Farrow at 7.30 at the B.B.C. Training School, Wood Norton,

FARNBOROUGH 17th. I.E.E.---" Modern ferromag-netic materials" by Dr. F. Brailsford at 6.15 at the Technical College.

24th. Brit.I.R.E.—"Masers and parametric amplifiers" by Dr. W. A. Gambling at 7.0 at the Technical College.

### FAWLEY

6th. Society of Instrument Techno-logy.—" Electronics and instrumenta-tion in the glass industry" by J. R. Beattie at 5.30 at the Administration Building, Esso Refin'ery.

### GLASGOW

12th. Brit.I.R.E.—"A survey of microwave valves" by C. R. Russell at 7.0 at the Institution of Engineers

at 7.0 at the institution of Engineers and Shipbuilders, 39 Elmbank Crescent. 16th. I.E.E.—"Applications of microwaves" by Prof. A. L. Cullen at 6.0 at the Institution of Engineers and Shipbuilders, 39 Elmbank Crescent.

#### LEEDS

26th. I.E.E.—Faraday Lecture on "Transistors and all that" by L. J. Davies at 7.0 at the Town Hall.

#### LIVERPOOL.

18th. Brit.I.R.E.—" Microminiatur-ization" by H. G. Manfield at 7.0 at

the Adelphi Hotel. 23rd. I.E.E.—"Cybernetics" by Prof. J. C. West at 6.30 at the Royal Institution, Colquitt Street.

### MALVERN

26th. Brit.I.R.E. — "Stereophonic broadcasting" by G. D. Browne at 7.0 at the Winter Gardens.

### MANCHESTER

MANCHESTER 10th. I.E.E.—"Radio communica-tion in the power industry" by E. H. Cox and R. E. Martin at 6.15 at the Engineers' Club, Albert Square. 11th. I.E.E.—"The ultrasonic microscope" by Dr. C. N. Smyth at 6.15 at the Engineers' Club, Albert Square

Square.

Square. 17th. I.E.E.—"A universal non-linear filter, predictor, and simulator, which optimizes itself by a learning process" by Prof. D. Gabor, Dr. W. P. L. Wilby and Dr. R. Woodcock at 6.15 at the Engineers' Club, Albert Square

Square. 24th. I.E.E.—Faraday Lecture on "Transistors and all that" by L. J. Davies at 7.30 at the Free Trade Hall.

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MIDDLESBROUGH 4th. I.E.E.—"The Fylingdales early warning station" by D. R. Evans at 6.30 at the Cleveland Scientific and Technical Institution.

### NEWCASTLE-UPON-TYNE

NEWCASTLE-UPON-TYNE 9th. I.E.E.—Hunter Memorial Lec-ture on "The application of electronics to the electricity supply industry" by Dr. J. S. Forrest at 6.15 at the Neville Hall, Westgate Road. 11th. Brit.I.R.E.—"The applications of photo-multipliers in industry and re-search" by J. Hambleton at 6.0 at the Institution of Mining and Mechanical Engineers, Neville Hall, Westgate Road. 16th. I.E.E.—"Precision measure-ment" by G. H. Rayner and A. Felton, with supporting papers, at 6.15 at the Rutherford College of Technology, Northumberland Road. 19th. Society of Instrument Techno-

19th. Society of Instrument Techno-logy.—"Recent advances in photo-electronic instruments" by H. Loebl at 7.0 in the Conference Room, Roadway House, Oxford Street.

### NOTTINGHAM

19th. Society of Instrument Techno-logy.—" The electrical synthesis of music" by A. Douglas at 7.15 at Not-tingham & District Technical College, Burton Street.

### SHEFFIELD

4th. I.E.E.—Christmas Holiday Lec-ture on "Colour television" by Dr. R.

ture on "Colour television" by Dr. R. Feinberg at 3.0 at the City Hall. 18th. I.E.E.—Hunter Memorial Lec-ture on "The application of elec-tronics to the electricity supply industry" by Dr. J. S. Forrest at 6.30 at the Memorial Hall, City Hall.

SOUTHAMPTON 4th. I.E.E.—"The planning and economics. of telecommunication plant" by C. J. Stubbington at 7.0 at the

University. 10th. I.E.E.—Discussion on "New semiconductor devices," at 6.30 at the University.

### WEYMOUTH

12th. I.E.E.—"Electronic aids to banking and commerce" by Dr. R. Feinberg at 6.30 at the South Dorset Technical College.

WOLVERHAMPTON 11th. Brit.I.R.E.-" An equipment for automatically processing time-multi-plexed telemetry data" by N. Purnell and T. T. Walters at 7.15 at the Col-lege of Technology.

18th. Institution of Production Engi-neers.—" The application of electronic computers to production control" by B. L. J. Hart at 7.0 at the College of Technology, Wulfruna Street.

### CONFERENCES AND EXHIBITIONS

Latest information on forthcoming events both in the U.K. and abroad is given below. Further details are obtainable from the addresses in parentheses.

LONDON		
Jan. 16-20	Physical Society Exhibition (Exhibition Secretary, 1 Lowther Gardens, S.W.7.)	R.H.S. Halls
Mar. 8		s Head, Harrow
Mar. 21-25	(A.S.E.E. Exhibition Ltd., Museum House,	.) Earls Court
Apr. 5-7	Museum Street, W.C.1.) Electrical Contacts Symposium	Brunel College
	(The Institute of Physics and the Physical Society, 47 Belgrave Square, S.W.1.)	
Apr. 6-9	Audio Festival and Fair (C. Rex-Hassan, 42 Manchester Street, W.1.)	Hotel Russell
Apr. 20-21	Television and Film Techniques Convention (Television Society, 166 Shaftesbury Avenue, W.C.2.)	Savoy Place

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Apr. 20-	Engineering, Marine, Welding and Nuclear Energy	y Olympia	-
May 4	Exhibition (F. W. Bridges & Sons, Grand Bldgs., Trafalgar Sq., W		
May 30- June 2	Components Exhibition (R.E.C.M.F., 21 Tothill Street, S.W.1.)	Olympia	A CONTRACTOR OF A CONTRACTOR O
June 12-17	Components and Materials used in Électronic Engineering (Conference) (I.E.E., Savoy Place, W.C.2.)	Central Hall	
June 21- July 1	INTERPLAS, International Plastics Exhibition and Convention	Olympia	
July 7-29	(British Plastics, Dorset House, Stamford Street, S.E.1. Soviet Trade Fair (Industrial & Trade Fairs Ltd., Drury House, Dersel Vice Colored (Colored Street, S	) Earls Court	TWO
Sept. 6-8	Russell Street, W.C.2.) Microwave Measurement Techniques Conference	e Savoy Place	
Oct. 4-12	(I.E.E., Savoy Place, W.C.2.) Computer Exhibition and Symposium (E.E.A., 11 Green Street, W.1.)	Olympia	USEFUL
Nov. 8-10	Non-Destructive Testing in Electrical Engineering (Conference) (1.E.E., Savoy Place, W.C.2.)	Savoy Place	
Nov. 13-18	(I.E.E., Savoy Place, W.C.2.) Factory Equipment Exhibition (Industrial & Trade Fairs Ltd., Drury House, Russell Street, W.C.2.)	Earls Court	AMPLIFIERS
FARNBORG			
Sept. 4-10	Farnborough Air Show (Society of British Aircraft Constructors, 29, King Street, London, S.W.1.)		
NEWCASTL			
Feb. 28- Mar. 2	North East Electronic Engineering Exhibition (Secretary N.E.E.E., c/o N.E.I.D.A., 9 Eldon Squa Newcastle-upon-Tyne 1.)	ire,	0.000
OXFORD July 5-9	Communications and Space Research Convention	a	
3	(Brit. I.R.E., 9 Bedford Square, London, W.C.1.)		
OVERSEAS Jan. 9-11	Reliability and Quality Control Symposium	Philadelphia	Model B100
	(R. Brewer, G.E.C. Research Laboratories, Wembley, Middx.)		Transistorised Amplifier for 12 volt
Feb. 15-17	International Solid State Circuits Conference (J. J. Suran, General Electric Co., Syracuse.)	Philadelphia	operation. Output 12 watts. Inputs for microphone and music. Minimum
Feb. 17-21	International Components Exhibition (Fédération Nationale des Industries Électroniques,	Paris	battery consumption-maximum effici- ency.
Feb. 20-25	23 rue de Lubeck, Paris XVIe.) International Symposium on Semiconductors (Fédération Nationale des Industries Électroniques, 23 rue de Lubeck, Paris XVIe.)	Paris	
Mar. 9-14	International Hi-Fi and Stereo Exhibition (Fédération Nationale des Industries Électroniques,	Paris	
Mar. 20-23	23 rue de Lubeck, Paris XVIe.) I.R.E. National Convention	New York	
Apr. 30-	(Dr. G. K. Teal, I.R.E., 1 E.79 St., New York 21.) German Industries Fair	Hanover	
May 9	(Schenkers Ltd., 13 Finsbury Square, London, E.C.3.)		and the second
May 2-4	Electronic Components Conference (I.R.E., 1 E. 79 St., New York 21.)	San Francisco	
May 9-17	Measurement, Control, Regulation and Automati Exhibition and Conference	on Paris	
May 19- June 4	(MESUCORA, 40 rue du Colisée, Paris 8.) British Trade Fair (Industrial & Trade Fairs Ltd.,	Moscow	Model GP100 AC operated general purpose high
	Drury House, Russell Street, London, W.C.2.)	Chierry	quality Amplifier. 4-way Input Selec-
May 22-24	National Telemetering Conference (I.R.E., 1 E. 79 St., New York 21.) Global Communications Symposium	Chicago Chicago	tor—Bass and Treble controls. 10/12 watts output.
May 22- <b>2</b> 4	(I.R.E., 1 E. 79 St., New York 21.)	-	
May 23-25	Large Capacity Computer Memories Symposium (Miss J. Leno, Office of Naval Research, Washington 25, D.C.)	1 Washington	Full details available on request
June 26-	International Measurement Conference	Budapest	
July 1 June 28-30	(Prof. J. F. Coales, The University, Cambridge.) Joint Automatic Control Conference	Boulder	
July 16-22	(Dr. R. Kramer, M.I.T., Cambridge 39, Mass.) Medical Electronics Conference (Dr. H. P. Schwan, School of Electrical Engineering, University of Pennsylvania, Philadelphia)	New York	TOL
Aug. 1-12	International Sydney Trade Fair (Industrial & Trade Fairs Ltd., Drury House,	Sydney	
Aug. 22-25	Russell Street, W.C.2.) Western Electronics Show and Convention (Wescon, 1435 LaCienega Blvd., Los Angeles.)	San Francisco	
Aug. 25-	German Radio Exhibition	Berlin	1-3 FIATLE FEACE, LOTTOOT, FIT
Sept. 3 Sept. 11-15	(Berliner Ausstellungen, Charlottenburg 9, Berlin.) International Cybernetics Congress (International Association of Cybernetics	Namur	Tel.: Museum 5817 (6 lines) Grams: Trixadio Wesdo London
	(International Association of Cybernetics, 13 Rue Basse-Marcelle, Namur, Belgium.)		A REAL PROPERTY OF THE REAL PR
Oct. 9-11	National Electronics Conference (N.E.C., 228 N. LaSalle St., Chicago.)	Chicago	

## RANDOM RADIATIONS

### Good Shot, Sir

IUST how electronic translating machines work I don't know; but work they certainly do, as is shown by the number now in use. There's no polished prose about their translations, which are more or less literal. Their vocabulary is necessarily somewhat limited, but they undoubtedly serve a most useful purpose by making books and articles understandable to people who don't know a word of the language in which they're written. Sometimes the machine is completely baffled by words which aren't in its vocabulary; but it always has a shot-and a pretty good one too as a rule. One machine kept on writing "water sheep" in the course of a translation on which it was engaged. The words occurred again and again. At first their meaning was dark to the mind of the would-be reader. Then light dawned. What it had been trying to set down was "hydraulic ram "!

### Tiros II

THE latest American satellite, Tiros II, started to do useful work very soon after it had been launched. One of its jobs is to televise pictures of cloud formations back to earth and this it is doing most satisfactorily. It travels in an almost circular orbit a little over 400 miles aloft and en-

ables a kind of weather forecast map to be made for all the areas within its range. It will thus become possible to track hurricanes with great accuracy. The satellite, besides carrying two television cameras, also has seven infra-red detectors for measuring the heat reflected from the earth, which again provides information about the gathering and the advancement of severe storms. The chief of the U.S. Weather Bureau says that information such as Tiros II sends to earth could have saved thousands of lives lost in the recent tidal waves in Pakistan. This satellite is the forerunner of the complete weather system "Nimbus" which is planned for 1962.

### Underwater TV

ONE of the many interesting items shown at the recent Industrial Photographic and Television Exhibition was the latest model of the Marconi-Siebe Gorman underwater television camera. It can be so adjusted that it is almost weightless and can be carried by a diver, but it can also be lowered from the surface. The camera will be used by the North of Scotland Hydro-Electric Board chiefly to examine the big protective grilles (some 120ft below the surface) which prevent young salmon and other fish being carried into the turbines. Experiments have shown that at consider-

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ILIFFE BOOKS LTD., Dorset House, Stamford Street, I	London,	S.E.1

### By "DIALLIST"

able depths and in water that is far from clear the TV camera can "see" a good deal better than a diver. The equipment will also be used for investigating fish life and behaviour in the various lochs. Closed-circuit TV has already become an invaluable aid to many branches of science and industry and new uses are always being found for it.

### Scaring 'em Off

STARLINGS are a major menace on aerodromes used by jet planes. A recent appalling crash was almost certainly caused by their being sucked into the intakes of the engines and another disaster was averted by the presence of mind of a pilot who saw flocks of them as he was taking off and switched off his engines. Feathers and bits of starling were later found in the intakes. How to get rid of starlings is a real problem and so far it has proved baffling. A new approach to finding a solution is being made by the Ministry of Aviation in conjunction with the Trix Electrical Company. It consists in making hi-fi recordings of the alarm calls of starlings and other birds which frequent airfields and transmitting them from strategically placed loudspeakers. It is said to have given very promising results during the experimental stages. And it's not only aviation people who are interested in large-scale bird scaring. Farmers, fruit growers and others who suffer from damage to their crops would have a ready welcome for a system proved to be successful. And so, one imagines, would authorities in London, Birmingham and other cities in which huge flocks of starlings arrive to roost every evening.

### TV in N.Z.

AT the moment I think I'm right in saying that New Zealand has only one television station in action, the 500-W experimental transmitter in Auckland. Three others, though, should be at work early this year. These are to be 5-kW stations at Christchurch, Wellington and Dunedin. All the equipment, except the aerials, which are to be supplied by the New Zealand Broadcasting Service, will be made in this country by Marconi's. It's going to be a difficult task to cover the whole of that mountainous country as the system grows and it'll certainly take some time to complete.

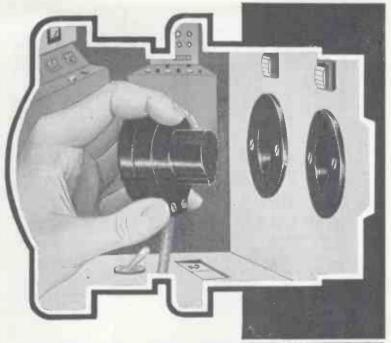
### Talking Books

OURS is, I believe, the first country to introduce books recorded on tape for the blind. An entirely novel feature of the new Talking Book machine being introduced by the National Institute for the Blind is a tape cassette containing two spools mounted one above the other and also housing and protecting the playback head. Half-inch tape is used, the 18 tracks recorded on it provide up to 20 hours of playing time. Near the end of each track the user is warned and all that he has to do to start on the next track is to turn over the cassette and press a button. And there are safety devices to prevent damage. The new tape equipment will gradually replace the disc machines now in use. Assistance in installing and servicing the machines is needed from volunteers in most parts of the country. Any reader who is able and willing to give it should write to: E. Read-Jones Manager, Nuffield Talking Book Library for the Blind, Mount Pleasant, Alperton, Wembley, Middx. Offers will be most warmly appreciated.

### A Long Time Coming

IT appears that we are to adopt the metric system for some purposes, at any rate. The old and quite outdated apothecaries' weight, with its scruples and drams and that sort of thing, is to be replaced fairly quickly by the metric system of weights. Scientists have, of course, largely adopted metric methods already and they're widely used in wireless. No one would now dream of expressing a wavelength in yards, feet and inches. But we've been as a country a very long time in making up our minds to discard our antiquated weights and measures in favour of the more sensible system. Some of our units are just nonsensical: why, for instance, are there 112 lbs in a hundredweight? I suppose that originally it deserved its name by consisting of 100 pounds; then the extra twelve were added for good measure. Now that they're about to adopt decimal coinage in the Antipodes I hope we shan't be long in following their lead. The metric system would save an immense amount of the time spent now in intricate calculations,

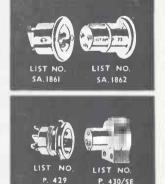
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### UNBIASED

### Exhibitiana

I HEAR the organizing committee is already holding meetings to discuss plans for the 1961 National Radio Show. If the committee has not advanced too far with its plans, I do hope it will consider making an innovation which I think would be welcomed by many.

My suggestion is that in addition to the individual exhibitors' stands, to which we are all accustomed, there should be special stands on which is grouped equipment of a like kind by all exhibitors. Thus I should like to see a stand on which there was nothing but tape recorders of all makes, another stand where there were shown all the available transistor midget sets, and so on.

At present if anybody wishes to examine the various types of tape recorders or other things before making a choice, it is necessary to go to umpteen stands which may be widely separated. It is, in fact, necessary to use tooth and claw, or at any rate, elbows and umbrella, to fight one's way through dense crowds of people who seem to have come to the show with the sole purpose of getting in the way of those who, like myself, have come with a serious purpose.

A splendid example of the usefulness of the grouping of exhibits in the manner I suggest was shown in the pages of our associate journal *Amateur Photographer* in its report of the world-famous biennial Photokina exhibition at Cologne. Although none of the stands at this exhibition were actually grouped in the manner I have suggested, the Editor of *Amateur Photographer* considered the interests of his readers by treating the stands as though they were so grouped. Thus, instead of a standto-stand report, he gave what I will call a group-by-group one.

### BY "FREE GRID"

Doubtless I shall be told that at Earls Court there is no room for these "group" stands. That is, of course, utter nonsense as can be seen from the accompanying photograph I myself took of one of the empty spaces in the gallery at last year's show.

[Logically, "Free Grid's" scheme would require additional floor area comparable with that at present allocated to firms' stands. Anything very much less would call for selection (by whom?) of representative exhibits. ---Ed.]

### Unwitting Offence

WE are often told that ignorance of the law is not a good defence if one is charged with a breach of it. But I have sometimes wondered if that would hold good in all circumstances. The thing which has brought this matter to my mind once more is the case recently reported in the Press of a young lady driving her invalid chair who, late at night, found her vehicle stuck in the mud in a rather lonely spot.

Because of a disability she was unable to get out of the chair and so was faced with the possibility of spending the night in this position. She remembered what she had been taught in the Girl Guides and endeavoured to attract attention by sounding on her electric horn the international distress signal in morse. She naturally expected the sound to reach the ears of some passer-by.

What actually happened was that her signals appeared as long and short flashes on the screen of a television set in a house not very far away. After a time the viewers realized that what they saw was a deliberate signal and when they went outside the house to consult the neighbours they could hear the actual sound of the horn although its signals were very faint.



The lady's rescue was obviously due to the fact that the radio signals she unwittingly sent out were picked up. Even had she known that the mechanism of the electric hooter formed a short-range spark transmitter, and had deliberately used it as such, it is hardly likely that any proceedings would have been taken against her, although it would certainly have been a technical offence as she had no transmitting licence.

Personally speaking, this incident has taught me a lesson and I intend to make provisions for the connection of my horn mechanism to my car-radio aerial so that I can send out a signal for help if I ever break down on a lonely road in evening broadcasting hours. I think that by making this provision I shall be guilty of establishing a radio transmitter even though I never have occasion to use it. I shall await with interest the service of a summons,

It is high time, I think, that all cars were equipped with short-range transmitters and all A.A. and R.A.C. 'phone boxes with receivers and other apparatus to enable an emergency call to be put through to the nearest depot of the motoring organizations. I have bitter memories of having to walk over two miles one stormy night to the nearest 'phone box after I had broken a half-shaft.

### I Smell, You Stink

THE words in the title of this note are, of course, those used by Dr. Johnson when he was trying to teach good English to somebody who seemed to think that the verb "to stink" was merely a cacophonous vulgarism for "to smell." As the good doctor explained, the first verb in my title could only be used of the person at the receiving end of an olfactory system, and the last verb only of the person at the transmitting end.

I mention this as I think it is high time we had another Dr. Johnson to rebuke the B.B.C. for the misuse of words which it permits its announcers to make. For instance, the word "nostalgia" used to mean homesickness but the B.B.C. announcers—and in particular the ladies seem to have changed all that. When they use the adjectival form of it, as they so often do, they seem to regard it as a synonym for reminiscent, sentimental, or half a dozen other things.

There are many other instances of this sort of thing, and I don't think Dr. Johnson would have allowed the B.B.C. the same poetic licence which he granted to Shakespeare. As the good doctor is no longer with us, could not A. P. Herbert or Eric Partridge be persuaded to undertake the task of preparing for us a dictionary of B.B.C. misusages.

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An Esso Photograph G24

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A

JANUARY, 1961

## A new Grommet development

### THE DOUBLE SEALING EMPIRE RUBBER GROMMET

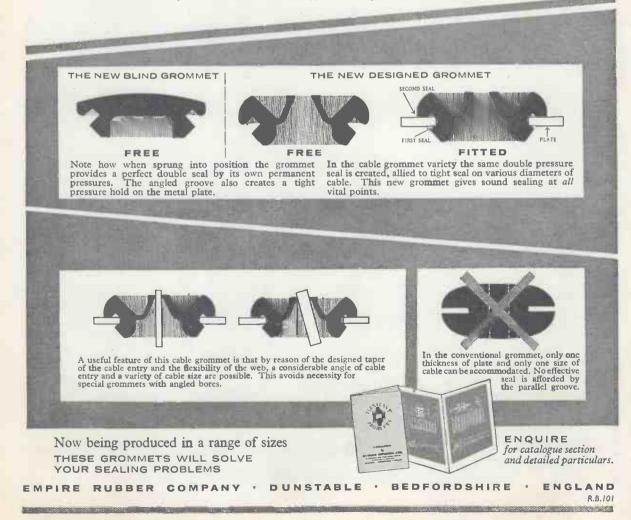
infinitely accommodating in use: considerably reduces range of sizes because the same grommet can be used with several plate thicknesses or cable sizes

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This newly developed self-conforming grommet, because it is immediately self-locking against the elements, is the solution to many of an engineer's scaling problems.

Any one size will not only accommodate itself to a variety of mounting plate thicknesses, but (designed for cable or control rod) will take these in a variety of sizes and be weather-, water- and dust-proof at a variety of angles to the cable or rod.

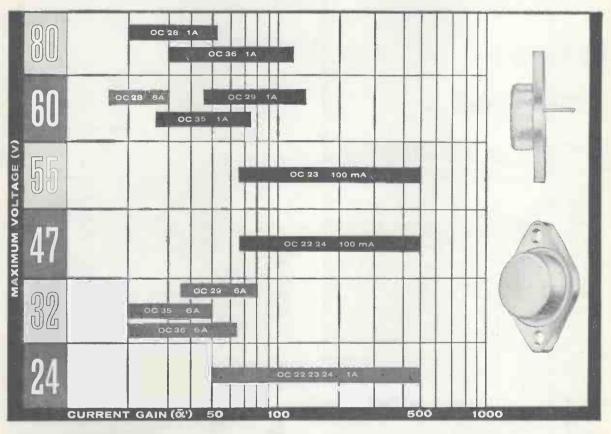
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JANUARY, 1961



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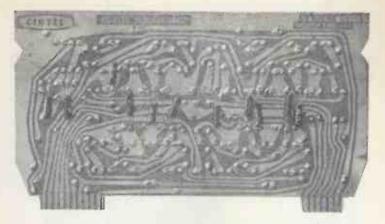
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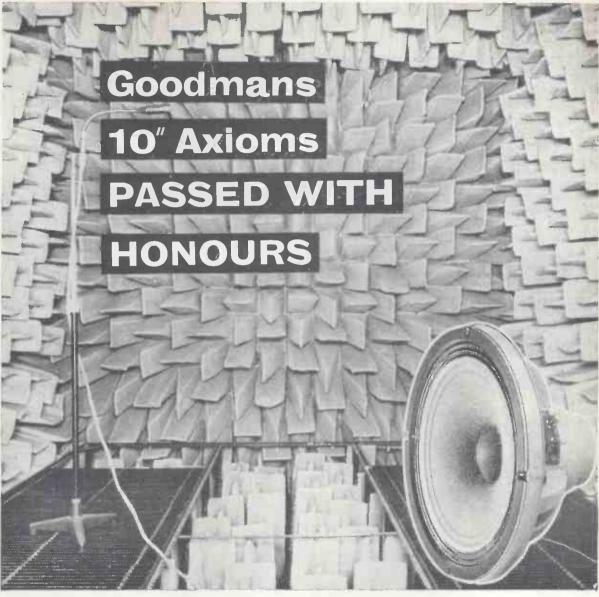
UNDRILLED: 2" 12/6: 3" 13/-: 4" 13/6:  $6^{"}$  15/-. With panel punched to take one  $50\mu A$ meter, add 1/6, or to take two meters 2/6.







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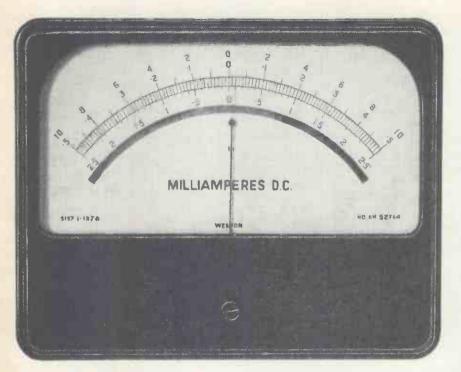
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This popular meter is used in many recording studios and broadcasting stations as a monitor as well as for servicing purposes. Dissipation rating up to 25 w. continuous, 50 w. intermittent. £13.18.6

### VALVE VOLTMETER KIT Model V-7A

The world's most popular valve The world's most popular valve voltmeter, with printed circuit and 1 per cent. precision resistors to ensure consistent laboratory performance. It has 7 voltage ranges measuring res-pectively d.c. volts to 1,500 and a.c. to 1,500 r.m.s. and 4,000 a.c. to 1,500 r.m.s. and 4,000 peak to peak. Resistance measure-ments from 0.1 ohm to 1,000 M ohms with internal battery, D.C. input impedance is 11 Megohms and dB measurement has a centre-zero scale. Com-plete with test prods, leads and standardising battery.



£13.0.0

R.F. PROBE KIT Model 309-CU

This complete probe kit will extend the frequency range of the V-7A Valve Voltmeter to 100 Mc/s. and will enable useful voltage indication to be obtained up to 300 Mc/s. £1.5.6



DEPT. W.W. I, GLOUCESTER, ENGLAND

6

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JANUARY, 1961

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### "HAM " TRANSMITTER KIT Model DX-40U



Covers all amateur bands from 80 to 10 metres. Power input 75 watts C.W. 60 watts peak C.W. 60 watts peak controlled carrier phone. Output 40 watts to aerial. Provision for V.F.O. Filters minimise T.V. inter-**£29.10.0** watts to

### BALUN COIL UNIT KIT

Model B-IU



Useful transmitter accessory. Will match unbalanced co-axial lines, used on most modern transmitters, to bal-anced lines of either 75 or 300Q impedance. Can be used with transmitters and receivers without adjust-ment over the frequency range of 80 through 10 power inputs **DA** £4.4.6

meters, and will handle power inputs up to 200 watts.

### MATCHED HI-FI STEREO KIT

4-speed Transcription Record Player Model £12 10 0 RP-IU

6 w. Hi-Fi Amplifier, Model S-33 ..... £11 8 0 Twin Stereo Speaker Systems Model SSU-1... £20 11 0

Total cost if purchased separately ..... £44 9 0

### YOURS for £42.10.0

if all ordered together, or  $\pounds 8/8/-$  deposit and 9 monthly payments of  $\pounds 4/3/-$ . Pedestal speaker legs  $\pounds 2/14/$ optional extra.

### PERSONAL TRANSISTOR RADIO KIT Model UJR-I



Operated by a 4.5V. torch battery, this sensi-tive dual-wave head-phone set is a fine introduction to elecintroduction to elec-tronics for young and old. In Polystyrene moulded plastic case

THE WORLD'S LARGEST-SELLING ELECTRONIC KIT-SETS 

which accommodates battery (and amplifier £2.16.6

Additional Amplifier Stage Model UJR-IS will enable the UJR-I to work a loudspeaker under favour-able conditions. 16/6 extra.



DAYSTROM

WIRELESS WORLD

★ Latest Additions ★

### TAPE AMPLIFIER UNITS Models TA-IM and TA-IS

This Combined Record/Re-Tape play Amplifier available in b is in both monophonic and Stereophonic models. Model TA-IM can be modified to the stereo version with modification kit TA-IC.

£16/14/-; TA-IS, £22/4/-; MI-AT TA-IC, £6.

### PORTABLE RADIO

three diodes this highly sensitive set is specially de-signed for Short and Medium wavebands (200-550, 90-200, 18-50 and

### **R.F. SIG. GENERATOR** Model RF-IU

Provides extended frequency cover-age on six bands from 100 kc/s.-100 Mc/s. on fundamentals and up to 200 Mc/s on cali- **C11 11** 200 Mc/s on cali- £11.11.0

> **GRID-DIP METER** Model GD-IU

Functions as oscil-lator or absorption wave meter. With coils for plug-in continuous quency coverage from 2 Mc/s. to from 2 250 Mc/s.



Two Additional Plug-in Coils Model 34I-U extend coverage down to 350 kc/s. With dial correlation curves, 15/-.

DEPT. W.W.I, GLOUCESTER, ENGLAND

### AMATEUR TRANSMITTER KIT Model DX-100U

\*\*\*\*

dependable

The world's most popular "Ham" T.X. Kit



- Completely self-contained, compact "Ham" Transmitter
- Built-in, high stable VFO and all Power Supplies. TVI: Careful design has reduced TVI to a minimum by use of effectively screened frequency-generating stages and pi tuned circuits at the input and output of the PA stage, and by 11 chokes and pi network filters to all outlets from the cabinet. No fewer than 35 disc-ceramic by-pass capacitors help to achieve the exceptional stability and high-performance for which this Transmitter is noted which this Transmitter is noted.
- The KT88 high-level anode and screen modulator stage gives over 100 watts of audio from less than 1.5 mV. input.
- Adjustable drive and clamp control ensure that valves are only driven sufficiently to maintain the required output.
- Keying on CW is via the VFO and buffer amplifier cathodes; the other RF valves are biased beyond cu-off. When zero-beating the TX with incoming signals, the exciter stages only may be run without the final amplifier being switched on.
- Provision has been made for remote control operation.
- VFO slow-motion drive is very smooth and back-lash free.
   VFO or Crystal control.
- Covers all Amateur bands up to 30 Mc/s. phone or CW £78.10.0

### VARIABLE FREQUENCY OSCILLATOR KIT Model VF-IU

Specially designed to meet the demand for the maximum possible flexibility from an ama-teur Transmitter which would otherwise be subject to certain limitations imposed by crystal con-trol. For all Amateur Bands 160-10 metres. Ideal for Heathkit DX-40U and similar transmitters ......

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Price less valves £8/19/6.

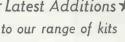


£10.12.0

N.	Our Technical Consultation and Service Departments are always ready to help in the unlikely event of your experiencing	d=+
L	any difficulty.	
Pleas	se send me FREE CATALOGUE (Yes/No)	
Full	details of Model(s)	Ì
NAM	ME	.,

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4-wave TRANSISTORISED

### Model RSW-1

Using 7 latest type transistors



II-18 m.). In solid leather case fitted with retractable **£20.18.6** 

....

JANUARY, 1961



because of THIS

# GOOD STRONG SIGNALS CLEAN PICTURES ELIMINATION OF AERIALS

An occasional intermediate amplifier and neat house-tohouse wiring are the only visible evidence that, in this area, for the first time T.V. is being enjoyed at its best, with clean pictures, no interference and no aerial replacement and repair costs.

Built with the future in mind, the installations of to-day are capable of handling a third or fourth channel—or more, colour, "coin-in-the-slot" T.V. or any foreseeable development in television and sound techniques.

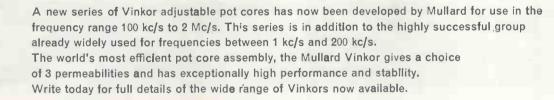
# E MI COMMUNITY TELEVISION SYSTEM

Full technical particulars together with any planning assistance that may be required can be obtained from THE GRAMOPHONE COMPANY LIMITED. Recording and Relay Equipment Division, HAYES, MIDDLESEX, ENGLAND. - Telephone: Southall 2468 (One of the E.M.I. Group of Companies)

GCICTVIa

# **NEW VINKOR SERIES**



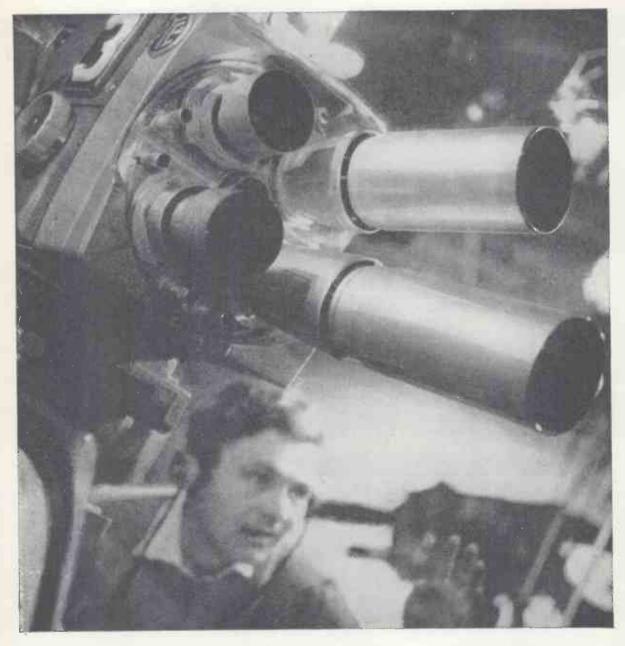






MULLARD LTD., COMPONENT DIVISION, MULLARD HOUSE, TORRINGTON PLACE, W.C.1. MCI(C)

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# **BUILDS COMPLETE STUDIOS**

Many countries owe their first glimpse of TV to the resource and enterprise of Pye. Today Pye transmission equipment is used by the television services of more than twenty-eight nations throughout the world. Even in such countries as America and Canada, where television is highly developed and keenly competitive, the Pye product is increasing in demand. The company's pioneering work in the field of TV transmission, and the succession of major developments which they have introduced, have given Pye a place of leadership in the industry.

For full technical details, please write to

PYE TVT LTD., CAMBRIDGE, ENGLAND



# MINIATURE THERMOSTATICALLY CONTROLLED OVEN

### GENERAL DESCRIPTION

This oven was designed specifically for accurate temperature control of Style "J" Quartz Crystal Units but is ideally suited for other components such as transistors, diodes, thermistors, etc., which are sensitive to ambient temperature variations. Heat control is achieved by a bi-metallic thermal switch. Capacity of the inner shell  $\frac{1}{16}$ " x  $\frac{3}{8}$ " x  $\frac{3}{2}$ " (2.0 x 0.9 x 2.3 cms). Overall dimensions  $1\frac{1}{4}$ " x  $\frac{1}{18}$ " x  $2\frac{5}{8}$ " long (3.2 x 2.0 x 6.6 cms). Weight 1 § oz. (43 grms).



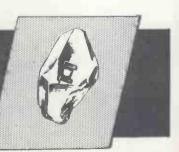
ACTUAL SIZE

### OPERATING DATA

6v or 12v supply. Power consumption 4.6 watts. Temperature settings between  $50^{\circ}C$  and  $85^{\circ}C$ . Standard tolerance on setting is  $\pm 2^{\circ}C$  but  $\pm 1^{\circ}C$  can be supplied.

Temperature differential over the operating temperature range is within  $\pm 2^{\circ}C$  of the setting-up temperature. The heating-up time is less than 5 minutes from  $\pm 20^{\circ}C$  to  $\pm 85^{\circ}C$ .

CATHODEON CRYSTALS LIMITED



**TELEPHONE LINTON 501 (4 lines)** 

# NEW! PORTABLE! LOW DISTORTION Transistor Oscillator

# R-C Oscillator CO 1008

- 20 c/s to 200 kc/s, continuously variable
- Outstanding Frequency and Amplitude stability
- D<sub>TOT</sub> <0.25% above 100 c/s, D<sub>TOT</sub> <1% below 100 c/s</p>
- Battery powered
- Weighs only 4½ lbs (2 kg).

The new Solartron R-C Oscillator, based on an original design by the Royal Radar Establishment, is fully transistorised and is powered by two internal batteries.

Frequency may be set to an accuracy of  $\pm$  5% with fine control dial, and is held to within 0.2% of the set frequency over an 8 hour period.

BRIEF SPECIFICATION Frequency stability:

Frequency	accuracy:
	Output:
Amplitude	stability:
	Zout:
Power requ	lrements:

0.2% over 8 hour period  $\pm 5\%$ 0-1 Volt (loads > 50 k $\Omega$ ) 0-0.5 Volt into 600 $\Omega$ ,  $\pm 2\%$  at full output 600 $\Omega$  at full output,  $\Rightarrow 2 k\Omega$  on any setting Two 9V EVER READY PP9 (or equivalent) batteries. 3 months battery life, based on 4 hours/day usage.

Write now for full details: SOLARTRON LABORATORY INSTRUMENTS LTD.

COX LANE, CHESSINGTON, SURREY.

Telephone: LOWer Hook 2150 Cables: SOLARTRON CHESSINGTON Telex: 23842 SOLARTRON T. DIT

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WIRELESS WORLD

How Compact is a Capacitor ?

### What capacitor?

Any capacitor!

They're all different.

Well, which is the best then?

SUFLEX, of course!

Polystyrene capacitors can be made smaller, pF for pF, than any other condenser of comparable performance.

Illustrated: 250 pF. 125 Volt Capacitor.

SUFLEX

### Suflex?

Yes, they're the most reliable because Suflex are Masters of the Empirical Art of Heat Shrinking (for resistance to humid conditions) and Ageing (for stabilization of parameters).

They should be swollen up about all that!

Perhaps, but they're not. In fact they're about the compactest capacitor to be had. We've used them for ages.

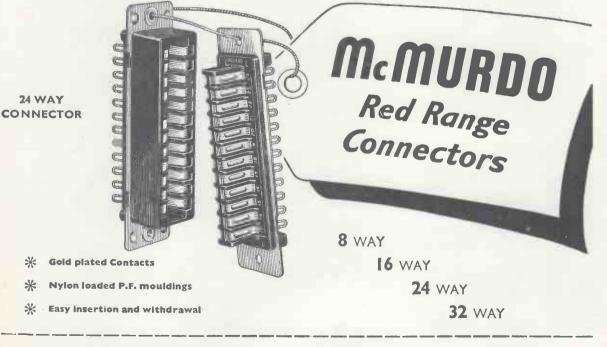
SUFLEX LTD. BILTON HOUSE, 54 UXBRIDGE ROAD, EALING, W.5. Tel: EALing 7733

### **Precision Wire-Wound Resistors**

If you are a designer or engineer in computer circuitry, instrumentation or any other field of electronics you will find this McMurdohm Resistor technical catalogue invaluable. Send for your copy.

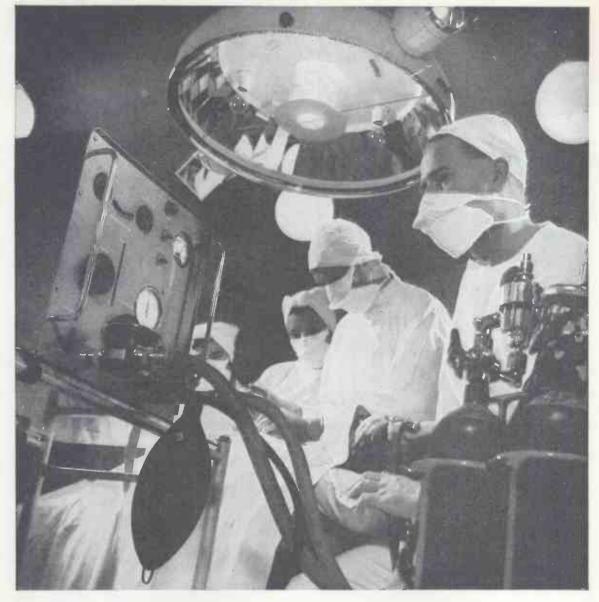
### THE MCMURDO INSTRUMENT CO. LTD.

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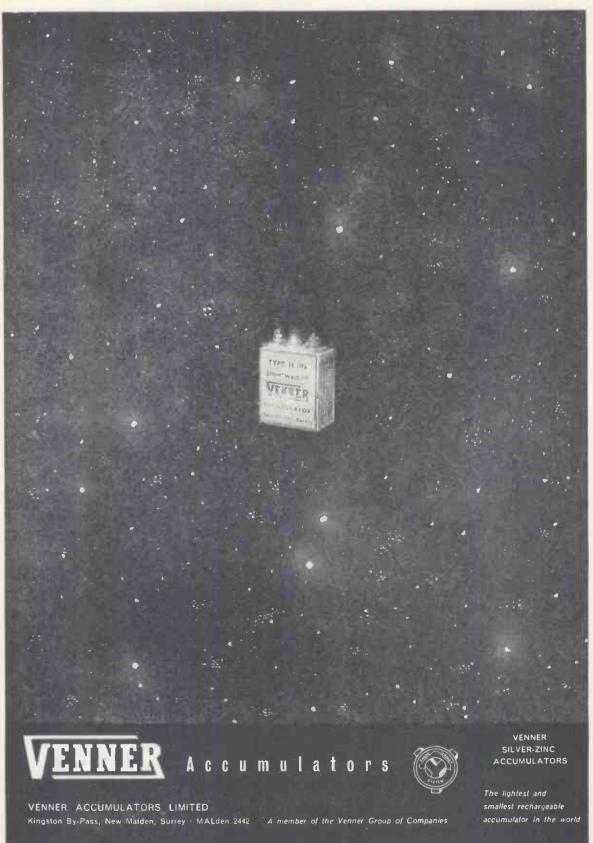
## The Electronic Lung

A member of the Pye Instrument Group W. Watson & Sons Ltd. has produced an electronic lung which is capable of replacing an iron lung. The Barnet Ventilator, as the instrument is called, is transistorised and is easily portable in cases of emergency. It is shown here in its application in an operating theatre for the administration of anæsthetics.

The Pye Instruments Group Consists of: Pye Atomics Division. Pye Industrial Television Division. Faraday Electronic Instruments Ltd. Labgear Ltd. W. G. Pye & Co. Ltd. Pye Telecommunications Ltd. Unicam Instruments Ltd. W. Bryan Savage Ltd. W. Waison & Sons Ltd.







JANUARY, 1961

WIRELESS WORLD



JANUARY, 1961



the most significant

development in Audio

for 10 years

AVEL TOROIDS

## Anew shape:

## A NEW STANDARD OF PERFORMANCE IN HIGH FIDELITY AUDIO AMPLIFIERS

Sector Sector

The AVEL TOROID DIVISION of Aveley Electric Ltd., now offer the type of transformer which the high fidelity user has been waiting for. The unique properties of the geometry of the toroid, with the latest grain-oriented strip-wound magnetic cores and superior synthetic insulating materials have skilfully been combined to give:

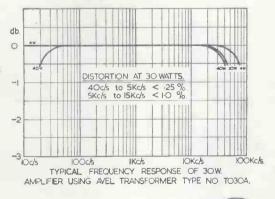
Two TO30A outputs and one TM250A Mains transformers mounted on Stereo twin channel 30-50 watt Two EL34 test amplifier. are used in the output and silicon diodes in the 500 v. 450 mA power supply.

The AVEL TM120D mains transformer and the AVEL TO30D/2 output transformer mounted on a 30 watt Hi-Fidelity power amplifier.

Write for fuller details TOROID DIVISION



### **\*WIDER BANDWIDTH \* LOWER DISTORTION** ★ SMALLER SIZE ★ LOWEST MAGNETIC FIELD HIGHER EFFICIENCY





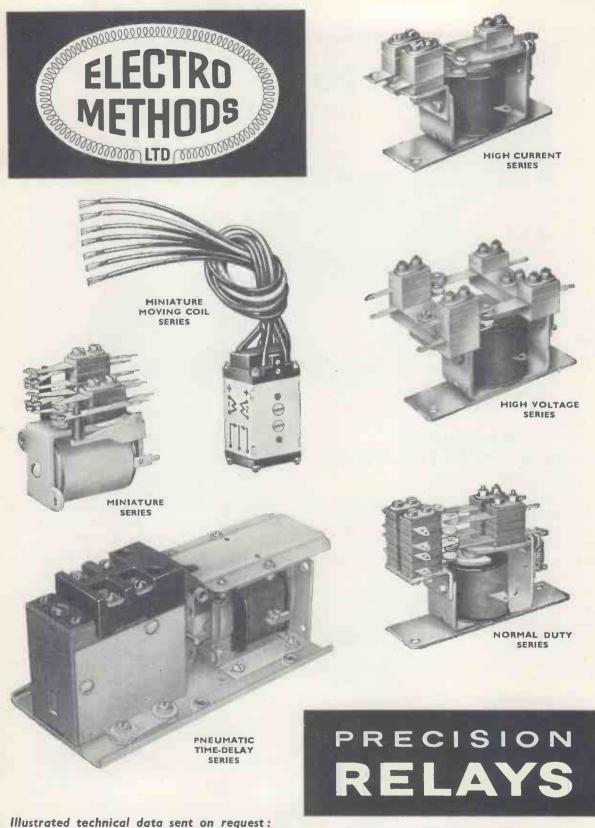
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## **Aveley Electric Limited**

SOUTH OCKENDON, ESSEX Telephone: SOUTH OCKENDON 3444 Telex: 24120 AVEL OCKENDON

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**JANUARY**, 1961



ELECTRO METHODS LTD., General Products Division, CAXTON WAY, STEVENAGE, HERTS Telephone: Stevenage 2110-7

## TRANSISTOR Converters and Inverters

for fluorescent lighting and general applications. The range of operating voltages, powers and frequencies is under steady development.

Inverters for fluorescent lighting from 12 volt d.c. supplies cover the range from a single 6" tube to six 24" tubes or equivalent. Inverters for 12v. d.c. to 50 c/s or 400 c/s a.c. up to 100 W. Constant frequency and locked frequency inverters for camera and tape recorder operation. Inverter-rectifier systems for d.c. to a.c. conversion.



Magnetic amplifier intermediate stage, saturable reactor power stage. A temperature controller for use with a platinum resistance thermometer to provide power control up to 60 KW 3-phase. No mechanical switches. Constant current characteristic for platinum furnaces.

PLUGS - LEVER KEYS - CABLES FUSE MOUNTINGS - AMPLIFIERS CONTROL PANELS - MOULDINGS COUNTERS - PLUNGER SWITCHES MAGNETIC AMPLIFIERS - RELAYS SATURABLE REACTORS - JACKS TRANSFORMERS - INSTRUMENTS CORDS - INTERNAL TELEPHONES TRANSISTOR INVERTERS - BELLS PROTECTORS - WIRES - BUZZERS

THE PHOENIX TELEPHONE AND ELECTRIC WORKS LIMITED, THE HYDE, LONDON, N.W.9. Telephone COLINDALE 7243

JANUARY, 1961

Advance 1

it's the inside that counts ... in this case up to 1 Mc/s. And the inside of the Advance time and frequency measuring counter type TC1 is rather exceptional. Over 145 transistors and 21 printed circuit boards have been combined in an advanced design to provide the full facilities and high performance expected

### Nett Price in U.K. £375

Leaflet Y101 will be forwarded on request.

	SIX FIGURE DISPLAY	accurate to $\pm 1$ count					
*	FREQUENCY MEASUREMENT	from 10 to 1,000,000 c/s					
	TIME MEASUREMENT	from 1 µsec. to 2,777 hours					
	PERIOD MEASUREMENT	1 or 10 periods of input wave- form down to 10 c/s					
	RANDOM COUNTING	totalling over any period					
	OUTPUT TIMING PULSES	from 10 <sup>-1</sup> to 10 <sup>s</sup> p.p.s.					
	INTERNAL STANDARD	oven controlled 1 Mc/s crystal					
	STABILITY	±1 part in 10° at 25°C					
	SELF-CHECK FACILITIES	to internal crystal standard					
	FREQUENCY MEASURING PER	RIOD 0.1, 1.0 or 10 seconds					
	REPETITION OF COUNT	manual or automatic					
	POWER CONSUMPTION	3 W (battery), 14 VA (mains)					
	DIMENSIONS	length 12in., height 9in., depth 6in., weight 12lb.					

AND NOW-Virtually to d.c. with V.L.F. Converter type C.A.I.



ROEBUCK ROAD . HAINAULT . ILFORD . ESSEX . TELEPHONE : HAINAULT 4446

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IT/AB

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Microwaves

RADAR: Fire Control · Navigation of Aircraft and Small Ships · Automatic Landing · Missile-Guidance · Transponders • COMMUNICATIONS: Multichannel Radio Links for telemetering Data and Speech • VALVES: Klystrons and Magnetrons for 35Gc/s and 75Gc/s bands · Monitor Diodes for IGc/s to 35Gc/s • INSTRUMENTS : Comprehensive Waveguide measuring circuits covering 6 to 75Gc/s • RESEARCH: Outstanding Research and Development of the latest techniques.



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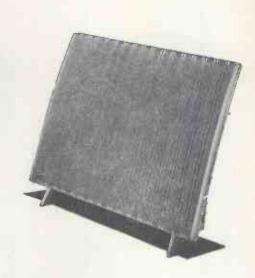


From very small beginnings, Hivac has grown to a company large enough to maintain the highest technical standards for subminiature valves, cold cathode tubes, numicators, neon indicator lamps, filamentary indicator lamps for telephone switchboards and other purposes; and a variety of small, specialised electronic components. The first company in the world to design and manufacture subminiature valves, Hivac now occupies one of the most modern electronics factories in the world. Yet Hivac still remain small enough to maintain a close and personal customer service which is second to none.

# YEARS OF HIVAC SERVICE

# **TO INDUSTRY**

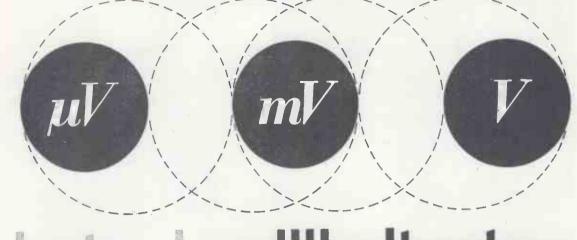
HIVAC LIMITED · STONEFIELD WAY · SOUTH RUISLIP · MIDDLESEX A member of the A.T.E. group FOR THE FULL APPRECIATION AND ENJOYMENT OF YOUR KIND OF MUSIC...



The complete Quad range of matched units comprises :

QUAD 22 Control Unit QUAD II Amplifier QUAD Electrostatic Loudspeaker QUAD F.M. Tuner QUAD A.M. II Tuner Send a postcard to Dept. for illustrated leaflet.

**QUAD**-for the closest approach to the original sound



# electronic millivoltmeters

## internal calibration

VHF Voltmeter, type GM 6025

PHI

Frequency range:	0.1 Mc/s - 800 Mc/s flat from 1 Mc/s - 300 Mc/s 1 dB at 0.1 Mc/s +- 1 dB at 800 Mc/s
Measuring range:	10 mV (f.s.d.) - 10 V dIvlded into 7 ranges in a 1-3-10 sequence
Overall accuracy:	$< 50/_0$ with respect to full scale
	65 k $\Omega$ at 1 Mc/s ; 50 k $\Omega$ at 100 Mc/s ; 35 k $\Omega$ at 200 Mc/s
Input capacitance:	ι μμε
Replacement of the	probe crystal:
the probe crystal c	an be easily replaced and the instrument rapidly

re-calibrated by the user For measurements on  $50 \Omega$  -coaxial lines the T-connector, type GM 6050T,can be ordered It should be noted that all Philips electronic voltmeters contain calibration standards which enable the user easily and rapidly to check, and, if necessary, to re-calibrate his voltmeter at any time without the use of additional instruments.

## **PS** electronic measuring

Sold and serviced by Philips Organizations all over the world Sole Distributors in the U.K.: Research & Control Instruments Ltd., 207 King's Cross Road, London W.C. I Overseas enquiries please, to the manufacturers,

N.V. Philips, EMA-Department, Eindhoven, the Netherlands.



#### Input 1 Input II 10 mV (f.s.d.) -Measuring range: 100 µV (f.s.d.) -10 V in II steps 1000 V in 11 steps 100 M $\Omega$ (± 1.5%) in parallel with Input impedance : $I M \Omega (\pm 1.5 \%)$ in parallel with 10 µµF 20 µµF Overall accuracy: with respect to full scale $\pm 30/_0$ Pre-deflection: < 5 µV Drift: 1 µV per hour after 1 hour of warming-up

DC Microvoltmeter, type GM 6020

Automatic polarity indication DC currents may be measured directly from 100  $\mu\mu$ A (f.s.d.) up to 10  $\mu$ A



## L F Millivoltmeter, type GM 6012

. . . . . . . . . . . . .

Frequency range:	2 c/s - 1 Mc/s
Measuring range:	I mV (f.s.d.) - 300 V in 12 steps
dB scale:	- 80 dB up to +52 dB
	$(0 dB = 1 mW into 600 \Omega)$
Input Impedance:	4 M $\Omega$ In parallel with 20 $\mu\mu$ F (up to 3 V);
	10 M $\Omega$ in parallel with 10 $\mu\mu$ F (in the other ranges)
Overall accuracy:	with respect to full scale
	± 2.5%, 5 c/s - 100 kc/s
	± 50/0, 2 c/s - 1 Mc/s
Pre-deflection :	< 100 µV

# 

#### without with pre-attenuator pre-attenuator

HF Millivoltmeter, type GM 6014

Frequency range:	l kc/s-30 Mc/s	10 kc/s-30 Mc/s
Measuring range:	1 mV (f.s.d.) -	100 mV (f.s.d.) -
	300 mV in 6 steps	30 V in 6 steps
dB scale:	-80 dB up to -8 dB	-40 dB up to $+32$ dB
Damping at 10 kc/s:	1 MΩ	50 M Q
1 Mc/s:	700 k Ω	10 M Ω
30 Mc/s:	50 k Ω	2 MΩ
Input capacitance:	7 μμF	2 μμF
Pre-deflection :	compensated by ele	ectrical zero setting

#### Variations of the frequency characteristic:

 $<50\!/_0$  over the whole range, with respect to the response at the frequency of the calibration voltages

Overall accuracy:

 $\pm$  3% with respect to full-scale and with reference to the frequency characteristic

instruments:



## quality tools for industry and research

JANUARY, 1961

## DOUBLE ENDED STAINLESS STEEL VACUUM OVENS

Single Vacuum Oven double-flanged with glove boxes each end.

\* Made throughout in polished stainless steel.

- \* Single action door openings.
- \* Rectangular with shelf spacings to suit.
- + Double ended controls.
- \* Electrical interlocking of air inlet and isolation valves.
- \* Outer cover hermetically sealed.
- ★ Temperature range 0°-300°C or equivalent F.
- ★ Temperature Control: Normal ±75°C. Special ±1C.
- Internal Spacing 7in. x 8in. x 18in. (can be altered to special requirements).
- \* Vacuum Range: To 10-4
- \* Respective Vacuum Gauges incorporated.
- \* Automatic air inlet valve on Backing Pump.
- \* Visual indicators and fuses on all switches.
- ★ Flanged for fitting into Dry Box.



View showing automatic interlocking of unloading compartment on glove box.

We design and manufacture Ovens to Customers' special requirements. Should you have any problems in this field our Technical Department is always willing to help you solve them.

Vacuum Ovens with temperatures of up to 600°C are also manufactured by us on similar lines but with Sectional Heating and Water-Cooled Ends.



WILLOW LANE, MITCHAM, SURREY

## Phone : MITcham 8211 (3 lines)

MONACO

Both you and the racing driver would agree that all circuits have their own characteristics and problems.....

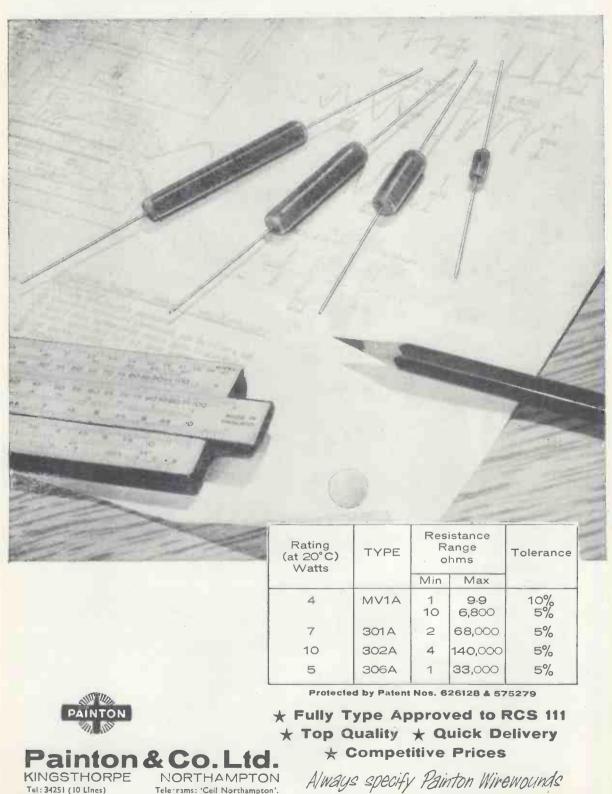
but for *you*, when selecting a valve for all round optimum performance . . . . . better make it BRIMAR

(BVA)

B

BRIMAR LIMITED + FOOTSCRAY + SIDCUP + KENT + FOOTSCRAY 3333

## **PAINTON Miniature Vitreous Wirewound Resistors**



Telex: 31576

JANUARY, 1951

JANUARY, 1961

WIRELESS WORLD

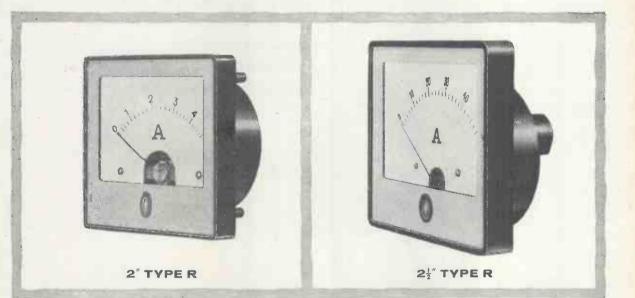
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## PANEL MOUNTING INSTRUMENTS

new Type R

## 2"-2½"-3½"-5"



A unique feature of this new range is the front plate of anodised aluminium, supplied in any colour to match surrounding apparatus. The case, of clean rectangular styling is moulded in phenolic material, and fixing is easy and adaptable.

Instruments can be supplied for the measurement of all A.C. and D.C. currents and voltages.

The legibility of the scale is exceptional and all unnecessary markings have been eliminated. Scales are white enamelled on metal with black markings.

Send for leaflets' today

MOVING COIL TYPE MOVING COIL WITH RECTIFIER MOVING COIL WITH THERMO COUPLE MOVING IRON TYPE



SALFORD ELECTRICAL INSTRUMENTS LTD.

PEEL WORKS, SILK STREET, SALFORD 3, LANCASHIRE. Tel: Blackfriars 6688 London Sales Office: Magnet House, Kingsway, W.C.2. Tel: Temple Bar 4668 A-Subsidiary of THE GENERAL ELECTRIC CO. LTD. OF ENGLAND

tested to withstand overloads of 10 times nominal value 上

e've come a long way since the crystal set era.

V.H.F., frequency modulation, stereo, big screen television — constant developments in the air create a constantly growing market. It pays to advertise to this market on television — a fact endorsed by the dealer himself.

> In a recent survey commissioned by Associated-Rediffusion, dealers mentioned television as being more successful than any other medium in helping them sell more television sets. In addition, more and more people are buying replacement television sets on styling. There is no better medium than television for actually demonstrating

your model or a particular feature of its design.

An early valve receiver, 1923.

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Look at the size of the replacement market for television sets in the London area covered by Associated-Rediffusion. Nearly *half* of the homes viewing Independent Television have sets three years old or more. The following table shows the breakdown in detail.

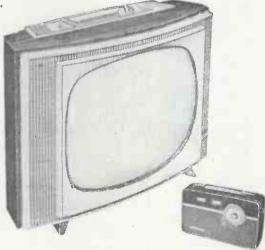
	AGE OF PRESENT SET	TOTAL
Base of %-No. of informants		2068
Percentage who had present television set	Three years	14%
	Four years	12%
	Five years or more	18%

There is no better medium than television for driving home your sales message in the face of increasing competition. Manufacturers of all electrical goods advertising on Associated-Rediffusion can reach over 8 million people in their homes, in London. This represents a prosperous mass market in an area where there is a huge replacement trade waiting to be developed. Advertise on television in London—

vertise on television in London--

on Associated-Rediffusion.

The above figures are quoted from the London Viewership Survey No. 4, commissioned by Associated-Rediffusion. John Talbot (HOLborn 7888) will be pleased to let you have full details.





## ASSOCIATED-REDIFFUSION

#### London's Television, Monday to Friday

Television House, Kingsway, London, W.C.2. Tel: HOLborn 7888 also Norfolk House, Smallbrook, Ringway, Birmingham 5. Tel: Midland 9151/2 and Peter House, Oxford Street, Manchester 1. Tel: Central 9867/8

## ALPHASIL-the modern core material

The Inset curves illustrate the superior magnetic properties of Alphasil cold-reduced grain oriented silicon steel over those of a typical hot-rolled grade (Ferrosil 80). Alphasil has a maximum permeability four times that of the hot-rolled transformer sheet and its core losses are approximately one-third. Initial and incremental permeability, stacking factor and ductility are considerably better than those of hot-rolled sheet.

Alphasil .013" thick is produced in coil 30 inches wide, and can be supplied slit to narrower widths by arrangement.

TABLE OF WATTS LOSSES	
Frequency Guaranteed max. total los cycles/second at B. Max. 15 Kilogaus	
HASIL 44 50 62 watts/lb.	5
HASIL 40 50 .56 watts/lb.	
HASIL 37 50 .51 watts/lb.	
HASIL 33 50 .46 watts/lb.	

ABOVE—A 4,000-lb. coil of 30° wide x ·013° thick, ready for despatch. RIGHT—Core-loss testing of Alphasil by the 'double-lop' Epstein method.

Thin Alphasil for high frequency work is also available in coil in .004" thick in widths up to  $5\frac{1}{2}$  inches, and in  $\cdot 002''$  thick, in widths up to  $4\frac{1}{2}$  inches.

ALPHASIL -004HF

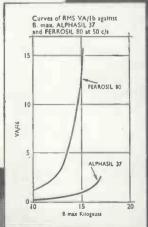
Frequency cycles/second 400 8,000

Guaranteed max. total losses 8.00 watts/lb. at B Max 15 Kilogauss 9.50 watts/lb. at B Max 2 Kilogauss Full technical data will be supplied on request

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Our Cookley Works is one of the largest in Europe specializing in the manufacture of laminations for the electrical industry.





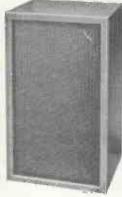


# WITH CLEAN BASS

In each of these models L.F. output is produced by a 12*i*n. unit type WLS/12 fitted with a soft fibrous cone for smooth response. The special roll surround permits large distortion-free excursions with fundamental resonance below 25 c/s.



A two-speaker model complete with treble volume control. Cabinet size 23 × 14 × 12in. Weight 37 lb. complete. Impedance 15 ohms. Max. input 15 watts. £29/10/- complete, tax free





A three-speaker system complete with midrange and treble volume controls. Cabinet size, 28 × 14 × 12in.

Weight 48 lb. complete. Impedance 15 ohms. Max. input 15 watts. £39/10/- complete, tax free

## WLS/12

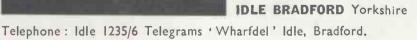
Each model is available in choice of walnut, oak or mahogany veneers. Also available in whitewood slightly cheaper. Tropical models with resin-bonded plywood approximately £2 extra.

Catalogue giving full technical details, response curves and oscillograms of the above models available on request.





A four-speaker system complete with mid-range and treble volume controls. Cabinet size 35 × 24 × 12in. Weight 65 lb. complete. Impedance 15 ohms. Max. input 15 watts. £49/10/- complete, tax free.





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If your problem is the design of Thermostatic or pressure sensitive systems, small shaft couplings, pressure seals or vacuum seals, then the versatile metal bellows will provide the solution. HYDROFLEX are, of course, to be preferred. Please write to the Bellows Division for your copy of the catalogue, quoting reference "WW".

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D.C. voltage 1mV to 1100 V. Accuracy  $\pm$  0.1% of full scale. A.C. voltage 1mV to 1100 V. Accuracy  $\pm$  0.25% of full scale. Resistance 10 to 1.1M0. Accuracy  $\pm$  0.25% of full scale.

DVOM uses well-tried Venner packaged circuits, making for reliability in use and ease in servicing. Stepping switches, relays and need for frequent calibration have all been eliminated.

DVOM has been developed by Venner Electronics in collaboration with Epsco Incorporated of Boston U.S.A.

See it on Stand No. 19 at the Physical Society Exhibition 16th - 20th January.





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VENNER ELECTRONICS LIMITED Kingston By-Pass, New Malden, Surrey. MALden 2442

JANUARY, 1961

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FIRST AGAIN IN BRITAIN

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E704 Island Storage Tube will store about 500 bits of information and is designed for analogue and digital storage and for radar data handling applications, where it gives improved signal to noise ratio. The electrical input and output may be applied or taken out at the same or at different rates; the two processes may be simultaneous or sequential.

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E 704

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# MODEL 5G-5 MODEL 6G-62 MODEL 58C- 8x5" MODEL 8C-8

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We have made this selection from our wide range of speakers as they cover practically all the requirements of the replacement trade.

The new prices are now operative.

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6 <u>1</u> in.	6Ġ	6500 g	21/6	6/11	
7 x 4in.	47G	6500 g	20/6	6/7	
7 x 3in.	37G	6500 g	20/6	6/7	
8 x 3in.	38G .	6500 g	20/6	6/7	
8 x 5in.	58C	8500 g	24/6	7/10	
8in.	8C	7000 g	25/6	8/2	

All loudspeakers have Standard 3 Ohm impedance. Higher impedances can be supplied at an extra cost of 3|- plus 1|- Purchase Tax.

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V <sub>a1</sub> (kV)	0,5	1.0	1.5	0.3	2.0	1.5	1.4	2.0	1.5	1.5
V <sub>a3</sub> (kV)	0.5	1.0	4.5	0.3	2.0	1.5	1.8	2.0	4.0	1.5
V <sub>a4</sub> (kV)	-	_		1.5	4.0	3.0	4.0	4.0	8.0	15
V <sub>a5</sub> (kV)	-	-	unterm		uniterus		10	_		15
Y scan (mm)	28	55	70	50	80	75	60	95	95	60
Y sensitivity (V/cm)	45	11.5	16	3.0	23	27	12.5	17.5	35	2.7
X scan (mm)	28	55	90	70	90	90	95	115	115	· 100
X sensitivity (V/cm)	53	20	23	5.0	36	27	26.5	29	57	11.2
Screen diameter (mm)	30	71	94	78	108	108	137	137	137	137
SCREEN TYPES:						4				
Medium persistence	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes
Long afterglow	No	Yes	Yes	To order	Yes	Yes	Yes	To order	To order	To.orde
Blue photographic	To order	To order	Yes	To order	Yes	Yes	To order	To order	To order	To orde
Short persistence	To order	No	No	No	To order	No	No	No	To order	No

\* Data is given for each gun.

Please send me data on the	e types ticked	
Name		
Position		4LP31
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	3BLP3I	5BVP3IA
	4EP31	5CLP31

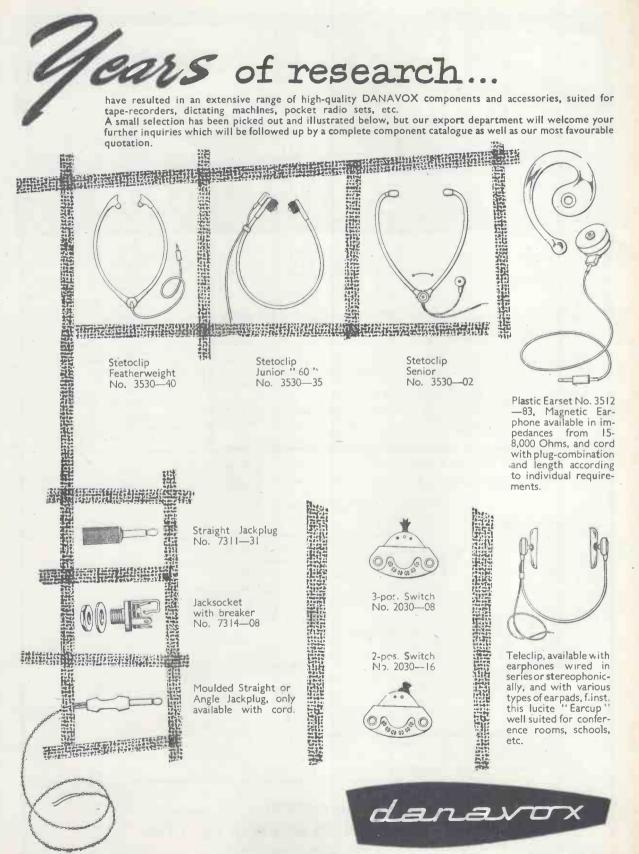
ETL20a

Brief details of the range and typical operating characteristics are given here—for full data please use the coupon adjoining.



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with sharp, controlled heat for transistor and other small assemblies, Location of element under soldering tip produces 30 watt capacity for only 15 watts consumption. Available for 230/240v. 220, 200, 110 and low voltages 6, 12, 24, 28, 50v. List prices range from 25/- to 29/6 (all prices subject).

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Vide Percy Wilson, The Gramophone, Sept. 1960

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#### JANUARY, 1961



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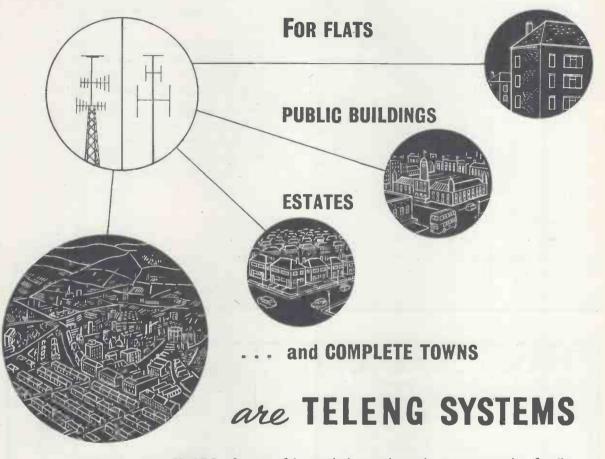
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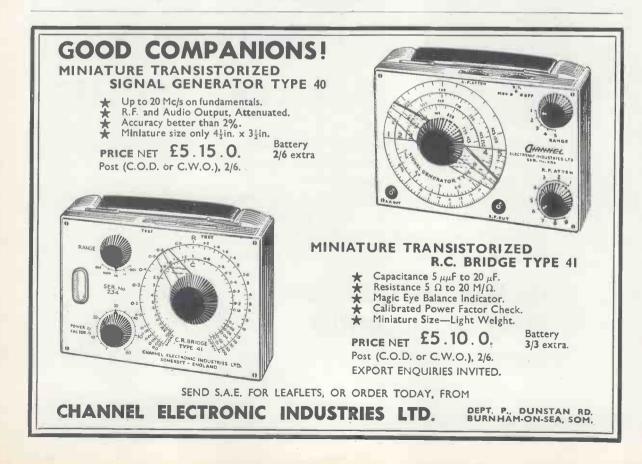
163 Mains transformers for valve and contact-cooled rectifiers, audio output transformers and chokes and fully described in Gardner's new "S/M" Catalogue available on request.

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The new range of Newmarket industrial transistors is specially designed for operating under the most rigorous industrial conditions where requirements exist for:

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Look for the gold top-





Newmarket Transistors Limited Exning Road, Newmarket, Suffolk Tel: Newmarket 3381/4 Cables: Semicon Newmarket

JANUARY, 1961





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We manufacture a large range of hand and automatic Eyeletting and Piercing Machines and also stock eyelets which we can supply in small or large quantities. Full illustrated brochure of the "Phoenix" machines, write for leaflet W.W.2.

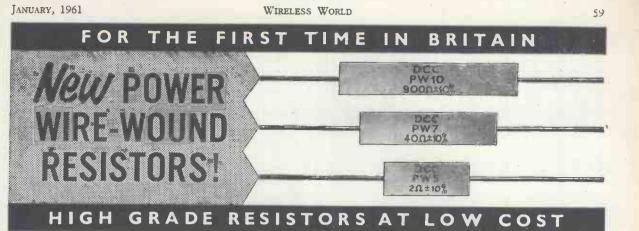


PHOENIX WORKS, 114-116 EUSTON ROAD, LONDON, N.W.1

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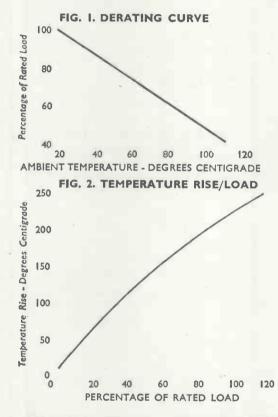
- Resistance change less than 5% after 100 hours at 40°C. ambient temperature and 95% relative humidity.
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- \* Resistance change less than 1% and no physical effects due to soldering.

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PW5	0.5Ω to 2.5Ω	<b>2</b> .5Ω to 2.0kΩ
PW7	$0.5\Omega$ to $8.0\Omega$	<mark>8.0</mark> Ω to 6.5kΩ
PWIO	1.0Ω to 10Ω	10Ω to 10kΩ

# DUBILIER

DUBILIER CONDENSER CO. (1925) LTD TELEPHONE: ACORN 2241 DUCON WORKS



5.0	7.0	10.0
0.5Ω	0.5Ω	1.0Ω
<b>2.0</b> kΩ	6.5kΩ	l0kΩ
7."	25/64"	1 7 "
	0.5Ω 2.0kΩ <del>3</del> ″	0.5Ω 0.5Ω 2.0kΩ 6.5kΩ

Catalogue R15A available on request.

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N.	10	82	CI	21	<u>_</u>	<b>A</b>	۳.	i.	0	1	J
2	5.0	55		E.E	The s	200		I.	v	18	W

Dimension         25x16x49mm         31x22x66mm         38x24x80mm           Weight         TRC-9-281         35g         TRC-9-281         35g           Weight         TR-9         37g         TRC-9-281         35g           Voltage         AType         BType         BType           Voltage         A-C.         100-130V         A.C.         200-130V
Weight         TR—9         37g.         TRC—9—2A1         23g.         TRC—9—2B24         40g TRC—9—2B25         40g TRC—9—2B1         40g TRC—9—3B1         40g           Volume         A—Type         B—Type         40g         40g
A.C. 100-1304 A.C. 400-2404

MANUFACTURERS & EXPORTERS TOREI INDUSTRIAL CORPORATION No. 150, KUGAHARA-CHO, OHTA-KU, TOKYO, JAPAN Codie : "TOREINDUST" TOKYO



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20

# **RADIO** EXPORT

# TUBES ONLY

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**HALTRON RADIO TUBES** again achieved a greater turnover plus a bigger volume of sales, and this in spite of the fact that during the year prices generally had been reduced. What is the secret of this success? We think the answer is that we have the most comprehensive stock in the world (over 3,000 types), really competitive prices and quick execution of orders. We are all set to follow the same pattern in 1961.

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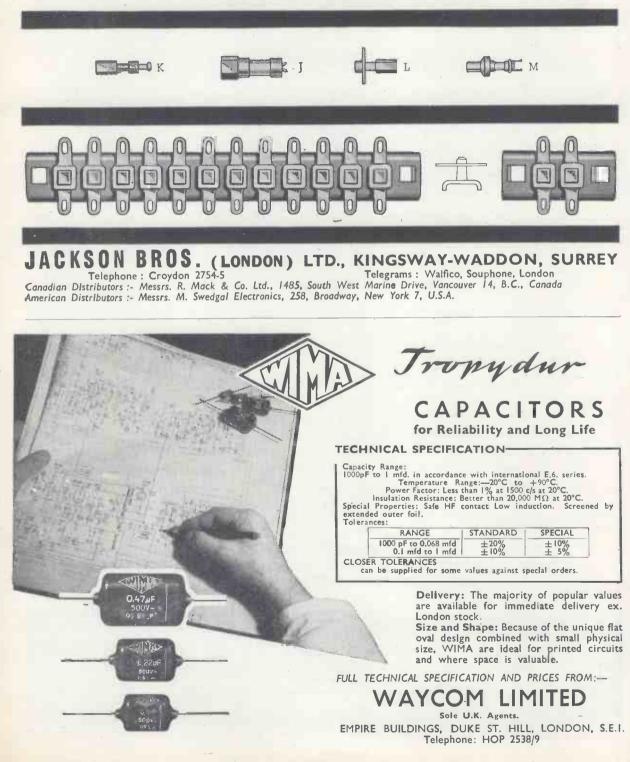
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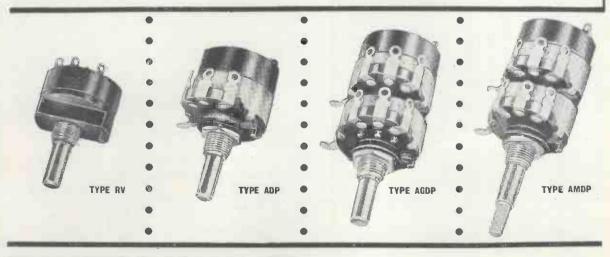
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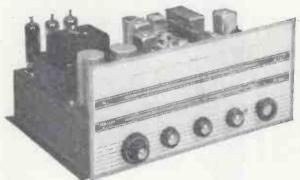


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Qty.	Type	Circuit	Maximum Input Volts (r.m.s.)	Max. Output Current mA (mean)	16	l D.C. O mfd. . Cap.   Full   Load	60 1 Resvr Half Load	nfi
1	C2H	Half-Wave	125	60	135	115	135	
1	C3H	16 15	125	120	120	85	130	
1	C2D	11 11	250	60	275	245	280	
1	C3D	11 10	250	120	275	245	290	
1	C20	Volt-Doubler	125	60	275	245	280	
1	C3D	11 11	125	120	260	205	285	

Qty.	Туре	Circuit	Maximum Input Volts (r.m.s.)	Max. Output Current mÅ (mean)	Typica 16 r Resvr Half Load		utput V 32 r Resvr Half Load	nfd.
1	C2V	Push-Pull	125-0-125	120	140	120	140	130
2	C2D		250-0-250	120	275	250	280	255
1	C3V	49 -49	125-0-125	240	130	115	140	130
2	C 3 D		250.0.250	240	280	250	280	260
1	C38	Bridge	250	120	275	250	280	255
2	C3D		250	240	280	250	280	260

### Send for leaflet MF/102

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**JANUARY**, 1961

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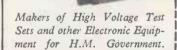
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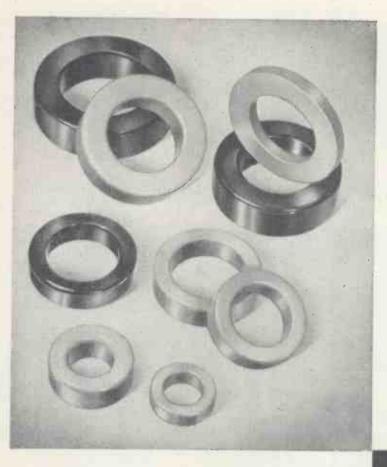
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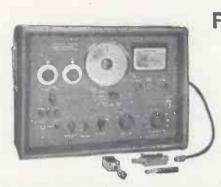
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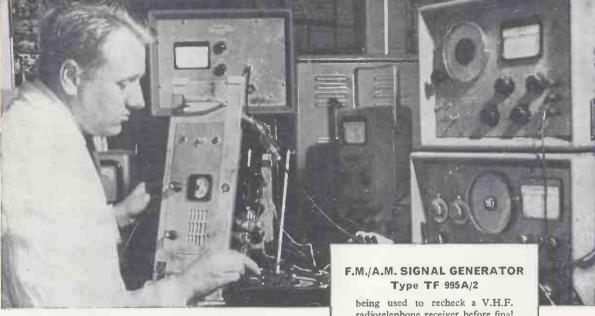
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For mobile radio testing

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JANUARY, 1961





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FOR

TEMPLE BAR 0201-4

### JANUARY, 1961

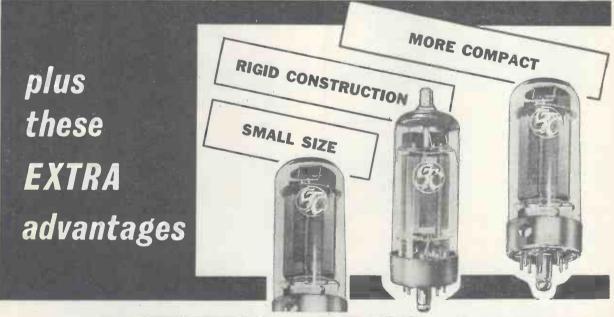
FOR THE

WIRELESS WORLD

# Electrical equivalents

These STC valves are loctal-based, indirectly-heated, beampower amplifiers electrically similar to the U.S.A. type 807 but with extra advantages. Little more than half the size of the 807 the electrode assemblies are more compact . . . stronger and more rigid . . . and the glass bulbs less vulnerable. Shorter grid and anode leads, and glass base with miniature-type pins reduce inter-electrode capacitance and improve performance at the higher frequencies. Thus, both mechanically and electrically, they offer a high standard of reliability, which, together with their small size, makes them particularly suitable for use in mobile equipment.

71



The range of	5B/254M double-ended	5B/255M single-ended	5B/257M single-ended	5B/258M double-ended
four valves offers a choice of heater voltage:	6-3V	6·3V	127	197
	0·9A	0.9A	0·47A	0-3A

VALVE DIVISION: FOOTSCRAY

Where the overall seated height must be kept as low as possible, the single-ended type 5B/255M may be used, but under conditions of high altitude or anode modulation the double-ended type 5B/254M is to be preferred because of insulation considerations at the valve base. A flying lead version is available under the code 5B/254G.

KENT

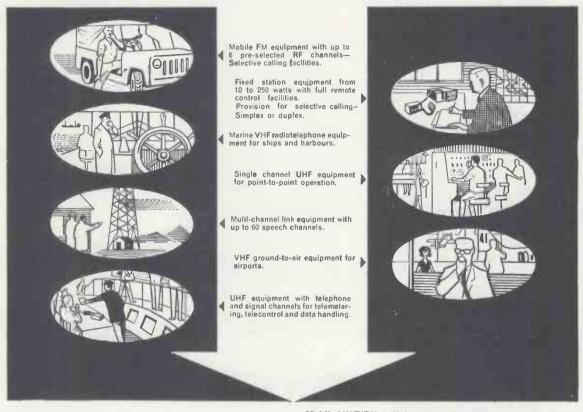
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60/7MS

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# No. 1 IN THE SERIES

# **Technical Information for the Transistor Circuit Designer**

### Switching Performance of Alloy Junction Transistors

There are several possible ways of specifying the switching performance of a transistor, but the most elegant of these involves the parameters  $\tau_C$  and  $\tau_S$ .  $\tau_C$  is a measure of the charge, or quantity of minority carriers, stored on the base of a transistor when it is operated in the active region;  $\tau_S$  is a measure of the charge stored when the transistor is operated in the saturated, or bottomed, condition.

For example, if  $\tau_{C}$  is 0.03  $\mu$ s, the collector current  $I_{C}$  is 10 mA, and the transistor is being operated in the active region, the charge stored is:

 $\tau_{\rm C} I_{\rm C} = 0.03 \times 10^{-8} = 0.3 \times 10^{-9}$  coulombs.

If  $I_C = 10 \text{ mA}$ ,  $\tau_C = 0.03 \mu \text{s}$ ,  $h_{FE} = 40$ ,  $\tau_S = 1.5 \mu \text{s}$  and  $I_B = 0.5 \text{ mA}$ , there is an excess base current equal to  $I_B - \frac{I_C}{h_{FE}} = I_B! = 0.5 - 0.25 = 0.25 \text{ mA}$ . This excess base current acts to drive the transister into the structure region and the charge stored in them.

drive the transistor into the saturation region and the charge stored is then:

 $I_{B}^{\dagger} \tau_{S} + I_{C} \tau_{C} = 0.92 \times 10^{-9}$  coulombs.

With a knowledge of  $\tau_C$ ,  $\tau_S$  and the collector junction capacitance, the total charge stored in the base region and depletion layer of the transistor can be calculated. Comparisons can then be made between the relative merits of different types of transistors when operated in a specific circuit.

For further details see STC application report No. MK/146, "The Junction Transistor as a Charge Controlled Device" by R. Beaufoy and J. J. Sparkes, A.T.E. Journal, Vol. 13, No. 4, October 1957, and "Transistor Switching Circuit Design using the Charge-Control Parameters" by R. Beaufoy, Proc.I.E.E., Paper No. 2970, May 1959, Vol. 106, Part B.

	Туре	<b>τC(</b> μs)	τ S(μs)	Cbic(pF)	hFE range	f <sub>hb</sub> (Mc/s)
STC						
Switching	TK28C	0.03	1.5	18	25-54	-
Transistors	ткзос	0.053	1.6	20	15-95	6
	ткзіс	0.025	1-3	18	25-125	н

### Basic circuit for measurement of TC and TS

CB Adjusted to give 'Best' Square Wave Output at the Collector.

With optimum\* positive going edge at the collector the charge required to switch on the transistor is

$$Q_{ON} = V_B C_{B1} \simeq I_C \sim C + 1.7 C_{blc} V_{CC}$$

With optimum\* negative going edge at the collector the charge required to switch off the transistor is

 $Q_{OFF} = V_B C_{B2} \simeq I_C \tau_C + I_B' \tau_S + 1.7 C_{b'c} V_{CC} = Q_{ON} + I_B' \tau_S$ \*Adjusted for minimum rise or fall time with no overshoot.

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### Radio Engineering and Electronics

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SOMETHING

### to appear shortly

### THEORY OF MICROWAVE VALVES

### by S. D. Gvozdover

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### TV TUNER HIGH SLOPE VHF TETRODE

### EDISWAN MAZDA 30F27

The 30F27 is a frame grid VHF tetrode having a mutual conductance of 15 mA/V at an anode current of 13.5 mA and a screen current of 1.7 mA with variable-mu characteristics to reduce cross-modulation effects at high signal levels.

screen current of 1.7 mA with variable-mu characteristics to reduce cross-modulation effects at high signal levels. This tetrode used in the RF stage of a television tuner offers certain advantages over the more conventional double triode cascode arrangement. For instance, the number of circuit components required is smaller, the layout is simpler and for valves having comparable slopes the single cathode type valve can be manufactured more economically.

Normally the noise performance of a tetrode is inferior to that of a triode due to the presence of partition noise arising from the screen current. However the 30F27 has been specially designed to provide a low ratio of screen to anode current to minimise the effect of partition noise while still retaining good screening between control grid and anode. When this is used in conjunction with frame-grid techniques a high slope per milliamp of anode current can be obtained resulting in a high gain RF valve with a noise performance much superior to that of a conventional pentode and equal to that of a double triode cascode amplifier such as the 30L1.

Heater Current (amps) I<sub>h</sub> 0.3 Heater Voltage (volts) V<sub>h</sub> 3.7

TENTATIVE RATINGS AND DATA

IENIAIN	E KA	111463	AND DA	A A A
Maximum Design Cent				
Anode Dissipation (watts)	)		pa(max)	2.5
Screen Dissipation (watts	)		Pg2(max)	0.4
Anode Voltage (volts)			Va(max)	250
Screen Voltage (volts)			Vg2(max)	230
Heater to Cathode Voltag	ge (volts	rms)	Vh-k(max)	rms 90*
Cathode Current (mA)			Ik(max)	18
* From cathode	to highe	r poter	tial heater	pin.
Inter-electrode Capacit	tances (	(pF)†		
Input Capacitance		100 P -	Cin	6.3
				1.8
Grid 1 to Anode			Cg1-a	0.027
0 11 1 0 11 0			Cg1-g2	2.0
				4.0

### Maximum Dimensions (mm)

Overall Length	·	÷.	56
Seated Height			49
Diameter			22.2



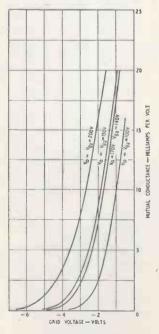
TYPICAL OPERATION

		Self Bias Circuit	Current Bias Circuit
Supply Voltage (volts)	Vb	200	200
Anode Voltage (volts)	V a	170	150
Screen Voltage (Initial) (volts)	Vg2		105
Anode & Screen Common	v g 2	110	103
Decoupling Resistor $(k\Omega)$			3.3
	• •	2.2	
Anode Decoupling Resistor $(k\Omega)$		33	22
Screen Decoupling Resistor $(k\Omega)$	n		33
Cathode Bias Resistor $(\Omega)$	Rk	82	
Grid Current Bias Resistor $(k\Omega)$	Rg1		330
Grid Bias Voltage approx. (volts)		-1.25	
Anode Current (mA)	I a	13.5	14
Screen Current (mA)	I <sub>g2</sub>	1.7	1.4
Mutual Conductance (mA/V)	gm	15	15.5
Inner Amplification			
Factor $(g_1 \text{ to } g_2) \dots \dots \mu$	4g1-g2	60	
Equivalent Grid Noise			
Resistance $(\Omega)$	Reg	450	
Input Loss at 50 Mc/s (k $\Omega$ )		) 6.8¶	
Input Capacity Working (pF)			
Change in Input Capacity produce	d by	2010 313	
bigging value to get off (nE) A c		2.041	

biasing valve to cut-off  $(pF) \triangle c_{1n}(w)$  2.9¶ — ¶ Measured at 50 Mc/s with the three cathodes strapped and taken directly to earth.

§ Inter-electrode capacity with holder capacity balanced out.

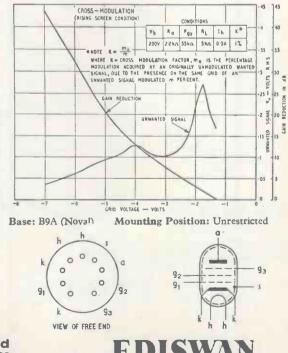
Tentative Characteristic Curves of Ediswan Mazda Valve Type 30F27





ANGOE CURRENT

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MAZDA

Grid

JANUARY, 1961



### NEW B9A EFFICIENCY DIODE

### **EDISWAN MAZDA U193**

The U193 is a new Efficiency Diode, designed for use in line deflection stages of television receivers using 110° and 114° cathode ray tubes and has a maximum peak heater to cathode voltage rating of 5.5 kV as compared with 5 kV for the U191. The valve is mounted on a B9A (noval) base with the cathode brought out to a top cap connection.

MAXIMUM DIMENSIONS (mm)

Diameter	22.2
Seated Height	75
Overall Length	82

### **TENTATIVE RATINGS AND DATA**

Heater Current (amps) I<sub>h</sub> 0.3 Heater Voltage (volts) V<sub>h</sub> 19

### **Maximum Design Centre Ratings**

Mean Anode Current (mA) Peak Anode Current (mA) Peak Inverse Voltage (kV)	Is(sv)max is(pk)max PIV(max)	150 450 5.5*
Peak Heater to Cathode Voltage		
(heater negative) (kV)	Vh-k(max)	5.5*
*Rated for TV line scan where	the duty cycle	does not exceed
15% and the pulse duration does	s not exceed 15	$\mu$ s. An absolute
rating of 6.6 kV must not be exce	eded.	

Turing of 0.0 KT must not be exect	uou.	
Inter-Electrode Capacitances	( <b>pF</b> )	
Anode to Heater and Cathode	Ca-h, h	6.6
Cathode to Heater and Anode	Ck-h, s	7.9

### **APPLICATION NOTES**

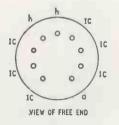
The Ediswan Mazda U193 is designed specifically for use as the efficiency diode in line defletion stages of television receivers using 110° and 114° cathode ray tubes. It would normally be used in conjunction with line scanning output valves such as the 30P4 and 30P19 in energy recovery types of line scanning circuits in a.c./d.c. television receivers. The heater may be directly connected in the normal series chain as the insulation between heater and cathode is adequate to withstand the pulse voltages encountered in auto-transformer scanning circuits. This high voltage insulation generally results in high thermal inertia of the cathode, giving a much longer heating time than the remaining valves in a television receiver. In the U193, however, the thermal inertia, has been reduced considerably so that there is only a delay of a few seconds between the warm-up of other valves in a receiver and the start of operation of the line scanning circuits. Besides reducing the waiting time for a picture to appear, this factor also alleviates problems of protection of valves where a line-gated system of automatic gain control is used.

is used. See "Aspects of Design No. 20 ('Wireless World' March 1960)—Efficiency Diodes for TV Line Output Stages" for notes on circuit considerations and limiting ratings.

### Mounting Position : Unrestricted

Base: B9A (Noval)

Top Cap: CT1-Cathode Connection

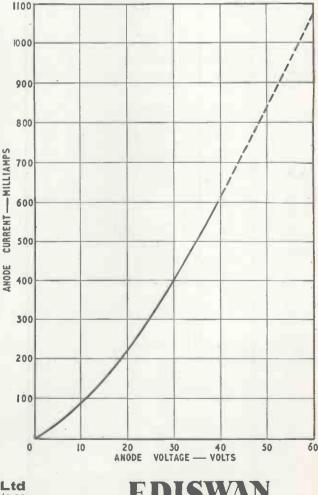




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Characteristic Curve of Average Ediswan Mazda Valve Type U193



MAZDA

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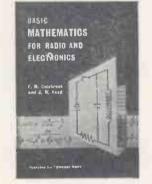
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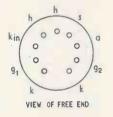
### TENTATIVE RATINGS AND DATA

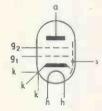
Maximum Design Centre Ratings	
Anode Dissipation (watts) pa(max)	2.5
Screen Dissipation (watts) pg2(max)	0.5
Anode Voltage (volts) Va(max)	250
Screen Voltage (volts) Vg2(max)	250
Heater to Cathode Voltage (volts rms) V h-k(max)rms	150*
Resistance Control Grid to Cathode	
(megohms) $R_{g-k(max)}$	1
*From cathode to higher potential heater pin.	
Inter-Electrode Capacitances† (pF)	
Input Capacitance Cin	8.5

Output Capacitance	2.7
Grid 1 to Anode Cg1- a	0.006
Grid 1 to Grid 3 C <sub>g1-g3</sub>	0.1
Grid 1 to Grid 2 Cg1-g2	1.8
Grid 1 to Cathode Cg1-k	6.0
Grid 2 to Anode Cg2-a	0.19
Grid 3 to Anode Cg3-a	0.45
Measured in fully shielded socket, without can.	

Base: B9A (Noval)

Mounting Position: Unrestricted





TYPICAL OPERATION

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90
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100
11.5
2.8
35
12.5

Tentative Characteristic Curves of Ediswan Mazda Valve Type 6F25

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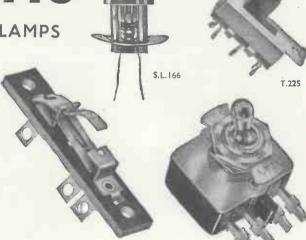


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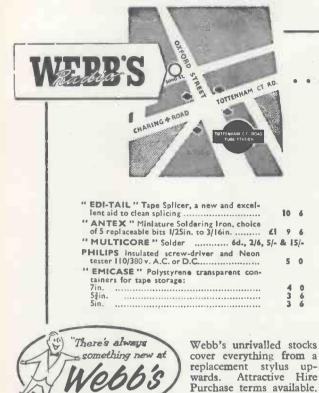
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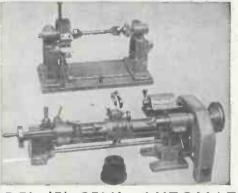
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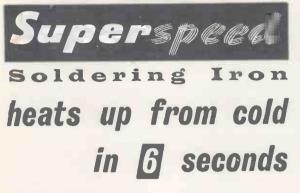
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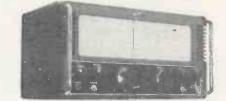
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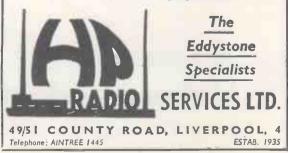
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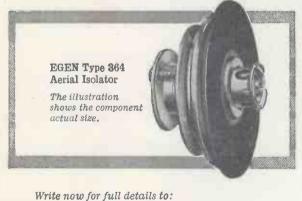


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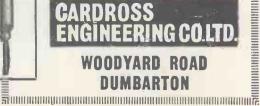
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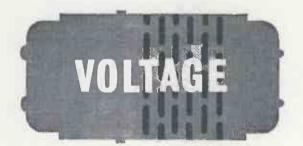
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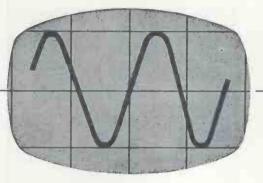
92 Wireless W	VORLD JANUARY, 1961
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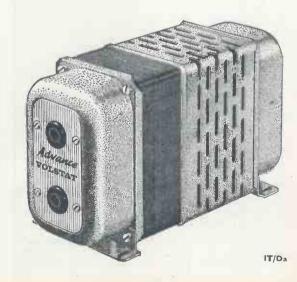
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Mechanical	Plug-in and solder-in, B7G and octal, printed circuit, sub-miniature; weight $\frac{2}{3}$ to $1\frac{1}{2}$ ozs.				
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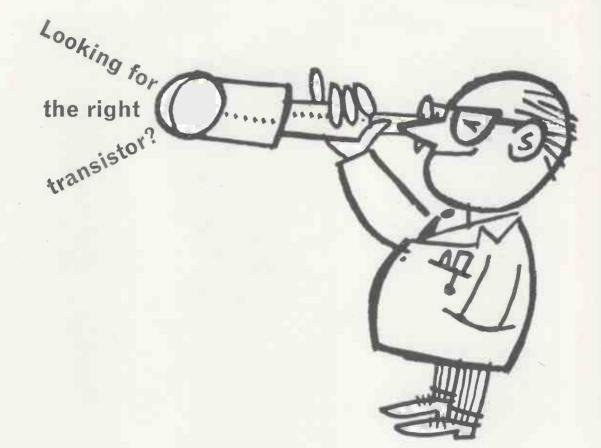
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Here is our new low impedance double triode, the 6080 WA. In the first place, you save space with these two valves in one envelope. In the second place, here is a really fine, special quality double triode, the result of our 40 years of experience in manufacturing valves of all kinds. Full information on all Series Stabilisers, including the 6080 WA is available, as well as a detailed report on the design of Series Stabiliser circuits.



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# G.E.C. valve

# 6080 WA

Maximum anode voltage 250V Maximum cathode current 2 x 125mA r<sub>a</sub> 280+2800 The full range of series stabilizers cover these powers:

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A tape recorder is only as good as its microphone. If you cut a top note off in the mike, you can't blame the recorder if the note isn't there. If there's distortion at the start, there'll be a din in the end. Give your recorder a fair chance. Give it a balanced, wide-frequency input. Give it a good microphone. Give it an Acos microphone.



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CHOICE OF DECOUPLING COMPONENTS for TELEVISION **IF STAGES** 

# Aspects of design

This is the thirtieth of a series of special features dealing with advanced problems in circuit design to be published by The Ediswan Mazda Applications Laboratory. We will be pleased to deal with any questions arising from this or other articles, the thirty-first of which will appear in the February 1901 issue.

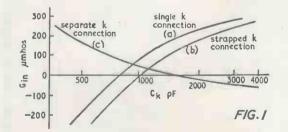
The performance of television IF stages is governed to a great extent by considerations of stability, and it is possible to design such stages with the highest gain consistent with adequate stability by careful choice of decoupling capacitors. Authough the method to be described is not directly applicable to every type of IF amplifier design, the underlying principles are useful and will therefore be dealt with in some detail.

### INPUT CONDUCTANCE

The input conductance of an IF amplifier stage is profoundly influenced by admittance in the cathode circuit and to a smaller extent by regeneration from g<sub>2</sub>. For typical values found in a sound IF stage, the cathode admittance produces an input conductance  $G_{in} \approx -\frac{g_k B_{gk}}{2}$  where  $g_k$  is the cathode current Bk

mutual conductance and Bgk and Bk are susceptances from mutual conductance and  $B_{kk}$  and  $B_{k}$  are susceptances from grid to cathode and cathode to chassis respectively. By using a sufficiently large value of cathode bypass capacitance  $C_k$  the cathode admittance might be expected to approach infinity at television IF, but in practice the inductance of the internal connection from the cathode itself to the cathode pin on the valve button cannot be bypassed, and results in degeneration, equiva-lent to some finite value of input conductance. All Ediswan Mazda television IF valves, including the 6F23, 6F24 and 6F25, have the cathode brought out to two separate pins on the valve button, and if these are strapped externally the unavoidable un-bypassed inductance is effectively halved, resulting in reduced degeneration.

There is, however, another way of reducing the degeneration, for whereas an inductive cathode load produces degeneration, a for whereas an inductive cathode load produces degeneration, a capacitive load produces regeneration (i.e. in the above expression  $G_{In}$  becomes negative). By reducing the value of  $C_k$  from that required to fully bypass the cathode resistor, the degeneration due to the unavoidable cathode lead inductance is opposed by the regeneration due to  $C_k$ , and for a particular value of  $C_k$  the two effects cancel, giving zero input conductance. The unwanted inductance has been "tuned out" by  $C_k$ . Further reduction in  $C_k$  provides a net regeneration; in by varying  $C_k$  and  $C_k$  and  $C_k$  and  $C_k$  and  $C_k$  and  $C_k$ . C<sub>k</sub>, a whole range of positive and negative input conductances can be provided. This can be seen in Fig. 1 which shows both the case of a single cathode connection (curve a) and strapped cathode connection (curve b) for the 6F24. The effect of strapping the cathode is to shift the curve along the C<sub>k</sub> axis so that a larger value of Ck is required to give the same input conductance.



The fact that some regeneration may be provided by the cathode circuit does not necessarily mean that the stage will oscillate; this will only occur if the negative input conductance due to the cathode circuit exceeds in magnitude the conductance of the external grid circuit.

From the curves (a) and (b) in Fig. 1 it can be seen that the input conductance depends rather critically upon the value of  $C_k$ . However, by using the separate cathode connection, in which the earthy end of the grid tuned circuit is decoupled to one cathode pin whilst the other cathode pin is reserved for the cathode components, the input conductance is controlled prin-

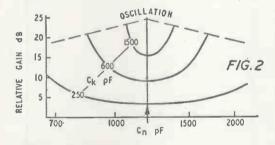
# **Associated Electrical Industries Ltd** Radio and Electronic Components Division Technical Service Department 155 Charing Cross Road, London, W.C.2 Tel: GERrard 9797.. Grams: Sieswan, Westcent, London

cipally by feedback via  $g_2$  and this circuit has been found to provide a more gradual control of input conductance with change in the value of  $G_k$ . This is shown in curve (c) of Fig. 1.

### NEUTRALISATION

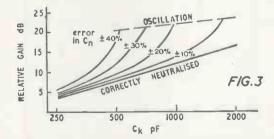
An earlier article in the series "Aspects of Design" (No. 14 Wireless World" September, 1959) described how the un-.. wanted feedback from output to input of an IF stage may be neutralised by means of a bridge circuit. Exact neutralisation by adjusting the neutralising component to suit each receiver will enable very high gain to be achieved, but individual adjustment is not usually possible in production receivers, so that in practice a fixed neutralising component must be used and a certain misneutralisation accepted.

The effect of errors in neutralising can be determined directly by plotting IF gain against the value of the neutralising com-ponent. The result of varying the neutralising component ponent. The result of varying the neutralising component (in this case a capacitor common to the anode and  $g_2$  circuit of a 6F24 sound IF (38 Mc/s) amplifier) is shown in Fig. 2. The correct value of  $C_n$  corresponds to the minimum of each curve, and in this case is 1200pF. The separate cathode connection circuit has been used here. Increasing the value of  $C_k$  lowers the input conductance and increases the IF gain, but at the same time it makes the choice of  $C_n$  very critical it makes the choice of Cn very critical.





CHOICE OF  $C_n$  AND  $C_k$ It is obvious that in the choice of  $C_n$  and  $C_k$  a compromise must be made between high gain and security against instability and it is perhaps easier to choose the most suitable values if the information in Fig. 2 is redrawn in the form shown in Fig. 3, where the parameter is the error in the neutralising capacitor Cn.



It will be seen that it is advantageous to keep the error as It will be seen that it is advantageous to keep the error as low as possible by using  $\pm 10\%$  tolerance capacitors for  $C_n$ and  $C_k$ . Stable Hi-k ceramic types are now available which can be recommended for this application. An allowance must be made for the spread of valve parameters and circuit strays. These can be regarded as having the same effect on the circuit as widening the tolerance of the neutralising component C<sub>n</sub> and cathode component Ck, and the final choice of suitable values of Cn and Ck must be made with this in mind.

For the Ediswan Mazda frame grid 6F24 used in a typical sound IF circuit, suitable values of  $C_n$  and  $C_k$  are:—

# $C_{p} = 1200 \text{ pF} \pm 10\%$

# $C_{k} = 470 \text{ pF} \pm 10\%$

These values may require adjustment to suit individual circuit design and layout.

**JANUARY**, 1961

# **NEW VERY HIGH SLOPE EDISWAN MAZDA 6F24** The 6F24 has a slope of 15 mA/V at 10 mA anode current and a high "Figure of Merit," both results achieved by incor-porating the new frame grid technique. This new screened HF Pentode enables improved performance to be obtained as an IF amplifier in a.c. or a.c./d.c. television receivers. Heater Voltage (volts) V<sub>h</sub> 6.3 Heater Current (amps) I<sub>h</sub> 0.3 TENTATIVE RATINGS AND DATA Maximum Design Centre Ratings Anode Dissipation (watts) Screen Dissipation (watts) Anode Voltage (volts) Screen Voltage (volts) 2.5\* Dat max) Dg2(max) 0.8\* 250 250 Va(max) Vg2(max) Heater to Cathode Voltage (volts rms) Vh-k(max)rms 150†

Control Grid to Cathode		
Resistance (megohms)	Rg1-k(max)	0.6‡
*With grid to cathode resistance not		
+From cathode to higher potential h		
+With m 2 W/: m a		iming a

a common anode and screen decoupling resistance  $< 2.2 \ k\Omega \pm 10\%$ .

1	nter-El	ectrode	Capacitances	( <b>pF</b> )§
1	manua Co	maaitamaa		

Inter - Diectroue Capacitances (pr.)3		
Input Capacitance	Cin	8.8
Output Capacitance	Cout	2.6
Grid 1 to Anode	Cg1-a	0.006
Grid 1 to Grid 3	Cg1-g3	0.1
Grid 1 to Grid 2	Cg1-g2	2.0
Grid 1 to Cathode	Cg1-k	6.2
Grid 2 to Anode	Cg2-a	0.15
Grid 3 to Anode	Cg3-a	0.47

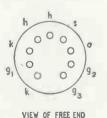
§ Measured in fully shielded socket, without can.

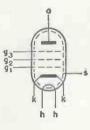
TYPICAL OPERAT	ION	
Anode Voltage (volts)	Va	170
Screen Voltage (volts)	Vg2	170
Self Bias Resistance (ohms)	Rk	150
Anode Current (mA)	I.a	10
Screen Current (mA)	Ig2	2.7
Mutual Conductance (mA/V)	S m	15
Inner Amplification Factor (g1 to g2)	μ <sub>g1-g2</sub>	65
Equivalent Grid Noise Resistance (ohms)	Reg	370
Input Loss at 38 Mc/s (Pins 1 and 3	~ *CU	
strapped) (k $\Omega$ )	Tg1-k(w)	8.5
Working Input Capacity** Measured	*81.E( W)	0.5
at 38 Mc/s (pF)	Cin(w)	13.7
Change in Input Capacity produced by	CIDIWI	2.3.11
biasing valve to cut-off. Measured		
at 38 Mc/s (pF)	40	3.4
Figure of Merit (Valve only) (Mc/s)	∠Cin(w)	375
Effective Figure of Merit (Valve and		515
Circuit) (Mc/s)		220
Trinton electrode conceitre with hel	der conscitu	

\*\*Inter-electrode capacity with holder capacity balanced out

 $g_m \times 10^3$ \_\_\_\_ see "Aspects of Design" No. 1 for ttGiven by further details. (Wireless World July 1958.)

Base: R9A (Noval) Mounting Position: Unrestricted



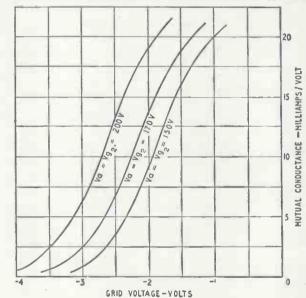


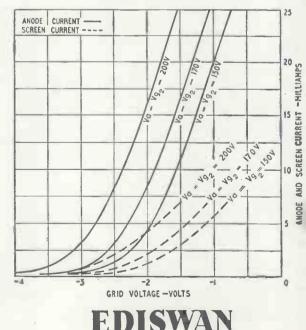
Overall Length 56 Seated Height 49 Diameter 22	

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Tentative Characteristic Curves of Ediswan Mazda Valve Type 6F24





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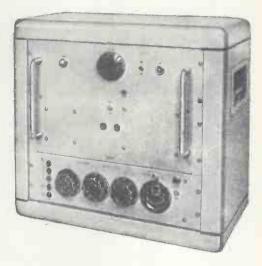


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It is available in our standard steel case with Baxendale tone controls

and up to 4 mixed inputs, which may be balanced line 30 ohm microphones or equalised P.U.s to choice.

# **ELECTRONIC MIXER/AMPLIFIER**

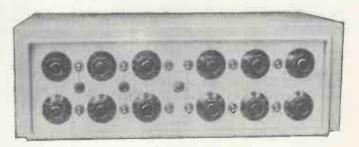
This high fidelity 10/15 watt Ultra Linear Amplifier has a built-in mixer and Baxendale tone controls. The standard model has 4 inputs, two for balanced 30 ohm microphones, one for pick-up C.C.I.R. compensated and one for tape or radio input. Alternative or additional inputs are available to special order. A feed direct out from the mixer is standard and output impedances of 4-8-16 ohms or 100 volt line are to choice. All inputs and outputs are at the rear and it has been designed for cool continuous operation either on 19 x 7in. rack panel form or in standard ventilated steel case.

Size 18 x 71 x 91 in. deep. Price of standard model £49.

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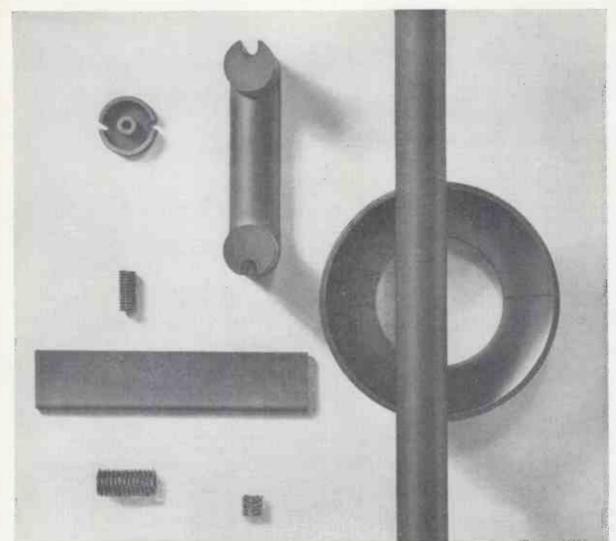


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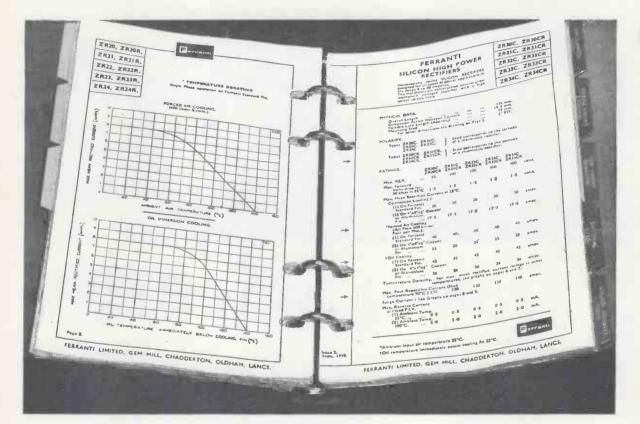
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**JANUARY**, 1961

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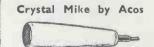
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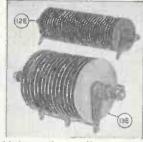
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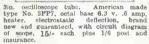
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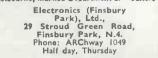
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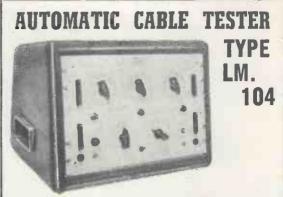
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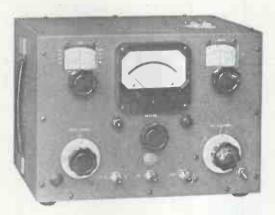
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EABC80 9/- ECL8210/6 EZ81 EAF42 9/6 EF41 9/6 FW4/ EB91 4/6 EF50 5/- GZ32		3Q5GT 9/6 6C5GT 6/6 6SL7GT 3S4 7/6 6C6 5/– 6SN7GT 3V4 8/– 6D6 5/– 6SQ7	8/- 12BA6 9/- 35Z4GT 8/- 7/6 12BE6 9/- 42 8/- 9/3 12C8 7/- 35Z5GT 9/-
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EBF89 9/6 EF8612/6 K133 ECCBI 8/- EF89 8/9 KT61 ECC82 7/6 EF91 5/9 KTW	13/6 PEN220A 4/- UY85 7/- /61 6/6 PENA4 12/6 VP23 6/6	5Y3G 8/- 6F1514/- 6X4 5Y3G 8/- 6F1514/- 6X5G .	7/6 12K8M13/- 142BT 3/6 7/- 12Q7GT 6/6 210DDT 4/6
SPECIAL MAINS DROPPER RE-	METAL RECTIFIERS.	5Y3GT 7/6 6J5GT 5/- 6X5GT . TRF KIT COMPLETE £5/10/	
SISTORS SMDI Philips 209U	Battery Chargers 6 and 12 volt	Line Output Transformer for Pye V4 etc., 55/2. REPANCO DRR2 DUAL RANGE	Elac 8in. x 5in. Loudspeaker 25/6. PIFCO ALL-IN-ONE RADIO
SMD7 Ultra Twin 50         5/3           SMD15 Philips 141U and         5/3           Stella ST105U         5/3           SMD27 Pye Piper         6/9	2 amp	COIL, WITH REACTION 4/ J.B. SL8 Spin Wheel Drive Assembly 27/6.	METER 32/6. Repanco TT 10 Push-Pull Output Transformer 12/6.
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METERS FOR STEREOPHONIC AMPLIFIERS, etc. 50K x 50K Log., 100K x 100K Log., 500K x 500K A/Log.,	WE STENTORIAN HEI012 10	ion Traps Type IT6, IT9, BCII, 5/6 each. TCC VISCONOL CONDENSER	4in. Square Tweeters, Elac or Plessey, 12/6. 100 ASSORTED RESISTANCES
1 Meg. x 1 Meg. Log., 250K x 250K Log., 1 Meg. x 1 Meg. Linear, 500K x 500K Linear, All 6/6 each.	BSR Monarch UA8, Record Changer £6/19/6.	.001 20KV 10/ Potentiometer SP5 for English Elec- tric T40 etc., 13/6.	<b>1-1 WATT 12/6</b> per 100. Crystal Cartridge BSR Type TC8, 15/6 each.
SWITCHES, WAVE CHANGE SWITCHES ETC. 3 POLE, 4 WAY	150/18, 7in. SPOOL 50/ "Wireless World" Radio Valve	ERIE 6 AND 10 WATT WIRE- WOUND RESISTORS 2/- each.	DIODES: OA81, GEX34, OA70, etc. 4/- each.
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2 POLE, 6 WAY 3/- ea. 4 POLE, 4 WAY	High Resistance Headphones 4000 ohms, 13/6.	RED, GREEN AND AMBER 2/3 each.	<b>2V, 6V and 12V, 4A, 21/9.</b> Rectifiers, RM1, 5/9, RM3 7/3, M1 2/8.
BEST QUALITY RECORDING	AERIAL 7/6. Multicore Savhit Solder 5/	3 way line cord2A, 100 ohms per foot, 1/9 yd. DROPPER RESISTORS FOR	12in. HEAVY DUTY 15 WATT SPEAKER, 105/ BSR Monarch UA14, in 2 tone grey,
1,200ft. on 7in. spool         21/-           850ft. on 5 <sup>3</sup> / <sub>4</sub> in. spool         18/6           600ft. on 5in. spool         13/9	HENLEY SOLON INSTRU- MENT IRON, MODEL 625 24/	ULTRA TWIN 50, 5/3 each. T.V. Electrolytic 100-200 mfd. 275 v. 7/6 each.	£8/19/6. ACOS DUST BUG FOR AUTO-
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IN CREAM, COMPLETE WITH PICK-UP FITTED TURNOVER, CARTRIDGE. Special Price 75/-	II/	Igranic type P50 Jack Plugs 2/9 each. OSMOR FERRITE ROD L & M WAVE 12/9 each.	ACOS PICK-UP ARM TYPE GP54, WITH CARTRIDGE 29/6. J.B. SLI5, FM/VHF Dial Kit 25/6.
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TUNING INDICATOR ESCUT- CHEON SUITABLE FOR EM80	BIB RECORDING TAPE SPLICER 18/6.	Acos Mic 40, with folding stand 19/6 each.	Transistor, 4 pushbuttons. Wave- change on/off switches. Overall size 3.4in. x 2.1in. x 1.8in. 90/3.
TYPE OF VALVES, 2/- each. TELEVISION TUBES REGUNNED, 12 MONTHS'	Mullard Circuits for Audio Amplifiers, 8/6. ELSTONE MULTI-RATIO OUT-	PICKUP MATCHING TRANS- FORMER BSR AF333 36/ Collaro Mark IV Tape Transcriptor	SET OF AERIAL COILS 9/ I.F. STRIP. 460 Kc/s. I.F. amplifier.
GUARANTEE MW31/74, £5/10/-; CRM152B, £6 MW36/24, £5/10/-; CRM141 £5/10/- CRM123, £5/10/-; MW43/69, £6	PUT TRANSFORMER 10 Watts 23/ TSL Crystal 'Stick' Microphone	£17/19/6. CELESTION 24in. SQUARE	Pair of OC169 Transistors. Overall size 3.5in. x 1.5in. x 1.5in. x 1in. 92/6 8in. x 3/8in. FERRITE ROD, 5/
IO/- extra. Allowance on old tube it		Rectifier RM4, 250 v., 275 mA. 18/6. MICROPHONE TRANSFORM-	I WATT A.F. AMPLIFIER. 2- OC71 Transistors direct coupled, cascade drive into a matched pair of
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R1155 RECEIVERS

# RCA AR88 RECEIVERS

One of the most renowned American Communications Receivers ever manufactured. Widely used by all the Armed Services to maintain World-wide Communications and Monitoring Posts under all conditions. Employs 14 valves, and has 6 switched overlapping wave bands for complete coverage. Refinements include Mechanical Band Spread with Logging Scale, Automatic or Manual Volume Control. Automatic or Manual Noise Limiter. BFO with pitch control, RF and AF Gain Controls, Variable HF Tone Control, Variable Selectivity with Crystal Filter, Aerial Trimmer, Choice of Headphones or Speaker. Has Internal mains power pack for nominal 115-230 volts A.C. In Black Crackled Case size 191 m. W. x 11 in. H. x 194 in. D. Thoroughly reconditioned, immaculate in appearance, and In perfect working order. "D " Model covers 500 kc/s-32 Mc/s, price £45." LF " Model covers 75-550 kc/s. and 1.5-30.5 Mc/s, Price £35 (add carrlage 30/- and 50/- deposit on returnable transit case). S.A.E. brings illustrated descriptive leadlet.

ROA Sin, P.M. SPEAKER in heavy black crackled metal case, size 11jin. x 10jin. x Sin. Designed for use with AB38 Receiver or any set with 3 ohms output. BRAND NEW IN MAKER'S CAETONS ONLY 45/-. (Post 3/6).

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OSCILLOSCOPE No. 11 by Cossor. A First Grade L.F. Oscilloscope incorporating a Hard Valve Time Base with speeds of 1-3-40 millinecounds but easily converted for a formation of the set of th

CARRYING GASES, solid leather. SLIGHTLY USED. Internal dimensions 8 $\mu$ m. H. × 8 $\mu$ m. W. × 4 $\mu$ m. D. Fitted lock and key, and shoulder strap. Ideal for Test Instrument, Camera and accessories, etc. ONLY 25'-(postage 2')-.

BC 342 RECEIVERS. A few only of these famous American sets covering 1.15-18.0 Md/s. in six bands. Internal 115 v. A.O. Mains pack. A super receiver in first-class condition and perfect working order. ONLY £25 (carriage 18/-).

HEO MAINS POWER UNITS. A.C. Input 115/230 volts, Output D.C. (fully smoothed) 230 volts 75 mA., and 6.3 volts 3.5 amps. Complete in black crackled case ONLY 59.6.

12-WAY SCREENED CABLE. In 10ft. lengths, fitted with plugs, originally made for No 19 Wireless Set. UNUSED. ONLY 15/- per lead.

P.M. SPEAKERS. 3in. 19/6, 614n. 17/6, 8in 21/-, 12in. 29/6.

SPRAGUE CONDENSERS. Metal cased wire ends. New. 01 mfd. 1,000 v. and 1 mfd. 500 v. 7/8 per dozen. Special quotes for quantities.

# HETERODYNE FREQUENCY METERS TYPE LMI4



Frequency range 125-20,000 ko/s. in 2 bands. This is the United States Navy Model of the well-known BC.221 Frequency Meter, but has many additional features which increase its usefulness. Voltage atabilisation eronis and Crystal control ensure extreme accuracy, and in addition it is fitted with an Internal Modulation switch to allow use as a Signal Generator. Size only 8§in. × 8in. × 8§h. Pull information on request.

# UNIVERSAL VOLT-OHM-MILLIAMETER

Reads A.C. and D.C. Volts up to 1,000 in 5 ranges at 1,000 o.p.v., D.C. Current. (3 ranges) to 500 nA. Resistance readings to 200 Kohms in 2 ranges. Basic movement  $300\mu A$  sensitivity. Easily read open scale. Dimensions 5 $\mu$ in. × 2 $\mu$ in. Seautifully made, and fully made, and fully maranteed. Complete with leads, prods and internal bat 59/6



# DOUBLE BEAM OSCILLOSCOPE TUBES

Type CV 1596 equivalent to Cossor O9D as used in oscilloscopes by Cossor (339 series). Hartley and Erskine (13 series). Listed at £12/10/-.

Our price £2/19/6 (carriage 5/6) Brand new in makers' crotes

# W 1191A WAVEMETER

Crystal controlled heterodyne frequency meter covering 100 ko/s to 20 Mo/s. in 8 switched bands and is virtually the British BC221 Fower requirements 2 \* L.T. and 40-60 voits H.T. Complete with Calibration Brock, Crystal, Operating Valves and full act of spares. BRAND NFW IN ORIGINAL TRANSIT CASES. 29/19/6 (carriage 15/-).

	METERS							
ł	F.S.D.	SI	ZE AND	ТҮРЕ	PRICE			
	25 microamps 50 microamps 50 microamps 100 microamps 1 milliamp 1 milliamp 1 milliamp 200 milliamp 200 amps 40 amps. 5 amps	D.C. D.C. D.C. D.C. D.C. D.C. D.C. D.C.	3 in. Flush 2 in. Flush 2 in. Flush 2 in. Flush 3 in. Flush 2 in. Flush 2 in. Flush 2 in. Proj. 2 in. Flush 2 in. Flush	circular circular circular square circular circular circular circular circular square	59/6 59/6 80/- 39/6 22/6 30/- 25/- 50/- 12/6 7/6 7/6 12/6			
	300 volts 500 volta	A.C. A.C.	21in. Flush 21in. Flush	circular circular	25/- 25/-			

The famous Bomber Command Receiver known the world over to be uprene an its class. Covers 6 wave ranges: 18.5-7.5 Mc/s 7.5 -3.0 Mc/s. 1,500-600 kc/s., 500-200 kc/s., 200-75 kc/s. and is easily and simply adapted for normal mains use. full details beins supplied. All sets thoroughly tested and in perfect working order before despatch, and on demoi.stration to callers. Fitted with latest type Super Slow Motion tuning assembly Have had some use, but are in excellent condition. ONLY £9/19/6. A.C. MAINS POWER PACK OUTPUT STAGE in black metal case to match receiver, enabling it to be operated immediately, by just plugging in, without any modification. Fitted with 8in. P.M. speaker £6/10-. DEDUCT 10/- IF PURCHASING RE-CEIVER AND POWER PACK TOGETHER. Send S.A.E. for illustrated lealet, or 1/3 for 14-page booklet which gives technical information, circuits, etc., and is supplied three with each receiver. Add carriage 10/6 for Receiver, 5/- for Power Unit.

HIGH FREQUENCY A.C. VOLTMETER. A first-grade moving iron instrument with 6in. Mirror Scale reading up to 150 volts A.C. at 400 and 1,200-2,400 cycles. In substantial oak case with removable 1id, overall size 8jim. x 8jin. x 6jin. Recently made for the Air Ministry by Everett Edgenmbe Lid, and in perfect order Brand new and unused. ONLY 27/10/-. Can also be supplied for 50 cycles use. either 0-160 volts or 0-300 volts, same price.

POWER UNIT TYPE 3. Primary 200/250 voite A.C., 50 cycles. Outputs of 250 voits 100 mA. and 5.3 volts 4 amps Fitted double smoothing and 2 meters to read H.T. ourrent and voltage. For normal rack mounting (or bench use) having are: front pane. Size 14th. x 7in BRAND NEW. ONLY 79/6 (carriage 7/6).

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Utiliaes 4 vaives, 1 each 5246, 6V6G 6J7G, AJ8G and high quality components such as "C' Core Transformers and Bick Paper Smoothing Condensers. A.C. Mains Pack for nominal 110 ×230 volts. Provision for 600 ohms or High Impedance input. Output to 600 ohm Line. For normal use only requires changing output Transformer. Output approximately 4 watts Designed for Standard Rack Mounting, having grey front panel size 19in. x 7in. All connections to rear panel, front having "OutOff" Switch Gam Control. Indicator Light. Fuses and Vaives Inspection Panel. BRAND NEW IN MAKEB'S PACKING. ONLY £4/9/6 (carriage 10/6).

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JANUARY, 1961



SELENIUM L.T. RECTIFIERS. Full Wave, bridge connected. 12/18 v. 1.5 A. 4/3; 12/18 v. 24 A. 6/9; 12/18 v. 4 A. 9/9; 12/18 v. 5 A. 12/6; 12/18 v. 6 A. 13/6; 24/36 v. 1 A 12/6; 24/36 v. 4 A. 22/6; 24/36 v. 15 A. 62/6. Please add postage

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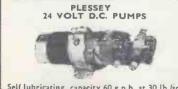
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Self lubricating, capacity 60 g.p.h. at 30 lb./sq. In. Will operate O.K. on 12 v. 4 BSP inlet/ outlet union. Only 15/6 ea. P/P 2/6.

# FIELD TELEPHONES TYPE F. Ideal for all Intercom. systems house, office. building sites, etc. Generator bell ring-ing, 2 line connection. Supplied complete with batteries and wooden carrying case, fully tested. £4/19/6 pair. P/P 5/-.

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R.1155 COMMUNICATION RECEIVERS Standard Model B. Frequency coverage 75 kc/s to 1,500 kc/s and 3 mc/s to 18 mc/s 75 kc/s to 1,300 kc/s and 5 mc/s to 16 mc/s on 5 bands. New improved geared slow-motion drive fitted. All receivers over-hauled, aligned and tested.  $\mathcal{E}8/19/6$  ea. P/P 7/6. Combined A.C. mains power pack and audio output stage supplied 85/- extra.

# BRAND NEW MEDRESCO HEARING

AIDS Supplied fully tested and complete with earpiece, leads and battery pouch. Incorporates 3 sub-miniature valves and sensitive crystal mic. Only 32/6 each. Batterles 5/- extra. mic. C P/P 1/-.

BC.221 HETERODYNE FREQUENCY

I25 kc/s to 20 mc/s. As new condition. Supplied complete with valves and crystal but no calibration charts. Only £14/10/- ea. P/P 7/6.

FIELD TELEPHONES TYPE H. Ideal for all intercom. systems. Generator bell ringing, two line connection. Supplied com-plete with batteries, ready to operate, 62/6 each, P/P 3/6.

SPARES KIT FOR CR.100 RECEIVERS Contains 15 valves: 2–U50, 2–DH63; 2–KT63; 2–K66; 7–KTW61. Condenser and resistor packs, pors, toggle switch, output transformer, etc. All brand new, 59/6. P/P 3/6.

R.C.A. LOUDSPEAKERS High-quality 3 ohm speaker housed in black crackle metal case to match AR-88 or H.R.O. receivers. Supplied brand new and boxed, 45/- ea. P/P 3/6.

COLLARO STUDIO TAPE TRANS-SCRIPTORS Latest 1960 model 3 speeds, 17, 37 or 74. Fitted with 3 separate motors, digital counter, press-button switching, provision for fitting extra stereo head. Supplied brand new and guaranteed complete with spare 7in. spool, £12/10/- ea. P/P 3/6.

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R.C.A. PLATE TRANSFORMERS Primary 200/250 v. 50 cycles. Secondary 2,000/1,500/0/1,500/2,000 v. 500 milliamps milliamps Supplied brand new and boxed, £6/10/- ea. PARMEKO TABLE TOP TRANSFOR-MERS. Input 230 v. 50 c/s. Output 620/550/ 375/0/375/550/620 v. 250 mA., 5 v. 3 amp, 5 v. 3 amp, Size 63 x 64 x 5½in. Brand new, boxed, 45/- ea. P/P. 5/-.

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E.M.I. 50 : I MICROPHONE TRANSFOR-MERS, 4/6 ea. P/P I/-.

**GERMANIUM DIODES.** General purpose type, 6d. ea. High quality type equivalent to OA81, 2/- ea.

TRANSISTORS. Red spot, 3/6 ea. White spot, 3/6 ea. Yellow/green, 4/6 ea. Red/yellow, 7/6 ea. P/P 3d.

	SPEAKER B	ARGAINS	
24in.	Perdio	3 ohm	17/6
21 in.	Perdio	15 ohm	17/6
3in.	Plessey	5 ohm	15/6
3in.	Rola	3 ohm	17/6
4±in.	Plessey	3 ohm	15/6
5in.	Goodmans	3 ohm	15/6
6 <u>‡</u> în.	Plessey	3 ohm+	17/6
Bin.	Elac	3 ohm	19/6
10in.	R.A.	3 ohm	27/6
12in.	Plessey	3 ohm	29/6
6 x 4in.	Plessey wafer	3 ohm	12/6
7 x 4in.	Plessey	3 ohm	15/6
8 x 6in.	Rola	3 ohm	17/6
10 x 7in.	Plessey	3 ohm	27/6
12 x 8in.	Plessey	3 ohm	49/6
8 x 22in.	Goodmans	3 ohm	17/6

AMPHENAL IS PIN UNITERS. Brand new 3/6 pr., ditto 18 pin, 3/6 pr. P/P 6d.

4 POLE II WAY METER SWITCHES. 4 bank 6/6 ea. P/P 9d.

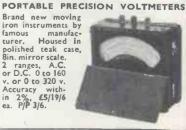
LEACH 12 v. DOUBLE POLE AERIAL CHANGEOVER RELAYS. 7/6 ea. P/P I/-.

SOUND-POWERED TELEPHONE HAND-SETS. Just connect with twin flex for complete telephone system. No batteries required, 15/- ea. P/P 1/6.

240 v. primary. Secondary 220 v. 85 mA and 6.3 v. 3.5 amps. New, 9/6 ea. P/P 1/3.

CONTACT-COOLED RECTIFIERS. 125 v. 85 mA 3/9; 250 v. 50 mA 5/6; 250 v. 85 mA 9/-; 250 v. 75 mA, full-wave bridge, 12/6. P/P 6d. ea.

HOOVER ROTARY TRANSFORMERS. Input 12 v. D.C.; output 310/360 v. 30 mA. New, boxed, 12/6 ea. P/P 1/3.



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# WIRELESS WORLD





BRAND NEW Boxed 100 MICROAMP METERS. Standard 24in. flush panel mounting. Scale calibrated 0-100 microamps. 42/6 each. P/P. 1/3. Also available 3±in. panel mounting 62/6 each.

MAINS PANEL NEON INDICATORS. Chrome escutcheon, flying lead connections. Available red, green or clear, 3/6 each. P/P. 6d. ALUMINIUM CHASSIS. 18 swg. four sided, reinforced corners. All sizes 2½in. deep. 6 x 4in, 3/6; 7½ x 5½in. 4/6; 10 x 7½in. 5/3; 11½ x 7½in. 6/-; 13½ x 9in. 6/9. Post extra.

PARMEKO MAINS TRANSFORMER. Fine heavy duty job. PrImary 0/110/230 volts. Sec. 350/0/350 volts 150 ma. 6.3 v. 4 amps, 5 volts 4 amps. New, boxed, 32/6. P/P. 2/-.

PRECISION WIREWOUND POTENTIO-METERS. Linear track, 34in. dla. Available 100 ohm, 500 ohm; 1k, 2.5k; 5k; 10k; 25k; 50k and 100k ohms. All 10/6 ea. P/P. 1/-.

750 WATT AUTO TRANSFORMERS. Fine heavy Admiralty type. Tapped from 110 to 230 volts to give any spot voltage. 69/6 each. P/P. 5/-.

POTTED "C" CORE CHOKES. 16 H. 150 ma.; 20 H. 100 ma.; 16 H. 120 ma.; 20 H. 80 ma.; 100 H. 30 ma. All 10/6 ea. 5 H. 500 ma. 17/6; 10 H. 500 ma. not potted, 25/- ea. Post extra. FERRANTI POTTED FILAMENT TRANS-FERRANTI POTTED FILAMENT TRANS-FORMERS. Primarles tapped 200/250 volts. 1. 6.3v. ct. 5.6 amp; 6.3v. ct. 3.3 amp; 6.3v. ct. 1 amp; 6.3v. ct. 9 amp; 6.3v. ct. amp; 6.3v. ct. mp; 6.3v. ct. 9 amp; 6.3v. ct. 6 amp; 15/6 ea. amp; 6. P/P. 2/-.

POST OFFICE JUMPER LEADS. Screened twin lead fitted with two standard jack plugs, 3/- ea. P/P. 9d.

SIEMENS HIGH-SPEED RELAYS. 1700 +1700 ohm colls. Single pole changeover. Miniature sealed type. New, boxed, 15/6 ea. P/P. 6d.

ROTARY TRANSFORMERS. 12 volt D.C. input. Output 250 volt 80 ma. 22/6 ea. Ditto 6.volt Input, 22/6. P/P. 2/6.

CHOKE BARGAINS. 4H. 22.5 ma. 2/6; 5 H. 60 ma. 3/6; 5 H. 200 ma. 5/6; Collins 8 H. 100 ma. 8/6; Rich & Bundy 50 H. 120 ma. 12/6. Post extra. HEAVY "C" CORE TRANSFORMERS. 230 volt primary. 725/700/675/0/675/700/25 volt 500 ma. 6.3 v. 6 amp. 6.3 v. 1 amp. 5 v. 6 amp. New, boxed, 72/6 ea. P/P. 5/-

FERRITE CORED LOOP AERIALS. Opera-tive up to 2mc/s. New boxed, 22/6 ea. P/P. 2/6. ADMIRALTY SLOW MOTION DRIVES, 180°, scaled 0 to 100. Fast and slow knob with lock and also flick mechanism for setting to fixed frequencies, new 7/6 ea. P/P. 1/3.

24 AMP. VARIAC TRANSFORMERS. Primary 230 volts. Adjustable sec. from 185 to 250 volts. 24 amps, £12/10/- each. P/P. 10/-.

"C" CORE POTTED L.T. TRANSFOR-MERS. Primaries all 230 volts. 1. 6.3 v. 3 amp. 6.3 v. 3 amp., 6.3 v. 3 amp., 6.3 v. 1.5 amp., 21/- ea. P/P. 2/-. 2. 6.3 v. 5 amp., 6.3 v. 4 amp., 2 x 6 6.3 v., 3 amp. 6.3 v. 2 amp., 6.3 v. 1.5 amp., 6.3 v. 1 amp., 35/- ea. P/P. 2/6.

VALVE VOLTMETERS No. 2. A labora-tory instrument at a fraction of cost. Five ranges A.C. and D.C. 1.5 v., 5 v., 15 v., 50 v. and 150 volts. Operation 200/250 volts A.C. Supplied as new, fully tested and com-plete with internally mounted H.F. probe. £17/10/- each. P/P. 10/-.

### HIGH FIDELITY P.V.C. BASE RECORD-ING TAPES



BARGAIN PRICES 3in. spool 225ft. L.P.

6/-. 5in. spool 600ft. std. 12/-. 5in. spool 900ft. L.P. 17/-. 5‡in. spool 1200ft. L.P. 19/6. 7in. spool 1200ft. std. 19/-. 7in. spool 1800ft. L.P. 29/-.

SPARE PLASTIC SPOOLS, 52in. 2/3; 7in. 3/6. New, Boxed, Guaranteed. Post extra.

HALLICRAFTER 6 VOLT VIBRATOR POWER SUPPLIES. Magnificent units housed in grey metal case and supplied with all necessary connectors, etc. Made for SX28, S27, S36 receivers, etc. Output 300 volts 170 ma., fully smoothed. Supplied new boxed, 29/6 each. P/P. 3/6.

AVO SIGNAL GENERATORS. Frequency coverage 95 kc/s. to 40 mc/s. Ideal for all general radio work. Supplied fully tested and checked. £7/19/6 each. Operation is from 2 v. and 60 v. batteries but original Avo mains units can be supplied at 19/6 ea. P/P. 3/6.

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# 8 RANGE SUB-STANDARD D.C. AMMETERS

Ranges 1.5, 3, 7.5, 15, 30, 60, 300 and 450 amps. 8in. mirror scale. Housed in polished teak case. Supplied complete with all shunts and leather carrying case, £15 each. P/P. 7/6.

### PHOTO VOLTAGE AMPLIFIERS

These special units contain a I microamp. Tinsley mirror galvonometer and a double selenium photo electric cell. Brand new, £9/19/6 ea. P/P. 7/6.

RECORD CHANGERS/PLAYERS Bargin prices, all new and guaranteed. COLLARO CONQUEST. 4 SPEED AUTO CHANGERS. £6/15/- ea. P/P. 3/6. B.S.R. UA8 4SPEED AUTO CHANGERS £6/15/- ea. P/P. 3/6. COLLARO JUNIOR 4 SPEED SINGLE PLAYERS. 79/6 ea. P/P. 2/6.





123

HALLICRAFTER S-36A U.H.F. COM-MUNICATION RECEIVERS. This Is the later version of the S-27. Frequency coverage 27 to 143 mc/s split on 3 bands, capable of receiving F.M. or A.M. sIgnals. Circuit incorporates calibrated S metrol Circuit incorporates calibrated S meter, B.F.O., noise limiter, etc. Operation 110/ 230 v. A.C. Output for phone or speaker. Supplied reconditioned and in superb con-dition. Price £27/10/- each. P/P, £1.

Attom. Frice 22/10/- each. F/F, EL. NEW BLOCK PAPER CONDENSERS. Nitrogol, Visconol types. .25 mfd. 4 kv. 3/6; .25 mfd. 7.5 kv. 10/6; .25 mfd. 10 kv. 15/-; 1 mfd. 600 v. 1/9; 1 mfd. 1 kv. 3/6; 1 mfd. 2.5 kv. 6/6; 1 mfd. 5 kv. 15/-; 2 mfd. 400 v. 2/6; 2 mfd. 600 v. 4/6; 4 mfd. 400 v. 3/6; 4 mfd. 600 v. 4/6; 4 mfd. 1,000 v. 6/6; 4 mfd. 1.5 kv. 8/6; 8 mfd. 400 v. 6/6; 8 mfd. 800 v. 8/6; 8 mfd. 1.5 kv. 15/-; 10 mfd. 1.5 kv. 17/6; 20 -cfd. 600 v. 17/6 32 mfd. 500 v. 17/6. Post extra.



POTTED TRANSFORMERS. 230 v. primary Secondary 350/310/0/310/350 v. 220 ma. Total of 6.3 v. 13 amps.; 5 v. 4 amps., 49/6 ea. P/P. 3/-110/230 VOLT AUTO TRANSFORMERS. 20 watt 9/-; 50 w. 12/6; 150 w. 18/6. Post extra. 100 AMP. A.C. MOVING IRON METERS. 6in. scale. Modern type, flush mounting ideal for switchboards, etc., new, boxed, 65/-. P/P. 3/6. "C"CORE E.H.T. TRANSFORMER. 230 v. primary. Secondary 3650 v. 55 ma. 4 v. 2.5 amp., 4 v. 1 amp. New boxed, 52/6 each. P/P. 2/6 JOHNSON VARIABLE INDUCTANCES. Large type 8ln. x 24jin. Supplied brand new boxed, Large type 8In. x 21in. Supplied brand new boxed, 22/6 each. P/P. 2/-.

METER BARGAINS	
20 mlcroamp D.C. M/C flush rd. 24in.	69/6
25 microamp D.C. M/C proj. rd. 21n.	59/6
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100 microamp D.C. M/C flush rd. 31in.	62/6
200 microamp D.C. M/C proj. rd. 21in.	29/6
300 microamp D.C. M/C flush rd. 23in.	29/6
1 milliamp. D.C. M/C flush sq. 2in	22/6
1 milliamp. D.C. M/C flush rd. 21in	25/-
1 milliamp. D.C. M/C flush sq. 4in	69/6
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2 in. rd.	9/6
15 amp. D.C. M/C proj. rd. 2in	8/6
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300 volt A.C. M/C rectifier flush rd. 21in.	25/-
300 volt A.C. M/I flush rd. 21in	25/-
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1,500 volts electrostatic proj. rd. 24in.	25/-
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JANUARY, 1961





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### **VOLTAGE TRANSFORMERS** VARIABLE



TYPE B5 5 amp. 260 v. output, as £8 10 0 illustrated . . . . . . . . . . . . . TYPE BIO 10 Amp. 260 v. output £17 15 0

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Large easily read Dial calibrated 0-260 v. Totally enclosed with Input and Output Terminals. Ideally suited for Laboratory experimental work and Schools.

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COMPLETE INSTRUMEN-TATION INSTALLATIONS FOR INDUSTRIAL PROCESS CONTROL.



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**EXPORT ONLY** PROMPT deliveries Mobile V.H.F. Radio Telephones. Frequency ranges on five bands (1) 36-44 Mc/s. (2) 65-78 Mc/s. (3) 78-100 Mc/s. (4) 118-132 Mc/s. (5) 156-174 Mc/s. R.F. output 10 watts. A.M. Single Channel, crystal controlled. To operate from 6 v. 12 v. or Mains supply sources. Reconditioned with same as new guarantee. Prices from £55 per complete station FOB U.K. Port, as Illustrated.



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# **RADIO CLEARANCE LTD.**

TRADE ENOURIES INVITED

27 TOTTENHAM COURT RD., LONDON, W.I The oldest Component Specialists in the Trade

Telephone: **MUSEUM 9188** EST. 30 YRS.



All Electrolytic Condensers as advertised in May 1950 issue still available.

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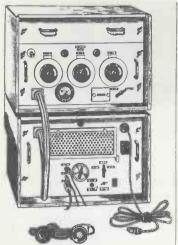


WIRELESS WORLD

127







TRANSMITTER TYPE 36. A complete 50 watt TX for phone or CW. Covers 10-40 Mc/s. (10-15-20m.), Crystal or stabilised VFO. Push-pull 807's plate and screen modulate parallel 807's. Tested and ready to plug into AC mains. Complete with 16 valves, handset, operating instructions and circuit. Wooden cabinets may be somewhat damaged. £15 Carr. England and Wales £2.

MOVING COIL PHONES. Finest quality Canadian with chamois ear-muffs Finest quality Canadian with chamois ear-muns and leather-covered headband. With lead and jack plug. Noise excluding and supremely comfortable. **19/6.** Post I/6. **MATCHING TRANSFORMER** (for Hi impedance) i.e. for HRO, CR100, etc., with standard jack plug. **4/6**. with standard jack plug, 4/6.

SELENIUM BRIDGE RECTIFIERS Funnel cooled. A.C. input 45 v. RMS. D.C., output 30 v. 10 amps. BRAND NEW. Boxed. 45/-, Post 3/6.

MARCONI IMPEDANCE BRIDGE. Type TF373. Measures, L, C & R at 1,000 Cycles. Accuracy 1%. 0-100 $\mu$ F; 0-100 $\mu$ F; 0-100 $\mu$ F; 0-200 $\mu$ F; 0-20

ADMIRALTY HT TRANSFORMERS Pri. 230 v. 50 c/s. Secs 620-550-375-0 375-550-620 v. (620 and 550 v. 200 m/amps.), 250 m/amps.), plus two 5 v. 3 amp. rectifier windings. Total rating 278 VA. Upright mtg. Vvt. 251b. Made 1953. BRAND NEW. Original boxes. 45/-. Carr. 5/-.

INSTRUMENT TRANSFORMERS 230 v. A.C. input. Outputs 0-65-130-195 v. 85 m/amps., 6.3 v. 5 amps., 6.3 v. 0.3 amps. Shrouded. Size 3½ x 3⅔ x 3∄in. high. 15/-. HALLICRAFTER VIBRAPACK. Input 6 v. output 300 v. at 170 mA. Designed for SX28 or S27. Size 6½ x 7 x 7in. BRAND NEW, BOXED. 29/6. Carr. 3/6.

AR88D MAINS TRANSFORMERS. Arood 1-1413 Interest 145-0-345 v. 125 m/amps., 6.4 v., 4.5 amps., 5 v. 2 amps. 4<sup>3</sup>/<sub>4</sub> x 4<sup>1</sup>/<sub>4</sub> x 5<sup>1</sup>/<sub>4</sub>In. high. Wt. 121b. Potted. Tag ends. RCA BRAND NEW. Boxed. 2014 area 316 29/6. post 3/6.



### TRIPLETT METER MOVEMENT This article consists of a basic 400 microamp meter movement mounted on a Bakelite panel 53 x 27. The dial is scaled as a 15 range

original instrument is supplied. BRAND NEW. Boxed. 35/-, post paid.

Testmeter. A circuit and parts list of the

QQV06-40 37/6 PV I-35 32/6, 2D21 7/6, OC3 6/-, PT I5 12/6, CV51(Y65) 5/-, 6F33 5/-, BRAND NEW in individual cartons. Bulk enquiries invited

# CANADIAN RECEIVER No. 52

CANADIAN RECEIVER NO. 32 1.75-16 Mc/s (19-170 m.) in three wavebands R.F., Mixer, Sep. Osc., 2 1.F.'s, Det/A.V.C., 1st Audio, Output, BFO (10 valves), plus a 3-valve dual Crystal Calibrator. Controls: R.F. Gain, L.F. Gain, Crash Limiter, C.W. Filter, Variable Selectivity, slow and Fast Tuning and Osc. Vernier Tuning. Man. or A.V.C. BFO pitch control. Internal 3in. speaker and valve check meter. Power supply required 160 v. H.T., 12 v. L.T. Data and Circuit supplied. A really excellent receiver, £8/19/6, carr. 15/6. Power supply Unit, 59/6, carr. 5/6.

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Type AN/APR4. Covers 38 to 1000 Mc/s. with 3 Plug-in R.F. Heads. TN 16 (38-95 Mc/s.) TN 17 (74-320 Mc/s.) and TN 18 (300-1000 Mc/s.). Self-contained power supply for 115v. 50-2,600 c.p.s. Thoroughly reconditioned as new. In absolutely 100 per cent mechanical and operational order. £100.

# MARCONI CRI00

Completely overhauled. In perfect working order. LOOK LIKE NEW. £21. Later model with Noise Limiter, £25. Carr. England Wales 30/-. Send S.A.E. for full details.

# **RECEIVERS R-1155B**

A first-class 10-valve Communications receiver, covering 75 Kc/s. to 18 Mc/s. (16.2-4,000 m.) in 5 bands. The large scale and superior dual ratio slow-motion drive make tuning easy and the R.F. stages ensure world-wide reception. All the receivers we sell have been thoroughly overhauled. completely realigned and are in first-class working order. ONLY £9/19/6. A.C. MAINS POWER PACK OUTPUT STAGE.

In handsome black crackled steel cabinet to match the R-1155. Fitted with RCA 8in. speaker. Just PLUG IN and switch on Only the finest quality components are used and we guaran-tee OUR power packs for 6 months. ONLY £6/10/-. Deduct 10/- when purchasing receiver and power unit cogether. Send S.A.E. for further details or 1/3 for 10-page Illustrated booklet giving technical data and circuits etc. (Free with each receiver). Add 10/6 carriage for receiver, 5/- for power unit.

# RCA AR-88 SPEAKERS

A high quality 3 ohm unit fitted into heavy gauge black crackled steel cabinet, size 104 x 114 x 6in. Fitted with rub-ber feet and 6tt. lead. Ideal for extension speaker. CR 100. etc. In original cartons. BRAND NEW. 45/-. Post 3/6.

MINIATURE 373 IF STRIPS. For FM tuner described in "Practical Wireless." Complete with 3 of EF91, 2 of EF92 and I of EB91. A fresh release enables us to offer these once again. BRAND NEW. Complete reprint of conversion instructions and circuit supplied free. 35/-, OR less valves 12/6. Post, either, 2/6.

# LOUD-HAILER EQUIPMENT

DEAL FOR GROWD CONTROL, FACTORIES, FETS, ETC. CON-BISTS OF 4 SPEAKER UNITS AND CONTROL UNIT. COMFLETE WITH MICROPHONE. HEADPHONE AND SPARES, OPERATES FROM 12 VOLTS D.C. (OR 6 VOLTS A.C. WITH SLIGHTLY REDUCED OUTPUT CONSUMING NOLY 3 AMPS, OUTPUT FOULS & WATTS. ALL TESTED AND WORKING, BUT SLIGHTLY SOLLED. A GENUINE BARGAIN. \$4/19/6. CARBIAGE 25/6.

T.C.C. VISCONAL CONDENSERS. 8 mfd. 800 v. D.C. wkg. at 71 deg. C. CP152V. Size 3 x 1½ x Sin. high. BRAND NEW. Boxed 8/6 each, post paid. 4 mfd. 600 v. wkg. CP 130T. 4/6 each, post paid.

GIANT COMPONENT PARCEL Contains 100 4 and 1 watt resistors, 50 Hi Stab resistors, wire wound resistors, carbon and W/W pots, 100 capacitors (mica, paper, Borague, bias, variable, etc.), valveholders, tag strips, metail rectifiers, leieving etc. All components are unused. GUARANTEED VALUE, 25/- plus etc. All 2/6 post.





CRYSTAL CALIBRATOR No. 10 **CRYSTAL CALIBRATOR No. 10** A crystal controlled heterodyne wave-meter covering 500 Kc/s. to 10 Mc/s. (Harmonics up to 30 Mc/s.) Requires 300 V. IS mA. and 12 V. 0.3 a. D.C., but can be easily modified for 120 V. and 1.4 V. work-ing. Size 7 x 7<sup>1</sup>/<sub>4</sub> x 4in. Good condition, complete with valves, crystal, instruction manual and circuit. ONLY 59/6. Post 3/6. **BRAND NEW** and with spare set of valves. £4/10/-, post 3/6.

CANADIAN CRYSTAL CALIBRA-TOR. Uses double crystal and multi-vibrator circuit to give "pips" at I Mc/s., 100 Kc/s. and 10 Kc/s. Incorporates Modulator Handbook supplied. 79/6, post 2/6.

PHILIPS RADIATION MONITOR. Type 1092C. A portable self-contained instrument for measuring radio-activity, uses the Mullard MX-115 Geiger counter tube, and is scaled 0-10 milli-Rontgens per Supplied complete with carrying k. BRAND NEW. £17/10/0. hour. haversack. BRAND NEW. £17/10/0. Carr. 5/-. Other types of radiation monitoring equipment in stock

SANGAMO WESTON ANALYSER E772. A useful multi-range meter. Thor-oughly overhauled and in perfect working order. For full details see previous adverts. £7/10/-. Carr. 4/6.

MARCONI TF987/1 NOISE GENE-RATORS. Range 100 Kc/s. to 200 Mc/s. Determines noise factor of AM and FM receivers. Fully stabilised H.T. supply A.C. mains operation. Brand new and A.C. mains operation. Brand no in original boxes. £15. Carr. 7/6,

HEAVY DUTY SLIDER RESISTORS. 1.25 \Overline 20 A., 12/6, post 3/6 | \Overline 12 A., 8/6. PRECISION RESISTORS. | Megohm. 1% I watt wire wound, Ex-U.S.A. BRAND NEW. 10/6 per dozen.

D.C./A.C. CONVERTERS. Input 12 v. D.C. Output 230 v. 50 c/s. A.C. at 135 watts. Fitted with 0-300 v. A.C. 24in. meter and slider resistor for voltage 2410. meter and slider resistor for voltage adjustment. In stout wooden carrying case with lid. Perfect working order. **59**/19/6. Carr. 10/6. 24 v. Input 230 v. A.C. 50 c/s. 100 watts output. In grey metal case. BRAND NEW. 92/6. Carr. 7/6

FIELD TELEPHONES. Army type D, Mk. 5. Buzzer calling. Ideal for building sites, farms, workshops, etc. Complete with handset and batteries. Tested. 32/6.

### MICROAMMETERS

R.C.A. 0-500 microamps. 24in. circular flush panel mounting. Dials are engraved 0-15, 0-600 volts. As used in the American version of the No. 19 set BRAND NEW. Boxed 15/-.

American 0-100 microamps. 24in. square flush panel mounting. BRAND NEW. flush panel mounting. Boxed, 42/6.

### FERRANTI VOLTMETERS

N.5. 0-300 volts, 25-0-300 voits, 25-100 c/s. Moving iron, 6in. scale. FI. mtg. Her-metically sealed, grade IN. Made 1955. BRAND 1955. BRAND NEW. Boxed. 79/6. Post 3/6



WIRELESS WORLD



TRANSISTOR AMPLIFIER K | 7 Printed circuit, 500 milliwatt push-pull output. Input and output transformers of 3 ohms impedance with two OC.71 and two OC.72. Supplied complete with all condensers, resistors, volume control, etc. 52/6 post paid Input 9 volts. for only 52/6 post paid

# MAINS TRANSFORMERS

	olt 50 c/s. post paid	
Type 1. 250-0-250 at 70mA.	6.3v at 2A. 4v at 2A.	9/-
2. 300-0-300 at 70mA.	6.3y at 2.5A. 5v. at 2A	10/-
3. 350-0-350 at 120mA.	6.3vat 3.5A. 5vat 2A.	16/6
4. 350-0-350 at 300mA.	6.3v at 8A. 5v at 2A plus 4v at 2A and	
	6.3v at 2A.	27/6
5. Filament only:	6.3v at 4A.	8/-



 Set of 80 covering 5/06.667 to 8340 kcs in 33.333 kcs steps.
 64/3/6 post paid

 Any one between 7006.667 & 7173 kcs, or above 7973 kcs
 6/6 each

 Any Three others
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 Set of 120 covering 5675 to 8650 kcs in 25 kcs steps.
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 Any one between 7000 to 7300, or 8000 to 8300 kcs
 6/6 each

 Any Three others
 6/6 each

 Low of 120 covering 5675 to 8650 kcs in 25 kcs steps.
 6/6 each

 Any one between 7000 to 7300, or 8000 to 8300 kcs
 6/6 each

 Any Three others
 10/- post paid

# AMERICAN 400 CYCLES INVERTER.



30/- post paid. Very neat unit indeed, only 21 dia. by 4in. long on 11 in. high pedestal base containing suppressor. Ball bearings. 24 volt D.C., input for 26 volt single phase A.C. 6VA output. Instrument quality — as used with Bendix Magnesyn compass system.

100 watt £18/0/0 carriage paid ARTIS TRANSMITTERS. 

SCR 522 TRANSMITTER RECEIVER Well known American airborne equipment covering 100 to 156 Mc/s. BC625 transmitter has a 6G6 Crystal oscillator with the 2nd harmonic fed to a 12A6 and an 832 tripler stage and finally to an 832 power amplifier giving 8 wats. The BC624 superhet receiver has a 9003 RF and Mixer, three 12SG7 12 Mc/s IF's, 12C8 Detector, 12J5 Audio, and 12AH7 oscillator using 8 to 8.72 Mc/s crystals. AVC, Noise Limiter, and Squelch provided. Output transformer provides 50, 300 and 4,000 ohms. Successful conversion to 2 metres was fully described in CQ. Circuit diagram (unmodified) supplied. In good used condition. • 

Transmitter, less valves, but including modulation trans-including Receiver, less valves, 10/-15'former, . post paid.

POST PAID VALVE BARGAINS

7Z4, EY91, 6AM5, 6AM6, 6C4, 6AL5, CV71 (neon stabiliser)..... ..... any 4 for 10/-POST PAID TRANSISTOR BARGAINS 

8 OC.170 27/6; OC.16 37/6; Goldtop V30/IODP ... 21/-Also leading make of 3v transistors ...... 4 for 10/-

WIRELESS WORLD



### POST FREE SNIPS

Double pole knife changeover switch on porcelain base. 2 for	5/-
G.P.O. 230 volt mains, twin six inch gong, outdoor bells 3	3/6
Siemens high-speed relays. 1,000-0-1,000 ohm coils	
Pyrex Aerial Insulators. Four 3in. OR one 8in	
U.S.A./British co-ax. adaptors. Four for	5/-
Neons. Ten 115 volt for 12/6; Six 80 volt for	7/6
G.P.O. mechanical counters. 0-9999	7/6

### RECEIVER TYPE 88 (R1475)

Highly stable, specially accurately calibrated, Marconi design, RAF communications receiver covering 2-20 Mc/s in 4 bands with built-in 600 kcs Xtal reference oscillator for checking dial which can be reset by special panel trimmer control. 11 Valves: 3x6K7, 6K8, 6J5, 3x6Q7, 6H6, Y63 tuning indicator and VR150-30 voltage regulator. Two stage IF with 8 tuned circuits, Xtal controlled B.F.O. Four position selectivity with audio filters for narrow bandwidth C.W. Fast and slow AVC, high and low noise suppression. A plug-in unit with additional mixer provides a "listening through" guard channel of either 2-4 or 4-74 Mc/s. Receiver 164 × 9 × 11in. Power pack 8×9 × 11in. Complete with 200-250 volt AC (or 12v. Used, but in very good condition. Guaranteed serviceable. A sound

buy indeed at £13,10.0 carriage paid.

### ETCH-YOUR-OWN PRINTED CIRCUIT KITS Post Free

132

Each contains over 60 sq. Each contains over 60 sq. in. of laminated board and sufficient chemicals to make dozens of printed circuits, plus comprehen-sive instruction book giving advice and examples on translat-

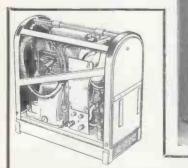
outs ready for etching. High-quality materials-completely safe to handle-carefully prepared to ensure fine definition and uniform results without laboratory control.

### ......... .... **£10 GEIGER COUNTER**

Circuit embodies U.K.A.E.A. patent. Specially moulded case. Cur-rently being supplied throughout the world. Three ranges—highly sensitive.—light—portable—visual and audible response—plus output socket. Ideal for introduction to radiation measurement and nucleonic circuitry. Specially written 40-page instruction manual supplied. Batteries £2/15/3 extra.

### KIT OF PARTS £4/17/6

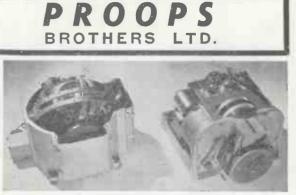
Identical parts. Guaranteed per-Manual and printed formance. circuit plates for battery pack supplied (assembled pack £2/15/3 extra). Fully illustrated assembly instructions. Spares and service permanently available.



PROOPS

### PORTABLE A.C./D.C. GENERATING SET

Self-contained 80 watt unit on compact chassis delivering 12 to 18 volts D.C. Size only 14×15×8in. Weight 46lb. Spring mounted air cooled petrol engine with fuel tank in base driving integral generator that has heavy duty bridge rectifier feeding D.C. terminal board. Miniature sparking plug. Filtered air intake. £9 plus 10/-Guaranteed serviceable.



D.C. GYRO & SERVO MOTOR-CI AUTO PILOT

D.C. GYRO & SERVO MOTOR-CI AUTO PILOT Beautifully engineered Minneapolis-Honeywell precision gyro, totally enclosed in sealed light-alloy housing about  $8^{\pm}_{1n}$ . cube. Automatic erection and precession correction. Large diameter Dessyn type trans-mitting potentiometers provide signals corresponding to the magnitude of the deviation of gimbal arms. Powerful D.C. motor coupled through a differential reduction gear to a 4in. spur driving gear integral with a 3in. dia. spiral groove cable driving drum. Two powerful solenoid clutches and corresponding brakes hold drum rigidly in position or set free for "neutral." Nominally for 26-volt operation, but operates at 12 volts. Size 10×6×8in. £10 each unit or £17/10/- pair, carriage paid.

This is the attractive lightweight American Inis is the attractive lightweight American Radio Altimeter that superseded the British version. A complete 14-valve radar set cover-ing 420-460 Mc/s it is ideal for conversion to radio control of models or 70 cm, work. It embodies three self-contained sub-units in separate detachable aluminium cases, follows: 28



#### TRANSMITTER

TRANSMITTER A push-pull feed-back oscillator tuncable either side of 445 Mc/s., frequency modulated at 100 c/s by a particularly robust moving coil transducer. Two 955 high frequency acorn valves. Case size only  $3\frac{1}{2} \times 6\frac{1}{2} \times 2in$ . plus  $2 \times 2\frac{1}{2}$  in. dia. for transducer.

#### RECEIVER

Tuneable to transmitter frequency. Size  $3\frac{1}{2} \times 6\frac{1}{2} \times 2in$ . Two 9004 acorn valves.

### AUDIO AMPLIFIER

AUDIO AMPLIFIER Self-contained RC coupled 12SH7, 12SH7 and 12SJ7. Size  $3 \times 5 \times$ 1§in. Amplifies the received signal which is passed to detector circuit giving a D.C. voltage proportional to the difference between the trans-mitted and received (reflected) signal to operate internal relays which pass appropriate correction signals to autopilot and supply external indicator (5 mA meter).

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MAIN CHASSIS The main chassis carries the 3 sub-units and has a further three 12SH7 one 12SJ7, two 12H6 and one VR150 regulator, three 1% wire-wound resistors, one 4-pole changeover relay, two SPCO relays, three twin-ganged pre-set potentiometers, trimmers, fuses, etc. Power supply is derived from a 27-volt dynamotor (charging rate for 24 v. supply) delivering 285 volts at 75 mA.

BRAND NEW, a very useful buy indeed at only £2 plus 7/6 carriage.



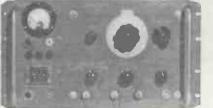
PORTABLE

RECEIVER

Neat, lightweight but really sturdy petrol engine. Completely self contained, air-cooled pedestal-based unit with 5in. dia.  $\times \frac{1}{2}$ in. Vee pulley for driving generator, pump, etc. Made by Lauson Engines in the U.S.A. for easy transport in a special lightweight con-tainer. Developing 1.8 h.p. at 2,700 r.p.m., this very fine unit is only 17in. high  $\times$ 14in.  $\times$  12in. and can be carried in one hand. It has stellited valves to suit any petrol, a totally enclosed carburettor with air filter and a mechanical fuel pump with glass bowl filter.

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### PRECISION SIGNAL GENERATOR CT53. A modern laboratory standard instrument still



### FEATURES

- Vernier tuned, Triple screened, 6-Band coil turret covering 8.9 to 300 Mc/s with 72 ohm output from 100 mV down to  $1\mu$ V. Precision decade ladder and silver slide wire attenuator calibrated in voltage and .
- 0-90db

- 0-90db. Variable carrier level monitored by cathode follower and VTVM. CW or modulated 30% by 1,000 c/s Sine or Square wave (variable mark/space ratio). External mod, by sine wave from 50 c/s to 10 kc/s. or pulses down to  $\frac{3}{4}\mu$  Sec. Seven B7G Valves, Potted "C" core transformers, Paper capacitors, Stabilised H.T. Selected spare oscillator, pre-aged spare monitor, 100 $\mu$ A meter. Mains, H.T., Bias and Filament supplies fully RF filtered. Combined cabinet/rack mounting case, Pressure sealed, Desicator, Panel Mains voltage adjustment, Triple fused, in fact, "the lot"! ٠

Offered straight from Service use, complete with calibration book, cables, circuit diagram and principal technical information, checked serviceable and fully guaranteed. Plus 15/- for careful packing and carriage £17.10.0



Attractive, lightweight, black crackle box  $11 \times 7 \times 13\frac{1}{2}$  in. deep with  $4 \times 2\frac{1}{4}$  in. and  $3\frac{1}{2} \times 3$  in. square windows on front panel for twin 5FP7 tubes. Neat arrangement of appropriate (independent) controls and variable scale illumination. Totally enclosed detachable magnetic focusing coils. All connections to rear sockets. Ideal TV monitoring unit as used by many amateurs. Used, but in very good condition, tubes guaranteed O.K.  $\pounds 1/10/0$  carriage paid.

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POWER

carburettor with air filter and a mechanical fuel pump with glass bowl filter. Flywheel cord start. Push-button stop. Adjustable throttle. Butterfly choke, etc. Standard 14 mm. spark plug with screen-ed H.T. harness. Crankcase oil bath. Supplied complete with 3ft. flexible exhaust pipe and detachable 9 × 3½in. dia. silencer, driving belt and 10ft. of high-grade flexible fuel hose. genuine quality engine offered at the remarkable price of only

3-INCH CIRCULAR SCALE MILLIAMMETER

American panel mounting "Radio Altitude" meter with modern (coil round magnet) movement giving beautifully steady deflection to reading on large dial boldly marked 1 to 4 with sub-divisions in tenths. Supplied with suppressed zero which requires 6.5 mA. for full scale deflection (0 = 1.5 mA.) but pointer is easily re-set to zero by moving conventional hair spring adjuster behind dial, when 5 mA. gives f.s.d. Rear housing incorporates on/off switch (operated by rotating small knob on front face) and 5-pin plug, two pins direct to meter and two to switch. Not new. 17/6, post free.





Remote indication to within Remote indication to within 1° on precision instrument type flush fitting black crackle indicator with 3in. dial calibrated in 2° steps plus the four cardinals. Simple D.C. wiring (6-30 volt) from specially wound potentiometer in sealed die-cast housing with ±in.

drilled spindle transmits accurate signal of horizontal or vertical bearing.

35'-Brand New, Post Free,

MEGISTORS, 125, 1,000 or 10,000 MEGohms

Glass encapsulated 10% tolerance high value resistors for minute grid current applications. Ideal for extending the range of sensitive meters or using in probes to provide a really high impedance input for VTVM's or 'Scopes. One of each value plus any chosen two, for VTVM's or 'Scopes. One of the 5 for 10/- post free by return.



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MODULATOR ... £1 5 0 AND CONVERSION ... £1 10 0 DETAILS.

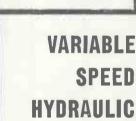
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A new two-wave band (L and M) Superhet using the latest miniature valves: ECH81, EF85 and ECL80, plus contact cooled

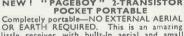
plus contact cooled Rectifier. Incorpora-tes Ferrite Rod Aerlal and Is of unit construction. Ex-ceptional sensitivity and selectivity. Out-standing performance and quality T.C.C. condensers throughout. Easily constructed in one evening. Brown or ivory Bakelite or wooden Walnut finish cabinet. A.C. mains 200/250 v. All necessary components at special inclusive price of £7/19/6 plus 3/6 P. & P. Instruction Book with itemised price list available separately at 1/6 post free. Also available In De Luxe Cabinet (as "Economy Four" at 5/- extra).

SUPER PERSONAL POR-TABLE. A wonderful little set that you can take anywhere. Ideal for camping, anywhere, totasi for camping, picnics, etc. Detachable aerial rod supplied. Covers Medium waveband 200-500 metres. Can be built in approx. I hour. All necessary components available at the approx. I hour. All necessary components available at the following SPECIAL INCLUS-IVE PRICES: I-valve version ONLY 35/- plus 2/- P. & P. Super 2-valve version ONLY 41/-. Plus 2/- P. & P. Send for point-to-point wiring diagram and parts price list 2/- post free. Extra for use with the above DLRS balanced armature headphones, 7/6 pair

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7/6 pair.



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Total cost of all (19) necessary components, Including transistors, wiring wire and even solder ONLY 32/6 plus 1/6 P. & P. Battery 3/- extra. Ardente type deaf-ald earpiece complete with cord and plugs extra at 12/6. Parts price list and Easy Lay-out Plans 2/- post free. Callers welcome to hear this set demonstrated at any of our branches. Our reputation is your guarantee

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400 MILLIWATTS OUTPUT To build yourself Medium and Long waves—Push-Pull Super-her A.V.C. Perfect Car Radio reception. Size 10in. x 63in. x

het A.V.C. Perfect Car Radio reception. Size 10in. x 64in. x 44in. at base tapering to 4in. at top. Very attractive two-tone grey Vynlde covered cabinet with black and gold printed escutcheon plate, cream and gold knobs, handle and cabinet fittings. Weight-complete with long-life 7½ volt battery-44lb.  $\bigstar$  Mazda high-grade transistors throughout.  $\bigstar$  High-flux 7in. x 4in. Elliptical Speaker.  $\bigstar$  Slow motion tuning.  $\bigstar$  Co-axial socket at rear for direct connection to Car Radio Aerial.  $\bigstar$  Improved reception by use of seven-section plated telescopic aerial disappearing into Cabinet when closed; 34in. above Cabinet when fully extended. when fully extended.

when fully extended. Construction simplified by Bakelite chassis board with the following components already mounted: I.F. Transformers (3). Oscillator Coil, Trimmer Bank, Output Transformer, Interstage Trans-former, Aerial Brackets and Earth Bar. SPECIAL INCLUSIVE PRICE for all required components, full assembly instructions—nothing more to buy—is £10/19/6 plus 3/6 P & P. Alignment service available. Full assembly instructions and Individually priced parts list, all of which are available separately, 2/6, post free.

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JANUARY, 1961

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CABY UNIVERSAL TEST METERS These pocket-size multi-range test meters are of excellent quality and cover all the most use-ful ranges (A.C. Volts, D.C. Volts, resistance and current). Supplied complete with test prods, in-struction book and batteries. Model A.10 (2,000 ohms per volt) £4/17/6

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1	2	7
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C.R.T. BOOSTER TRANSFORMERS	196
For Cathode Ray Tubes having Heater/Cathode short circuit and for C.R. Tubes with falling emission. Fuil	
installation instructions supplied. Type A. Low Leakage windings. Optional Boost 25% and 50%. Tapped mains primaries:	1
2 volt	1
	13
OUR LATEST SUPERIOR PRODUCT. Type A2. High Quality. Low capacity. 10/15 pf. Optional hours 158/ 508/ 758/	
13.3 voit 12/10 each OUR LATEST SUPERIOR PRODUCT. Type A2. High Quality. Low capacity, 10/15 pt. Optional boost 25%, 50%. conacity. Multi-Ontput 2, 4, 6.3, 10 and 13 voits. Optional boost 25% and 50%. Sultable for all Cathode Ray Tubes 21/	
50%. Suitable for all Cathode Ray Tubes 21/ RESISTORS. All preferred values. 20% 10 ohms to 10	THREE V
RESIGNORS         All preferred values.         20% 10         ohms to 10           meg., ‡ w., 4d.; ‡ w., 4d.; 1 w. 6d.; 14 w., 8d.; 2 w. 1/-         HIGH STABLLITY.         ‡ w., 1%         § 2/         Freierred values           10 ohms to 10 meg.         Dito 5% gd.; 100; to 5 meg.         5 watt         WIRE-WOUND RESISTORS         [ 3/2	S.W. 16 1 M.W. 200 L.W. 800
To watt > 20 oning-10,000 oning. 1/0	Short-Med
	Glass Dia
Knurled Slotted knob. Spindle High Grade. All	Lamps.
12,500 ohms50,000 ohms. 10 w	BRAM
30 K. to 2 Meg., 3/ CONTROL 10Ω, 3/ O/P TRANSFORMERS. Heavy duty 50 mA. 4/6. Multi-	
50 K. 10 2 meg., 3/2, 1 COMMON 1011, 3/2, 10/	MONA
1.1 HI20 mA., 12/6. 10H 150 mA., 14/	
MAINS TRANSFORMERS 200/250 v. A.C. STANDARD 250-0-250, 80 mA., 6.3 v. 3.5 a. tapped 4 v. 4 a. Restlifer 6.3 v. 1 a., tapped 5 v.	
tapped 4 v. 4 a. Restlifter 5.3 v. 1 a., tapped 5 v.         22/6           or 4 v. 2 a. Ditto Sol0-350.         22/6           MINDET. 200 v. 45 un A. 6.3 v. 1 a	0.011
SMALL, 200-0-200 50 mA., 6.3 v. 2 A 176 STANDARD, 250-0-260, 65 mA., 6.3 v. 3.5 a 176	SAVI
HEATER TRANS., 6.3 v. 1 a., 7/6; 3 amp 10/6 GENERAL PURPOSE LOW VOLTAGE. Outputs 3, 4, 5, 6, 8, 9, 10, 12, 15, 18, 24 and 30 v. at 2, 4 22/6	POUN
AUTO, TRANS. 150w, 0, 10, 120, 200, 230, 250v 22/6	FUUN
23in. sq. x 21in. or 1in. sq. x 11in., 2/- with cores.	
SLOW MOTION DRIVES. Epicyclic rath 6:1, 2/3. SOLON. Midget Soldering Iron, 220/40 v. 25 w., 24/-, REMPLOY MATRUMENT IRON. 220/40 v. 25 w., 17/6.	
MAINS DROPPERS. 3 × 14in. Adj. Bliders .3 amp. 1,000 ohms 4/32 amps. 4/31 amp. 2,000 ohms, 5/	BUIL
ALADDIN FORMERS and cores, in. 8d.; in. 10d. 0.3in. FORMERS 5937 or 8 and Cans TV1 or 2jin. sq. x 2jin. or 1in. sq. x 1jin. 2/2 with cores. SLOW MOTION DRIVES. Epicyclic ratio 6:1, 2/3. SOLON. Misses Boldering Iron, 220/40 v. 25 w., 24/-, REMPLOY MESTRUMENT IBON. 220/40 v. 25 w., 21/7 MAINS DROFPERS. 3 x 1jin. Ad. Bilders. 3 amp. 1,000 ohms 4/32 samps 4/31 amp. 2,000 ohms, 5/- LINE CORD3 amp. 60 shows per foot. 2 amp. 100 ohms per foot, 2-way, 6d. per foot. 3 way 7d. per foot.	MONA
CRYSTAL MIKE INSERT by Acos 6/6 Precision engineered. Size only 1 × 10 m.	HAND
ACOS CRYSTAL STICK MIKE 39-L. Bargain 35/	HIGH
LOUDSPEAKERS PM. 3 OHM. 5m. Bela, 17/6. 6in. x 4in. Rola, 18/ 7in. x 4in. R.A., 21/	FULL
10in. x 6in. Rola 27/6.         8in. Plessey, 19/6.           61in. Rola 18/6.         8in. Rola, 21/           10in. R.A. 30/         10in. R.A. 30/           HI-FI TWERTERS. 4in. 25/         12in. Plessey, 30/	Carr
MIKE TRANSF.         50:1, 3/9 ca., 100:1         Potted 10/6- LOUDSPEAKERS PM.           LOUDSPEAKERS PM.         3 OHM.         6hn.         Bela., 17/6.           Gin x 4in. Rola, 12/         7in. x 4in. R.A., 21/         10in. R.A., 30/           Join x 6in. Rola 27/6.         8in. Pleasey, 19/6.         10in. R.A. 30/           HI-FI TWENFTERS, 4in. g5/-,         12in. Pleasey, 30/         12in. Pleasey, 30/           12in. Baker 15 wt. 3 ohm. and 15 ohm models, 90/         12in. Baker foam suspension 15 w. 15 ohm. £6.         12in. HIGH-FI Master £17/10, 20 c.p.s. to 17 k.c.s.	RECO
I.F. TRANSFORMERS 7/6 pair	1
1.F. TRANSFORMERS 7/6 pair 465 ke/s. slng tuning miniature can 11 × 1 × 1m. High Q and good bandwith. By Pye Radio. Data sheet supplied.	1 6
Weymouth I.F. Standard size 465 ko/s., 12/6 pair. CRYSTAL DIODE G.E.C., 2/-, GEX34, 4/-, 40 Circuits 3/-,	
H.R. HEADPHONES. 4.000 chms, brund new, 15/- pair. SWITCH CLEANER Finid, squirt spout. 4/3 tin.	6
x 11 in x 12 in, 10/0005 Standard with trimmers, 9/-; less trimmers 8/ Midget 7/6; Single 50 pf. 2/6;	The
100 pf., 150 pf., 5/6. Sund dielectric 100, 300, 500 pf., 3/6. VALVE MOLDERS. Pax. int. Oct. 4d. EF50, EA50, 6d. B12A. CRT. 1/3. Eng. and Amer. 4. 5, 6, 7 pin, 1/	
MOULDED Mazda or Int. Oct. 6d. B7G, B8A, B8G, B9A. 9d. B7G with can, 1/6; B12A, 1/3. B9A with can, 1/9. DEPAMORE FURD B7G B04 Oct 11 B7G B04 Con 1/	4 Speed Stereo o
SPEAKER FRET. Gold Cloth 17th. × 26th., 5/-; 25th. × 35in., 10/ Tygan 54in. wide, 10/- ft., 27in. wide 5/- ft.	Collar Garra
Brown, Green or Red. Samples S.A.E. WAVECHANGE SWITCHES 2 n. 2-way, or 3 n. 2-way: short spindle	4 speed or Mona
5 p. 4-way, 2 wafer, or 3 p. 11-w. 3 wafer, long spindle 6/8 2 p. 6-way, or 4 p. 2-way, or 4 p. 3-way. long spindle 3/6	Gar Gar
Weymouth I.F. Standard size 485 ke/s., 12/6 pair. GRYSTAL DIODE G.E.C., 2/-, GEX.34, 4/-, 40 Chronts 3/-, H.R. HEADPHONES. 4.006 ehms, breach new, 15/- pair. SWITCH CLEANER Frid, squirt spout. 4/3 tin. TWIN GANG CONDENERS. "See N. Ministure, 14in. x 14in. x 14in., 10/ 0066 Skandard with trimmers, 9/-; ieas trimmers 8/ Midges 7/61 Kingle 69 ft. 2/6; 100 pf., 150 pf., 5/6. Switd dielectric 100, 300, 500 pf., 3/6. VALVE HOLDERS. Tax. int. Oct. 4d. EFS6, EA50, 6d. B12A. CBT. 1/3. Eng. and Amer. 4, 5, 6, 7 pin. 1/ MOULDED Mazda or int. Oct. 6d. B7G, BSA, B86, B9A. 9d. B7G with can. 1/6; B12A, 1/3. BYA swith can. 1/9. GERAMIG. EF50, B7G, B9A, Oct. 1/ B7G, BSA Cans. 1/ BYEAKER FREET. Gold Cloth 17in. x 25in., 5/-; 25in. x 30in., 10/ Tygan 54in. with. 30/- 16. 27in. wide 5/- ft. BY 2-vay. of 3. p. 2-way. short spindle	Guilden
TOGGLE SWITCHES. S.P., 2/-; D.P., 3/6; D.P.D.T., 4/-; MORSE KEYS, good quality 2/6. SUB-MUNATURE ELECTROLYTICS (IS -) 1 2 4 5 9	Suitable Amplifier 2.valve an
25. 50 mfd., 100 mfd., 3/- each.	Ready mo
THE HI-GAIN BAND 3 PRE-AMP Cascode circuit using Valve ECC84. 17db	200-250 v output. I volume co
gain. Kit 29/6 less power; or 49/6 with power pack. Plans only 6d. Also Band I version same prices.	with de l 3 ohm, and
Also Band I version same prices. (PCC84 Valve if preferred)	chassis siz supplied.
	TAP



CYLDON THODET TELETINED
CYLDON TURRET TELETUNER LF. 33/38 mess, complete with trame-grid values. 30(1) and 301.15. (LT 16v. 3a.) With couls for channels 1 to 5 & 8 to 11. Latest model. Brand new price. 45-, oper- ating data & circuit supplied. Ideal for "F.T." Offsmile.
VOLUME CONTROLS 80 Ohm Coaxial
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Semi air spaced, Hn. dia. 1 year. All values. 5 K. ohms up to 2 Meg. No switch. D.P. Sw. 3/- Linear or Log Tracks. Semi air spaced, Hn. dia. Losses cut 50% FRINGE QUALITY AIRSPACED 1/- yd.
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AMERICAN MAGNETIC RECORDING TAPE FERRODYNAMICS ' BRAND FIVE'' 5in 600 feet
5in.         600         feet         16/-         MYLAR         DUPONT           5in.         900         feet         18/6         Super         High         Fidelity           5in.         12.00         feet         23/6         Double Piay           7in.         1.200         feet
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NEW ELECTROLYTICS. FAMOUS MAKES
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50(50 v 2)-(32+32/350v. 4/5) 100+200/278v.12/6 FULL WAVE BRIDGE SELENIUM RECTIFIERS. 2.6 or 12 v. 14 amp., 8/9; 2 a., 11/3; 4 a., 17/6; 6 a., 22/6.
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6AM6         5/-12A6.         7/6         ECL80         10/6         PEN25         .6/6           6BE6         7/6         12AT7         8/-         ECL82         10/6         PEA2         .10/6           6BE6         9/6         12AU7         8/-         ECL82         10/6         PEA2         .10/6           6BE6         9/6         12AU7         8/-         EF39         .5/6         PY80         .7/6
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JANUARY, 1961



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AVANTIC PL621 20-watt monaural Amplifier, frequency response 10 c/s-30 Kc/s. 1bB. L.S. Impedance, 4, 8 or 16 ohms. Dimensions 14in. x 84in. x 73in. Orlginal price 29 Gns. P. & P. 7/6. OUR PRICE 81 in. x 19 Gns.

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SP21. 6 inputs for each channel, bass, treble, volume control, on/off stereo/3D/reverse stereo switch, stereo phase switch, low pass filter. Power requirements 6.3 v. at 1.3 A., A.C. 350 v. at 5 mA. D.C. Dimen-sions 144 x 9 x 4in. Original price £28/10/-.

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Brief specifications: DL7/35. Power output 54 watts peak; L.S. impedance 4, 8 or 16 ohms. power inputs 105-250 v. Valve line-up GZ34, 2-EL34, ECC83, EF86. Dimensions 14½ x 9 x 8¼in. Original price 30 gns.

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AMERICAN C.B.S.

**RECORDING TAPE** 

31 l.p.s. W

\*Latest BSR Tape Deck, with inter-locking device to prevent accidental erasure.

Single speed 31in. per sec. ★ Playing time 5% in. std. tape-1% hours. L.P. tape-2 hrs. 8 mins.

★Volume on /off and tone control. ★Power output 3 watts.

Finput sockets for Microphone, Radio/ Gram.

\*Extension speaker socket. Size: 13} × 9} × 6in., weight 17 lb.

#### SINGLE PLAYERS

Collaro Junior 4-speed rlayer construction Pick-up. 23 15 0 Garrard 48P 4-speed Player, complete with Pick-up and automatic stop. 26 19 8 Garrard TA Mk. 2, 4-speed Player, wired for sterco, with plug-in Head. 28 10 0 Philips AG2009, 4-speed Player, with diceasi turntable and Microlift, wired for stereo £10 10 0 31 I.p.s. Will take 5fin spools. 25/19/6. F. & P. 5/. COLLARO STUDIO TAPE TRANSCRIPTOR. S speeds 14, 33, 71 I.p.s. 3 motors. Fush-button controls. Will take 7in. spools. 212/19/6. P. & P. 7/6. COLLARO MK. 4 TAPE TRANSCRIPTOR. COLLARO MK. 4 TAPE TRANSCRIPTOR. Twin track operation, 3 speeds, 37, 74, 15 I.p.s. Will take 7in. spools. £17/19/6. TAPE RECORDER AMPLIFIER specially designed to match the Collaro Studio Tape Deck. £10/18/6. P. & P. 4/. Size 114 × S × 3m., uses 3 valves, magic eye, contact cooled metal rectifier. Incorporates mike/ pram/radio inpute, ext. i.s. jack, superim-posing switch, with matching knobs.

P. & P. 3/6 on above units.

### **RECORD CHANGERS**

BSR UA12, 4-speed, wired for stereo a complete with Stereo cartridge 28 19 Collaro Conquest, 4-speed Chauger 6 and 6

27 19 Collaro RC457, latest type 4-speed change 6

0666

Garrard EC111 3-speed Changer 25 10 ( Garrard EC121 Ms. 2, 4-speed 25 19 ( Garrard EC122 Mk. 2, 4-speed 35 19 ( Garrard EC121/4D, 4-speed ... 29 19 ( Garrard RC121 Mk. 2, 4-speed, whred for steree and with plug-in Head 210 19 ( P. & P. 5/- on above units. 6

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 Brand new, fully guaranteed and with Leader Tape:- No

 600ft. on 5h. Spool
 17/6

 1,200ft. on 5fin. Spool
 25/ 

 1,800ft. on 7fin. Spool
 25/ 

 1,900ft. Spool 
 Garrard 301
 £22
 7

 Garrard 301 (Strobe turntable)
 £23
 18

 Garrard 4HF (Stereo)
 £19
 4

 Garrard 4HF (GC3)
 £18
 9

 P. & P. 7/6 on above units.
 218
 9
 3480

### TAPE RECORDER RADIO JACK

May be bu it for 29/6 plus 1/6 p. and p. Tape Recorder Plug Extra. Improve the quality of your recordings with the most inexpensive Radio Jack available, suitable for any type of Tape Recorder, only a short external Aerial required for full medium waveband coverage. Phono Plugs-9d., Jack Plugs-3/-,

**INSTANT BULK TAPE ERASURE** 

Erase complete Reels of Tape in a matter of seconds. PRICE 27/6 post paid.



### MODEL 379 AM/FM RADIO-**GRAM CHASSIS BY FAMOUS** MANUFACTURER PRICE £12.12.0 plus 7/6 p. & p.



Price £32.0.0 Good quality Crystal Microphone with Lead and Jack Plug mited, 51" Reel of Standard Tape and Spare Reel, spare Lead fitted with Jack Plug and Wander Plugs for recording from Radio.



£2.19.6



### Plus 21/- P. &

Collaro 3-speed Tape Deck.

The Magnaphon

- Separate Input for Microphone and Gram Recording.
- \* Separate Volume Controls for recording. \* Volume On/Off and Tone Control for
- replay. ± 3 watts output.
- \* Housed in smart two-tone Blue/Beige Cabinet with detachable Lid.

Œ



Plus 2/- P. & P.

BABY ALARM Battery extra 2/9 79/6 Plus 3/- P. & P. (Ever-Ready PP1 6 volt or equiv.)

(Ever-Ready PPI 6 volt or equiv.) The answer to the modern Parents' problem for "Baby Sitting," this extremely efficient Unit is completely safe of the house. Extra Microphones may be used in different rooms without impairing the efficiency of the Unit. It is the most economical Unit of its kind and will run on one Battery for approximately two months of continuous day and night use. It is housed in an attractive bakelite Cabinet in either twory or pastel blue. The price includes one Microphone, extra Microphones can be supplied at 12/6 and Microphone Lead at 5d.





A truly High-Fidelity Ultralinear Amplifier with a push-pull output of 10 watts and incorporating negative feedback. Provision for Tuner, also bass and treble control and 6-position selector switch for Microphone, Radio Tape and L.P. and Standard Recordings. Finished in an attractive greygreen store enamel. FOR A LIMITED PERIOD ONLY £14/19/6

140



OPERATING MICRO PYË LEVER SWITCHES. Single pole change over. Brand new. 4/- each or 42/- dozen, p. paid.

TANNOY P.A. LOUDSPEAKER. For out-door use, metal exponential horn with 20in, square flare. Overall length 30in. Speech coil 15 ohms. Guaranteed in working order and good condition. Price £7/10/-. Carriage 10/-.

PACKARD BELL BRAND NEW RELAYS. 2 pole c.o. 6 volt 80 ohms. 7/6 each. P. & P. 6d.



DIAL THERMO-METER. Made by Short & Mason, Calibrated 0-160 degrees Fahrenheit. 4 dial. 6in. rim 41in. n for dial. 6in. rim for flush mounting with 6in. long rod pro-truding at the back. Brand new. Manu-facturer's packing. Price 22/6. P. & P. 3/-

SOLAR OIL-FILLED CON-DENSER. 240 mfd. for 230 V.A.C. or 600 volt D.C. Overall size 14in, x 9in. x 5½in. plus feet. Weight 46 lb. Brand new. Guaranteed new. Guaranteed perfect. Manufacturer's packing Price £7/10/-, car packing.

100 YARD DRUMS GLASS BRAIDED FLEX, 10/.010. New. 10/6 per coil. P. & P. 2/-

18-WAY P.V.C. COVERED 14/36 WIRE, screened overall, covered with P.V.C., all colour coded, 3/6 per yd.; £15 reel of 100 yds. Carriage paid.

A.R.B. U.S.A. RECEIVERS. 24 volt. Covering 195-9,050 k/c. in 4 bands. As new, suitable for use on boats, etc. Price £6 including carriage.

NEW UNCHARGED UNCHARGED UNCHARGED UNCHARGED UNFILLED 12 9 ampere

unspillable plastic cases. Comprises 6 x 2 v. separate cells connected by terminal strips. 6 x  $5\frac{1}{2}$  x  $4\frac{1}{4}$  in. over terminals. Price 19/-, plus P. & P. 2/9.

Price £7/1 riage 10/-.



245 AMP. 2 VOLT ACCUMULATOR Admiralty type in wooden casing. Size I5 x  $7\frac{1}{2} \times 7\frac{1}{2}$ in. Weight 601b. Unfilled, uncharged. Naw. Price £4. Carriage 10/-.



MINIATURE P.M. MOTOR. 12/24 volt, re-12/24 volt, re-versible. I≩in. dia. New. Price 10/6 each. P. & P. 1/-.

### AIRCRAFT CINE CAMERA G45B Mk. III.

00 incare of per sec. Brand me

Fully modified, fitted with f/3.5 triple anastigmat lens, takes 25ft. of 16 mm. film, fitted with 24 v. motor. 16 exposures

Brand new, original packing, £4/10/-





EX OSITION DICATOR.

. 1/6.

R.A.F. AIR con-





latest type, sealed. H96E, 1,700 ohms plus 1,700 ohms, single C.O. contacts. Brand new with fixing clip. In maker's cartons. Price 16/6 each, plus 1/- P. & P.

Siemens sealed similar relay to above, but 2.2 ohms plus 2.2 above, but 2.2 ohms plus 2.2 ohms. Minus clips, 12/6 each. Plus 1/- P. & P.

SUPERIOR BRAND NEW RELAY. 7.000 SUPERIOR BRAND NEW RELAY. 7,000 ohms coil. Will pull in at 750 microamp. and out at 450 microamp. Change-over, platinum contacts. Vacuum sealed, will therefore not be affected by oil, moisture or water and never needs adjusting. Weight  $2\frac{1}{2}$  oz. Price 18/6. needs adjusting. P. & P. 1/-.



p. way, or, Two contact current 100 mA at 50 V A.C. or

Size 1 x 3 x 3 in. Price 22/6 each.

G.E.C. SEALED RELAY. Type M.1090. 180 ohms coil. 6/12 volt. 4 C/O. Brand new. 18/-. P. & P. 1/-.

G.E.C. SEALED RELAY. Type M.1092. 670 ohms coil. 12/24 volt. 4 C/O. Ex new equipment. Unused. 10/-. P. & P. 1/-.

G.P.O. 600 TYPE RELAY. 400 ohms coil. 24 volt. 2 C/O plus 2 M. New 7/6. P. & P. 1/-.

MINIATURE OP5N TYPE RELAY. 700 ohms coils. 24 volt. 2 C/O. Ex new equip-ment. Unused. 7/6. P. & P. 1/-.

ROTARY RELAY. 12 volt. Heavy duty change-over contacts and one low current for external circuit, plus one break set. Price 7/6.



MINIATURE UNISELECTOR SWITCH. Two banks of ten plus home contacts one bank continuous of normal. 30 ohm coil for 24 volt operation. Brand new, manufac-turer's packing. Price 22/6 each. P. & P. 2/6. As illustrated.



CLASS D WAVE METER. Latest release of these famous Hetrodyne wave meters with directly calibrated illuminated dial, most suitable for amateur transmitters, covers two ranges 1.9-8.0 Mc/s, and 4.0-8.0 Mc/s. Complete with reference crystals for zero settings, two valves,  $2 \times 6$  volt

and 4.0-8.0 Mc/s. Complete with reference crystals for vibrators, MAKER's instruction book and matched set of headphones for monitoring. Designed for 6-volt D.C. operation, can easily be modified for mains and suitable transformer supplied for 7/6. In spot-on condition as tested by R.E.M.E. In transit case. Price 5 gns. each, plus 6/6 carriage.



WIRELESS WORLD

SOUND POWER TELEPHONE HAND. SETS. Each couple connected by ordinary 2 core lighting flex will secure instant and reliable intercommunication. No batteries required. Price per set of 233/-, plus P.&P. 3/-.

**ENGINE SPEED INDICATOR.** On the basis of a special ex-R.A.F. meter which we are able to supply and a few small linking parts which can be purchased anywhere, an inexpensive engine speed indicator can be made up which works on simple pulse counting principles in conjunction with the contact breaker on the distributor. Will give direct reading in R.P.M. Full conversion instructions are supplied by us. Additional standard parts required easily obtainable for about 15/-. R.A.F. meter as offered by us 16/6, plus 2/6 P. & P.

ONE ONLY, "CINTEL" DEMONSTRA-TION OSCILLOSCOPE. ISin. double beam. Condition as new. Guaranteed. Complete with full maker's instruction book. Price £85.



NEW IMPORTED EX-TREMELY EFFICIENT MOTOR with tremendous power weight ratio. For 12 volt D.C. but very efficient on 6 volt. Three position switch. Weight 2.1 oz., size Ifin. x Ifin. dia. Speed 7,000 r.p.m. Self lubricating. 15/-, plus I/- P. & P.



TYPE R.Q.R., reversible. 37 r.p.m. overall size Sin. x 4in. x 5½in. Weight 4½ lb. Ex brand new equipment. Un-used. Price £3/17/6. used. Price £3/17/6. P. & P. 3/-.



MAINS POWER Potted and sealed transformer and

choke by famous maker. Mounted on metal chassis 6½ x 7½in., complete with 5Z4 rectifier valve and full smoothing.

Input tapped 220–230–240 volts. Output: 300 V. D.C. at 100 mA. 6.3 V. A.C. at 4.5 amp. 6.3 V. A.C. at 4.5 amp. Rectifier supply 5 V. A.C. at 3 amp. Very conservatively rated. Price 47/6 plus P. & P. 6/6.

Postages and carriage shown above are inland only. For overseas please ask for quotation. ERVICE TRADING CO.



24 amp. D.C. M.I. 2in. fl. rnd. 5 amp. D.C. M.I. 24in. fl. rnd. 74 amp. D.C. MI.I. 34in. proj. rnd. 9 amp. D.C. Hot Wire W.R. 24in. fl. rnd. 15 amp. D.C. M.C. 2in. rnd. 30 amp. D.C. M.C. 2in. fl. sq. 100 amp. A.C. MI. 44in. fl. rnd.	7/6 11/6 12/6 6/6 10/6 12/6 32/6
Voltmeters           12 v. D.C. M.C. 2‡in. proj. rnd.           20 v. D.C. M.C. 2in. fl. sq.           25 v. D.C. M.C. 2in. fl. rnd.           30 v. M.I. 3in. proj. rnd.           40 v. M.C. 2in. fl. sq.           300 v. A.C. M.C. 2½in. fl. rnd.           300 v. A.C. M.C. 2½in. fl. rnd.           300 v. A.C. M.L. 2½in. fl. rnd.           300 v. A.C. M.I. 4½in. rnd.           90-180 v. A.C. M.I. 4½in. fl. iron	8/6 10/6 10/6 10/6 27/6 22/- 35/- 25/-
Milliammeters           I mA. M.C. 2½in. fl. rnd.           200 mA. M.C. 2½in. fl. rnd.           500 mA. M.C. 2½in. fl. rnd.           500 mA. M.C. 2½in. fl. rnd.	12/6
<ol> <li>microamp., scaled 0-100, M.C. 2±in. fl. rnd.</li> <li>scales</li> <li>microamp., M.C. 2in. rnd. F.L. scaled 15/600 volt. NEW</li> </ol>	42/6 35/ 16/6
Postage on all meters 1/- each.	

Miniature latest type mov-Ing coil 0-5 milliamp meter, Izin. diameter, flush fitting. complete with fixing clip. Price 17/6. P. & P. 1/-.



CRYSTAL CALIBRATOR No. 10.



crystal controlled 4-valve high-grade instrument in the same category as the famous B.C. 221. Directly calibrated, does not require cross reference or charts — functions cross charts — function a s follows:— (1) A controlled which

Meg. and up to 30 Meg. (2) A variable oscillator from 250 KC to 500 KC, this enables all intermediate frequencies between 250 Kc/s. and 30 Meg. to be produced

and modulated. Supplied complete with 3 spare valves, all leads and

maker's instruction book in carrying ersack. The complete outfit is brand newhaversack. repeat NEW. Price £4/19/6. Carr. 3/-.

PERSONAL CALLERS ONLY: 9 Little Newport Street, London, W.C.2 TEL: GER 0576 ALL MAIL ORDERS. ALSO CALLERS AT: 47-49 High Street, Kingston-on-Thames Telephone: KINgston 4585



For factory or outdoor use Tannoy 7.5 ohms 8 watts 25/9.

23.93. Parmelo horn type, highly efficient. Handles up to 10 watts. 15 ohm 200 ohm and 600 ohm mat.hing 59/8.  $\mathbf{K}$  C.A. 20 watt rating, 3 ohm, 15 ohm. 200 ohm. and 600 ohm matching **6 gms**.



Type BM2. Bize 8×0 \*×2 in. Supples 120 v. 90 v. and 60 v., 40 mA. and 2 v. 0.4 a. to 1 amp., (u) smoothed. THEREBY COMPLETELY REPLACING BOTH H.T. BATTERIES AND H.T. 2 v. ACCUMULATORS when connected to A.C. mains supply 200-250 v. 50 c/s. SUITABLE FOR ALL BATTERY RECEIVERS normally using 2 v accumulator

Complete kit with diagrams and instructions. 49/9 or ready for use 59/6.

TAPE RECORDERS AT WHOLESALE PRICES. Leading makes. 26 gn. model 18 gns. 29 gn. model 20 gns. 40 gn model 27 gns. 42 gn. model 28 gns. 63 gn. model 45 gns. 64 gn. model 44 gns. All brand new with manufacturer's guarantee. Make and model No. on request. No. H.P.



TWEETERS, 4in. Plessey, 3 ohms, 18/9, R.A. 15 ohms 25/9.

3.00

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JANUARY, 1961



Terms of business:--Cash with order or C.O.D. only. Post/Packing oharges 6d. per item. Orders over 23, poet free. C.O.D. 2/6 extra. Any parcel insured against damage in transit for 6d. extra. We are open for personal shoppers. Mou-FM. 82:0-5.30, Sats 8.30-1 p.m.

Latest catalogue of over 1,000 different valves, also metal rectifiers, volume controls, electrolytic condensers, transistore, germanium diodes, valve holders, and Hivac miniature valves, with full terms of business, price 64. All valves boxed, fully guaranteed, and new manufacturers' stock or covernment stores aurplus. First-grade goods only, no seconds or rejects. Please enquire for any type not listed. S.A.E. please

TANUARY, 1961



Quality Tape Recorder at this amazing reduced price. 7in. 1,200ft. Standard reels. Latest Studio 3-speed Deck, 14, 31, 74 I.P.S. Includes Twin Tracks, Reverse counter, Pause control and magic eye recording indicator. Volume and tone control, super-impose switch. 3 watts output. Attrac-tive design cabinet in beige. Size 19in. × 13in. × 8in, Ins. & Carr. 12/6.

EXTRAS: Microphone 27/6. Tape 25/-.

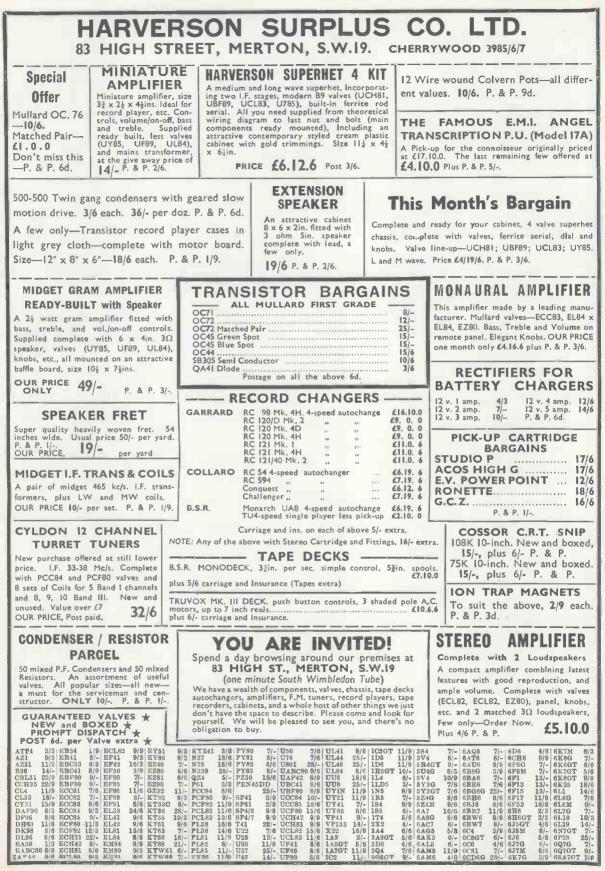
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# ANNOUNCING ... THE OFFER OF THE YEAR! HARVERSON'S SUPER STEREO KIT

The product of a world-renowned manufacturer, this stereo amplifier is composed of a number of compact "ready-built" units, only requiring interconnection. This system has the big advantage of being easily adaptable to fit any cabinet. Each unit is extremely well made from first-grade components, and all valves employed (ECL82, EZ80 range) are genuine Mullard. The comprehensive instructions supplied with each kit make the simple interconnection of units easy even for the novice.

### THE KIT COMPRISES . .

**TWO MIDGET AMPLIFIERS** each capable of 3W output. The reproduction is good, enabling you to get the best from both your stereo or monaural recordings. Both amplifiers are complete with well-designed output transformers providing perfect matching to standard 3-7 $\Omega$  loudspeakers, and have remote bass, treble and volume controls. Size 5in. x  $2\frac{1}{2}$ in. x 3in. high (each amplifier).

**CONTROL UNIT**, this is a flying panel fitted with three 2-gang potentiometers, enabling the bass, treble and volume controls of each amplifier to be positioned in the most convenient place in your layout. These dual controls are equipped with attractive cream and gold knobs and an escutcheon is provided for the complete panel.

**SEPARATE POWER PACK** complete with valve rectifier, although of midget size (5in. x 2in. x  $3\frac{1}{4}$ in. high), provides power for complete amplifier equipment.

**ISOLATED MAINS TRANSFORMER** of robust construction is a separate unit and may be mounted independently.

**VOLTAGE SELECTION PANEL.** Consisting of a panel fitted with the "valve base" type of mains input selector and a channel output socket.

**ONE LOUDSPEAKER**, a good quality 5-inch speaker, specially selected for this equipment. (Note: The second speaker may be purchased from us for an additional 14/6.)

**CREAM DOUBLE PUSH BUTTON SWITCH** of attractive design gives positive on/off switching action.

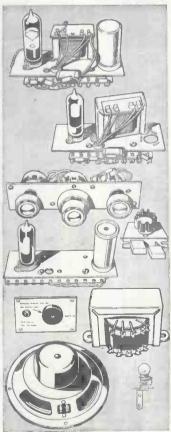
**INDICATOR LIGHT.** This pilot light provides visual indication that the equipment is operating, and is complete with an attractive gold-finished escutcheon.

This kit, which is complete in every way, is exclusive to HARVERSON'S, who are proud to present it at the amazing price of ... ... ... ...

PLUS -6/6 POST, PACKING & INSURANCE

FOR MORE BARGAINS SEE OUR OTHER ADVERTISEMENT

## HARVERSON SURPLUS CO. LTD. 83 HIGH STREET, MERTON, S.W.19 CHErrywood 3985/6/7







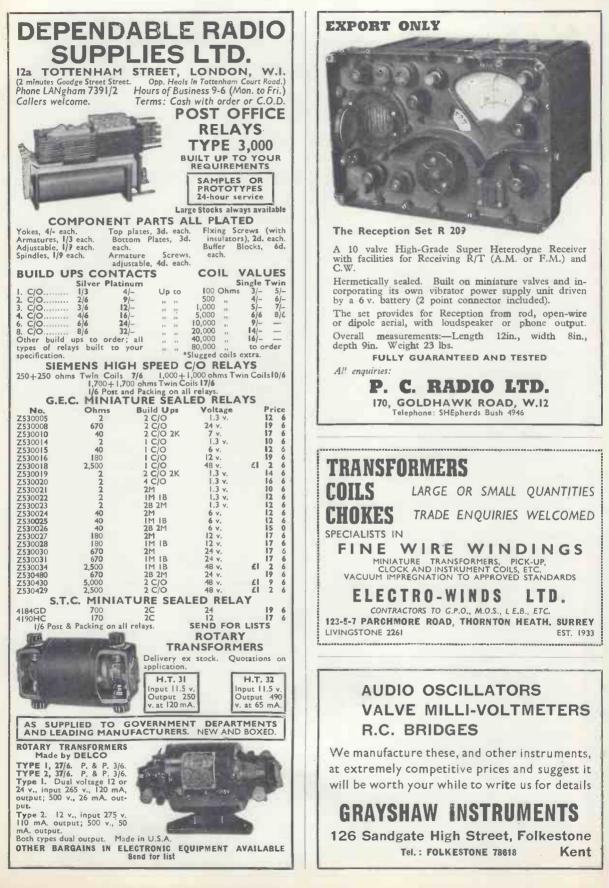


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154

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**IANUARY**, 1961

### WIRELESS WORLD

155



S.T.C. SELENIUM METAL RECTIFIERS. F/B, FOR BATTERY CHARGERS, ETC. 6 or 12 v. 1 amp. 5/-; 24 v. 1 amp. 10/-; 12 v. 2 amp. 7/6; 24 v. 2 amp. 15/-; 12 v. 2<sup>1</sup> amp. 12/6; 24 v. 3 amp. 23/-; 12 v. 4 amp. 15/-; 24 v. 4 amp. 30/-; 12 v. 4 amp. 15/-; 12 v. 6 amp. 20/-; 12 v. 10 amp. 35/-; v. 4 amp. 24 24 v. 6 amp. 32/6; 24 v. 10 amp. 70/-;

**NEW AND UNUSED ACCUMULATORS** 

12 v. 25 A.H. (as illus.) 45/-. Carr. 7/6. (Ideal for use with our Amplifier in centre column). 2 v. 100 A.H. 75 actual (ex-Govt.) with carrying handle. Size  $6\frac{1}{2} \times 6\frac{1}{4} \times 3\frac{1}{2}$ in., 15/- each. Carr. 3/6.

2 v. 16 A.H., as above. 7½ x 4 x 2in., 5/- each. P. & P. 2/-.

6 for 24/-. P. & P. 10/-.

2 v. 14 A.H., as above (less handle). 7 x 21 x 21 in., 5/- each. P. & P. 2/-.

### TRANSFORMERS POTTED C CORE

Pri.: 230 v. 50 c/s. Sec.: 450-0-450 v. 220 mA. 5 v. 3 amps., 6.3 v. 5 amps., 6.3 v. 3 amps. £2/10/-. Pri.: 230 v. 50 c/s. E.S. Sec.: 500-0-500 v. 500 mA. 6.3 v., 500 mA. 6.3 v. 5 amps., 5 v. 6 amps. £3/10/-.

6 amps, £3/10/-. Pri.: 230 v. 50 c/s. Sec.: 6.8 v. 5 amps., 6.3 v. I amp., 6.3 v. 3 amps., 6.3 v. 1.5 amps., 6.3 v. 2 amps., 6.3 v. 3 amps., 6.3 v. 4 amps. £1/12/6. Carr. 5/- each item. MAINS ISOLATING TRANSFORMER

Gresham. Pri, 230/250 v. Secs. 240-0240 v. 1.5 amps., 5 v. 12.5 amps. Potted. Size 7in. x 74in. x 104in. Weight 50 lb. Ideal for obtaining TWO ISOLATED 240 v. lines at 360 watts each. Perfect condition. 80/-. Carr. 10/-.

Perfect condition. 80/-. Carr. 10/-. L.T. TRANSFORMERS for Battery Chargers etc. All Pri. 200/250 v. Tapped 50 cycles. Type 048B. Sec. 24, 30, 36 v. 6 amps. 4x4x4in. £2/9/6. Type 066A. Sec. 18, 24, 30, 36 v. 8 amps. 4x 4 x 5in. £3/19/6. Type 053A. Sec. 12, 24, 30 v. 10 amps. 4x5 x 5in. £4/4/-. Carr. 3/6 each

item AUTO TRANSFORMERS. 0-110, 205, 225, 245 v. Fully shrouded. Terminal block

225, 245 v. Fully shrouded. connectors.

Connectors. Type 063A. 500 w., 4 x 5 x 5in. £3/7/6. Carr. 3/6. Type 064A. 750 w. 4 x 6 x 5in. £3/17/6. Carr. 3/6. Type 065A. 1000 w. 4 x 7 x 5in. £4/17/6. Carr. 5/-. 3 kV/A. AUTO TRANSFORMER. 250/110 v. 50 cycles (fully tapped primary and secondary). Capable of 25% over actual rating. Brand new and unused, £12/10/-. Carr. 20/-. Also 6 kV/A. as above. £18. Carr. 20/-.

20 kV/A. AUTO-TRANSFORMER. 230/115 v. 50-60 cycles. by Jefferies Transformer Co., U.S.A. Perfect condition, £20. Carr. 20/-.

U.S.A. Perfect condition, £20. Carr. 20/-CONSTANT VOLTAGE TRANSFORMER 190-260 v, primary, sec. 115 v. at 14 kV/A. (listed at 2 kV/A). Brand new and unused. £25 or £45 per pair. Carr. 10/- each. E.H.T. TRANSFORMER. 8,000-0-8,000 at 400 mA. Primary 230 v. 50 cycles. Oil filled. New and in original crates. £25. Carr. 10/-. E.H.T. TRANSFORMER. 1,800-0-1,800 at 1 kV/A. 230 v 50 cycles primary. Fully troplcal-ised. New and boxed. £6/10/-. Carr. 10/-. E.H.T. TRANSFORMER. 1,100-0-1,100 at 250 mA, plus 4 v. LT. Pri. 200/250 v. at 50 cycles. £5. Carr. 10/-.

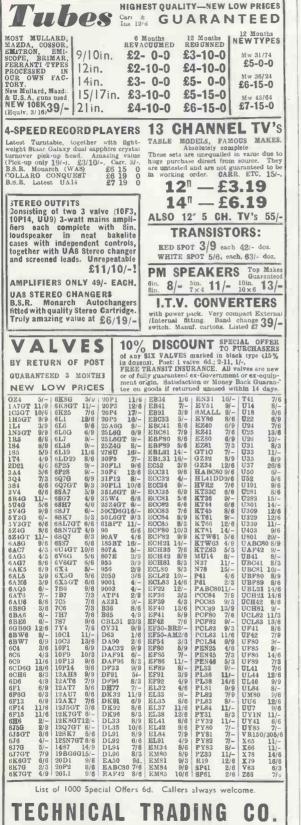
HEAVY DUTY REGULATING RESISTOR. 0.25 ohm. 200 amps. Wheel control. £4/15/-. Carr. 10/-.

CONDENSER, oil filled. 240 mfd. 230 v. A.C. or 600 v. D.C. Made in U.S.A. Size 21in. x 5≵in. x 9in. Brand new in original cases. £7/10/-. Carr. 5/-.

ROTARY CONVERTER. 24 v. D.C. input. 230 v. A.C. output at 250 watts. Complete with starting switch. New and unused. £15. Carr. 7/6.

Carr. 7/6. ROTARY CONVERTER. 24 v. D.C. to 230 v. A.C. 50 cycles, 150 watts. Brand new and unused. £8/10/-. Carr. 7/6. Ditto, 100 watts £6/9/6. Carr. 7/6. ROTARY CONVERTER. Ex-Govt. 12 v. D.C. input, 230 v. A.C. output 50 cycles at 135 watts. Complete in carrying case with lid. Voltage control, sliding resistance, mains switch and 0-300 v. A.C. flush meter. In good condition. £10. Carr. 10/-.

Motor only, without case, etc. Brand new and unused. £8/10/-. Carr. 5/-.



350-352 FRATTON ROAD, PORTSMOUTH POST: 2 lbs. 1/6, 4 lbs. 2/-, 7 lbs. 2/9, 15 lbs. 3/6. No C.O.D. JANUARY, 1961

T.O.C. "CATHODEAY "VISCONAL TYPES. 1 mfd., 2 kV. wkg., 7/6 each. 0.25µF., 4 kV. wkg. 6/- each. 0.05µF., 8 kV. wkg., 7/6 each. 0.1µF., 5 kV. wkg. 6/6 each. 0.03µF., 5 kV. wkg., 6/6 each. 0.1µF., 6 kV. wkg., 7/6 each. 0.3µF. 2.3 kV. wkg., 6/6 each. 0.25µF., 2.5 kV. wkg., 6/- each. 0.0025µF., 6 kV. wkg. 5. - each. 0.0025µF., 8 kV. wkg., 4/6 each. 0.0425µF., 5 kV. wkg. 3 kV. wkg. 4/- each. 0.025µF., 2.5 kV. wkg. 4/6 each. 0.0025µF., 3 kV. wkg. 4/- each. 0.005µF., 2.5 kV. wkg. 4/6 each. 0.0025µF., 3 kV. wkg. 4/- each. 0.005µF., 2.5 kV. wkg. 4/6 each. 0.0025µF., 3 kV. wkg. 4/6 each. 4/1 the above are tubular and mounting.

BLOCK PAPER TYPES, 10 mfd., 1,600 v. wkg. 15/- each, post 3/6. 3 mfd. 1,200 v. wkg. 11/6 each, 8 mfd. 300 v. wkg. 5/- each. 6 mfd. 500 v. wkg. 5/6 each. 4 mfd. 500 and 730 v. wkg. 4/6 each. 4 mfd. 1 kV. 5/6 each. 4 mfd. 1 kV. 5/6 each.





#### POWER UNITS

100-250 v. A.C. input, 24 v. D.C. at 3 amps, or 12 v. twice at 3 amps. each winding. Continuous tropical rating switched and fused, etc. In metal case that fits 19in. rack, size 19  $\times$  7  $\times$  7in. Brand new £3/15/-. Carr. 7/6 (with circuit).

#### SMOOTHING UNIT

for the above power supply. 2 chokes and 0-1 mA meter (grade 1), metal case, same as the p.u. £2. Carr. 7/6.

### **RANGE CONVERTOR**

large dial with a Muirhead slow motion drive. Valves EP39, ARTH2, the set can be used with R107, R208 and many other types of receivers.

115 v. A.C., 1/6th H.P., variable speed box 0-166. Size of unit  $14\frac{1}{2} \times 9\frac{1}{2} \times 8in$ . £8/10/-. Carr. 10/-.

INDICATOR UNIT Type 1-152-c (U.S.A.) 3in. tube 3DP1, 1 rectifier  $2\times2$  and  $3\times$  6AG3, with controls, etc. In a neat metal box  $11\times6\times6\,{\rm jm.},\,50/{\rm -}$  each. Post 2/6. ROTAX CONVERTORS Type 8A, 24 v. D.C. input, 115 v. A.C. at 1.8 amps. 400 c. 3-phase. Just the job for the laboratory or experimenting. £6/10/- each. Carr. 7/6

U.S.A. R-B/APN-4 RECEIVERS, designed for RDF. Valves 1-68N7, 1-46K7, I-VE105, I-5U4G, 1-58L7, 3-684G, 2-879/2, 1-68J7, all GT type. Complete unit with transformers, condensers, etc. As new, first-class condition. Price 65/- each. Packing and carriage 7/6.

MODDLATOR UNITS. MD 7/ARC5. 2×1625, 12J5. VR150, Modulation trans-former. 5 Kelays, etc. 32/6 each. Post 3/6. MOVING IRON METERS. 0-100 amps. 6in. scale, at £2; 90-180 v., 4in, scale at

35/-Post 3/

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No. 13. Pri, 230 v. Sec. 10 v. C.T. 10 amps. and 4 v. 7 amps., 32/6. P.P. 3/6. No. 14. Pri, 200-240 v. Sec. tapped 30, 32, 34,

No. 15. Pri. 200-240 v. Sec. tapped 10, 17, 18 v. 10 amps., 57/6. Carr. 4/-.

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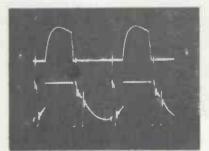
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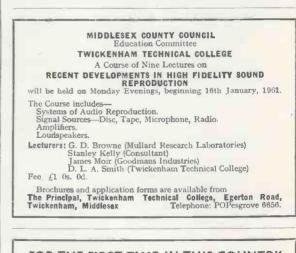
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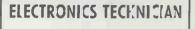
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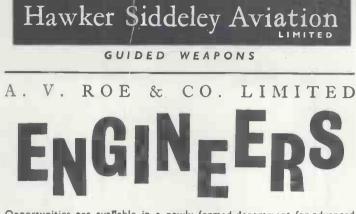
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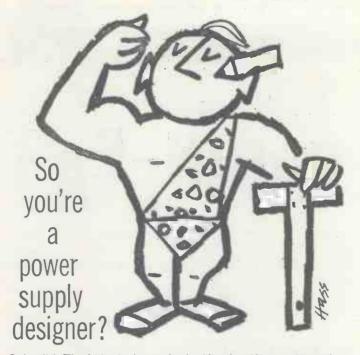
Chief Engineer, Alpha Television Services (B'ham) Ltd., Aston Road North, Birmingham, 6

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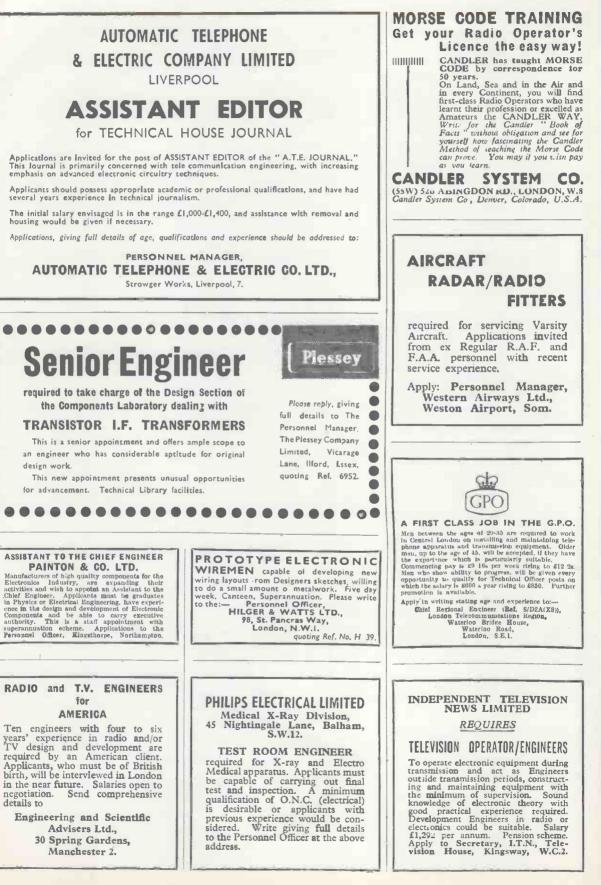
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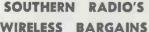


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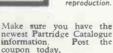
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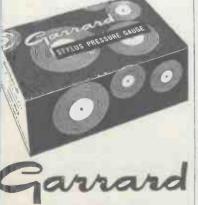
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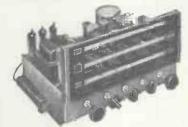
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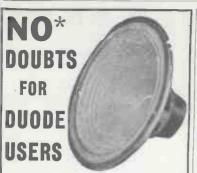
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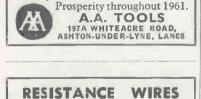


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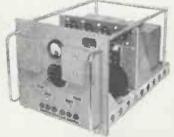
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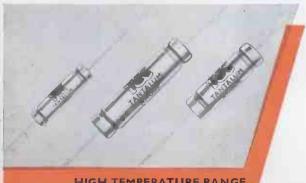
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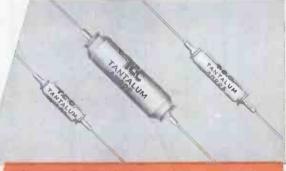
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MINIATURE RANGE



# STANDARD RANGE

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