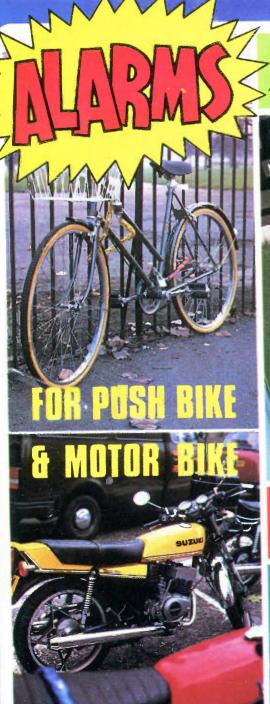
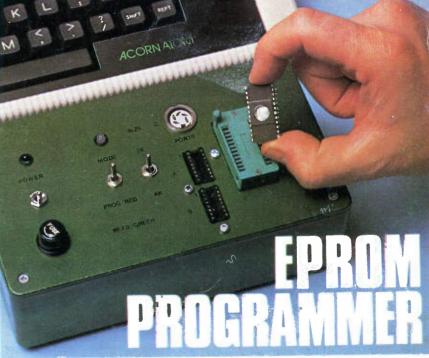
Easy to build projects for everyone

Everyday FEB. 83 80p ELECTRONICS



TEMPERATURE METER



ZX81 SPEED COMPUTING SYSTEM

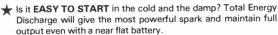


INTERVAL TIMER

electronize

ELECTRONIC IGNITION KITS OR READY BUILT





- ★ Is it ECONOMICAL or does it "go off" between services as the ignition performance deteriorates? Total Energy Discharge gives much more output and maintains it from service to
- ★ Has it PEAK PERFORMANCE or is it flat at high and low revs. where the ignition output is marginal? Total Energy Discharge gives a more powerful spark from idle to the engines max. (even with & cylinders).
- ★ Is the PERFORMANCE SMOOTH. The more powerful spark of Total Energy Discharge eliminates the 'near misfires' whilst an electronic filter smooths out the effects of contact bounce
- ★ Do the PLUGS and POINTS always need changing to bring the engine back to its best. Total Energy Discharge eliminates contact arcing and erosion by removing the heavy electrical load. The timing stays "spot on" and the contact condition doesn't affect the performance either. Larger plug gaps can be used, even wet or badly fouled plugs can be fired with this

Most NEW CARS already have ELECTRONIC IGNITION. Update YOUR CAR with the most powerful system on the market - 31/2 times more spark power than inductive systems -31/2 times the spark power of ordinary capacitive systems, 3 times the spark duration.

Total Energy Discharge also features: EASY FITTING, STANDARD/ELECTRONIC CHANGEOVER SWITCH, LED STATIC TIMING LIGHT, LOW RADIO

SPARK POLARITY and INTERFERENCE. CORRECT DESIGNED IN RELIABILITY.

★ IN KIT FORM it provides a top performance system at less than half the price of competing ready built units. The kit includes: pre-drilled fibreglass PCB, pre-wound and varnished ferrite transformer, high quality 2uF discharge capacitor, case, easy to follow instructions, solder and everything needed to build and fit to your car. All you need is a soldering iron and a few basic tools.

FITS ALL NEGATIVE EARTH VEHICLES

6 or 12 volt, with or without ballast.

OPERATES ALL VOLTAGE IMPULSE TACHOMETERS: (Older current impulse types need an adaptor).

STANDARD CAR KIT £15.90 Assembled and Tested £26.70

TWIN OUTPUT KIT £24.55 For Motor Cycles and Cars with twin ignition systems

Assembled and Tested £36.45

Prices include VAT

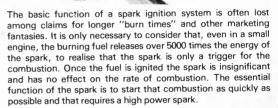
PLUS

P. & P. £1 (U.K.)



ELECTRONIZE DESIGN

Dept. C · Magnus Rd · Wilnecote Tamworth B77 5BY tel: 0827 281000



The traditional capacitive discharge system has this high power spark but, due to it's very short spark duration and consequential low spark energy, is incompatible with the weak air/fuel mixtures used in modern cars. Because of this most manufacturers have abandoned capacitive discharge in favour of the cheaper inductive system with it's low power but very long duration spark which guarantees that sooner or later the fuel will ignite. However, a spark lasting 2000µS at 2000 rev/min, spans 24 degrees and 'later' could mean the actual fuel ignition point is retarded by this amount.

The solution is a very high power, medium duration, spark generated by the TOTAL ENERGY DISCHARGE system. This gives ignition of the weakest mixtures with the minimum of timing delay and variation for a smooth efficient engine.

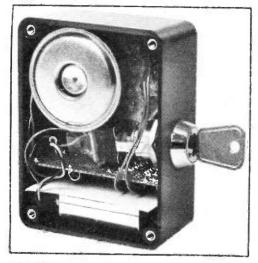
- * SUPER POWER DISCHARGE CIRCUIT A brand new technique prevents energy being reflected back to the storage capacitor, giving 31/2 times the spark energy and 3 times the spark duration of ordinar, C.D. systems, generating a spark powerful enough to cause rapid ignition of ever the weakest fuel mixtures without the ignition delay associated with lower power 'long burn' inductive systems.
- ★ HIGH EFFICIENCY INVERTER A high power, regulated inverter provides a 370 volt energy source - powerful enough to store twice the energy of other designs and regulated to provide sufficient output even with a battery down to 4 volts.
- ★ PRECISION SPARK TIMING CIRCUIT This circuit removes all unwanted signals caused by contact volt drop, contact shuffle, contact bounce, and external transients which, in many designs, can cause timing errors or damaging un-timed sparks. Only at the correct and precise contact opening is a spark produced. Contact wear is almost eliminated by reducing the contact breaker current to a low level - just sufficient to keep the contacts clean.

TYPICAL SPECIFICATION	Total Energy Discharge	Ordinary Capacitive Discharge
SPARK POWER (Peak)	140W	90W
SPARK ENERGY	36mJ	10mJ
STORED ENERGY	135mJ	65mJ
SPARK DURATION	500µS	160µS
OUTPUT VOLTAGE (Load 50pF, equivalent to clean plugs)	38kV	26kV
OUTPUT VOLTAGE (Load 50pF +500k, equivalent to dirty plugs) VOLTAGE RISE TIME TO 20kV	26kV	17kV
(Load 50pF)	25µS	30µS

TOTAL ENERGY DISCHARGE should not be confused with low power inductive systems or hybrid so called reactive systems.

VOL. 12 NO. 2 FEBRUARY 1983

PROJECTS ... THEORY ... NEWS ... COMMENT . . . POPULAR FEATURES . . .







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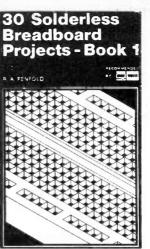
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Audio Amplifiers
Light & Dark Activated Switches
& Alarms & Alam Timers Metronome Metronome Oscillators & Tone Generators Warbling Door Buzzer Two-tone Train Horn Touch Switch Reaction Game Sound Activated Switch

BRAND NEW VEROBLOC KIT!!!

Just published by Babani, Mr. R. A. Penfolds new book, "30 SOLDERLESS BREADBOARD PROJECTS" – this book-features 30 different projects for assembly on a Verobloc, and the kit contains all pairs necessary to make:

Sound Activated Switch
Radio Receivers
Fuzz Unit ... + lots more!!
The introduction shows all the different components and explains how to use the breadboard. The Verobloc layout is shown for every project together with the circuit diagram and an explanation of how it works. Ideal for beginners in electronics, but also suitable for more advanced students.

beginners in electronics, but also suitable for more advanced students. The complete kit is contained in an attractive plastic case, which can be divided up into 15 compartments in which your components may be stored. Complete Kit, including book, Verobloc & all parts £24.95; Book only £2.25; Kit without Verobloc £20.45.



REDUCED TO £3.95

ELECTRO-DIAL

ELECTRO-DIAL

Electrical combination lock-for maximum security-pick proof. I million combinations!! Dial is turned to the right to one number, left a second number, then right again to a third number. Only when this has been completed in the correct sequence will the electrical contacts close. These can be used to operate a relay or solenoid. Overall dia. 65mm×60mm deep. Only £3.95.

STABILIZED PSU PANEL

A199 A versatile stabilized power supply with both voltage (0-30V) and current (20mA-2A) fully variable. Many uses inc. bench PSU, Ni-cad charger, gen, purpose testing. Panel ready built, tested and calibrated. 47.75. Suitable transformer and pots, £6.00. Full data supplied.

HEAT SINKS

Redpoint 4W type drilled for $2 \times TO3$. Size $130 \times 100 \times 32$. £2.50.

OPTO SCOOP!!!

7 Seg Displays from 40p!!!

0.3" CA MS133A 50p; 4 for £1.60

0.6" CA Dual digit MS261 90p

0.6" CC 4 digit MS261 90p

0.6" CC 4 digit MS4631 £1.75

0.21" CA TIL302 55p; 4 for £2.00

All cupplied with bispert data All supplied with pin-out data.

AA NI-CADS - 10 for £9.95 Brand new nickel cadmium batteries by GE, standard 1.2V @ 450mA/H. Professional quality with solder tags both ends. Special price, £1.40 ea; 10 for £9.95; Box of 80 £65. Ni-cad Charger: Charges up to 4 AA, C or D cells + PP3. Only £7.95.

ROBOTICS

Motor drive + gearing, as used on "Big Trak" computerised tank. 2 motors with magnetically linked clutch. Supplied with data + circuits. €4.95.

SIMON GAME

is back again. Another supply of ready built PCB's for this flashing light/pulsating tone computerised game is now with us. Supplied tested and working with speaker & instructions £4.95.

REED RELAYS

Manufacturers rejects R — DIL and other PCB mounting types. SP, DP and 4P — make, break & c/o contacts. Not tested, may be only partially working or o/c etc. so very low price – pack of 10 assorted £1; 25 £2.00; 100 £7.00

IN4148 – BEST PRICE EVER Supplied in packs of 100, by Toshiba £2 per pack; 3 packs £5.50; 10 packs £15; 25 packs £32; 100 packs £115.

FERRIC CHLORIDE

New supplies just arrived – 250mg bags of granules, easily dissolved in 500ml of water. Only £1.15. Also abrasive polishing block 95p.

I 982/3 CATALOGUE
Bigger! Better!! Buy one!!!
Only 75p inc. post → Look what you get!!!
★ Vouchers worth 60p.
★ I st class reply paid envelope.
★ Wholesale list for bulk buyers.
★ Bargain List with hundreds of surplus lines.
★ Huge range of components.
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Sent free to schools, colleges, etc.

IN4007 1000V IA RECTS ii loo Motorola bandoliered-lowest ever price!! 100 £2.95; 300 £8.50; 1k £27; 3k £72; 10k £220.



LIE DETECTOR

Not a toy, this precision instrument was originally part of an "Open University" course, used naily part of an "Open University" course, used to measure a change in emotional balance, or as a lie detector. Full details of how to use it are given, and a circuit diagram. Supplied complete with probes, leads and conductive jelly. Needs 2 4½ batts. Overall size 155×100×100mm. Only £9.95—worth that for the case and meter alone!!

COMPONENT PACKS

K503 150 wirewound resistors from IW to 12W, with a good range of values. £1.75. K505 20 assorted potentiometers, all types including single, ganged, rotary and slider.

£1.70.
K514 100 silver mica caps from 5pF to a few thousand pF. Tolerances from 1% to 10%.
£2.00.
K520 Switch pack-20 different rocker, slide, rotary, toggle, push, micro, etc. Only £2.00.
Pack of disc ceramics, assorted values and voltages – 200 for £1.00.
K501 Approx 300 long leaded ½ & ½W carbon resistors – wide range of values. Only £1.00.

STARBIRD

Gives realistic engine sounds and flashing laser blasts – accelerating engine noise when module is pointed up, decelerating noise when pointed down. Press contact to see flash and hear blast of lasers shooting. PCB tested and working complete with speaker and batt clip. (needs PP3). PCB size 130 × 60mm. Only £2.95

1000 RESISTORS £2.50

We've just purchased another 5 million preformed resistors, and can make a similar offer to that made two years ago, at the same price!!! K523–1000 mixed to ½W 5% carbon film resistors, preformed for PCB mtg. Enormous range of preferred values 1000 for £2.50; 5000 £10; 20k £36.

SWITCH BARGAIN

Push-on, push-off "table lamp" type, rated 2A 250V ac. 10p ea, 15 for £1, 100 for £5.

443D MILLBROOK ROAD, SOUTHAMPTON SOI 0HX All prices include VAT - just add 50p post. Tel. (0703) 77250 i

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SUPERIOR QUALITY CARBON FILM RESISTORS, HI STAB LOW MOISE 4 W 100. 10 MOISE 2 So 10 MOISE 12 SO	TRANSISTORS 2N540 Probably tine largest retail N5416 variety in UK 2N5418 variety in UK 2N5448 variety in UK 2N5449 variety in ub 2N5450 variety in ub 2N545	359 8C212 109 8F437 1100 8C212A 129 8F438 11.54 8C212A 129 8F438 8C212A 139 8F438 199 8C212A 139 8F438 219 8C212A 139 8F318 230 8C213A 119 8F64 259 8C213A 119 8F64 259 8C213B 129 8F68 259 8C213B 129 8F68 259 8C213B 129 8F68	order by phone quoting Access, Visa or American Express no or by nominal mail
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VQ Board 1.50 Dp Board 3.25 Frack Cutter 1.15 Pin Insertor 1.59 10 Way 84p 100 Pins 50 124 Way 160 Verobico 3.70 0 Way 270p War Wrining 1.00 Class fuses Fin 6.50 120 Class fuses Fin 6.50 Fin 6.50 120 Class fuses Fin 6.50	2N4903 3.24 8C178 2N4904 2.75 8C178A 2N4905 3.25 8C178B 2N4906 3.42 8C179 2N4907 3.20 8C179A 2N4908 3.70 8C179A 2N4908 2.90 8C179A 2N4918 956 8C182	100	140 4 TIC106C 48p LM381N 7.40 VOLTAGE 4.894 39 74184 50 7418324 1.79 1.79 1.79 1.79 1.79 1.79 1.79 1.79
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2.20 mm slow 17p PCL85 2.20 Volr Reg Data 20 mm guick 10p PCL85 2.20 Volr Reg Data 20 mm guick 10p PCL85 2.20 PCL805	2N5190	10p 8F246A 39p 11P137 13p BF247A 54p 17P140 14p 8F247B 56p 11P145 10p 8F254 39p 14P147 12p 8F255 42p 11P3055 13p 8F256A 36p 11P3055 10p 8F256B 46p 11S43	TRIACS T
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4081	14
4082	14
4093	20
4510	46
4511	46
4511	120
4516	53
4518	40
4520	60
4543	120
4583	130
All above	
are NE	
shown in	pence

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709C5 709C14 723C14 741C8 741C8 741C14 748C8 1458C5 1458C14 7106 7107 555 7555 556 CA3080E CA3130E CA3130E CA3140E LM301AN LM308N LM308N LM348N LM348N LM348N LM388N LM388N

1		
14 14 20 46 46 120 53 40	LM382N LM3900N LM3914N LM3915N NE555V NE556A NE567N RC4151NB SO41E	11: 50 20 20 2 4: 10: 8 29
120		
130	S041P	12
130	S042E	36
rices	S042P	13
		11.2
and		31.0
ence		13.2
- 1	S566B	21
-	S576A	23
	\$576B	23
UE	S576C	23
	S576D	22
49	S1469	46
44	SAB0600	30
36	SAB3209	49
57	SAB3210	31
18	SAB3211	16
47	SAB3271	32
65	SAB4209	49
35	SAB8256C	
62		35.4

115	TAB1041W	/ 18
50	TBA120AS	. 6
200	TBA120U	7
200	TBA400	100
23	TBA800	7
45	TBA10S	
104	TBA820	7
80	TBB1458	- 6
290	TBB1458B	- 4
121	TCA105	12
364	TCA105B	10
138	TCA105G	14
11.23	TCA250A	18
31.00	TCA205K	20
13.28	TCA335A	6
214	TCA345A	10
235	TCA345W	17
225	TCA671	12

SAB8256C 35.40 SAD1024 15.50 SAJ131 238 SAJ141 224 SAJ205 810 SAS231W 260 SAS251 142 SAS580 196 SAS590 196 SDA5680A 15.55

115 50 200 200 200 23 245 104 290 102 103 290 121 138 121 138 214 235 235 225 246 225 246 225 246 225 246 226 226 226 226 226 226 226 226 226	TAB1041W, TRA1200 TRA4200 TRA4200 TRA1200 TRA400 TRA100 TR	62 75 100N 755 75 62 2 40 120 120 120 120 120 120 120 120 120 12
5.55 104 49 62 103 59 72	UAA170 UAA170L UAA180 UAA190 XR2206 ZN414 ZN424P	46 165 152 165 141 520 85 99
120	ZN425E	350

ı	CRYSTA	LS		
П	(in MHz)		4.433	1
	0.032768	102	4.915	1
i	0.100000	453	5.000	1
	1.000	453	5.026	1
ı	1.8432	320	6.000	1
1	2.000	268	6.144	1
1	2.4576	268	6.5536	1
1	3.2768	188	8.000	1
П	3.579	128	8.867	1
Н	4.000	102	10.000	1
1	4.194	128	18.432	1
ľ	756	IFR	DIODE	_

400mW/2.7-30V 1-3W/3.3-100V

20W/7.5-75V

JACK PLUGS, SKTS Chrome P1 53p 34p 42p plastic P2 32p 14p 20p STEREO 1"

| D - SERIES | Plug sk1 | hood | 9 | 78p 104p 920 | 15 | 106p 140p 95p 25 | 141p 205p 110p 37 | 199p 298p 122p | 10C Connectors

ΠA	LS		
z)		4.433	128
68	102	4.915	157
00	453	5.000	157
	453	5.026	128
	320	6.000	157
	268	6.144	157
	268	6.5536	128
	188	8.000	1.88
	128	8.867	128
	102	10.000	157
	128	18.432	188

CONNECTORS

14 42p 16 20p 18 20 22 52p 24 20u 28 40 14p TERM 42p Plastic

Very wide range of many

other types in stock

- see latest price list.

RF Chokes
1, 1.1, 4.7, 10, 47μH ea
42p, 100, 22μH 53p,
470μH, 1mH 73p.

CAPACITORS Polystryens. Siernens 5% Tolerance 160V 5, 7, 10, 12, 15, 18, 22, 27, 33, 39pf 15p; 47, 56, 68, 82, 100, 120, 150, 180, 220, 720, 330, 390, 470, 560, 680, 820pf, 1n, 1n2, 1n5, 1n8, 2n2, 2n7, 3n3, 3n9, 4n7, 10p; 5n6, 6n8, 8n2, 10n, 13p Ceramic Very small 1, 8, 2, 2, 2, 7 etc., up to 1n 5p each. In5, 2n2, 3n3, 4n7, 6n8, 5p, 10n, 22n, 6p; 33n, 47, 7p; 100n, 8p Polysster, Siernens Layer Type 7, 5mm lead spacing 100V 1n, 1n5, 2n2, 3n3m, 6p; 4n7, 6n8, 8n2, 10n 12n, 15n, 18n, 22n, 33n, 47n, 7p 56n, 68n, 7p; 82n, 100n, 9p 120n, 150n, 15p; 160n, 220n, 12p; 270n, 330n, 330n, 390n, 470n, 15p; 550n, 680n, 24p; 10mm spacing 1uf 25p; 15mm spacing 2u2 36p; 22.5mm, spacing 1uf 400V 80p; 3, 33uf 100V 89p; in depth stocks. ELECTROLYTICS

CAPACITORS

	FLECT	KULI		
	1/63	10	1000/10	19
DIODES	2.2/25	13	1000/16	26
	2.2/63	10	1000/25	36
V 7p	4.7/63	11	1000/40	44
15p	4.7/100	14	1000/63	76
£1.98	6.8/40	10	2200/6	09
11.30	10/6	13	2200/16	44
	10/25	10	2200/25	
	10/40	11	2200/40	73
ECTORS	10/63	14	4700/16	72
	10/100	15	4700/25	90
PARALLEL TYPE	22/10	10		
Plug Skt	22/25	11		
2 8 wey 192 184 2 - 10 way 225 205	22/40	14	TANTALL	ING.
2 - 13 way 279 258	22/63	15	BEARS	
2 · 20 way 382 315	22/100	16	0.1/35	13
Plags have retenness Socke's have strain relief	47/3	03	0.22/35	13
IC Sockets	47/10	11	0.47/35	13
Pres each 25 for 8 80 1 60 N	47/10	12	1.0/35	13
14 120 2 40 N	47/40		2.2/16	13
16 14p 2 80 N 18 16p 2 20 N		15	2.2/35	16
20 18p 3.60 N	47/63	17	4.7/16	16
22 21p 4 20 N 24 21p 4 20 N	47/100	18	4.7/35	18
28 22p 4 40 N	100/3	03	6.8/16	16
TERMINALS	100/10	13	6.8/25	
Plastic screw 15A several col- ours 24	100/16	14	10/6.3	24 16
S14.4mm speket 12	100/25	15		
P14 4mm plug 10 Wander plugs & sockets	100/40	15	10/16	18
red/black pnly ea 14	100/63	20	10/25	18
PC Mountang terminal blocks 60am	100/100	27	22/6.3	18
2 way 17p 4 way 33p, 5 way 48p 8 way 66p, 10 way 82p	220/10	16	22/16	30
12 way 98p	220/16	16	22/25	30
Mains connector - block	220/25	18	33/6.3	24
fused 525	220/40	20	33/10	30
	220/28	28	47/6.3	30
	220/100	42	100/10	80.
inge of many	470/10	19		enges
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ca	st.	plain	or st	ove (grey.					
Į	L	w	D	ΑE	S	Plai	n	Sto	ve C	àr.
	50	50	25	_		5001P	90p	500	1 12	3 ₀
1	00	60	25	2002	96p	5002P	117p	500	2 15	4n
1	113	63	31	2003		5003P				
1	21	66	40	2004		5004P				
1	152	82				5005P				
1	92	113				5006P				
VE	RO	RAN	GE pla	stic b	oxes					
. 1	_	W	0						NGE	pro
	72	47	25	2	1024	51		ases		
1	20	50	35	- 2	1390	83	, 1	34	90	44
1	80	110	55	- 5	1391	151	. 4		140	64
			0.0	•		1019	- 3	02	170	84

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PANEL LAMPS LED chrome, red 54 LEO yellow or	5V filament 3015F 8 mA 255 3015F 15 mA 256
green 70 (high efficiency)	KNOBS black
Filament in red,	plastic
amber, green or clear	K1 skirted with spot 16

		plastic K1 skirted with spot 16 K2 as K1, 35mm 21			
		K3 pointer 15			
LED's		K7 19 mm			
	10	knurled 23			
	19	K8 as K7 +			
green 1	16	skirt 23			

green 16 Mountings, 865	skirt 23
DISPLAYS 7-segment Common anode MAN72A red 75 MAN82A yellow 110 MAN52A green 1110	RECHARGEA- BLE CELLS AA 99p C 227p D (4AH) 376p PP3 cell (7 2V) PP3 char ger 495N 410 CHARGERS FOR ABOVE
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			490N4	TIP150	75
3F244B	40	C407	17	TIP2955	55
3F254	14		126N1	TIP3055	55
3F255	14	C762			50
3F420			40	TIS43	
3F421	31				
	34	C1406	77	U7 6 3	50
	34 35	C1406 D4 clip	for	VN10KM	55
BF457	34	C1406 D4 clip C1406	for and	VN10KM VN46AF	55 93
BF457 BF458	34 35	C1406 D4 clip C1406 E2506	for and 10	VN10KM VN46AF VN66AF	55 93 105
BF457 BF458 BF459 BF597	34 35 35	C1406 D4 clip C1406	for and	VN10KM VN46AF	55 93

MAIL ORDERS: Unit 3, Hill Farm Industrial Estate, Boxted, Colchester, Essex CO4 5RD. **TELEPHONE ORDERS:** Colchester (0206) 36412.

TRANSFORMERS

HARDWARE

Miniature mains: 606V, 909V, 12012V all @ 100mA 100p each.



BES PRI	IICRO ST ICES YWHEI	271 253 273 411	2 2 6 P20 6-P3 150	75 205 340 340 70 0nS 365 440		P10 266 CTC 266 S10 906 DMA 1156 DART 506	0 81L 0 148 0 148 0 Eps 0 Cor	on Printer inectors n at low lo	ow ava	5 100 5 1MH 5 1.84 D 2.0M iil 2.45	1z 32M 4 76M 9M	275 6.0N 200 7.0N 225 8.0N	4 150 4 140 M 170 I 170
4000 4001 4002 4006 4007 4008 4010 4011 4011 4013 4014 4015	10 10 12 50 14 36 24 24 10 15 20 45 40	4016 4017 4018 4019 4021 4022 4023 4024 4025 4026 4027 4028 4029 4030 4031	20 30 45 25 42 40 45 18 33 12 75 20 40 45 14 125	4034 4036 4039 4040 4041 4042 4043 4044 4046 4047 4048 4049 4051 4051 4052 4053	140 249 280 40 40 38 40 40 40 35 38 21 21 42 48 48	4054 4055 4059 4060 4063 4066 4067 4068 4069 4070 4071 4072 4073 4075 4076	78 80 430 42 80 22 225 14 13 13 13 13 13 14	4081 4082 4085 4086 4089 4093 4095 4097 4099 40106 40109 40163 40173 40173	12 12 48 50 125 18 68 65 290 70 70 40 110 60 100 75	40193 4502 4503 4507 4508 4510 4511 4512 4514 4515 4516 4518 4520 4521 4526 4527	65 60 32 35 110 45 40 40 115 115 55 40 50 130 60 50	4528 4529 4532 4534 4538 4549 4553 4555 4556 4559 4560 4584 4585 4724	45 150 60 400 50 360 215 35 35 390 140 35 60
LS00 LS01 LS02 LS03 LS04 LS05 LS08 LS09 LS10 LS11 LS12 LS13 LS14 LS15	11 11 11 12 12 12 12 12 12 12 12 12 12 1	LS20 LS21 LS22 LS26 LS27 LS30 LS32 LS37 LS38 LS40 LS42 LS47 LS48 LS51 LS55 LS73 LS74	12 12 12 14 12 13 14 15 13 28 35 45 14 14 18 17	LS75 LS76 LS78 LS83 LS85 LS86 LS90 LS92 LS93 LS95 LS95 LS107 LS109 LS112 LS113 LS114 LS112	20 17 17 35 48 16 24 25 24 38 95 40 21 21 21 22 35	LS123 LS125 LS126 LS136 LS138 LS139 LS146 LS147 LS148 LS151 LS155 LS155 LS155 LS156 LS156 LS156 LS157 LS158	34 24 25 35 26 30 30 70 150 75 38 38 75 33 36 26 29	LS160 LS161 LS162 LS163 LS164 LS165 LS166 LS170 LS173 LS174 LS175 LS199 LS19 LS1	35 35 35 35 40 55 60 45 45 35 35 35 35 40 45 45 35 35 40 45 45 45 45 45 45 45 45 46 46 46 46 46 46 46 46 46 46 46 46 46	LS197 LS221 LS240 LS241 LS242 LS243 LS244 LS245 LS257 LS257 LS258 LS259 LS266 LS273 LS273 LS273 LS273	45 50 60 55 55 55 70 48 28 32 32 55 20 58 30	LS353 LS365 LS366 LS367 LS368 LS373 LS373 LS375 LS377 LS378 LS390 LS393 LS399 LS541 LS670	60 28 28 28 29 58 60 43 60 57 45 40 156 78 135
7400 7401 7402 7403 7404 7405 7406 7407 7408 7409 7410 7411 7412	11 11 11 12 12 14 19 19 13 13 13 15	7413 7414 7416 7417 7420 7421 7422 7427 7428 7430 7432 7433 7433 7437 7438 7440 7442	17 23 19 19 14 19 19 18 25 13 20 20 23 24 14	7444 7446 7447 7448 7450 7451 7453 7454 7472 7472 7473 7474 7475 7476 7480 7482	85 58 36 43 14 14 14 14 12 24 19 26 25 45 65	7483 7485 7486 7489 7490 7491 7492 7493 7494 7495 7496 7497 74100 74107 74109	30 60 19 180 19 34 24 24 33 33 38 86 78 22 24	74122' 74123 74125 74126 74132 74141 74145 74147 74148 74153 74154 74155 74156 74157 74160	38 38 33 33 30 54 48 75 60 48 38 47 36 36 28 55	74161 74162 74163 74164 74165 74167 74170 74173 74174 74175 74177 74179 74180 74181	46 46 46 46 150 115 58 53 45 35 42 75 38 100 55	74190 74191 74192 74193 74194 74196 74197 74198 74199	40 40 40 40 40 40 40 80 80
CAF	ACIT	ORS		1		NTIOME	Name of		T	OOLS			

	SWITCHES	
١	Submin toggle:	
ı	SPST 55p, SPDT 60p, DPDT 65p.	
ŀ	Miniature toggle:	
	SPDT 80p. SPDT centre off 90p.	
١	DPDT 90p. DPDT centre off 100p.	
1	Standard toggle:	
ı	SPST 35p. DPDT48p	
L	Miniature DPDT slide 12p.	
	Push to make 12p.	
ŀ	Push to break 22p.	
ı	Rotary type adjustable stop.	
ı	1P12W, 2P6W, 3P4W all 55p each.	
ı	DFL switches:	
ķ	4SPST 80p 6 SPST 80p. 8SPST	

VERO					
VEROBLOC +	4				350
Size 0.1 matri					
2.5 x 1					22
2.5 x 3.75					75
2.5 x 5					85
3.75 x 5					95
VQ board					160
Veropins per 1	100				
Single sided					50
Double sided					60
Spot face cutt	er				105
Pin insertion t	ool				162
Wiring pen and	d spo	ol			310
Spare spool 75	5p	Co	mb)5	6

Speaker cable
Standard screened
Twin screened
2.5A 3 core mains
10 way rainbow ribbon
20 way rainbow ribbon.

The state of the s	Residence of the Control of the Cont
VERO VEROBLOC 350 Size 0.1 matrix 25 x 1 25 x 1 25 x 2 25 x 5 35 3 375 x 5 95 VQ board Veropins per 100 Single sided 50 Double sided 50 Spor face cutter 105 Spor face cutter 105 Spor face out the first out of the first ou	PP3 battery clips Red or black crocodile clips Black pointer control knob 19 FU Ultrasonic transducers 50 V Electronic buzzer 50 12 V Electronic buzzer 50 12 V Electronic buzzer 60 12 V Electronic buzzer 61 12 V Electronic buzzer 62 12 V Electronic buzzer 63 12 V Electronic buzzer 64 12 V Electronic buzzer 65 12 V Electronic buzzer 66 12 V Electronic buzzer 67 12 V Electronic buzzer 68 12 V Electronic buzzer 69 12 V Electronic buzzer 69 12 V Electronic buzzer 60 12 V Electronic buzzer 61 12 V Electronic buzzer 61 12 V Electronic buzzer 62 12 V Electronic buzzer 63 12 V Electronic buzzer 64 12 V Electronic buzzer 65 12 V Electronic buzzer 66 12 V Electronic buzzer 67 12 V Electronic buzzer 67 12 V Electronic buzzer 68 12 V Electronic buzzer 68 12 V Electronic buzzer 69 12 V Electronic buzzer 60 12 V Electronic buzzer 61 12 V Electronic buz
DIODES	4½x3x1½" 88 6x4x2" 120 7x4x2 160 6x4x3" 150
BY127 12 ▶1N4001 3 OA47 10 1N4002 5 OA90 8 1N4006 7	SCRs
OA91 7 1N4007 7 OA200 8 1N5401 12 OA202 8 1N5404 16 1N914 4 1N5406 17 ▶1N4148 2 400mWzen 6	BRIDGE 2A 200V 4 RECTIFIERS 2A 400V 4 6A 100V 8 6A 400V 95 1A 50V 20 VM18 DIL 0 9A

SCRs 400 V 8A 400 V 12A
0.0001
BRIDGE 2A 200V
RECITIE ERS 2A 400V
6A 100V
6A 400V 9

1A 400V 35 200V

N	ACCE	ESS	AN	ID ·
ВА	RCL	AY	CA	RD
	WE	LC	MC	3

Please add carriage charges to our normal post charges

600 V, 909 V, 120 V a 1 e 100 MA 100 p each. PCB mounting. Miniature: 3VA 0-6, 0-6 @ 0.25A; 0-9, 0-9 @ 0.15A; 0-12, 0-12 @ 0.12A 200 p each. 6VA 0-6, 0-6 @ 0.5A; 0-9, 0-9 @ 0.3A; 0-12, 0.12 @ 0.25A 270 p each.

6V4 0-6, 0-6 @ 0.5A. 0 9, 0-9 @ 0.3A, 0-12, 0.12 @ 0.25A 270p each. High qualifty. Split bobbin construction of VA 0-6, 0-6 @ 0.5A, 0-9, 0-9 @ 0.4A, 0-12, 0-12 @ 0.3A 220p each. 12VA 0-6, 0-6 @ 1-A, 0-9, 0-9 @ 0.8A, 0-12, 0-12 @ 0.5A. 0-15, 0-15 @ 0.4A 295p (plus 40p carriage) 25VA 0-6, 0-6 @ 1.5A, 0-9, 0-9 @ 1.2A; 0-12, 0-12 @ 1A, 0-15, 0-15 @ 0.8A 330p each (plus 60p carriage) 5VA 0-12, 0-12 @ 2A, 0-15, 0-15 @ 1.5A. 440p each (plus 75p carriage)

C	ONN	ЕСТ	ORS
DIN		Skt	Jack

1		
DIN Plug Sk	t Jack	Plug Skt
2 pin 9p 9	p 2.5mm	10p 10p
3 pin 12p 10	p 3.5mm	9p 9p
5 pin 13p 11		
Phono 10p 12		
1mm 12p 13		18p 17p
UHF (CB) Con	nectors:	
PL259 Plug 40	p. Reduce	er 14p.
SO239 square	chassis skt	38p.
SO239S round	chassis sk	t 40p.
IEC 3 pin 250	V/6A.	
Plug chassis mo	ounting .	. 38p
Socket free ha	nging	. 60p
Socket with 2r	n lead .	. 120p

MULTIMETERS

HT-120 4,000 apv
A smart looking 11 range pocket sized multimeter with an impressive spec. Complete with battery, etc.

spec. Complete with battery, etc. 650p each. HT-320 20,000 opv. Htghly sensitive 19 range multi-meter including transistor tester. Overload protection. DC volts – 1000, AC volts – 1000, DC current 0.25A, 4 resistanceranges, Complete with batteries, leads, etc. 1395p

MIN D CONNECTORS

The second second					-
	9 way	15 way	25 way	37 way	5
Plugs				S	1
Solder lugs	75p	110p	150p	240p	11
Right angle pcb mount	150p	200p	240p	350p /3//	e
Sockets					
Solder lugs	105p	200p	240p	350p	
Right angle pcb mount	160p	210p	290p	4400	
Covers	100p	90p	100p	110p	

Polyester, radial leads, 250v. C280 type: 0.01, 0.015, 0.022, 0.033 -6p; 0.047, 0.068, 0.1 - 7p; 0.15, 0.22 - 9p; 0.33, 0.47 - 13p; 0.68 -20p; 1u - 23p.

322 - 39, 0.33, 0.47 - 13p, 0.68 - 20p, 10 - 23p, 14 - 13p, 0.68 - 23p, 14 - 23p, 14 - 23p, 16 -

4 70n 26p; 680n, 29p; 1u 33p; 2u 2, 50p.
Tentallum bead:
0.1, 0.22, 0.33, 0.47, 1.0 @ 35V - 12p. 2.2, 4.7, 10 @ 25V - 20p;
15/16V - 30p; 22/16V - 27p; 33/
16V - 45p; 47/6V - 27p; 47/16V - 70p, 68/6V - 40p; 100/10V - 90p,
Cer. disc. 22p-0.01u 50V, 3p each,
Mullard ministure caramic plate:
1.8pf to 100pF 6p each,
Polystyrane, 5% tot: 10p-100p, 6;

Polystyrene, 5% tol: 10p-1000p, 6p; 1500-4700, 8p; 6800 0.012u, 10p. Trimmers. Mullerd 808 series: 2-10 pF. 22p; 2-22pF, 30p; 5.5-65pF, 35p

RESISTORS

SOCKETS	Low	Wire
8 ppn	6p	25p
14 pin	80	35n
16 pin	9p	420
18 pin	12p	520
20 pin _e	130	60n
22 pm.	16p	70p
24 pin	18p	700
28 pin,	230	800
40 pin	25p	98p

Rotary, Carbon track Log or Lin 1K - 2M2, Single 32p, Stereo 85p, Single switched 80p, Slide 60mm travel single Log or Lin 5K - 500K

63p each. Preset submin, hor, 100 ohms -1M-7p each. Cermet precision multiturn, 0.75W %" 100 ohms to 100K - 88p each.

REGULATORS

78L05	30	79L05	65
78L12	30	79L12	65
78L15	30	79L15	65
7805	35	7905	40
7812	35	7912	40
7815	35	7915	40
LM309K	130	LM723	35
LM317K	270	LM338K	475
LM317T	120	78H05 5A	
LM323K	350	▶5V	550
EAVE T	-	CYMPRO	-

SOLOERING IRONS

Antex CS 77W Soldering iron	46
2.3 and 4.7mm bits to suit	6
CS 17W iron: 450, element:	21
Antex XS 25W	48
3.3 and 4.7mm bits to suit .	6
Solder pump desoldering tool.	48
Spare nozzle for above	74
10 metres 22swg solder .	10

PCB MATERIALS

Altac transfer sheets - please s	state
type (e.g. DIL pads etc.)	45
Dalo etch resist pen	100
Fibre glass board 3.75"x8"	80
Ferric chloride 250ml bottle.	100

Small trimming tool	22
Small pocket screwdriver	16
Large pocket screwdriver	13
6 piece precision screwdriver	set
in plastic case	170
Low cost side cutters	160
High quality side cutters	650
Low cost pliers	160
High quality pliers	650
Wire strippers	120
Expo reliant driff	695
Expo Titan drill	1025
Drill stand	1200
Reduced shank drill bits for	
above 0.8mm, 1mm, 1.4mm	60

OPTO

▶3mm red	7	▶ 5mm red	7
▶3mm gree	n 10	▶5mm green	10
▶3mm yelic	w10	₱ 5mm yello	w10
Clips to suit	- 3p	each	
Rectangular		TIL32	40
▶red	12	T1L78	40
green	17	▶TIL111	60
yellow	17	ORP12	85
	40	T1L100	90
2N5777	45	Dua! colour	60
Seven segme	nt di	splays:	
Com cathod	e	Com anode	
DL704 0.3"	95	DL707 0.3"	95
▶FND500		FND507	
0.5"	100	0.5"	
		TIL3120.3"	
T1L3220.5"	115	TIL3210.5"	115
LCD: 3½ di	git 58	Op. 4 digit 62	20p.
Desire Control		THE REAL PROPERTY.	
TRIACS		400 V 8A	65

TRIAC	s	400 V 8A 400 V 16A	6
_	_		
400 V 4 A	50	BR100	2
MONOT PERSONAL IN	PACCE	MORE AND PAGE	150457

COMPONENT KITS

An ideal opportunity for the beginner or the experienced constructor to obtain a wide range of components at greatly reduced prices. "W" Resistor kit. Contains 10 of each value from 4.7 ohms to 1 M (total of 650 resistors).

Ceramic Cap. kit. 5 of each value 22p to 0.01u (135 caps)

Polyester Cap. kit. 5 of each value from 0.01 to 1uF (65 caps)

Preset kit. Contains 5 of each value from 100 ohms to 1M (total 65 brasets)

65 presets Nut and Bolt kit (total 300 items): 180p

25 6BA ½" bolts 25 6BA ½" bolts 50 6BA nuts 50 6BA washers 25 4BA ¼" bolts 25 6BA ½" bolts

TRANSI	ETOL	ie.	BC517		BF337	40 1	MPSU56	60 ZTX	108	8 2N3055	50
▶CA3240E	110	LM3352	125	▶LM3909	- /0	NE565	110	TL062	60	ZN1034E	200
CA3189	290	LM334Z	100	LM3900	45	►NE556	45	TL061	40	ZN459	285
CA3161E	100	LM324	40	LM2917	200	►NE555	16	TDA1024	125	ZN428E	480
▶CA3140E	36		120	LM1468	40	NE544	205	▶TDA1022	490	ZN427E	650
CA3130E	85	£M311	70	LM747	60	NE531	150	TDA1008	320	ZN426E	330
CA3090AQ		LM301A	25	LM741	14	NE529	225	T8A950	220	ZN425E	350
CA3089	180	LM10	360	LM733	75	MM5387	465	TBA820	70	ZN424	135
►CA3080	65	LF356	90	LM725	350	ML929	140	TBA810	96	ZN423	135
CA3046	60	LF353	85	LM711	60	ML928		TBA800	75	ZN414	100
AY-3-8912		▶ LF351	45	LM709	25	ML927	140	T8A120S	70	XR2206	290
AY-3-8910		ICM7555	80	LM393	100	ML926	140	SP8629	250	ULN2004	90
AY-3-1270		1CM7224	785	LM387	120	ML925	210	▶SN76477	380	ULN2003	85
9400CJ	350	ICLE211A		LM386	65	ML924	195	SL76018	150	UA2240	120
748 *	* 35	ICL8038	295	LM384	130	ML922	400	SL490	250	TL170	50
₱741	14	ICL7622	180	LM382	120	►MF10C		SL480	170	TL082 TL084	95
709	25	ICL 7621	180	►LM381	120	MC1496 MC3340	68 135	▶RC4558	60	►TL081	25
556CMOS	150	#CL7611	95	►LM380	65	LM13600		▶RC4136	55	TL074	95
555CMOS	80	ICL7106	790	LM358	170	LM3915	195	►NE570 NE571	370	TL072	50
	-			LM348 LM358	60	LM3914	175	►NE567	100	TL071	30
LINEAR	3			LM339	45	LM3911	1 20	NE 566	140	TL064	96

TRAN	20.0	100		BC547	7	8FR40	23	TIP29A	30	ZTX109	12	2N3442	120
				BC548	10	BFR80	23	TIP29B	55	ZTX300	14	► 2N3702	
	35	BC149	9	BC549	10	▶BFR81	20	TIP29C	37	ZTX301	16	2N3703	9
AC126	25	BC157	8	BC558	10	BFX29	25	TIP30A	35	ZTX302	15	▶2N3704	4 6
AC127	25	BC158	10	BCY70	18	BFX84	25	TIP30B	50	ZTX304	17	2N3705	9
► AC128	20	BC159	8	BCY71	18	BFX85	25	TIP30C	37	ZTX341	30	2N3706	. 9
	25	BC160	45	BCY72	18	BFX86	28	TIP31A	35	ZTX500	15	2N3707	10
AC187	22	BC168C	10	BD115	55	BF X87	25	TIP31C	37	ZTX501	15	2N3708	10
AC188	22	BC169C	10	BD131	35	BFX88	25	TIP32A	35	ZTX502	15	2N3709	10
AD142 1	20	BC170	8	BD132	35	BFY50	23	TIP32C	37	ZTX503	18		170
	80	BC171	10	BD133		BFY51	20	TIP33A	50	ZTX504	25	▶2N3773	
	40	BC172	8	BD135	40	BFY52	23	TIP33C	75	2N697	20	► 2N3819	
	40	BC177	18	BD136	30	BFY53	32	TIP34A	60	2N698	40	2N3820	40
	60	BC178	18	BD137	30	BFY55	32	TIP34C	85	2N 706A	20	2N3823	65
	50	BC179	18	BD138	30	BFY56	32		105	2N708	20	2N3866	90
	40	BC182	10	▶BD139	35	BRY39	40		126	2N918	35	2N3903	10
	70	▶BC182L		▶BD140	35	BSX20	20	TIP36A			22	2N3904	10
	75	BC183	10		110	BSX29	35	TIP36C		2M1613	30	2N3905	6
	10	BC183L	10		110	BSY95A	25	TIP41A	45	2N2218A		2N3906	10
	12	BC184	10	BD222	85	BU205	160	TIP42A	45	2N2219A			45
▶BC108	9	BC184L		BF180	35	BU206		T#P120	90	2N2221A		2N4058	10
	12	BC212	10	BF 182	35	BU208	170	TiP121	90	2N2222A		2N4060	10
	12	BC212L	10	BF184	25	MJ2955	99	TIP122	90	12N2368	25	2N4061	10
▶BC109	9	BC213	10	BF 185	25	MJE340	50	TIR141	98		16	2N4062	10
	12	BC213L	10	BF194	12	MJE520	65	TIP142	98	2N2484	25	2N5457	36
	18	BC214	10	BF 195	12	MJE521	95	TIP147	110	2N2846	#	2N5458	36
	22	▶BC214L		BF196	12	MJE3055	70	TIP2955	60	2N2904	20	2N5459	30
	18	BC237	8	BF197	12	MPF102	40	TIP3055	55	2N2904A	20	2N5485	36
	35	BC238	14	BF198	10	MPF 104	40	T1S43	40	2N2905	22	2N5777	.45
	40	BC308	12	BF199	18	MP\$A05	22	TIS44	45	2N2905A			30
	40	BC327	14	BF200	30	MPSA06	25	TIS90	30	2N2906	25	40360	40
	28	BC328	14	▶BF244B	22	MPSA12	30	TIS91	30	2N2906A		40361	50
	30	BC337	14	BF 245	30	MPSA55	30	VN10KM		2N2907	25	40362	50
	25	BC338	14	BF256B	45	MPSA56	30	VN46AF	75	2N2907A		40408	70
	25	BC477	30	BF257	32	MPSU05	55	VN66AF	85	2N2926	9		
BC147	8	BC478	30	BF258	25	MP\$U06	55	VN88AF	95	▶2N3053			
BC148	8	BC479	30	BF259	35	MPSU55	60	2TX107	8	2N3054	55		

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4 AMP - 400v - T0202 - TAG 136G 50 OFF £17.50 40p £3.75 £17.3 8 AMP 400v — T0220 — TAG 425 £3.75 £30.00 £5.75 60a £27.50 £50.00

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SX66 5 x 22k Log

50p PER PAK SX67 5 x 47k Lin SX68 5 x 47k Log SX69 5 x 100 Lin SX63 5 x 470 ohms Lin \$X64 5 x 1k Lin \$X65 5 x 22k Lin

ALL AT

\$X70 5 x 1 meg Lin

250 Silicon Diodes—Switching like IN4148 DO 35 All good—uncoded Worth doubleour price 45v 75mA £1.25 250 Silicon Diodes—General Purpose like QA200/202 BAX13.16 Uncoded £1 25 30-100y 200mA DO-7

10 5A SCR's TO64 3 x 50v. 3 x 100v. 2 x 200v. 2 x 400v. Super value less than 1: 10 5A SCR s TO66 2 x 50v. 2 x 100v. 4 x 200v 2 x 400v All coded Brand new a give away at



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insulated handles 4 inch length. Order

Miniature long nose pliers - insulater handles 5 inch length Order No:YD44.

handles 54inch length Order No YO45

Miniature end nippers - insulated handles 4 inch length. Order No YO46. de

Miniature snipe nose pliers with side handles 5inch length. Order No YO42.

1.25 All with insulated handles

PLASTIC

SIZE " I W

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ALUMINIUM BOXES

Made with Bright Aluminium toided

214 1 1/2 161

construction with deep lid and screws

ALL AT

Order No

164

166 167

ength Bilinch, Order No:FS-1 Flat blade 4mm FS-2 Cross point no.1 £1.76 au



Ginch long screwdriver with spring loaded grip on end to hold screws in position while reaching into those difficult places. Order No SD-1 Flat blade 4mm SD-2 Cross point no.0. 95p and

INEXPENSIVE TOOLS OF IMMENSE VALUE Combined wire stripper, cutter, crimper incl. 25 asst terminals for crimping Order No.WS2. Our low price £1.20 oor

Plastic Boxes

Flanged Lid, fixing screws into brass bushes

Order No

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144

146

Price

£1.00

€1.30

€1.50

£1.40

£2.14

Coloured Black Close fitting

4 3/4 21/2 11/2

4 234 1

51/2 41/4 21/4

Plastic sloping front

31/4 2

Plastic as above but with aluminium

slope

1 1/2 148

*L W

Set of 5 BA spanner shafts plus universal handle in roll-up wallet. Sizes 08A 2-4-6-88A Order no: T192 62 75 cat

NEON SCREWDRIVER

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RANGES DC Voltage 0-200mV 0-2-20-200-1000V Acc 0.8% AC Voltage 0 - 200 - 1000 V Acc. I 2% DC Current 0 290uA 0.2.20.200mA 0-10 A Acc 1 2%

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In the cut-throat world of consumer electronics, one of the questions designers apparently ponder over is "Will anyone notice if we save money by chopping this out?" In the domestic TV set, one of the first casualties seems to be the sound quality. Small speakers and no tone controls are common and all this is really quite sad, as the TV companies do their best to transmit the highest quality sound. Given this background a compact and independent TV tuner that connects direct to your Hi-Fi is a must for quality reproduction. The unit is mains operated.

This TV SOUND TUNER offers full UHF coverage with 5 pre-selected tuning controls. It can also be used in conjunction with your video recorder. Dimensions: 1134" x 81/2" x 31/4"

E.T.I. kit version of above without chassis, case and hardware. £12.95 plus £1.50 p&p.

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· NOISE REDUCTION SYSTEM · AUTO STOP · TAPE * NOISE REDUCTION SYSTEM * AUTOSTOP * TAPE COUNTER * SWITCHABLE E.Q. * INDEPENDENT LEVEL CONTROLS * TWIN V.U. METER * WOW & FLUTTER O.1.% * RECORD/PLAYBACK I.C. WITH ELECTRONIC SWITCHING * FULLY VARIABLE RECORDING BIAS FOR ACCURATE MATCHING OF ALL TAPES.

BIAS FOR ACCUMALE MALCHING OF ALL LINES. Kit includes tape transport mechanism, ready punched and back printed quality circuit board and all electronic parts. i.e. semiconductors, resistors capacitors, hardware top cover, printed scale and mains transformer. You only supply solder and hook-up wire.

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2 WAVE BAND MW -- LW

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Amplifier for your personal stereo cassette player as featured in January issue of Everyday Electronics, Turn your personal stereo into a mains powered home unit



Parts: Stereo power amp PCB with all components, £3.50 + 75p p&p. Power supply unit, £1.95 + £1.50 p&p. Pair of 4%" eliptical speakers, £1.50 the pair, £1.00 p&p. Input & output sockets & plugs, £1.50. Recommended case (for the power supply and amp only), £2.95 + 80p p&p. P&P inclusive price of £1.75 for two or more articles

P.E. STEREO TUNER



This easy to build 3 band stereo AM/FM tuner kit is designed This easy to build 3 band stereo AMI/FM tuner kit is designed in conjunction with Practical Electronics (July '81 issue). For ease of construction and alignment it incorporates three Mullard modules and an I.C. IF. System. FEATURES: VHF, MW, LW Bands, interstation muting and AFC on VHF. Tuning meter. Two back printed PCB's. Ready made chassis and scale. Aerial: AM - ferrite rod, FM - 75 or

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Self assembly simulated wood cabinet sleeve to suit tuner only Finish size: 114"x 8½"x 3¼". £3.50 Plus £1.50 p&p

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Suitable LS coupling electrolytic: £1.00 plus 25p p&p



SPECIFICATIONS:

Max. output power (RMS): 125W.
Operating voltage (DC): 50 - 80 max.
Loads: 4 - 16 ohms.
Frequency response measured @ 100 watts: 25Hz - 20KHz.
Sensitivity for 100 watts: 400mV @ 47K. Typical T.H.D. @ 50 watts, 4 ohms: 0.1%. Dimensions: 205 x 90 and 190 x 36 mm.

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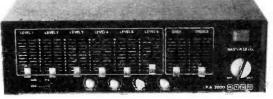


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THE COMPUTER AS ALLY

ITHOUT doubt, 1982 was the year of the home computer. The only other Vinnovatory electronic consumer product in a similar price bracket likely to have equalled it in popularity was the video-recorder; but the latter is more rightly considered as an extension of home entertainment, and not an inter-active device opening up an entirely new field like a computer.

The sales of home computers must have gladdened the hearts of manufacturers and retailers alike, providing (as they did) welcomed bright spots in an otherwise dull scene. It speaks volumes for the fascination of computers that in a time of recession thousands of customers have emerged to buy a piece of equipment they can have initially little understanding about and are prepared to venture upon a journey into the unknown, seeking-what? For many new computer owners, the destination will be simply "games". Just how many will persevere with the writing of programs for personal needs and thereby justify the considerable financial outlay is a matter for conjecture.

Be that as it may, what is now happening on a grand scale is the development of computer consciousness and "hands on" experience amongst a broad selection of the general public. For the younger generation in particular, all this can be extremely valuable since many of the jobs likely to be on offer in the future will require

familiarity with computers and computing.

While there is bustle in the computer field, the home construction scene appears to be a trifle stagnant. One cannot be too sure and claim that the one is a consequence of the other. But even if so, it is likely to be but a transitory fall-off; in the longer term some of the interest now developing in computers will percolate into electronic technology itself, and produce new recruits for the hobby of circuit construction. Curiosity in the technology behind the keyboard could encourage a wish for practical involvement with electronics in general.

Support for this view was to be found at the Electronic Hobbies Fair last November (see review in this issue). It seemed that a very considerable proportion of visitors were interested in electronics, but indirectly. That is to say they were interested in (or attracted by) the ends, rather than the means—computers, of course, providing the chief and most striking example. But during their tour of the Fair awareness of the scope and possibilities of electronic construction must have been created in the minds of many non-technical visitors. Proof that converts were made is found in the large sales of educational kits reported by exhibitors. Similarly, much interest was shown in the Introducing Electronics series as featured on the EE stand. The solderless technique employed was favourably commented on and seems likely to win quite a few new recruits to our hobby.

The coming of the home computer and the new field of interest it creates should in no way detract followers of electronics from the practical business of designing and building circuits. In fact, the computer can become a valuable aid and ally to the electronics experimenter, by testing ideas and solving problems—as well as to the practically inclined at large, by controlling small machines such as the wood-turning lathe demonstrated on the EE stand.

Readers' Enquiries

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We cannot undertake to engage in discussions on the telephone.

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Readers should note that we do not supply electronic components for building the projects featured in EVERYDAY ELECTRONICS, but these requirements can be met by our advertisers.

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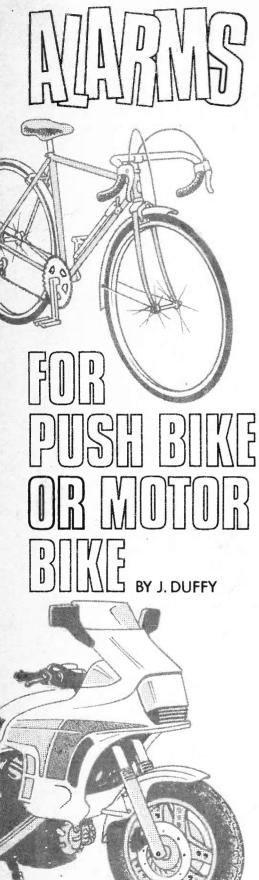
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Today, when enough money has been saved to buy and insure a bike the last thing on the owner's mind is theft. Unfortunately, it is all too common but this project should set the minds of some readers at rest.

Two bike alarms are described—one for a motorbike and the other for a pushbike. They both use the same circuit employing a mercury tilt-switch, which triggers the horn and lights of the motorbike version, or an integral alarm unit on the pushbike version.

The motorbike unit also has an external trigger facility from a microswitch

for further protection.

A pulsed output is more effective than a steady output and, apart from being easier to locate, reduces battery consumption.

CIRCUIT DESCRIPTION

The circuit diagram of the two versions of the Bike Alarm is shown in Fig. 1. The system has three basic sections: the trigger; low frequency oscillator; and the alarm output.

The first section consists of a monostable multivibrator which is triggered by the mercury switch, S2 (and in the case of the motorbike version, microswitch S3 will also trigger it). The time constant of this monostable circuit is governed by R3 and C2 and it functions as follows:

The monostable consists of two CMOS 2-input NOR gates and in the quiesent state, the output of IC1a (pin 3) is at logic 1 as both inputs (pins 1 and 2) are at logic 0. As a result, the output of IC1b (pin 4) is held low as both its inputs (pins 5 and 6) are high (this gate is acting as an inverter)

Upon receipt of a positive pulse to pin 1 from the mercury switch, the output of IC1a will go low causing the output of IC1b to go high. The low on pin 1 provides a charge path for C2 via R3 and when this capacitor has charged to a sufficiently high voltage, the input of IC1b reads this as a logic 1 and the output (pin 4) of this inverter is therefore returned to a logic 0.

With the values given, the period of this monostable is in the region of one minute.

OSCILLATOR

When the output of the monostable is high, it enables a low frequency oscillator consisting of NOR gates IC1c and IC1d. The output of this section is a square-wave of approximately one hertz. This frequency is controlled by R4 and C3.

When the output of the monostable is low, the oscillator cannot function since the time constant capacitor C3, will not charge up as current flows through D1 to the low (effective earth) on pin 4 of IC1b.

ALARM OUTPUT

On the motorbike version, the alarm output is in the form of a relay, activated by a high pulse from the oscillator and

driven by TR1. D3 protects the transistor from the back e.m.f. from the coil.

The contacts of this relay are used to control the horn and headlight from the host motorbike. Note that the positive supply to the light and horn is taken to the relay before the Alarm unit on/off switch.

COMPONENTS

MOTORBIKE

Resistors

R1,2 $100 \mathrm{k}\Omega$ (2 off) R3,4 $10 \mathrm{M}\Omega$ (2 off) R5 $10 \mathrm{k}\Omega$ All $\frac{1}{2} \mathrm{W}$ carbon $\pm 5\%$

Capacitors

C1 100µF 16V elect. axial

C2 15µF 16V elect. axial lead

C3 0.1µF polyester

Semiconductors

D1 1N4148 D2,3 1N4001 (2 off) TR1 BC107 npn silicon IC1 4001UB cmos quad 2-input NoR gate

Miscellaneous

S1 s.p.s.t. key-switch S2 mercury tilt-switch S3 microswitch RLA miniature relay, 12V,

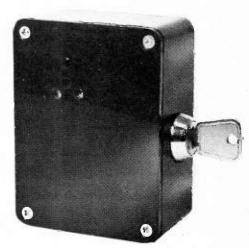
A miniature relay, 12V, 130Ω coil with 30A rated contacts

0.1 inch matrix stripboard 11 strips by 28 holes; plastic case; 100 × 76 × 41mm (ABS case type MB2); M2 mounting screws 13mm long; 14-pin d.i.l. i.c. holder; terminal block 6-way; 7/0.2mm connecting wire; 24/0.2mm wire.

Approx. cost Guidance only

E12

Completed alarm for a pushbike with the key-operated switch mounted on the side.



The pushbike version of the alarm has its own integral buzzer (WD1), and this is driven, again on a high pulse from the oscillator, by Darlington pair, TR1 and TR2. This is shown on the additional section of the circuit diagram, Fig. 1.

This unit has its own supply, a 9V PP3

type battery.

COMPONENTS

PUSHBIKE

Resistors

R1,2 100kΩ (2 off) R3,4 10MΩ (2 off) R5 10kΩ All $\frac{1}{2}$ W-carbon $\pm 5\%$



Capacitors

C1 100µF 16V elect, axial lead

C2 $15\mu F$ 16V elect. axial lead C3 $0.1\mu F$ polyester

D1 1N4148 D2 1N4001

TR1 BC441 npn silicon
TR2 BC107 npn silicon
IC1 4001UB cMos quad
2-input NOR gate

Miscellaneous

S1 s.p.s.t. key-switch S2 mercury tilt-switch WD1 9V buzzer

WD1 9V buzzer
B1 9V PP3 battery
0.1 inch matrix stripboard,

strips by 28 holes; battery clip; plastic case, 100 × 76 × 41mm (ABS case type MB2); M2 mounting screws 13mm long; 14-pin d.i.l. i.c. holder; 7/0-2mm connecting wire.

Approx. cost Guidance only £10



CIRCUIT BOARD

A plastic case measuring about $100 \times 75 \times 41$ mm is used for both the motor-bike and pushbike projects. Although any plastic or metal boxes around this size should be acceptable.

The component panel for both versions is a 0·1 inch matrix stripboard having 28 holes by 11 strips, and these are shown in Figs. 2 and 3. Construction of the board follows the normal pattern with the breaks in the copper strips being made first. Next, solder in all the links and the i.c. socket, after which the components

The motorbike alarm with lid removed to show positioning of the terminal block, circuit board and key-switch. The mercury "trip" switch is mounted in the bottom right corner.



may be soldered into place and Veropins fitted where connections to off-board components are to be made. Take care with the CMOS i.c. since it may be destroyed by static.

In both units the board fits easily into the slots in the plastic case. In the motor-bike version the holes for the key-switch, grommet, and the mounting holes for the terminal block should be made first. In the pushbike version only sound holes for WD1 and a key-switch hole are required. Care must be taken not to let the metal case of the buzzer touch any connections or switch terminals.

MERCURY SWITCH

The mercury switch, S1, should be soldered to the component board on leads approximately 100mm long and temporarily attached to the side of the box with a small piece of Plasticine or Blu-Tak.

This is necessary, as the final position of this switch can only be determined after the Alarm unit has been mounted on

Layout of components inside the case of the pushbike alarm. The siren is temporarily held in position by Blu-Tak. The leads to the mercury switch can be seen top right.

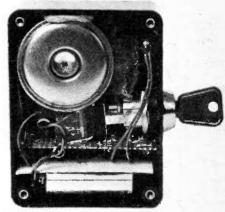
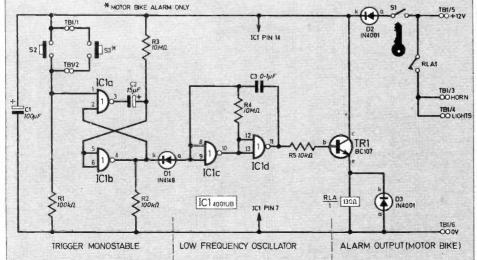
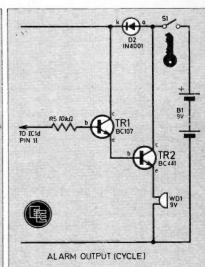


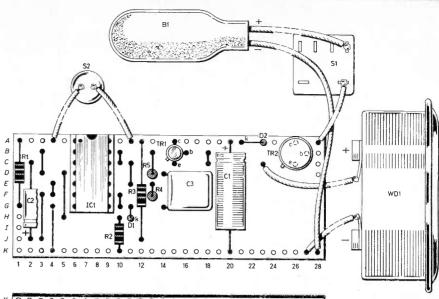
Fig. 1. Circuit diagram of the two versions of the Bike Alarm. The siren circuit for the pushbike is shown on the right.

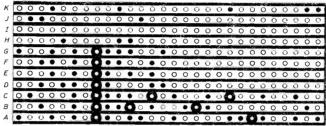




FOR POST BIXE OR MOTOR BIXE BY J. DUFFY

Fig. 2 (Right). Component layout, underside, showing breaks in the copper strips and interwiring details for the Pushbike Alarm.





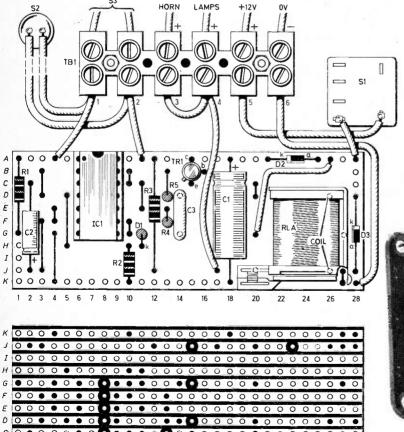
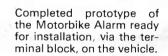
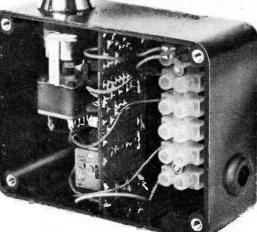
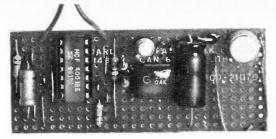


Fig. 3 (Left). Circuit board for the Motorbike Alarm showing component layout and wiring to the terminal block for connecting to the bike. The relay is directly wired to the circuit board by connecting pins. These are shown dotted.

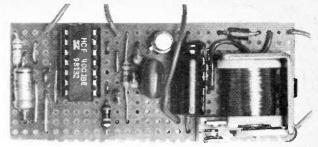








Finished circuit board for the Pushbike Alarm. The leads to the mercury switch, shown left, should be approximately 100mm long.



The completed Motorbike Alarm circuit board.

the bike frame. A certain amount of experimentation will be required in order to achieve the optimum switch position to trigger the alarm when the bike is moved by an unauthorised rider.

The final location of the mercury switch is largely dependant on two factors, whether the alarm is to be triggered by a side-to-side movement or a forward and backward motion. Bearing in mind that a bicycle (or motorbike) is often left in an inclined position, for example against a wall or on a kick-stand, the former consideration can be utilised so that the alarm triggers when the bike is returned to the upright position.

Mercury switches of this type activate

Mercury switches of this type activate at around 10 degrees from the horizontal (that is, at this angle the mercury blob will roll to the end of the device and short out the contacts) so it will need to be mounted at something like this angle inside the case.

This type of triggering does mean that the bike will always have to be inclined to the same side.

MOUNTING

When locating the Alarm unit on the bike, it is important to keep it as discrete as possible; a box that looks like an alarm could be removed and rendered inoperational by the prospective thief.

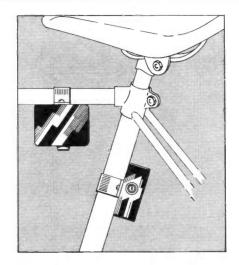


Fig. 4. Two suggested mounting positions for the completed Pushbike unit. Final arrangement will be governed by placement of the mercury switch.

For this reason, the lock of the keyswitch should not be visible but it must be accessible. The ideal orientation of the key-switch is downward facing. The finished Alarm unit should be

The finished Alarm unit should be securely fixed to a frame member with a fixing bracket obtained from the local

bicycle shop. A suitable position is on the down tube just below the saddle or under the cross-bar as shown in Fig. 4.

For additional reliability, the Alarm unit can be water-proofed. The simplest way to do this is with a plastic bag or "Clingfilm" wrapped around the case. Another alternative would be to use a more expensive enclosure with a sealing gasket or to use a commercially available sealing compound.

MICROSWITCH

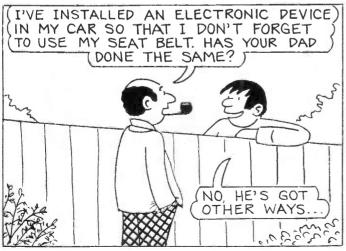
The motorbike version of the Alarm has the option of an additional microswitch to trigger the circuit and this is connected in parallel with the mercury switch (via TB1 terminals 1 and 2).

The normally open contacts are used and the switch is placed where it will be activated when the motorbike is moved. One position could be under the seat so that when someone sits on it, the lever of the microswitch is depressed.

The additional wiring required for this version, that is the wires to the headlight and horn and to the 12V supply, must be carried out with a stranded wire of sufficient current rating. A 24/0·2mm wire is suitable and should be fitted with the correct type of connector to mate with the terminals on the electrical system of the bike.

JACK PLUG & FAMILY...

BY DOUG BAKER







BY D. WILKES & A. WILLIAMS

ONE PIECE of laboratory apparatus still frequently used in experiments concerning the general equations of motion is the "ticker-tape timer". This electromechanical device consists of a mains solenoid inducing a pointer in its magnetic field to oscillate at 50Hz. A strip of paper passes between the oscillating pointer and a disc of carbon paper to produce a series of dots, the distance between each consecutive dot representing the distance the paper strip has travelled in 1/50th of a second.

In this way, a plot of distance against time can be recorded for an object rolling down an inclined surface, when that object is secured to the length of paper tape and is pulling it through the ticker-tape timer as it falls.

INACCURATE

This method of timing motion does, however, have inherent inaccuracies. Firstly, the oscillating pointer does not always make positive contact with the ticker-tape. Secondly, and perhaps more

importantly, two dots in rapid succession caused by slackness in the tape could be interpreted as a single mark. Finally, the tape itself could cause some resistance to the free motion of the object pulling it.

These three factors, combined with the cumbersome operation and wasteful use of paper necessitated the design of an electronic method of timing the motion. Without changing the basic experiment, that is, a trolley (the "object") rolling down an inclined surface, the ZX81 Speed Computer System was designed.

The trolley is linked to the main unit via a pulsed ultrasonic transmitter thus eliminating the errors associated with the ticker-tape method. The ultrasonic receiver amplifies the pulses and feeds them to a Sinclair ZX81 microcomputer to be processed and the results displayed on a 4-digit 7-segment display on the main unit.

The system has been code named G.E.M.N.I.F. for General Equations of Motion Notational Information Finder.

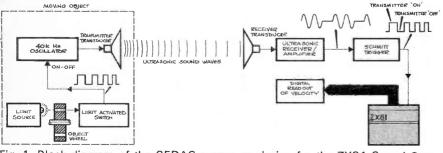


Fig. 1. Block diagram of the SEDAC runner-up design for the ZX81 Speed Computing System (G.E.M.N.L.F.). Note that the moving object represents the trolley.

BASIC THEORY

The trolley provides the information with which the ZX81 will perform the calculations to determine velocity and acceleration.

The basic formulae of motion are as follows:

$$velocity = \frac{distance travelled}{time taken}$$

$$acceleration = \frac{change in velocity}{time taken}$$

So, to calculate the velocity, the computer needs to know the distance travelled (this will be constant) and the time taken to travel that distance. In order to do this, the trolley will send 30 ultrasonic pulses and the computer measures the time taken for them to arrive. As previously stated, the distance the trolley moves to generate 30 pulses is always constant.

For the acceleration calculation, as the trolley starts from standstill, that is 0m/s, the change in velocity will equal the total velocity and since this has already been calculated by the computer it can now calculate the maximum acceleration.

However, computers can be programmed to do much cleverer things so in this circuit, on obtaining the set number of pulses, it can time each individual pulse and provide acceleration figures for any point during the trolleys voyage, thus enabling a pupil to plot a graph from selected notational information; which is neatly displayed on the digital display.

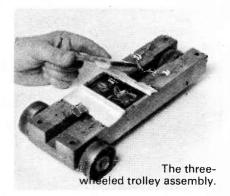
THE TROLLEY

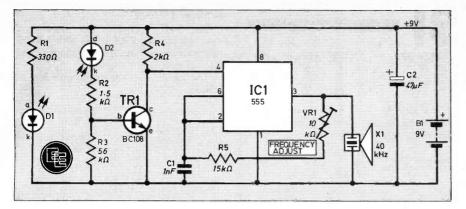
As already mentioned, the microcomputer relies on a set number of pulses which are linked directly to the distance the trolley moves.

The trolley produces pulses whose rate of production is proportional to the speed of the trolley, that is, the faster the trolley moves, the more pulses are produced per second. It achieves these linked pulses by utilising its back wheel. First, however, the basic circuit of the pulse producer and ultrasonic transmitter will be explained. See Fig. 2.

Ultrasonics are a range of frequencies just beyond the limit of human hearing, normally about 30 to 50kHz. This circuit transmits 40kHz pulses to the receiver.

The 555-timer IC1, is connected in an astable mode, oscillating at approxi-





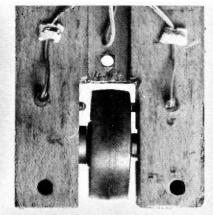
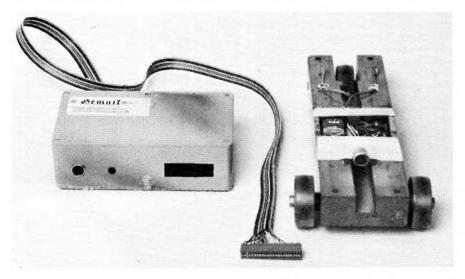


Fig. 2 (Above). Circuit diagram of the trolley mounted ultrasonic transmitter. D1, D2, TR1 and associated components form an infra-red light activated switch to enable IC1.

Photograph (left) shows a close up of the rear wheel of the trolley. The infra-red emitter and receiver D1 and D2 can be seen set into the wooden body of the trolley either side of wheel.

(Below). The main unit and trolley showing the mounting of the ultrasonic transducer X1 and X2. Note how SK1 is soldered to a length of 14-way ribbon cable.



mately 40kHz. An infra-red l.e.d. and an infra-red photodiode (D1 and D2) are positioned either side of the back wheel. The wheel itself has a hole bored through it. D2 is designed so that when infra-red light reaches it, it switches on allowing current to flow through it. In this circuit the light comes from D1.

As both D1 and D2 are on either side of an opaque material (in this case the rear wheel), the only time D2 conducts is when the hole in the rear wheel aligns itself between them. When this takes place, TR1 switches on generating a negative pulse. It is only a pulse since the light is cut off by the wheel's rotation.

This negative pulse gates the 555 off,

stopping momentarily the 40kHz transmission. Thus, as the wheel rotates the hole aligns and dis-aligns, the 555 astable switches on and off. An ultrasonic transducer X1 is used to transmit the pulses.

The ultrasonic transducer is mounted, forward facing, on the front of the trolley as shown in the photographs. A compartment can be made for the electronics in

the wooden body of the trolley.

When mounting D1 and D2, the single rear wheel of the trolley is drilled and the hole elongated (so as to produce a longer pulse) and the infra-red devices are mounted in the trolley body so as to align with the hole.

COMPONENTS

Resistors

R1,13-19 330Ω (8 off) R2 1.5kΩ R3 $56k\Omega$ R4 2kO 15kΩ (2 off) R5,10 R6 $4.7k\Omega$ R7 $1M\Omega$ R8,9 560Ω (2 off) R11 $10k\Omega$ 4700 R12 All \(\frac{1}{4}\text{W carbon } \pm 5\)%

Capacitors

1nF polystyrene
47μF 16V tantalum
22pF ceramic (2 off)
10nF polyester (3 off)
10pF ceramic
0·1μF polyester

Semiconductors

D1 D2 D35	Infra-red emitter Infra-red receiver BAX13 silicon (3 off)
TR1-3	BC108 silicon npn (3 off)
IC1	555 timer
IC2	TAA960 triple
	amplifier
IC3	74LS132 TTL low
	power Schottky quad
	2-input NAND
	Schmitt gate
IC4	74LS04 TTL low
	power Schottky hex
	inverter/buffer
IC5	74LS08 TTL low
	power Schottky quad
	2-input AND gate
IC6	74LS10 Tt low
	power Schottky triple
	3-input NAND gate

power Schottky triple
3-input NAND gate
1C7,8 74LS373 TTL low
power Schottky octal

latch (2 off)
4-digit, 7-segment common cathode display (R.S. 587–507)

X1 40kHz ultrasonic transmitter transducer X2 40kHz ultrasonic receiver transducer

Miscellaneous

LED1

VR1	10kΩ miniature
	preset
VR2	1MΩ miniature
	preset
VR3	100kΩ miniature
	preset
B1	9V PP3 battery
SK1	23 + 23-way double
	sided 0-1in. pitch
	edge connector (to
	suit finger set on
	ZX81)



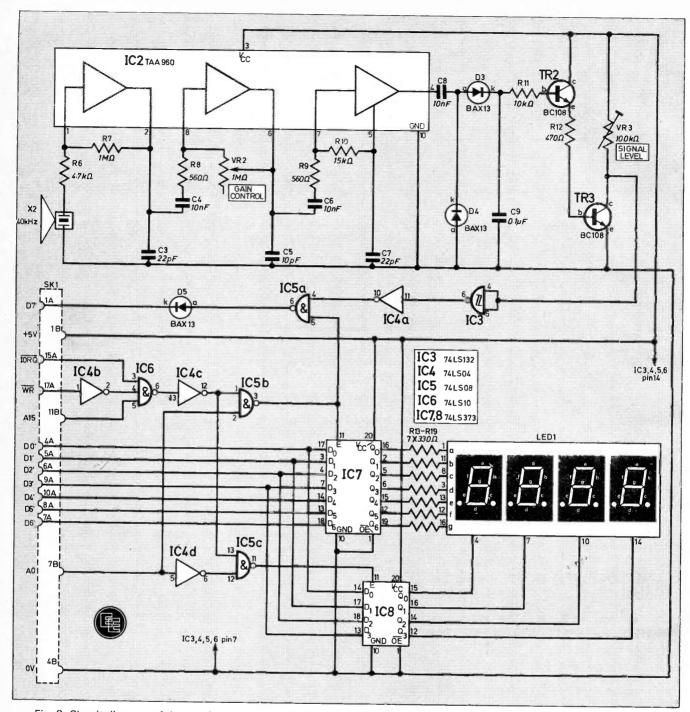


Fig. 3. Circuit diagram of the receiver, computer interface and display sections of the ZX81 Speed Computing System.

THE RECEIVER

The complete circuit diagram of the receiver unit and the ZX81 interface is shown in Fig. 3. The Ultrasonic pulses from the trolley-mounted transmitter are received by a second transducer, X2, and this is coupled to the first stage of IC2, a triple amplifier. This i.c., the TAA960, is specially designed for use with high impedance receivers.

The input impedance is high to ensure that the transducer has peak response at its anti-resonant frequency.

The combined three stages and sup-

porting circuit gives the amplifier a gain of about 100dB, and the receiver is efficient up to a range of 8 metres.

The power supply for this circuit and for the interface circuitry is taken from the ZX81 finger set at the back of the computer. The +5V and 0V come from pins 1B and 4B, respectively.

The output from the receiver amplifier is fed to TR2 and TR3 and the resulting pulses are fed to a Schmitt trigger NAND gate, IC3. This gate produces a clean 5V pulse to be fed to the ZX81 to be processed. Fig. 1 (block diagram) shows the resulting waveform.

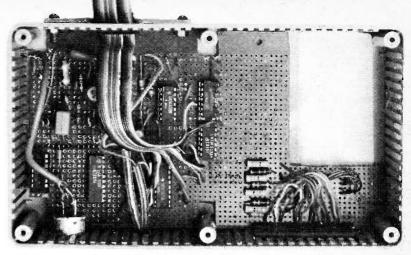
COMPUTER INTERFACE

The G.E.M.N.I.F. unit is connected to the ZX81 computer via the 23 + 23-way finger set on the printed circuit board at the back of the computer.

A program, written in machine code, instructs the computer to read the data being inputed from the ultrasonic receiver and calculate the trolley velocity. The computer will only accept a set number of pulses (in this case, 30) as the number of pulses corresponds to the distance travelled and this must be constant in order to perform the calculations.



Main unit plus Sinclair ZX81 computer.



View inside prototype showing circuit board and ribbon cable connections.

When the 30 pulses have been received, the counting procedure stops and the display routine commences operation.

Six values of velocity of the trolley are taken at intervals of multiples of five pulses. So, for example, after five pulses the first velocity is calculated, again after 10 pulses, after 15 pulses and so on up to 30 pulses.

Once received, the stored values are used to provide the outputs to the display. They are both on screen (if a TV is being used) and on the integral 4-digit display on the main unit (LED1).

DISPLAY

Initially the display reads four flashing "eights". Whilst the trolley is actually rolling down the track and the computer is receiving data, the display will automatically blank. When all 30 pulses have been received, the display will read "VEL1", whereupon it will-go on to dis-

play the velocity calculated from the first five pulses. It will proceed to display the next five values under the headings of "VEL2", "VEL3" and so on up to "VEL6".

It continues in a loop displaying these values of velocity until the unit is reset. It will then show flashing "eights" until the trolley starts sending pulses once again.

The gates IC4, IC5 and IC6 and the octal latches IC7 and IC8 are used to decode and display the information received from the ZX81 on the multiplexed display.

PROGRAM

As previously stated, the computer program for use on the ZX81 in conjunction with the G.E.M.N.I.F. Speed Computing System is written in machine code but unfortunately space does not permit the reproduction of it here.

However, interested parties can contact the editorial offices of EVERYDAY

ELECTRONICS and arrangements can be made to supply a transcript or tape copy of the program along with more detailed instructions.

CONSTRUCTION

No detailed constructional information is given for the building of the Speed Computing System as it was felt that, armed with the circuit diagram and background information, it should not prove too difficult to complete.

The original prototype was assembled in a plastic case $190 \times 110 \times 60$ mm housing a single stripboard circuit panel. The transducer was mounted in one side and the 23 + 23-way edge connector was soldered to a length of 14-way ribbon cable.

The trolley was converted from a standard three-wheel trolley of the type found in the physics laboratory and runs on any flat track of minimum length two metres.

BOOK REVIEWS

TELEVISION AND RADIO 1983

Editor E Price f Size

Eric Croston £3.50 limp

Size 190 x 230mm. 224 pages Publisher Independent Broadcasting Authority

ISBN 0 900485 43 3

W TTH our television viewing habits about to be changed beyond recognition with the advent of the fourth channel, breakfast viewing, satellite broadcasting and cable TV, this annually produced volume makes interesting reading.

1982 saw the birth of the first new national television service for 20 years—Channel 4—and 1983 will be the year of Tv-am, an additional three hours of programmes to wake-up to. This handbook discusses both these new services as well as the established ITV companies and sets out to illustrate the function of a nation-wide broadcasting network.

Many pages are dedicated to the programmes themselves, with sections on sport, drama, science, the arts and of course, news coverage, to name but a few. With over 400 illustrations,

many in full colour, the scope and quality of Independent Television can be seen.

Not to be forgotten is the constantly expanding independent local radio coverage, currently at 47 stations and new contracts are being granted all the time. Other chapters in this compulsive coffeetable book include advertising, finance, Oracle (teletext) and working in broadcasting.

G.P.H.

ELECTRONICS FEBRUARY 1983

PROJECTS

Twilight Warning
Radio Booster
Ultimum Computer
Interface Part 4

12V to 240V Inverter

FEATURES

Into The Real World—
interfacing micros
Programmable Unijunction
Transistors—a few PUTs
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PLUS

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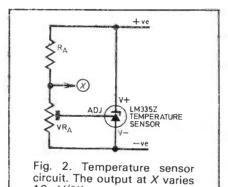
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Completed prototype of the Beehive Temperature Meter. The frame selector switch is on the right and the remote sensor input socket can be seen on the side of the case.

and increases by 10mV per degree Celsius. So at 0°C, the output would be 2.73V. Each output needs to be trimmed in using its associated preset, VR2 to VR11. All X's are commoned at the negative end of R6, but only one sensor is "on" at one time; which one is on, is determined by the setting of S1, which connects the sensor V- terminal to 0V.

The sensors should be set up after the monitoring unit and remote sensors are in



COMPONENTS

esistors		See
R1	1-8kΩ	Chan
R2	120Ω	SUOD
R3	10kΩ	
R4	330Ω	alk
R5	2.7kΩ	page 85
R6	5.6kΩ	
All ¼W ca	arbon ±59	6

Potentiometers

Resis

VR1 2.2kΩ miniature horizontal preset VR2-11 10kΩ miniature horizontal presets (10 off)

Semiconductors

TL430C programmable IC1 Zener diode IC2-11 LM335Z precision temperature sensor (10 off)

Miscellaneous

ME1 1V d.c. digital (Digitron model 8000) SK1 21-way socket 21-way plug 1-pole 10-way rotary PL1 S1 miniature on-off toggle S2 B1 6V, 4 × 1.5V (HP11) Pointer knob for S1; 21-way cable; battery holder for 4 x HP11

Guidance only excluding Approx. cost connectors and meter

later

REFERENCE VOLTAGE

Resistor R1 and IC1, a programmable

Zener diode, provide the reference voltage

for the main circuitry. VR1 is set

(programmed) to provide a stabilised

voltage across IC1. R2 and R3 are padding resistors for VR1 to give finer

R5 across IC1 produces a value of 2.73

volts at their junction. This feeds the

negative input of ME1, a 3-digit, 999mV

full-scale digital voltmeter. The signifi-

cance of 2.73 volts will become apparent

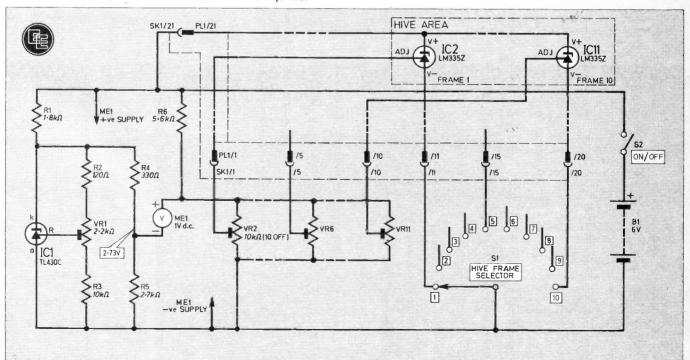
The potential divide effect of R4 and

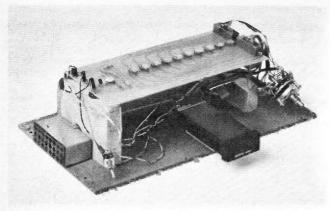
sensitivity to VR1 wiper variation.

SENSOR CIRCUIT The other side of ME1, the +ve input. is fed by the outputs (one at a time) from each temperature sensor circuit. Each sensor is wired as seen in Fig. 2. The output at X varies linearly with temperature,

Fig. 1. Complete circuit diagram for the Beehive Temperature Meter. The circuit is drawn to include ten temperature sensors, but this number may be reduced or increased as required.

10mV/°K.





The "baseboard" removed from the case to show positioning of the main components. Four HP11 batteries are located in a holder which is sited across the rear of the baseboard

Access to the temperature "trimming" potentiometers is

gained via a removable lid in the top of the case. The reference voltage "programming" preset VR1 can be seen on the far right.

their final positions. The sensor temperatures should be given time to stabilise. Place a thermometer beside a sensor when it is being calibrated, and adjust its preset until the readout on ME1 equals the thermometer reading. Repeat for all other sensors. Leave for a while and repeat the operation.

To read the temperature of each sensor, set S1 to the appropriate position. In the prototype, the reading was given directly in degrees Celsius, where the position of the decimal point could be selected on the rear of the specified meter. Any digital voltmeter could be used here which is capable of displaying up to 1V in 10mV steps. Sockets could be fitted in place of ME1 to allow any suitable meter to be plugged in when required.

USES

By monitoring the temperature in the hive, it is expected to be able to determine cluster movement (if any) during cold winter conditions. In such environments bees must cluster to keep alive. Disturbances, such as those from a nearby busy woodpecker could easily upset the bees and cause them to disperse and so lose the heat generated by clustering.

Regular monitoring of the hive temperature in various positions within the hive would alert the keeper who could then take early measures to remove the disturbance and pacify the bees.

An interesting plan now being considered by the team who produced this unit is to connect it directly to a personal computer. With suitably developed software there are many possibilities including a display of the hive spatial density.

For a detailed booklet on remote control — send us 30p and S.A.E. $(6'' \times 9'')$ today.

HOME LIGHTING KITS

These kits contain all necessary components and full instructions & are designed to replace a standard wall switch and control up to 300w. of lighting

TDR300K Remote Control £14.30 Dimmer Transmitterforabove £ 4.20

TD300K Touchdimmer £ 7.00 TDE/K

Extension kit for 2-way switching for TD300K £ 2.00 LD300K



THERMOMETER KIT

3-NOTE DOOR CHIME

Based on the SAB0600 IC the kit is supplied with all components, including loudspeaker, printed circuit board, a pre-drilled box (95 × 71 × 35mm) and full instructions. Requires only a PP3 9V battery and push-switch to complete. AN IDEAL PROJECT FOR switch to complete. AN BEGINNERS. Order as XK102

XK113 MW RADIO KIT

Based on ZN414 IC kit includes PCB, wound serial and crystal earpiece and all components to make a sensitive miniature radio. Size: 5.5 × 2.7 × 2cms. Requires PP3 9V battery. IDEAL FOR BEGINNERS. £5.00

COMPONENT PACKS

PACK 1 650 Resistors 47 ohm to 10 Mohm - 10 per

PACK 2 40 \times 16V Electrolytic Capacitors $10\mu F$ to $1000\mu F - 5$ per value £3.25 PACK 3 60 Polyester Capacitors 0.01 to $1\mu F/250V$

5 per value £5.55 PACK 4 45 Sub-miniature Presets 100 ohm to 1 Mohm 5 per value £2.90

PACK 5 30 Low Profile IC Sockets 8, 14 and 16 — pin — 10 of each £2.40

PACK 6 25 Red LEDs (5mm dia,) £1.25

MINI KITS

MINI KITS

MK1 TEMPERATURE
CONTROLLER/THERMOSTAT
Uses LM3911 IC to sense temperature (80°C max.) and triac to switch heater. IXW £4.00
MK2 Solid State Relay lideal for switching motors, lights, heaters, etc. from logic. Opto-isolated with zero voltage switching, solided with zero voltage switching. Solided with zero voltage switching. MK3 BAR/DOT DISPLAY
Displays an analoque voltage on a lineer 10 element LED display as a ber or single dot. Ideal for thermo-meters, level indicators, etc. May be stacked to obtain 20 to 100 element displays. Requires 5-20V supply.
MK4 PROPORTIONAL £4.50
TEMPERATURE CONTROLLER
Based on the SL441 zero voltage switch, this kit may be wired to form a "burst fire" power controller, enabling the temperature of an enclosure to be maintained to within 0.5°C. Max. load 3KW £5.55
MK5 MAINS TIMER
Based on the ZN1034E Timer IC this kit will switch a mains load on (or off) for a preset time from 20 mins. to 35 hrs. Longer or shorter periods may be realised by minor component

LCD 31/2 DIGIT MULTIMETER

LCD 37/2 DIGII MIUL INVEL ET 6 ranges including DC voltage (200 mv-1000) and AC voltage, DC current (200 mA-10 A) and resistance (0-2 M) + NPN 8 PNP transistor usin and diode check. Input impedance 10M, ize 155/x88x31 mm. Requires PP3 9v bettery, est leads included

ELECTRONIC LOCK KIT XK101

This KIT contains a purpose designed lock IC, 10-way keyboard, PCBs and all components to construct a Digital Lock, requiring a 4-key sequence to open and providing over 5000 different open and providing over 5000 different combinations. The open sequence may be easily changed by means of a pre-wired plug. Size: 7 x 6 x 3 cms. Supply: 5V to 15V d.c. at 40uA. Output: 750mA max. Hundreds of uses for doors and garages, car anti-theft device, electronic equipment, etc. Will drive most relays direct. Full instructions supplied.

ONLY £10.50

Electric lock mechanisms for use with

DISCO LIGHTING KITS

value-for-money This value-for-money kit features a bi-directional sequence, speed of sequence and frequency of direction change, being variable by

change, being variable by means of potentiometers and incorporates a masteridimming control

£13.50

£14.60

Only £8.00

DVM/ULTRA SENSITIVE



This new design is based on the ICL7126 (a lower power version of the ICL7106 chip) and a 3 ½2 digit liquid crystal display. This kit will form the basis of a digital multimeter (only a few additional resistors and switches are required-details supplied), or a sensitive digital thermometer (-50°C to +150°C) reading to 0.1°C. The basic kit has a sensitivity of 200mV for a full scale reading, automatic polarity indication and an ultra low power requirement-giving a 2 year typical battery life from a standard 9V PP3 when used 8 hours a day, 7 days a week

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DL 1000K

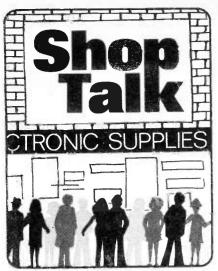
DLZ100K

This 3 channel sound to light kit features zero voltage switching, automatic level control & built in mic. No connections to speaker or amp required. No knobs to adjust - simply connect Only £11.95 (1Kw/Channel)

DL2100K
A lower cost version of the above, featuring undirectional channel sequence with speed variable by means of a pre-set pot. Outputs switched only at mains zero crossing points to reduce radio interference to a minimum.

60p

Optional opto input DLA1 Allowing audio ("beat")



By Dave Barrington

Sounds of ZON

All the latest reports indicate that probably the most popular gift this festive season will have been one of the ZX81 or Spectrum home computers. It is equally likely that thousands of people spent most of their holiday period mastering the wonders of their new acquisitions.

Having been hooked on their possible capabilities, owners will, by now, have soon spot-lighted any limitations and are eagerly seeking the add-on units available to expand their system. These will include printers and increased memory or RAM packs.

Although it does not expand the system, a wide range of sound effects can be added with the ZON X-81 Sound Box from Bi-Pak Semiconductors.



The unit is based on a three-channel-plus-noise sound chip and is so designed that the pitches and volumes of the three channels and overall attack/decay envelope can be controlled by simple BASIC instructions to the computer.

This means that piano, organ, bells, helicopters, lasers, explosions, space invaders, and so on, can be simulated and easily added to existing programs.

The circuit board is housed in a black plastics case with loudspeaker and manual volume control (in addition to programmed volume) and simply plugs in between the rear ZX81 "finger" set and its RAM pack and/or printer. The power supply for the unit is taken from the internal power supplies of the computer, via the computers bus outlet or finger set.

For use with the ZX Spectrum a special adaptor is required. This is available for the sum of $£5.50 \, \text{plus VAT}$.

Included with the ZON X-81 are operating details which take the user through its operation step-by-step and also contain a number of example programs of useful sound effects.

The ZON X-81 Sound Unit costs £25.95 including postage and VAT. For more details, readers should contact: Bi-Pak Semiconductors, Dept. EE, P.O. Box 6, Ware, Herts.

Drilling Machine

With the increase in designs built on printed circuit boards (p.c.b.), many readers must have tried their hand at making their own p.c.b.s and encountered particular problems. From the letters received, the most common and annoying of these would appear to be the accurate and time consuming task of drilling the component holes in the board.

This can now easily be accomplished with the latest addition to the Toolrange stocks. It is the ORYX Model B10 drilling machine designed for small-scale production work on p.c.b.s.

A feature of the drill is a built-in mains power supply which provides a low voltage source to an illumination "torch" for the work surface. The power supply also feeds the 24V motor, which direct drives a quill spindle at a speed of 14,500 r.p.m.

An adjustable magnifier enables accurate location of the workpiece to drill point and there are adjustable depth stops for hand feeding. The maximum chuck-to-table distance is $1\frac{1}{2}$ in (32mm).

Drills of up to $\frac{1}{8}$ in shank diameter can be held in the interchangeable collet chucks. Three collets are supplied as standard, allowing drills from 0-6mm to 2mm shank to be used.

More details and price can be obtained from Toolrange Ltd., Dept EE, Upton Road, Reading RG3 4JA.

CONSTRUCTIONAL PROJECTS

Alarms for Push-Bike and Motor Bike

When ordering the 4001 i.c. for the Alarms for Push-Bike and Motor Bike project, be sure to ask for the "unbuffered outputs" type. This will have the suffix UBE or UB after the type number. This device should be stocked by most advertisers but in case of difficulty it is listed by Maplin.

The mercury switch used in our models are available from Magenta Electronics. This switch is also available from Tandy shops (order code 275-025).

Note that mercury is a toxic substance, therefore be careful not to break the glass encapsulation when installing in the case.

The 9V buzzer used in the prototype was purchased from Tandy and is listed as D.C. Buzzer 273-052. Other suitable mechanical sirens are stocked by Electrovalue, Magenta, TK Electronics and J. Bull.

The 12V 30A miniature relay is available from Maplin and listed as YX99H (12V 30A

Relay). This relay is listed for use in automotive applications.

The type of case used is not critical, but should be made weather-proof from the elements. An old cycle lamp could be adapted and the bulb reflector area used to house the siren; the glass being replaced by some gauze.

Interval Timer

The 4-digit common cathode multiplex display, Type DL340M, used in the *Interval Timer* is available from TK Electronics.

They also supply a 9-digit Bowmar display that could be used in this project for the sum of 55p plus VAT. However, the pinning for this device is different and will have to be "hard-wired" to the board.

ZX81 Speed Computer System

The ultrasonic transducers for the ZX81 Speed Computer System are available from Electrovalue, Magenta, Maplin and Rapid Electronics. When ordering ask for transmitter and receiver transducers with pin terminations.

The 4-digit 7-segment common cathode display is a RS Components device, stock code 587-507. The diodes D3 to D5, Type BAX13, are stocked by Cricklewood Electronics.

The 23+23-way ZX81 edge connector is available from Maplin.

Beehive Temperature Meter

The "programmable" Zener diode called for in the *Beehive Temperature Meter* is likely to cause purchasing problems. This is a three-terminal type with excellent temperature stability. Two external resistors set the Zener voltage to any desired value in the range 3 to 30V.

The only source we have been able to locate for the TL430C is Maplin Electronic Supplies: Order code YY77J. However, RS Components stock an equivalent to this device which is listed as a programmable Zener diode and should be ordered as RS 283-227.

Once again, we would point out that RS Components will only supply to *bona fide* traders and readers should order this item through their component supplier.

The temperature sensor Type LM335Z is available from Europa, Maplin and Rapid Flectronics

A 12-way rotary switch with an adjustable end-stop will need to be used to make up the single-pole 10-way switch S1. Alternatively, a single-pole 12-way switch can be used with two tags of the switch left unconnected or joined to the tenth position.

EPROM Programmer for the Acorn Atom

It is not essential to use a zero insertion force (z.i.f.) socket for the *EPROM Programmer for the Acorn Atom* "Program" socket. This socket plugs into the low profile socket mounted on the circuit board and protrudes through the top of the case. The Vero 24-pin Miniwrap i.c. socket, type 200-2133B, has been found to have a low insertion and withdrawal force and could be used as a cheaper replacement for "expensive" z.i.f. sockets.



AST month we built amplifiers and oscillators. This time we'll extend our experiments from audio to radio frequencies and make a rather crude but reasonably effective a.m. receiver.

DISTORTION

To make it work we'll turn to good account something which is usually regarded as an unmitigated nuisance: distortion. Distortion, in amplifiers, means a departure from perfection. In a perfect amplifier the output would be an exact reproduction, in enlarged form, of the

input.

Let's take a closer look at what this means. The "signals" which most amplifiers have to deal with are mixtures. They contain many different frequencies (in a hi fi music signal, from about 20Hz to 20kHz). They contain mixtures of intensities (from pianissimo to fortissimo). There are all sorts of time relationships between components of the mixture and these may be important for stereo reproduction, for example. All these characteristics would be preserved in an ideal amplifier.

No real-life amplifier is perfect. It may not amplify all frequencies equally well. It may not preserve the original time ("phase") relationships between components of the mixture. Above all, it may not reproduce the intensities faithfully.

In electronic terms this means that the shapes of the signal waveforms may be distorted. For hi fi, this amplitude distortion is by far the most serious kind.

OVERLOAD

When we turned amplifiers into oscillators by feeding back the output to the input positively we saw how oscillations

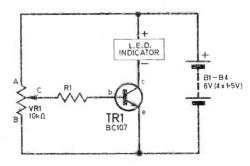


Fig. 5.1. A low value for R1 allows the transistor to be driven into overload.

can build up rapidly, starting with only the tiny amount of "noise" present in all circuits. This is amplified, fed back, amplified again, and so on, getting bigger and bigger.

Why doesn't this process of magnification go on for ever, eventually producing infinitely large voltages and currents? Because a point is soon reached where the transistors just can't handle larger signals. They overload. They can't deliver the output demanded of them.

If you repeat (in thought, at least) an earlier experiment (Fig. 5.1) you'll see why. Suppose we give R1 a lowish value, say $1k\Omega$. Then with VR1 slider at the positive end, several milliamps of current flow into TR1 base. TR1 will try to pass several hundred milliamps of collector current. It can't, because at a mere 20mA or so all the battery voltage is used up in the Indicator, leaving nothing for the transistor. So as you turn up VR1 the lamp reaches full brightness and gets no brighter as VR1 is turned up further. TR1 is no longer following the input signal (the base current).

LOW DISTORTION AMPS

In low-distortion amplifiers the transistors are given enough steady (d.c.) bias to turn them part-way on. The a.c. signals are added to the bias and as they swing positive or negative the transistors are turned on further or the reverse. Provided they don't get too large the output current and voltage can follow the input as required.

Figure 5.2 shows how the collector current and voltage of a transistor with a resistance load $(1k\Omega)$ vary as the d.c. base-to-emitter voltage is varied. With no input there is no collector current, hence no voltage drop across R1 so the collector voltage is the same as the battery voltage. As the collector current approaches 6mA nearly all the battery voltage is dropped by R1 and the collector voltage falls to zero, nearly. It can go no further, whatever you do to the base.

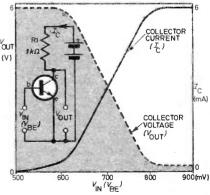


Fig. 5.2. How collector current and voltage change as the base voltage changes.

For minimum distortion the transistor is biased to say 700mV and a.c. signals added to the d.c. bias. So long as they are not too large the collector current never gets increased to its maximum (here 6mA) or reduced to zero and distortion is minimised (as shown in Fig. 5.3).

DETECTION

In amplitude-modulated radio there is a high-frequency "carrier wave" whose strength (amplitude) is made to vary in sympathy with the audio (programme) signals (Fig. 5.4 a and b). (For clarity, far fewer cycles of carrier are shown than with real radio signals.)

with real radio signals.)

To recover the audio from the modulated carrier, two steps are taken. First all the negative half-cycles of the carrier are suppressed (c). This leaves a train of pulses whose amplitudes are in step with the original audio. They form a sequence of samples of the audio.

To re-form the audio the gaps between samples must be filled in. This is done with the help of a capacitor which is charged by the pulses and retains charge when a pulse has ended.

There's a snag. A perfect capacitor will charge to the biggest pulse amplitude and stay like that, ignoring any smaller ones which follow. Goodbye, audio. To avoid this a resistance is connected across the capacitor, to allow it to discharge slowly. This gives it enough "memory" to bridge the very short time gaps between sample pulses while allowing the slow audio

TRANSISTOR DETECTOR

changes to be followed.

To use a transistor as a detector it is biased to a point such as X in Fig. 5.3. Negative carrier half-cycles then reduce the collector current towards zero but positive ones increase it according to their size. The collector current is a train of sample pulses like Fig. 5.4c.

In our receiver (Fig. 5.5a) TR1 is the detector. Its collector current is low (about $20\mu A$). Gap-filling is done by C3. This is normally charged to about 1-3V in this circuit. If this voltage changes it can restore itself slowly by charging more via R1 or discharging via TR1 collector and TR2 base.

The incoming signal appears across L1. TR2 amplifies the programme audio.

CONSTRUCTION

The terminal-block connections (Fig. 5.5b) are similar to last month's amplifiers. You need resistors of $33k\Omega$ and $220k\Omega$ and polystyrene capacitors of 39pF, 390pF and 1nF (1000pF). The other components you should have from before, except L1, which you make. For this you need a rather long piece of insulated hook-up wire (4 metres) but it can be re-used later if need be.

Another new component is a ferrite aerial rod. Get a blank (unwound) rod 100mm long (or longer). (If you have two shorter rods tape them end to end.)

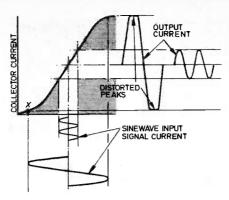


Fig. 5.3. Biasing the transistor to a partly-on state enables small a.c. signals to be amplified with low distortion, but large ones get "clipped".

The coil L1 is wound on a hollow cardboard or plastic tube wide enough for the ferrite rod to slide easily inside. I used an empty *Smarties* tube. Most ferrite aerial rods are 6mm or 9mm diameter, which is a lot less than my tube but that doesn't matter.

To wind L1, fix one end of your wire to your tube, leaving about 100mm free for connecting. Wind, spacing the turns to cover about 100mm. My coil has 42 turns but anything between 30 and 60 will do. Tape the loose end to the tube, again leaving enough spare for connecting to your terminal block.

If the rod goes right inside the tube tape a handle to it: I used the barrel of an old ball pen. Wood can also be used, but not metal.

OPERATING

Turn VR1 spindle fully clockwise. This puts the full $10k\Omega$ in circuit and TR1 oscillates. Moving the ferrite rod slowly into the coil should produce a sequence of whistles. Each whistle results from the

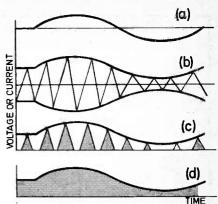


Fig. 5.4. Amplitude modulation of a radio transmission. (a) Audio (programme) signal; (b) Amplitude-modulated carrier frequency waveform; (c) First step in recovering the audio: the negative half-cycles of the a.m. waveform are suppressed; (d) Second step: The gaps are filled in, leaving the original audio plus d.c.

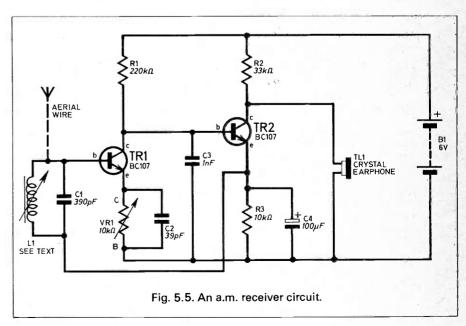
oscillation interacting with an incoming carrier frequency. This produces a *beat tone* whose frequency is the difference between the carrier frequency and the local oscillation.

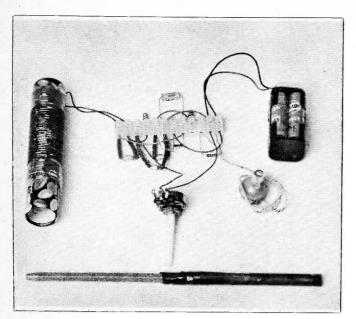
Tune in a strong beat then turn down VR1 until oscillation just stops. Slight retuning should then give you the station.

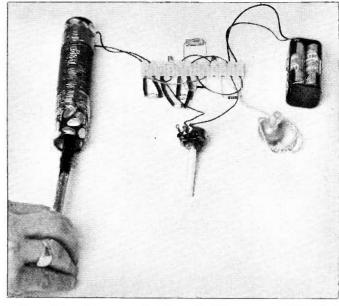
TUNING RANGE

My coil tunes from 1.2MHz to 2.5MHz, which embraces the high frequency end of the medium-wave band of 0.52MHz to 1.65MHz (550kHz to 1650kHz) where (in Britain) many of the local radio stations are to be found. After dark more distant ones can be heard.

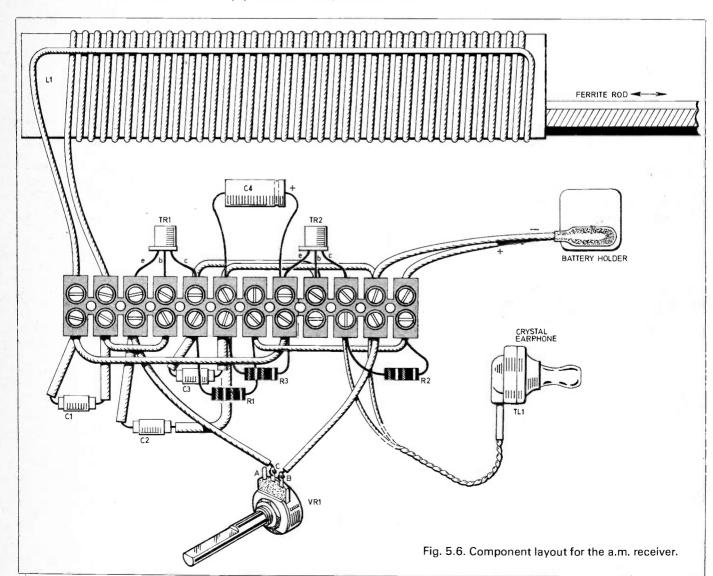
You may be able to pick up a few more with the aid of a metre of wire connected as a vertical aerial (dotted in Fig. 5.5). Have fun!







The a.m. receiver wired up on the screw terminal block with the ferrite rod taped to a plastic pen case. Moving the rod in and out of the coil which is wound on an empty "Smartie" tube, tunes the receiver.



TUNED CIRCUIT

L1 and C1 form a tuned circuit. Earlier we saw how inductors and capacitors can store energy as fields and charges. We learned that when combined with resistors these energy-storage devices can give time delays. Here, however, an inductor and capacitor are connected to one another. This has a striking effect on the time behaviour.

If C1 is charged, it discharges through L1, and a magnetic field is set up. When C1 is empty the field collapses, inducing a voltage which drives current from L1 into C1 (in the opposite direction from before). This to-and-fro oscillation of current goes on, gradually dying away because part of the energy is lost in the resistance of L1 each time current flows.

This damped (dying) oscillation occurs at one special frequency, the natural frequency of the tuned circuit. If the circuit is energised from outside, by a signal whose frequency matches the natural frequency the oscillation keeps on getting little pushes which keep it going. Just as somebody on a swing can go higher and higher by moving in step with the natural motion of the swing so small input signals can produce large voltages and currents in the tuned circuit.

As the signal frequency is removed from the natural frequency the effect diminishes. Only signals close to the natural frequency get built up. Hence a tuned circuit is frequency-selective. In our receiver we enhance the selectivity (and the sensitivity) by positive feedback (also called "reaction" or "regeneration").

SIDEBANDS

When a carrier frequency is modulated by an audio frequency extra frequencies on each side (above and below) the carrier frequency are created. These "side frequencies" are known collectively as sidebands and in an amplitude-modulated (a.m.) transmission they extend to each side of the carrier by the same amount as the audio frequency.

To transmit the full audio band, calls for some 20kHz of space on each side of the a.m. carrier. Tough, as the stations are spaced only 9kHz apart in the medium-wave band. Their sidebands overlap.

The only thing to do about it is to keep adjacent-frequency stations apart

geographically and remove the higher audio frequencies before transmitting, to restrict the sideways spread to, say, 10kHz, which still gives reasonable music

The carrier frequency itself contains no programme information. It is needed merely to "decode" the sidebands. Moreover, the sidebands duplicate the programme. Each one contains all the audio "information."

The carrier and one sideband can be removed before transmission leaving only a single-sideband. This halves the bandwidth of the transmission and avoids the "waste" of carrier frequency energy. The price is that special and at present expensive techniques are needed to get back the audio at the receiver. Ordinary detection doesn't work.

At very high frequencies (v.h.f.)

transmitters have a short range so if spaced well apart don't interfere with one another. It is then feasible to use a bandwidth-hungry but high quality form of transmission based on *frequency modulation*. The carrier has a constant amplitude but its frequency is made to vary in sympathy with the intensity of the audio signal. Again, ordinary detectors won't work but the solution is less expensive than for single-sideband (s.s.b.) modulation.

INVISIBLE COMPONENTS

There is a mystery about your receiver. Since it can be made to oscillate, positive feedback must be occurring. But how? Why should varying VR1 cause signals to be fed back from output to input?

Only TR1 is involved. The feedback is at radio frequencies. It is not from collector to base but from emitter to base. How? Through an invisible connection, formed by the internal impedance of the transistor between emitter and base.

Voltages across VR1 and C2 are fed back through this impedance to the tuned circuit. They get increased by a sort of transformer step-up property of the tuned circuit, which is just as well since the voltage at the emitter is less than the voltage at the base. This sustains the oscillation.

If the inductance of L1 is too large oscillation may only be obtained at the high-frequency end of the tuning range. The remedy is to reduce the number of turns or to reduce C1 (to, say, 150pF).

To be continued

COUNTER INTELLIGENC

The Future . . .

This is the time of year when Old Moore Young attempts to see the future. The delightful thing about forecasting the future is that you can go as far ahead as you wish.

I can imagine that no-one goes to work and we all sit at home and tap out instructions on our computer, with all the manual work being carried out by robots. A start has even been made in the manufacture of

Would the home constructor be able to tap out his requirements on his computer, and would we ever reach the stage where it could all be beamed down and land immediately on his bench, a la "Star Trek"? Then again with the progress of Physcokinetics he would not even have to press the buttons, only to think of the items he required. In time his arms and legs would atrophy, and he would move around in a programmed wheelchair, finally becoming a blob of jelly.

There are a few simple souls that worry about the computer taking us over completely, but just imagine the scientist programming his computer to carry out some original thinking, like discovering penicillin or cracking the D.N.A. code, the double helix, or the composer commanding it to compose a fugue that will excell all J. S. Bach's famous 48.

I will conclude by quoting a few words by Christopher Booker in a recent edition of the Daily Mail, talking about a friend. "Until a few years ago he was a brilliant computer expert. Financially, and in terms of job prospects he had the world of electronics at his feet. But he has thrown it all up to spend his time as a builder, as a potterand looking after his children-for no other reason that he eventually found the glitter-ing world of electronics utterly dead and

Now I don't subscribe to his view and neither, I am sure, do our readers but I do agree with the Irishman who said, predictions are unreliable, particularly those dealing with the future," but in case I have depressed our reader's, I decided I would cheer them up by looking backwards instead of forward.

. The Past

Looking for a "spark" from the past with an electronics connection that might interest readers is very difficult. But suddenly have an inspiration, Russian Ambassador I. M. Maisky.

He is particularly appropriate if only for the following reason. Ask any British schoolboy who invented radio and the answer will come back, "Marconi". Ask a Russian schoolboy the same question and he will answer with equal speed, "Popoff".

Now although the Russians may claim to have invented the aeroplane, the internal combustion engine and the telephone, their claim to the invention of radio has consubstance. Alexander siderable Stepanovich Popoff was transmitting radio messages up to 30 miles in 1898, and it was used by the Russian navy for communications between ships, and ship to shore. The Marchese Guglielmo Marconi established radio communication between France and England in 1899 and in 1901 succeeded in spanning the Atlantic.

In 1936 my father worked for a short time as a humble clerk in the Russian Trading Mission at Bush House, London. Perhaps the O.G.P.U. had a file on him, and noted that he had a son who was a radio technician at Philips. Now their ambassador in London his Excellency Ivan Mikhailovich Maisky had a Philips radio at the embassy and was getting bad interference on his reception from electrical apparatus in the vicinity. I was asked to provide a cure.

So one afternoon in the autumn of 1936 saw me clambering on the embassy roof in Kensington Palace Gardens armed with a long coil of screened cable. Fortunately I was able to greatly improve reception by cutting out the noise and I was asked to demonstrate this to his Excellency who thanked me warmly for my trouble.

SHORT INTERVAL TIMER POSITIVE OFF NEGATIVE ON BY A. P. DONIEAVY

THIS INSTRUMENT has been designed to measure time intervals of up to one second with a resolution of 0.0001 seconds, and with an accuracy of better than one per cent.

The stop/start can be activated by mechanical, electronic, or photoelectronic means, enabling the instrument to measure pulse widths, the time between two pulses of either polarity (the period), passage times of moving objects, and so on. The display is a 4-digit calculator type l.e.d. display.

CIRCUIT DESCRIPTION

A block diagram of the system is shown in Fig. 1. The incoming pulses are gated to produce the required polarity. Where the interval between two pulses is to be measured, the pulses are fed through a flip-flop. The resulting waveform is shown.

The rising edge of this pulse triggers a reset pulse which zeros the counter. The falling edge of the pulse (the end of the period to be timed) sends a pulse to the latch which then transfers the information in the counter to the display.

The clock oscillator, nominally at 9,999Hz, continuously feeds timing pulses to the counter. Thus the display at the end of the period shows the number of clock pulses counted during the timing pulse. For a 4-digit display, this will be a direct reading of the time in microseconds. The decimal point is to be before the most significant digit.

The reset and latch pulses are of microseconds duration, and therefore do not encroach significantly into the period to be timed.

POLARITY

Fig. 2 gives the circuit diagram. IC1 is a CMOS quad 2-input Schmitt trigger NAND gate. The input is fed in via IC1d and any effect of switch bounce or input signal jitter will be minimised by the switching characteristics of this type of

gate. S1 is set for the appropriate signal polarity, this switch brings in an additional Schmitt gate, IC1c, for the NEGATIVE input pulse position.

The signal is then fed either through IC2 or direct to IC3 depending on the position of S2. IC2 is a dual D-type flipflop, only one of which is used in this circuit. This flip-flop changes output state every time a rising edge signal is applied to pin 3. So if the time interval between two successive pulses is to be measured, the first pulse will switch the output to high, and the second pulse will switch it to low again, thereby activating the RESET and LATCH pulses in sequence.

RESET

The circuitry associated with S2a (C1 and R2) is designed to put the flip-flop into the right state when initially switched into this mode, by applying a momentary positive pulse to the RESET input pin 4.

The leading edge of the timing pulse is coupled via C3 and R3 to pin 12 of IC4 to produce a momentary RESET pulse. The trailing edge is fed to pins 11 and 5 of IC3, a CMOS dual monostable multivibrator, to produce the LATCH pulse from one of the monostables.

The other monostable has an l.e.d. connected to the output which illuminates when a new reading is made. The l.e.d. remains alight for a period determined by the values of R4 and C4, which for the

given values is a few tenths of a second.

IC1a and IC1b form the clock oscillator to provide the timing pulses. The frequency is set by VR1 and should be 9,999Hz.

The counting, latch and display drive functions are all done by IC4, a 74C925. IC5 contains five individual transistors, four of which are used as current sources for the digits of the display. The display is a dual-in-line (d.i.l.) 4-digit common cathode display for a calculator. The decimal points are not used, since extra circuitry would be required to drive them.

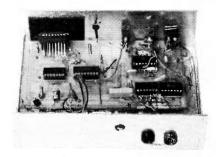
Finally, a stablised power supply of 5V is obtained using a voltage regulator, IC6. This is necessary to stablise the frequency of the oscillator and for IC4, which requires a +5V supply. The instrument is driven using a 9V PP3 battery. It can also be driven using a 9V mains adaptor via a suitable socket. The power supply circuit diagram is shown in Fig. 3.

COMPONENTS

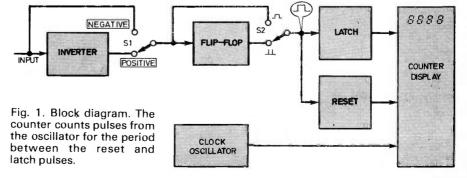
IC5 may be replaced by an LM3086, a pin-for-pin equivalent of the CA3046. VR1 should be a multiturn preset potentiometer.

The display used on the prototype is a DL-340M common cathode calculator display. This is a d.i.l. display with four digits, all of which were used in this project. Other displays may be used, the best being d.i.l. for easy mounting. The p.c.b. layout may need to be modified to accommodate different pin configurations. Individual common cathode displays may also be used.

With the exception of C2, the values of the resistors and capacitors used are not at all critical, and the other values may be used. R7 to R13 affect the brightness of the display. Increasing the value will decrease the brightness by reducing the current, but will increase the battery life.



Rear view with top cover removed.



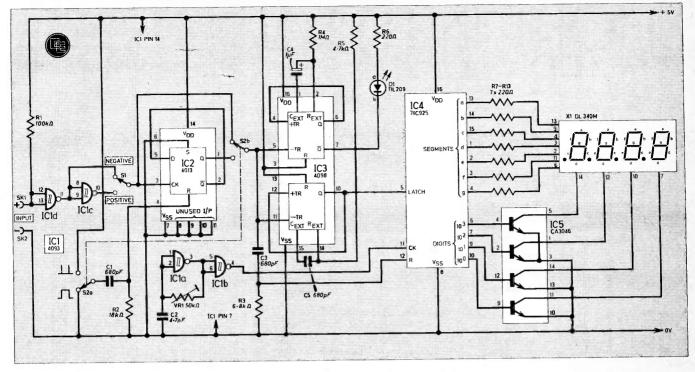


Fig. 2. Main circuit diagram of the Short Interval Timer.



CIRCUIT BOARD

The unit is built on a single sided p.c.b. 138×75 mm, the layout of which is shown in Fig. 4. Holders are used for the i.c.s as this prevents any soldering damage and facilitates any necessary debugging.

The display is mounted using a special 14-pin d.i.l. holder with both sets of pins bent at 90 degrees to allow the socket to sit at right angles to the board. As previously mentioned, if a different display is used, the p.c.b. connections may need to be altered. This should not be beyond the capabilities of the constructor.

If a multi-digit display is used, for example, from a calculator, this can be mounted behind the front panel of the instrument case and hard wired to the p.c.b. This type of display can have anything up to ten digits of which only four will be used.

CASE

The unit is housed in an aluminium instrument case, $45 \times 105 \times 150$ mm, with a vinyl covered lid, a type readily available. The p.c.b. is mounted using

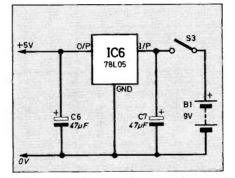


Fig. 3. Power supply circuit diagram. Optional external power socket not shown.

three spacers approximately 13mm long. Holes for the three switches and l.e.d. are drilled on the front panel, and a window for the display approximately 13 × 34mm must also be cut. Ensure that the position of this window is such that the display appears to be central.

The input terminals are two 4mm banana sockets located at the rear of the instrument. If the instrument is also to be powered using a 9V calculator adaptor, a hole for a 2.5mm jack socket will also be required.

The battery is secured to the box by a double-sided adhesive tab. D1 is mounted with the standard black bezel clip.

A red filter is glued to the inside of the front panel to increase the display contrast. The switches are labelled using "Letraset" or similar dry print transfers and secured with a clear varnish to make them more durable.

If a mains adaptor jack socket is used, ensure that the terminals are wired so that

the battery is disconnected when the jack plug is in place and that the polarity is correct.

A diagram showing all interwiring is given in Fig. 4.

CALIBRATION

If a calibrated frequency source is available, with an 0-10Hz range, then this can be used for calibrating the instrument. With S2 set to measure the time between two pulses, the period is measured. This period is the reciprocal of the input frequency.

So, for example, a known 2Hz signal is applied to the input and VR1 is adjusted until the display reads 5000 (0.5 seconds).

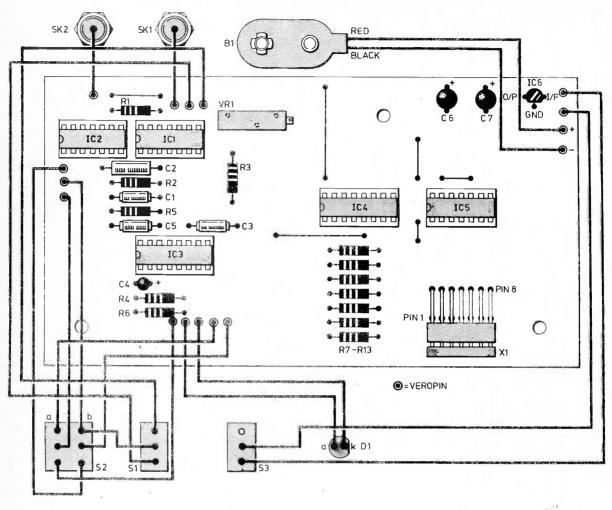
In the absence of a calibrated source, it is possible to use the 50Hz mains frequency. Using the unrectified secondary output of a low voltage mains transformer in the range 5 to 12V fed into the input via a 10 kilohm resistor, adjust VR1 until the display reads 0200 (0.02 seconds).

When using this latter method, the instrument is taking 50 readings per second, and the last digit may appear to change rapidly.



Front view clearly showing the labelling of the three switches.

SHORT INTERVAL TIMER



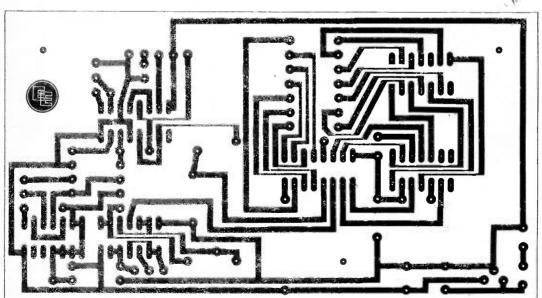


Fig. 4. Full size printed circuit board artwork and component layout. Topside also shows interwiring details. Note that switches are shown as if the front panel had been folded flat.

COMPONENTS

Resistors		See
R1 R2 R3 R4 R5 R6–13 All 1 /W ca	$\begin{array}{c} 100 k\Omega \\ 18 k\Omega \\ 6 \cdot 8 k\Omega \\ 1 M\Omega \\ 4 \cdot 7 k\Omega \\ 220\Omega \ (8 \\ rbon \pm 5\% \end{array}$	

Capacitors

C1,3,5	680pF polystyrene
	(3 off)
C2	4.7nF polyester
C4	1μF 6·3V tantalum
C6,7	47μF 10V tantalum
	or electrolytic (2 off)

Semiconductors

D1 IC1	TIL209 red l.e.d. 4093B cmos quad
	2-input Schmitt triggered NAND gate
IC2	4013B cmos dual D-type flip-flop
IC3	4098B cmos dual
IC4	monostable 74C925 cmos 4-decade counter/driver with multiplexed 7-segment
IC5	outputs CA3046 or LM3085 silicon npn transistor
IC6	array 78L05 +5V, 100mA regulator
X1	DL-340M 4-digit common cathode multiplexed display in d.i.l. package (see text)

Miscellaneous

S1,3

	(2 off)
S2	d.p.d.t. miniature toggle
VR1	50kΩ multiturn preset
SK1	4mm bañana socket red
SK2	4mm banana socket
	black
B1	9V PP3 battery
Aluminiun	n instrument case, 150
× 105 >	45mm; single sided
p.c.b. 140	x 76mm; red display
filter appre	ox. 50 x 30mm; battery
clip; 16-p	in d.i.l. holder (2 off);
14-pin d.i	.l. holder (3 off); vertical
mounting	14-pin d.i.l. holder (for
X1); 7/0	2mm wire; rubber feet
(4 off); mc	ounting hardware.

s.p.d.t. miniature toggle

Approx. cost **Guidance only**

CIRCUIT OPTIONS

The constructor may wish to extend the time range, for example, 0 to 10 seconds. This can be achieved by either changing VR1 or C2. The oscillator frequency is inversely proportional to the resistance and capacitance, so to have a maximum period of 9.999 seconds, C1 would have to be 0.047 microfarads or VR1 would have to be 500 kilohms.

Should more than one range be required, the appropriate number of preset potentiometers will need to be added, and another switch to select the range.

TROUBLE SHOOTING

The following should be of use to identify any problems which may occur when the instrument is first switched on.

The four digits should light, or possibly a random number appear, when first switched on. Any missing digit or segment will be either a wiring error or a fault in IC4 or IC5.

A momentary short of the input terminals should cause D1 to flash (S1 in NEGATIVE position, S2 in PULSE WIDTH position). The new display reading should coincide with this flash. If this does not occur and the oscillator is known to be working, the function of IC4 may be checked by putting pin 5 (LATCH input) of IC4 to $V_{\rm DD}$.

However it will be necessary to disconnect pins 10 and 12 of IC3. Do this by putting IC4 in another d.i.l. holder and bend pin 5 of this holder out so that it protrudes from the side, and plug into the board socket.

When the new protruding pin 5 is connected to $V_{\rm DD}$ the display will count continuously. If it does not, and assuming the RESET pin 12 of IC4 is at 0V, the i.c. may be considered faulty.

If the timer gives completely wrong readings, then it is likely that there is a wiring error to one or both of S1 and S2.

USING THE TIMER

The polarity of the pulse to be measured must be known before the measurement is taken.

The switches can now be set to the appropriate mode. These are as follows:

To measure the duration (pulse width) of a positive pulse, S1 is in the POSITIVE position (up) and S2 must be down. pointing to the pulse symbol.

To measure the duration (pulse width) of a negative pulse, S1 is switched to the NEGATIVE position (down). S2 is unchanged.

To measure the interval between two positive pulses (the period), S1 is set to the POSITIVE position (up) and S2 is switched to the up position, pointing to the symbol representing two pulses.

To measure the interval between two negative pulses (the period), S1 is simply switched to the NEGATIVE position (down) and S2 is unchanged.

SOURCE IMPEDANCE

If a measurement is being taken from a circuit or piece of equipment with an output impedance of greater than 33 kilohms, then R1 will have to be substituted with a higher value, for example one megohm.

The Short Interval Timer has been designed for pulses of 5V peak and signals of greater amplitude require an additional resistor in series with the input. A 10 kilohm resistor is sufficient, and the reason that it is required is that the input protection diodes on IC1d will start to conduct at voltages greater than the positive supply rail.

PHOTOCELL

If a phototransistor is to be used to trigger the timer, for example a TIL81, then all that is required is to connect the collector to the positive input terminal (SK1) and the emitter to the negative input terminal (SK2). R1 acts as the load resistor for this device and may need adjustment to suit the transistor used.

If the time interval to be measured is the interruption of incident light falling on the phototransistor then the Timer must be set up for the measurement of a

positive pulse width.

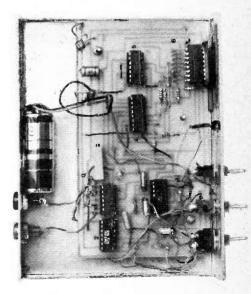
To measure the time interval of light actually falling on the phototransistor, the Timer is set to measure a negative pulse

A microswitch can also be used to trigger the timer. In this case the Timer would be measuring the interval of a mechanically moving object which activates the switch. Using the normally open contacts, put in the NEGATIVE position; and for the normally closed contacts, in the Positive position.

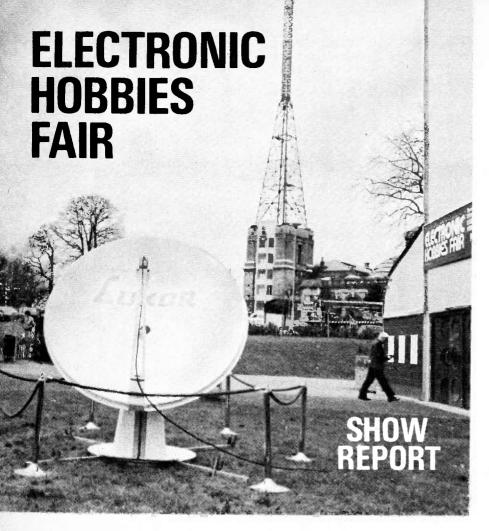
APPLICATIONS

An example is to use a phototransistor to measure the shutter speed of a nonautomatic camera. With a light in front of the lens and the phototransistor behind the lens, the shutter speed can be checked for accuracy. The waveform seen by the input would correspond to measuring a negative pulse width.

The Timer can also be used to measure the r.p.m. of a rotating object. For example, a white strip painted or stuck on the rotating object will reflect light to the photocell (held in close proximity) with every revolution, and thus giving a continuously up-dated display of the time between each revolution, which for a constant speed can be converted to r.p.m. by dividing 60 by the time displayed.



The finished prototype p.c.b. assembly mounted in the case. This layout differs slightly from that given in Fig. 4.



FROM the opening of the doors on Thursday, 18th November, the Alexandra Pavilion in North London became a bustling scene with visitors of all ages circulating among the stands which offered a wide range of products from components of all descriptions to complete units and instruments representative of all branches of electronic technology.

SPECIAL ATTRACTIONS

The crowds swelled appreciably on the Saturday and Sunday when the family presence was particularly noticeable. Non-technical members of visiting families soon discovered plenty to amuse and interest themselves amongst the special attractions. Handel's Water Music, Bach's Toccata and Fugue and Jeremiah Clarke's Trumpet Voluntary issued forth at regular intervals from the mighty four manual electronic organ built

by a member of the Electronic Organ Constructors' Society. There was no dearth of volunteer players hence the nearly continual flow of music from this example of the king of instruments.

Ham radio and holography may seem poles apart but they occupied adjoining stands as if to demonstrate how wide and divergent are the activities to be found within the sphere of electronics. Holographic Developments demonstrated 3D holograms and offered for sale products related to this the very latest field for the hobbyist to explore.

Radio is as old as electronics, and the Radio Society of Great Britain carried the banner for this most popular hobby, aided and abetted by two other satellite organisations, the British Amateur Radio Teleprinter Group and the North London Raynet Association. The latter participated actively in guiding fellow hams to the Fair via VHF and UHF links.

ELECTRONIC MILESTONES

Approaching the entrance to the Alexandra Pavilion visitors passed between two metallic objects each highly significant in technology terms. On the left, rising from the East Tower of the old Alexandra Palace, is the mast that radiated the world's first TV service. On the right, on the ground just before the entrance to the Pavilion, stands a 2-metre diameter dish aerial designed to receive Russian national TV programmes from a satellite hovering over the USSR.

Forty-four years of electronic history neatly symbolised by two aerials standing less than 100 yards apart in the grounds of Alexandra Palace, North London.

Nearby the Army had a display of its own very special kind of radio equipment and this was demonstrated by young members of the Royal Signals from the Army Apprentice College. Exhibits included an operational radio station, teleprinter and a "rolling road" on which visitors were invited to ride a bicycle and try their luck at beating the speed record.

Just across the way in a railed off arena, radio was being put to another use by Model Land. Model cars, buggies, helicopters and planes were put through their paces and demonstrated their manoeuvrability at high speed when under the control of a skilful operator.

On the first day Leonard Bliss of Model Land, reported a near disaster when his radio-controlled helicopter went out of control and "crash landed", by providence, on his stand. Damage to the stand was minimal and fortunately they had a standby helicopter to hand.

ROBOTS

Not far away stood a family of life size robots belonging to Advanced Robotics. Now and again the father of the quartet felt the urge to take a wander around the neighbouring area of the Pavilion, and in one instance even ventured outside to greet visitors to the show. During the course of these perambulations he chatted up visitors much to their amazement.

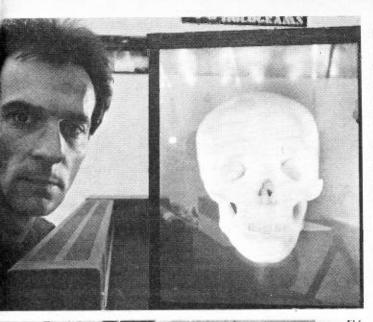
These robots were featured in the ITV programme, "The Six O'clock Show" on the Friday evening. The activities of the outside broadcast unit supplied an added attraction to visitors who stayed on to watch this live telecast from the Fair.

Over on "the other channel" the BBC's important role in education was illustrated by displays of the BBC Computer, electronic teaching modules and other items featured in BBC sound and TV programmes. The stand was crowded most of the time with visitors glued to the demonstration video film monitors.

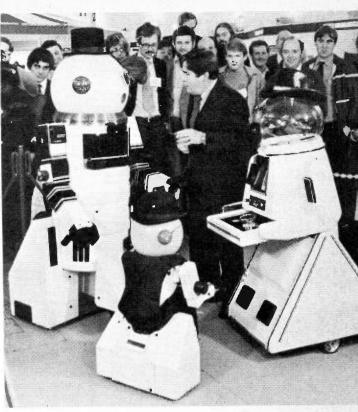
BBC representative Robin Gwyn said the numbers of visitors and the interesting enquiries concerning the BBC's activities in education was, in his experience, unprecedented. Many parents wanted guidance about computers, often because they felt the need to keep up with their children. Others involved in teacher/parent associations wanted advice regarding suitable computers for schools.

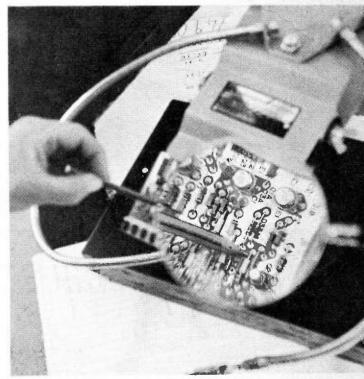
This general view of a thronged aisle indicates the popular appeal of the Fair.











(Top left). Skulduggery by Holographic Developments. Holographer Ken Harris is pictured beside a dramatically realistic 3-D hologram.

(Top right). The family of robots being interviewed during the ITV telecast from the Fair.

(Left). An Army P.E. instructor advises a visitor before she attempts to break the speed record! This racing cycle was linked to a microprocessor to record the rider's speed.



COMPUTERS

Talking of computers, the visitor could scarcely step in any direction without soon encountering the screen of a VDU alive with readout from one of the well-known personal computers.

There was no shortage of eager fingers to manipulate the keyboards, whether a game or a more professional problem was being tackled. It was not uncommon to see a youngster clutching his own hand written program eagerly awaiting the chance to try it out.

Maplin had a particularly impressive array of computers, including Atari, Vic and Dragon modules, and these attracted crowds. Roger Allen of Maplin reported a brisk trade in computers, software and publications.

Amongst other exhibitors featuring personal computers were Electrovalue, Chromasonic Electronics, Midwich, Kansas City Systems, SEDAC and Army Apprentice College, with some selling software and computer accessories.

COMPONENTS AND KITS

The constructor and practising designer or engineer alike had much to feast their eyes on. The whole gamut of electronic components was on display, as well as tools and instruments, materials and cases. If it was a new soldering iron you were after then a visit to the Light Soldering Developments stand was the place to call.

Bargain packs were offered by many components firms.

Calling in on the Bi Pak stand, Bill Baines informed us that apart from component packs, cases and their new ZON X-81 Sound Unit were in big demand.

The ZON X-81 is a sound effects box that plugs directly into the back of the ZX81 home computer. This allows the user to produce such sounds as: helicopters, lasers, explosions, space invaders, and so on.

Electronic kits provide a convenient alternative for the constructor in a hurry as well as for the less experienced to assemble. Some fine collections of kits were on display covering electronic gadgets such as musical doorbells to hi fi



Bargains galore! Brisk business in the market place area.

(Left). The Roadrunner wiring system being demonstrated.

equipment. Vellerman, for example, offered a good selection of kits and furthermore had a working model of each on display so that the intending purchaser could see and try for himself the operation of a completed assembly.

Electronic ignition is still extremely popular judging by the sales reported by Peter Biddle of Sparkrite. Apart from an ignition system, they also market a comprehensive anti-theft device.

CATALOGUES

Apart from cash sales, most retailers disposed of considerable numbers of their catalogues. Taken away by visitors, these will be consulted time and time again in the coming months and many an order will be placed by this means. Indeed, some firms have already reported a large amount of business subsequent to the Fair that must be directly attributed to the selling power of the catalogue.

Maplin inform us they sold a few thousand copies of their mammoth opus. Another household name, Vero Electronics, gave away several thousand copies of their catalogue. Mrs Mary Pearce, who was in charge of their stand, told us they have high hopes of extending the uses of their well-known products as a result.

Talking of follow-ups, Global Specialities were delighted with the large proportion of professional enquiries and business buyers amongst visitors to the Fair. Managing director Tina Knight told us they are confident that large amounts of business will materialise from the nature of the enquiries at their stand.

Global's breadboard system, for one, attracted the attention of the professional and looks like being introduced into more than one industrial R&D department as a result. Global also report lively business in their educational kits. Dads were much in evidence buying these kits for their sons (a likely story!).

Global seized the opportunity (and challenge) offered by their commanding site at the front of the house: a warm introduction to the Fair was induced by the carefully arranged and festively decorated

MARKET PLACE

One corner of the Pavilion was reserved for "small holders" of the electronic retail trade. Trestle tables laden with component goodies of all shapes and forms, and enticing bargain bags of capacitors and the like, attracted crowds.

Everyone likes a rummage, it seems! Except perhaps certain component retailers who felt that business on their stands had suffered as a consequence of the market trading area. The more typical view, however, was that the market area added a desirable touch to the whole scene, and since no-one can obtain all his requirements from these less orthodox trading pitches, the retailer with a wide selection of first class components will always be sought out.

WE THREE

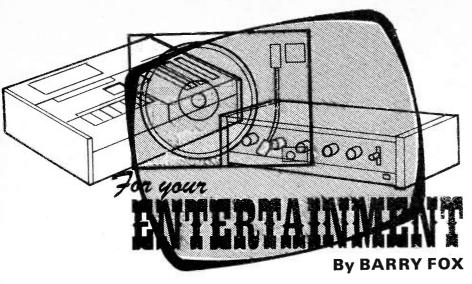
The centre piece in the Pavilion was a large circular stand shared by the three sponsoring magazines, PE, PW and EE. The EE stand featured the *Introducing Electronics* series for the beginner using solderless techniques, and this attracted considerable attention.

Another focal point was a small lathe under the control of the BBC Computer. Here visitors could see a length of plain wooden dowel being machined to a complex shape—the pattern being displayed on the VDU.

The five-digit combination lock provided a source of frustration to countless visitors as they strived to "crack the safe". However, we must congratulate Miss R. Mitchell of Gidea Park, Essex on showing her brother, and others on how it is done.

A new collection of test instruments for the hobbyist, musical effects units, computer add-ons and a variety of miscellaneous pieces of equipment were also on show.

Winning projects were demonstrated on the SEDAC stand by students from the successful schools in last year's competition. Considerable interest was aroused in this national competition for secondary schools sponsored by Mullard Ltd and Everyday Electronics, and requests for entry forms for the 1983 contest were frequent.



Mysteries Of Hi Fi

A fascinating gadget is being demonstrated by Sony at hi fi shows and trade exhibitions around Europe.

A transparent plastic case stands on a tape-recorder. Inside the case an ordinary audio cassette hangs suspended by a couple of thin black threads.

There's no visible drive for the suspended cassette, no playback heads and no electronic circuitry. But as you operate the controls of the tape-recorder underneath, the floating cassette runs, re-winds, fast winds and produces music through an unseen amplifier and loudspeaker. It's all very eerie.

In fact the demonstration is intended to show off the extraordinary small size of the tape drive mechanism used in a Walkman portable stereo. The two black threads suspending the cassette are electric wires which feed power to a Walkman drive motor attached to the underside of the suspended cassette. You can just see it if you peer underneath and up at the right angle.

The controls of the recorder underneath operate not only the main recorder drive, but also (via the threads) the unseen motor underneath the suspended cassette. So, as the main recorder re-winds, so does the suspended cassette, and so on.

The sound you hear is actually coming from the recorder, not the suspended cassette. So in some respects the demonstration is a phoney. But it's an interesting talking point, and it does prove the point that tape drive mechanisms are now incredibly small.

Jargon Generator

A few years ago I wrote about a computer jargon generator. Now there's a hi fi jargon generator. It's published, but so far only for the benefit of the trade, by Celestion Loudspeakers.

Like all jargon generators it's simple to use. There are three columns of technical terms, each with a number. At random you take one phrase from each column and drop it casually into a sentence.

Celestion reckon that any one of the 9000 permutations of hi fi expertise offered by their generator is enough to silence a hi fi salesman, confuse a friend, or serve as a basis for an impressive letter to a hi fi magazine. Here are a few examples:

"I am having trouble with my (1) transient (2) mosfet (3) bucket brigade". If that doesn't work you could try. "I've now got doubts about (12) digitally processed (16) comb filtered (7) linear interpolation.

Shopping In Tokyo

Digital Audio

I spent a few hours shopping in Tokyo recently, which is always an interesting experience.

Digital audio Compact Discs and players went on sale in early October. It's too early yet to say whether any are actually being bought. The price is low compared to that expected in Britain when Compact Discs go on sale in March 1983.

In Japan the price of a player is under £400 and discs around £7 each. It's likely that players will cost between £500 and £600 in Britain and the discs £10 each.

This is why many people feel that Compact Discs will take longer to take off here than the trade originally expected. It will be too expensive for all but the most dedicated hi fi enthusiast.

Video Disc

The Philips LaserVision video disc system has been on sale in Japan since October 1982. But it is backed over there only by Pioneer and it's very hard for any one company to do anything alone in Japan.

As a result LaserVision hasn't been selling well. This is also one reason why the rival VHD system, developed by JVC and backed by twelve Japanese companies, hasn't yet been launched.

Although many hi fi and video shops in Japan still have Pioneer LaserVision players on working demonstration, they don't seem to attract much interest. Neither, incidentally, did the Compact

Hot Shot

Pooled information will often solve problems you hadn't yet recognised.

Recently a chance remark by a studio engineer raised an interesting question. The studio had built a digital timer and clock from a package of chips. One day one of the junior engineers was idly playing with an anti-static pistol that's kept in the studio to clear the static from gramophone records after they have been wiped clean. He was "playing guns" with the pistol, pointing it close to the clock and squeezing the trigger. Soon afterwards the studio found that the clock wasn't working. All the chips had blown.

No-one will ever know for sure whether it was the ion stream from the pistol that blew the chips or whether there was some quite unrelated fault, like, for instance, a power supply failure. Since then I've tried gunning an old calculator with an anti-static pistol and it's done no damage. But there's a lingering doubt.

Have any readers ever encountered blown chips after using an anti-static pistol near them? If so, let's hear about it and pool information.

It's obviously important because more and more gramophone turntables now incorporate integrated circuits and microprocessors, and many people use anti-static pistols to clear the charge from a disc while it is sitting on a turntable.

Disc demonstrations I saw. But this is probably because many customers in department stores and record shops where Compact Disc is being demonstrated, didn't really recognise the significance of what they were seeing and hearing.

Solar Power

The most interesting gadget I saw was a solar power pack, costing around £25. It's a panel of solar cells, with rechargeable Nickel Cadmium batteries, and in sunshine it delivers enough power to drive a portable radio or cassette player.

One shop was demonstrating it with one of the new portable Sony flat screen televisions. These cost a little over £100 and use a flat, squashed cathode ray tube similar to the type originally proposed by Sinclair of Britain. The rechargeable Ni-Cads keep the set running while the sun goes behind a cloud.

In-Car TV

Probably the daftest new development in Japan, is the craze for in-car television. I saw several electronics shops showing in-car TV systems. One even had a mock-up of a car with a hi-fi stereo and TV installed alongside the driver's seat.

I've always thought that pocketportable TV's were a pointless extravagance. After, all who wants to watch TV while they are walking down a road. But the idea of anyone watching TV while they drive is really ridiculous:

By Pat Hawker, G3VA

Exit The Pirates?

In Radio World-November 1982, in discussing the complex legal maze surrounding so much of our use of modern electronics technology, I observed that the Wireless Telegraphy Acts had fallen into disarray although new legislation was threatened.

Over recent years "pirate" (unlicensed) use of radio transmitters has multiplied many times over. Although there have been a number of successful prosecutions brought by the authorities, these have been rendered very difficult by what amounts to the need to catch offenders "microphonehanded" so to speak.

The 1967 Act, for example, made it illegal to import certain types of 27MHz CB equipment, but there proved to be glaring loopholes in the legislation and there was nothing to prevent the open sale and advertising of equipment that would have been illegal to import in working order or manufacture. Then again the fines imposed on "pirates" have often made it hardly worth the bother of bringing prosecutions.

Piracy

Apart from the continued sale and use of amplitude-modulated CB equipment, and equipment capable of operating outside the 40 UK 27MHz CB channels, there are many other forms of "piracy" including illegal operation that causes interference to broadcast services or other licensed users of radio frequencies.

These include the many "broadcast" stations on both medium-wave and v.h.f./f.m. that can be heard most weekends and evenings in many cities. In London some have been active for many years despite occasional prosecutions and seizure of equipment, often using tapes which are produced openly.

There are also a number of so-called "international CBers" using modern h.f. s.s.b. transceivers on frequencies between 6600-6700kHz. Then again there are the pirate or "bootlegger" stations who operate inside the amateur bands using the callsigns of licensed radio amateurs. (I have had the experience of listening to someone calling CQ de G3VA in execrable morse!)

There are the CBers who have been infesting the 28 to 29.7MHz amateur band using equipment designed originally for both amateur and CB use, or else modified for this band. There are also a number of misguided individuals who for some five years or so have taken a delight in interfering with and abusing the use of amateur 144MHz "repeaters".

All such pirates may soon be confronted with altogether tougher opposition.

Telecommunications Bill

In November 1982, terms of a new "Telecommunications Bill" were published. This is an extremely long and complex Parliamentary Bill aimed primarily at preparing the way for the "privatisation" of British Telecommunications, abolishing many of its monopoly powers and setting up a Director General of Telecommunications to license firms wishing to provide telecommunications services. But one part of this Bill-Part V, which runs to 150 pages—is concerned with amendments to the Wireless Telegraphy Acts 1949 to 1967 "and to make further provision for facilitating enforcement of these Acts".

While, of course, the Act may be modified during its passage through Parliament, the Bill as published promises to be very tough indeed on pirates and those causing interference to other services. In particular it proposes that the restrictions on specified apparatus should be extended to cover not only use and importation but

Manufacture (whether or not for sale). Manufacture is defined as including "construction by any method and the assembly of component

Selling or offering for sale, letting on hire or offering to let on hire, or indicating (whether by display of the apparatus or by any form of advertisement) one's willingness to sell or let on hire.

Having in one's custody or control. This is thus far more wide ranging than Section 7 of the existing 1967 Act and creates entirely new offences

It is clearly aimed not only at stopping the sale of equipment which it would be illegal to use-but also at making "possession" of such equipment a breach of the law. It will also be possible to specify equipment by the use to which it is put, rather than the frequencies on which it can

There are clauses relating to seizure and disposal of equipment; arrest without warrant; a clause making it an offence to obstruct intentionally the seizure of equip-

There is little doubt that if Part V of the Telecommunications Bill becomes law in a, few months time it will immensely strengthen the power of the authorities to clamp down on radio pirates and the use of unauthorised frequencies. This is long overdue, although in giving the State so much more power it does, or should, impose on the authorities an equal obligation that the licensing process should be reasonable and

It also strengthens the case for amateur radio "novice" licences and legal low-cost "community" radio broadcasting.

Simple Radio—Costly TV

The advent of video cassette machines combined with rental of tapes (whether "pirated" or genuine) has tended to make viewers forget the very high cost of producing high-quality TV programmes. With VCR it may appear to cost only one or two pounds an hour to have your own programme.

In the talk about 30-channel cable TV networks, and whether these should be based on co-axial cable or optical fibres, most of the debate has been concerned with the cost of the network. Less attention has been given to the cost of worth while programmes and how much greater these would be than, for example, the making of radio programmes for local stations or national networks.

Production costs for cinema films can amount to millions of pounds for each hour of material; network television drama is costing £100,000 to £200,000 per hour or even more, while production costs of some TV commercials can reach £100,000 for 30-seconds, not including the cost of airtime. By comparison programme budgets for radio are usually modest in the extreme.

One reason is that radio programmes are produced much faster with far smaller production teams and far less hassle.

Recently I had personal experience of this while taking part in various TV and radio programmes. In each case what was required was a brief one-or-two minute session of off-the-cuff replies to questions.

The prime-time TV current affairs programme took up two mornings, including one morning with the full film production crew of eight people. network radio spot, on the other hand, required a ten-minute visit to my office from a presenter/researcher with his own portable recorder. The local radio recording was a matter of an interview conducted over a noisy telephone line.

With the arrival of the integrated TV camera/recorders such as the RCA "Hawkeye" and the Sony "Betacam" it would in theory be possible to produce TV news and current affairs almost as simply as with radio's portable audio recorder. But there are many reasons why, even with such equipment, production costs are likely to remain far, far higher for TV than radio.

This is one reason why truly local TV is never likely to be possible to full broadcastquality standards—even a tightly-budgeted TV national programme channel represents £100-million-plus per year in programme



MARCH 1983 ISSUE ON SALE FRIDAY, FEBRUARY 18

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An easy and economical way to equip your workshop. Start "collecting" these useful instruments. Full details will be published over the next six months. The first unit appears in the March issue and is a LABORATORY POWER SUPPLY. This provides dual outputs to cater for most needs of the

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THE ELECTRONICS OF



INFORMATION TECHNOLOGY

plexing signals L and R.

INTERLEAVING

PART FOUR

BY T.E. IVALL C.Eng., M.I.E.R.E.

Place your two hands together so that

the fingers of one hand are positioned be-

tween the fingers of the other. The finger

time slots are now interleaved and so

provide the timing conditions for multi-

A rather more advanced analogy will

show the interleaving of more than just

two signals. Imagine the heads of several

garden rakes laid on a table with the

prongs pointing upwards. Line them up

one behind the other so that the sets of

prongs are staggered with respect to each

other, as shown in Fig. 4.2(a). If you now

view all the prongs from the side at table

level they will appear interleaved as shown in (b).

the time slots available for each individual

signal are spaced some distance apart on

What emerges clearly from (b) is that

NE method of sending a number of different signals through a single common medium without mixing them up was described in last month's article on carriers. Frequency division multiplex (f.d.m.) makes use of the time of occurrence of electrical events (Part 1), as manifested in the different periods of the various oscillation frequencies.

Another technique for achieving the same result—putting byways on to highways, so to speak—uses time intervals more directly. Called **time division multiplex (t.d.m.)**, it is now taking over from f.d.m. in trunk telecommunications and becoming increasingly important in

local data transmission.

The whole purpose of multiplexing is to keep line and cable installations fully loaded with information so that they are utilised economically. In this way we save space, materials and money. But first a small digression on the subject of timing.

IMPORTANCE OF TIMING

In your bank account the presence or absence of a mere "0" at a particular position on paper can make a lot of difference to the size of your balance or your overdraft. All digitally represented information is sensitive in this way. When such information is electrically transmitted from place to place, what then becomes significant is the occurrence or non-occurrence of an electrical symbol (for example, voltage or current value) at a particular time. Small variations in the value of the electrical quantity itself make no difference at all to the situation: what matters is whether the symbol is there or not there.

So in the digital electronics of IT systems the timing of electrical events is all-important. Indeed it is only through the relative timing of the electrical events that digital signals carry any meaning at all. Some of these relationships are internal to a given item of information. For example, in transmitting sequentially the group of binary digits 1101, the whole

meaning of the group could depend on whether you start with the least significant digit or with the most significant digit.

The timing of the electrical events can also be in relation to some outside time reference such as an electronic clock.

The electronic clock is, in fact, a prominent feature of many digital IT systems. Often in the form of a crystal controlled pulse generator, it measures out the precise intervals of time in which the electrical values representing digits are permitted to occur. It's rather like an orchestral conductor—or more prosaically a metronome—defining the beat of a piece of music and hence the duration of its bars. Into the intervals so defined the performers insert notes and rests (the information) to produce sounds with a formal rhythmical structure and therefore a meaning as music.

Multiplexer for data communications made by Racal-Milgo. This is a statistical multiplexer (see text) which takes data from up to 32 sources and issues the interleaved information sequentially on a single line at speeds up to 19,200 binary digits per second.

Approximation of the second of

TIME SLOTS

A t.d.m. transmission system, as sketched in Fig. 4.1, makes use of these "time slots". It assigns the common path successively to the different signals by allocating different time slots to pieces of these signals. You have a simple mechanical analogy right in front of you. Spread out the fingers of your two hands. The left-hand fingers represent time slots available for signal L (imagine a line across the base of the fingers as the time scale) and similarly the right-hand fingers represent time slots for signal R.

the analogue of the time scale. How, then, can they be used to convey the information in the signal?

If we want to send an analogue signal like a speech waveform, it can be sampled at intervals corresponding to the occurrence of the time slots for that signal (see discussion on pulse code modulation in Part 2). But the t.d.m. system must be designed so that for each signal the intervals between time slots, and hence the sampling rate, will convey all the information we require (see relevant footnote in Part 2).

If, however, the signal consists of discrete binary digits, as in data transmission, this digital data can be generated at a rate that fits into the time slots available in the t.d.m. system (for example, eight digits per slot).

Figure 4.2 is only a rough analogy. In using the rake prongs to represent the time slots it suggests, wrongly, that there are gaps between the t.d.m. time slots. In reality the time slots follow directly after each other. In Fig. 4.2 you could convey this continuity by imagining the gaps to

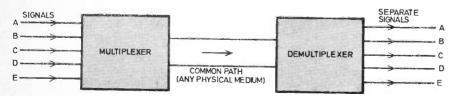
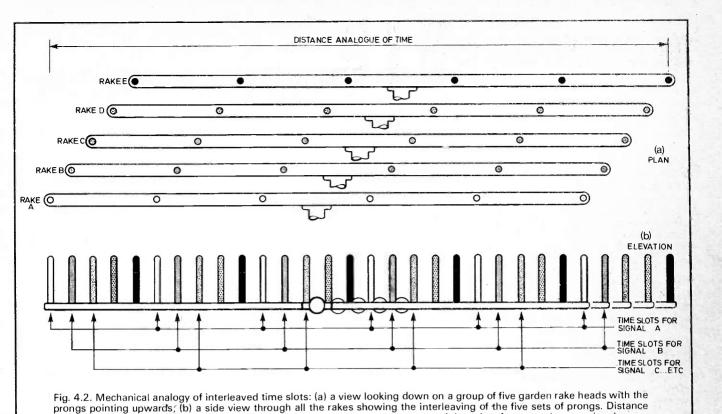


Fig. 4.1. Outline of a multiplex transmission system, for sending a group of separate signals (A to E) along a common path without mixing them up. In time division multiplex, portions of the separate signals are taken in turn, interleaved and transmitted in a continuous sequence along the common path.



along the rakes is an analogue of time. Each set of prongs represents a sequence of time slots for conveying one signal.

be completely closed up by fatter prongs—or more rakes. The actual electrical conditions on the common path of a t.d.m. transmission system are something like Fig. 4.3—though the voltage steps will not be so sharp, for reasons to be explained later in the series.

ELECTRONIC HARDWARE

But how are the time slots shown in Fig. 4.3 actually established in the electronic hardware? It is simply a matter of switching the common path in Fig. 4.1—whether twin-wire circuit, coaxial cable or optical fibre—to carry a piece of each signal in turn. The highway scans the byways, accepting their offerings one by one.

The principle can be illustrated by the analogy of a railway with a single-track section which takes trains coming from the "up" and "down" pairs of lines. This is operated by points, shown as selector switches in Fig. 4.4. Imagine, however, that the arrangement is not being used normally for trains going in opposite directions (trains P and X) but for trains going in the same direction (trains P and Q).

For the time that the points are switched to line 1 as shown, train P, representing an encoded signal sample, can travel from one end of line 1 to the other. This is one time slot. When train P has completed the single track part of its journey, both sets of points are switched over to line 2. For the time they are in this position, train Q—representing a piece of

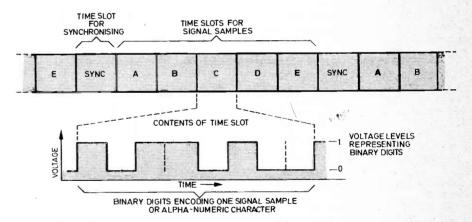


Fig. 4.3. Successive time slots for portions of the five signals A to E in Fig. 4.1, with one extra slot per group to carry synchronising information. Each signal's time slot contains a group of binary digits formed from two voltage levels. These groups are binary coded samples of analogue signals (for example voice) or codes for alpha-numeric characters (for example data transmission).

another signal—can travel along the common track in a similar way. This is another time slot.

And this imaginary railway could of course be extended to allow the common single track to take trains from further lines, as indicated by the chain lines in the diagram. The right-hand set of points acts like the de-multiplexer in Fig. 4.1.

So the time slots are defined by the actions of the points at each end of the single-track section. In the real t.d.m. system the time slots are similarly defined by electronic switches at both ends of the common highway in Fig. 4.1. But to

make sure that all the pieces of signals go to their right destinations (for example, that train P goes back on to line 1 and is not switched to line 2) the two electronic switches must work in exact synchronism.

SYNCHRONISING

In Fig. 4.4 the two sets of points could be automatically operated together, at regular intervals, by a common electrical equipment controlled by a clock, as shown at the bottom of the diagram. The t.d.m. system in fact uses its electronic

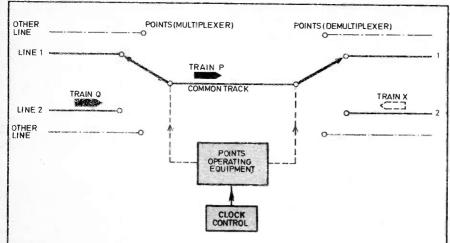


Fig. 4.4. Railway analogue showing how coded samples from several different signals are interleaved into successive time slots by electronic switching. The left-hand set of points corresponds to the multiplexer in Fig. 4.1 while the right-hand set corresponds to the demultiplexer.

clock to synchronise the switches. This process takes the form of synchronising pulses sent along the common path by time division multiplexing just as if they were portions of actual signals, as shown in Fig. 4.3.

In telecommunications this t.d.m. principle is employed throughout the world for pulse code modulation trunk transmission systems (see Part 2). More locally it is being used to allow a number of terminals to work into a remote computer via a single telephone line. A "statistical multiplexer" is one that takes advantage of any inactivity in signals to fill up the otherwise empty time slots with information from other, active signals, thereby increasing the number of signals that can be multiplexed for a given rate of transmission along the common path.

BUSES AND INTERFACES

To transmit information between electronic units we must make sure that what comes from the output of one unit is effective as an input for another unit. It is not only a matter of getting the wires, plugs and sockets right but ensuring that

the electrical representation of the information on these conductors is compatible with the functioning of the units connected. If communication is to be achieved the "talker" must first make himself heard to the "listener". Then he must use a language that the listener can understand.

If a single manufacturer designed and produced all the units concerned he could make sure of these requirements himself. But because IT equipment is introduced by different firms at different times in the development of the technology there is a real problem to deal with. For example, we might need to connect together a computer, modems (Part 3), a multiplexer and several terminals of different kinds—all of which could come from different makers.

STANDARD INTERFACE AND BUS

A sensible way of dealing with this problem is to have a standardised method of connection and information transmission that will cover all eventualities. This is the reason for the emergence of the

Videotex systems (formerly called Viewdata the UK) send digitally-encoded information along telephone lines using the serial method shown in Fig. 4.5(a). This videotex business terminal made by Pye TMC can be used with the public Prestel service or in private information retrieval systems. (In Prestel, digital information is sent from the computer data base to the terminal at a rate of 1200 binary digits per second; from terminal to data base at 75 binary digits per second.)

standard interface, as it is called, for connecting separate units, and the standard bus*, for connecting internally the different parts of a single equipment.

Typically, the interface is used for connecting terminals to a computer, while the bus is used inside a computer to transfer information between its central processor, its memory and its input and output devices. (The distinction between them is not rigid, however. One widely used connecting system, for example, is called the General Purpose Interface Bus.)

But because technology develops as an historical process influenced by commercial pressures, there is no single standardised interface and no single standardised bus. Instead we have a *number* of "standard" interfaces and buses, most of them originated by manufacturers or their trade associations. Nevertheless a few of them have been accepted world-wide and ratified by international bodies—which helps to avoid the utter confusion that would otherwise occur.

Although buses and interfaces can be very complicated in their electrical and mechanical detail they are simple in principle. They all depend on the three aspects of an electrical quantity—magnitude, position and time of occurrence—that are used throughout IT to represent and convey information (see Part 1). The position of the electrical quantity is, of course, the particular conductor on which it occurs; the magnitude is, typically, a voltage or current; and the time of occurrence is when that voltage or current acts as an electrical symbol.

In this article we are mainly concerned with digital information. For analogue signals it certainly helps to standardise your system of connection between units, but the magnitude and time aspects of the electrical quantity are not so critical. One need only specify ranges of, say, amplitude and frequency to transmit information successfully between units.

SERIAL AND PARALLEL SYSTEMS

So in digital systems, magnitude, position and time are manipulated by engineers to produce different kinds of "standard" buses and interfaces. For practical and economic reasons two main groups have emerged: serial and parallel. In serial systems the different magnitudes of the electrical quantity are sent one after the other. In parallel systems the different magnitudes are sent simultaneously.

As an analogy, consider the process of reading this magazine. The individual words (or perhaps small groups of words) forming a sentence enter the eye and brain serially. When one looks at a diagram, however, the component parts of its two-dimensional pattern are perceived simultaneously—in "parallel".

^{*} The term "bus" is of course an abbreviated form of the Latin *omnibus*, meaning "for all" (cf. "busbar" in electrical engineering).

For serial transmission only one electrical circuit is needed to connect electronic units—the type of simple information path discussed in Part 3. For parallel transmission there have to be several such circuits between the units. Fig. 4.5 is a comparison between these serial and parallel methods of representing and conveying an item of information—here a number, or other character, encoded as the binary digits 1011.

In the serial case (a) the electrical magnitudes representing these four digits can only be sent one after the other through the single circuit connecting Units A and B. In the parallel case (b), however, all four electrical magnitudes representing the digits can be sent simultaneously on the four circuits (which share a common return conductor) between Units C and D.

So if each of the successive electrical symbols for the digits has a duration of 1 microsecond, the serial system (a) will take 4 microseconds to send the whole character while the parallel system (b) will take only 1 microsecond to send it.

This illustrates the general point that serial systems are slower in transmitting information than parallel systems. But in practice the single circuits of serial systems are extremely useful as interfaces because our existing telephone networks, both public and private, usually end up in simple pairs of wires running to individual instruments in homes, offices and factories. These single pairs are used in IT, for example, to connect terminals to distant computers or to connect facsimile machines to each other.

SHORT DISTANCES

Where the distances between units are short and the cost of installing many parallel circuits is not very high, the system in (b) can be utilised, to take advantage of the high speed of information transmission it allows. In practice this means connections within a machine (for example, a computer) or within a room (example, test instruments working automatically together). The multiple conductors required are sometimes on printed circuit boards, sometimes in flat flexible cables.

Associated with these interfaces and buses one finds electronic devices which allow serial information to be converted to parallel information and vice-versa. An integrated circuit commonly used for this purpose is the "universal asynchronous receiver-transmitter (u.a.r.t.)".

Figure 4.5(b) could be an elementary bus. Practical buses in IT are, however, extremely complicated devices. One commonly used in microcomputers, for example, has 100 conductors altogether. Some of its circuits carry encoded characters (data), some convey encoded information on the storage locations of characters (addresses) and others carry control signals.

A bus provides a means for several different units to be connected to a common

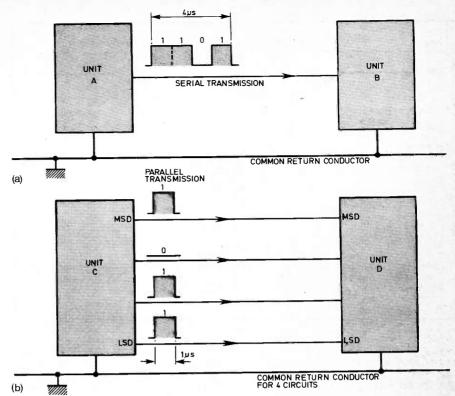


Fig. 4.5. Two ways of transmitting the group of binary digits 1011 between units: (a) serial transmission on a single circuit, the electrical values representing digits following one after the other; (b) parallel transmission, with the four electrical values sent simultaneously on four separate circuits. Assuming equal durations of digit signals, (b) is much faster than (a). Note in (a) that the least significant digit (l.s.d.) is sent first and the most significant digit (m.s.d.) last.

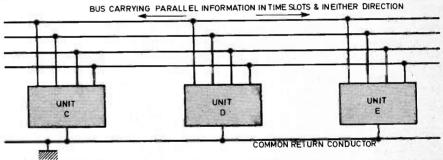


Fig. 4.6. Parallel connections between Units C and D in Fig. 4.5(b) are here extended to a further electronic unit, Unit E. The connections become a bus, or common highway, which, by suitable time-sharing arrangements, can carry parallel information in either direction between any of the units. More units could be connected to the bus on the same principle.

highway, as shown in Fig. 4.6, allowing two-way transfers of parallel information (in Fig. 4.5(b) form) between the various units. But electrically only one transfer can take place at a time, so there have to be careful arrangements for time sharing using clock-defined "time slots" as described earlier.

MICROPROCESSORS

Within a microprocessor, for example, where parallel information is continually being transferred between storage registers, a bus is electronically switched, in each time slot, to connect the output of one register to the input of another register. Buses are essential to IT systems using microprocessors, partly because these devices are designed to work with parallel information and partly because

the small physical size of integrated circuits limits the number of pins that can be put on them and so makes the time sharing of a common set of conductors absolutely necessary.

Even a standard interface based on the serial principle in Fig. 4.5(a) can be quite complicated in practice, as conductors for sending characters are not sufficient in themselves. Other circuits, carrying monitoring and control signals, are needed to ensure that the units at each end—say teleprinters—are in proper electrical contact and operating in synchronism (compare with time division multiplex). As an example, the V.24 international standard serial interface (known also as RS232C in the USA) provides 25 conductors to cope with all the functions that may be required.

To be continued

veryday N

HEADACHE CURE

A unique new instrument from Canada promises to assist sufferers from tension headaches and is now being marketed in the UK.

Working on biofeedback principles, the Antache instrument has been developed for research, clinical and home use. It consists of a pair of headphones and electronic circuitry to which is attached an elasticised

headband and electrodes.

In operation, the Antache continuously monitors and averages the wearer's electromyogram (EMG) and converts it into a pleasantly modulated tone. Using the pitch of this tone as a guide, it is claimed that the user soon learns how to relax the muscles in the scalp, face and neck which, when they become too tense, bring on the symptoms of tension headaches.

Providing the headache condition has first been diagnosed by a medical doctor and provided, too, that the doctor has approved the use of the device for the patient concerned, the Antache is quite safe for use

The major advantage of using biofeedback principles for the treatment of headaches is that, unlike drugs, there are no side effects and the user is not restricted in his activities immediately following treatment.

For more details of the Antache, readers should contact: Beam Components Ltd., Dept. EE, 108 High Street, Strood, Rochester, Kent ME2 4TR.



OVERTAKEN

Cars, until now the largest in money terms of Japanese imports into the UK, have been overtaken by video-cassette recorders. But discussions between the UK Department of Industry and Japan's Ministry of International Trade and Industry may result in establishment of manufacturing plants in the UK, possibly as joint ventures with UK companies, in the hope of stemming the import flood.

- Video games Shock --

Video games help kids co-ordinate manual and visual skills. No doubt. But games manufacturers in the USA are reeling after a shock attack by opponents who claim on psychiatric grounds that as most kids prefer violent rather than educational games. they breed acceptance of violence as the norm.

LASER-FI

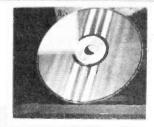
Playing your favourite pop or classical disc takes on a new meaning with the announcement from Pioneer of the introduction of the Laser-Disc LD1100 player.

Launched just in time for Christmas, the new player uses a laser beam to read a special disc and reproduce both hi fi stereo sound and "action" video pictures. It is claimed that the sound reproduction is on par with, or better than, f.m. radio and that picture quality is equal to 'off-air" broadcasts.

Over 54,000 individual video frames are contained on each side of the acrylic encased disc and a random access facility enables the user to locate and "freeze" any individual frame. Slow motion operation is also possible.

Unlike the audio stylus the optical laser system makes no physical contact with the disc so there is no deterioration in sound or visual quality, no matter how many times the disc is played.

Additional user benefits claimed for the player include a CX noise reduction system and the facility to link with teletex and interface with computers.





Cash Mountains

Leading British electronics manufacturers continue to beat the recession with growing order books and even money in the bank. Biggest cash mountain is at GEC with more than £1 billion. In comparison Plessey's liquidity is more like a molehill at £180 million following a 30 per cent rise in profits.

Cash in hand means takeover possibilities such as Plessey's recent acquisition of Stromberg

Carlson in the USA.

Solar Flight

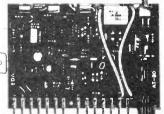
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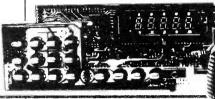


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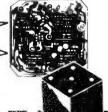
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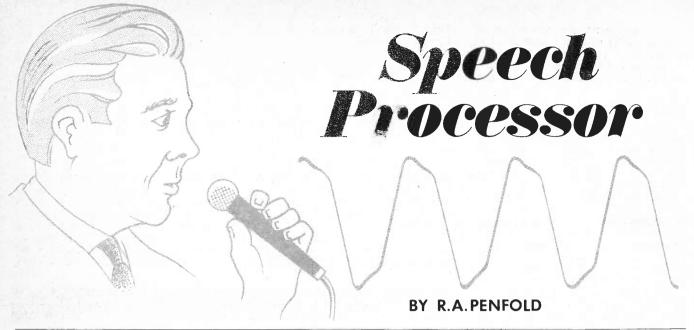






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BOOSTS AVERAGE LEVEL OF SOUND WITHOUT INCREASING PEAK LEVEL SUITABLE FOR CONNECTION BETWEEN MICROPHONE AND AMPLIFIER

A SPEECH processor is a device which is used to process a speech signal in such a way that the strength of the signal is effectively boosted without any increase in the peak level of the signal.

Units of this type rely on the fact that a speech signal has a rather high peak level when compared to the average level, and the increase in "talk power" can be obtained by boosting the signal but clipping the peaks so that there is no increase in the peak level. Thus the peak level remains unchanged, but the average signal level is greatly boosted, and the volume of the signal is effectively boosted by a substantial amount.

Speech processors are mainly used in communications equipment; sometimes as an integral part of a transceiver, and sometimes as an add-on unit which connects between the microphone and the transceiver. Speech processors can also be used to good effect in other types of equipment, such as a public address system.

The unit described here is a reasonably simple but effective device which is battery powered, and is simply connected between the microphone and the transceiver or other equipment. It is intended for use with a high impedance dynamic microphone or an electret type, having a built-in step-up transformer, and this should present no problems since most communications and PA microphones are the high impedance dynamic type.

PROCESSOR SYSTEMS

There is more than one way of obtaining the limiting of signal peaks, and one method is to use a form of automatic gain control circuit. Here the processor adjusts the level of gain so that it is automatically reduced during periods of high dynamic level and increased during periods of low dynamic level.

This obviously gives the required narrowing of the difference between the peak and average signal amplitudes, but the circuit must be designed to respond very rapidly to changes in signal level if it is to be of real benefit. This can easily result in the signal waveform being seriously distorted with a lot of distortion being evident on the output signal.

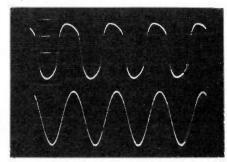
Slower response times give reduced distortion, but also give a reduction in effectiveness. Of course, a certain amount of distortion is quite acceptable in this application, but more than several per cent distortion would impair the intelligibility of the processed signal and obviously reduce the benefit of the unit.

DISTORTION REDUCTION

A very simple method of speech processing is to use a clipping circuit which prevents the output signal voltage from exceeding a certain level. The circuit is adjusted so that most of the signal is below the clipping level and is unaffected by the unit, but so that the high signal peaks are held down and kept well below their normal level.

This system has the advantage of simplicity plus instant attack and decay

Upper Trace: Soft-limited sinewave, Lower Trace: 800Hz sinewave input signal.



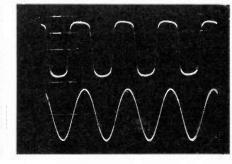
times, but in this basic form it produces quite high levels of distortion. The distortion products produced consist mainly of harmonics, and harmonics are simply multiples of the frequencies in the clipped signals.

In order to obtain really good results from a clipping circuit it is necessary to include additional circuitry to minimise the distortion generated. The most sophisticated method of achieving this is to use an r.f. clipping circuit, and with this system the input signal is first processed to raise all the input frequencies by a substantial amount so that they are increased into the radio frequency (r.f.) range.

For example, the input frequencies could be raised by 100kHz, and then they would be at frequencies from just over 100kHz to about 120kHz. After clipping, harmonics would still be generated, but these would be at frequencies of about 200 to 240kHz, 300 to 360kHz, 400 to 480kHz, and so on. These are well clear of the input frequency range, and can be filtered from the output to leave a distortion-free signal which is then processed to restore the original audio frequencies.

The severe drawback of this system is

Upper Trace: Hard-limited sinewave. Lower Trace: 800Hz sinewave input.



the cost and complexity, unless it is an internal part of an s.s.b. transmitter. For use with other types of equipment an addon processor of this type could cost more than the main item of equipment!

SIMPLIFIED DESIGN

What is needed is a simpler method of obtaining a similar effect, and a system of this type is used in the processor described here. Fig. 1 shows the arrangement used in this processor in block diagram form.

The microphone signal is first amplified and then fed to a further stage of amplification by way of a variable gain control. This amplification is needed because the output of a microphone is at a fairly low amplitude, and it is not easy to produce a clipping circuit which operates at such low signal levels.

The signal is therefore boosted to a level where clipping can be achieved more readily. The gain control enables the unit to be adjusted to give the desired degree of clipping.

The next stage is a high pass filter, and this removes the lower frequencies in the signal. Frequencies below about 300 hertz contribute nothing to intelligibility, and can even impede it. These frequencies would be most troublesome if not removed as they would produce numerous harmonics right through the middle and upper audio range when the clipping was applied, and the removal of these frequencies substantially reduces the distortion level at the output.

SOFT CLIPPING

Distortion can be further reduced by using a soft clipping circuit rather than a normal hard clipping type, and so a soft limiter is used here.

The difference between the two is that a hard limiter permits no significant increase in the output level once the clipping level has been reached, no matter how large the input signal is made, whereas a soft limiter permits a slight increase in the output amplitude as the input is increased above the clipping threshold.

The use of soft limiting gives only a very marginal reduction in efficiency, and the fundamental signal is significantly stronger and the harmonics significantly weaker when compared to results using a hard limiter.

A l.e.d. indicator is switched on while the limiter is driven beyond the clipping threshold, and this makes it much easier to adjust the gain control correctly.

Most of the harmonics on the output signal will be at frequencies of about 3kHz and above, and frequencies in this range aid the intelligibility of a speech signal very little. A low-pass filter at the output of the unit is therefore used to severely attenuate signals at these frequencies, thereby greatly reducing the level of distortion on the output.

The final section of the unit is simply an attenuator which reduces the output level to one that is comparable to the

Fig. 1. Block diagram of Speech Processor.

L.E.D. DRIVER

HIGH GAIN HIGH PASS SOFT LOW PASS FILTER

GAIN CONTROL

CONTROL

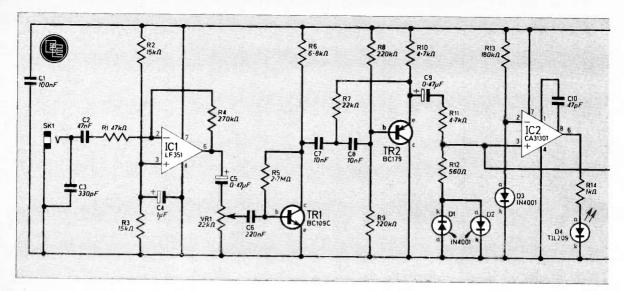
input signal level. This enables the processor to be connected between the microphone and the main equipment without introducing any compatibility problems.

This system does not completely eliminate distortion from the output since some input frequencies (those at about 300Hz to about 1.5kHz) will be fed to the limiter and will produce at least one distortion product that will not be removed by the low-pass filtering at the output. However, the distortion is kept down to acceptable levels provided an excessive amount of clipping is not employed, and the unit certainly seems to make a very

R1 R2 R3 R4 R5 R6 R7 R8 R9 R10	47kΩ 15kΩ 15kΩ 270kΩ 2·7MΩ 6·8kΩ 22kΩ 220kΩ 220kΩ 4·7kΩ 4·7kΩ	R12 R13 R14 R15 R16 R17 R18 R19 R20 R21 R21	560Ω 180kΩ 1kΩ 100kΩ 100kΩ 3.9kΩ 3.9kΩ 3.9kΩ 4.7kΩ 4.7kΩ	
All 1/3 watt car				
Potention	atore			
VR1	22k Ω log, carbon potential	ometer	Se	e
VR2	47kΩ 0-1 watt horizontal			hot
Capacitors				Tall
C1	100nF polyester (C280)			CLIP
C2	47nF polyester (C280)			page 85
C3 C4	330pF ceramic plate 1µF 16V elect.			
C5	0.47µF 10V elect. radial	leads 🙀	SERVER PLANER	M SP M N NO
C6	220nF polyester (C280)	4 2		
C7	10nF polyester (C280)	259		Eas .
C8 C9	10nF polyester (C280) 0-47µF 10V elect. radial	loade B		
C10	47pF ceramic plate	leads		
C11	100nF polyester (C280)		anet	
C12	22nF polyester (C280)	100	£1	0
C13	3-3nF ceramic plate	100		
C14 C15	10nF polyester (C280) 2-2nF polystyrene		*,	
CID	120pF ceramic plate			
	22nF polyester (C280)			
C16 C17	ZZIII poryodioi (ozoo)	,		
C16		/		
C16 C17		,		
C16 C17 Semicondu IC1 IC2	uctors LF351 j.f.e.t. op. amp. CA3130T MOS f.e.t. op. a	mp.		
C16 C17 Semicondu IC1 IC2 TR1	uctors LF351 j.f.e.t. op. amp. CA3130T Mos f.e.t. op. a BC109C npn silicon	mp.		
C16 C17 Semicondu IC1 IC2 TR1 TR2	LF351 j.f.e.t. op. amp. CA3130T Mos f.e.t. op. a BC109C <i>npn</i> silicon BC179 <i>pnp</i> silicon	mp.		
C16 C17 Semicondo IC1 IC2 TR1 TR2 TR3	uctors LF351 j.f.e.t. op. amp. CA3130T Mos f.e.t. op. a BC109C npn silicon	mp.		
C16 C17 Semicondu IC1 IC2 TR1 TR2	LF351 j.f.e.t. op. amp. CA3130T Mos f.e.t. op. a BC109C npn silicon BC179 pnp silicon BC109C npn silicon	mp.		64
C16 C17 Semicondo IC1 IC2 TR1 TR2 TR3 D1,2,3	LETOTS LETOTS	mp.	16	O

strips x 12 holes; two control knobs; panel holder for D4; battery connector; two

6BA fixings; two 6mm spacings; p.v.c. insulated connecting wire.



worthwhile improvement when used with communications equipment under adverse operating conditions.

Of course, if conditions are such that proper contact is easily achieved with no interference and good signal strengths, there is little room for improvement and a speech processor can be of little help. It is only when conditions are poor that the effect of a speech processor will become apparent.

THE CIRCUIT

Fig. 2 shows the complete circuit diagram of the Speech Processor.

IC1 is an operational amplifier used in the inverting mode and this acts as the microphone pre-amplifier. This stage has its voltage gain set at a modest level of about six times by R1 and R4, and R1 also sets the input impedance at a suitable level of about $47k\Omega$.

C3 is an r.f. filter capacitor and helps to prevent problems with r.f. breakthrough and consequent instability if the unit is used in a strong r.f. field.

IC1 is a low noise device having a j.f.e.t. input stage and this gives the unit a good signal-to-noise ratio.

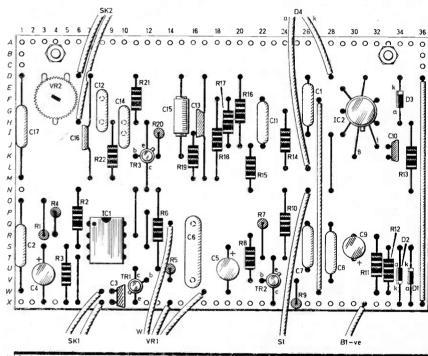
VRI is used to control the degree of clipping and the output from its wiper is fed to a high gain common emitter amplifier which uses TRI in the standard configuration. The gain of IC1 together with the gain provided by TR1 enables the microphone signal to be readily

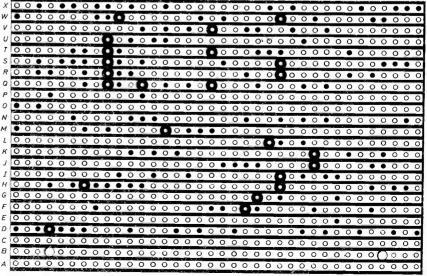
boosted to a level of several volts peak-topeak, and this is sufficient to drive the

clipping circuit.

SIGNAL CONTROL

The high-pass filter is an active type which uses the Sallen and Key configuration and has a nominal attenuation rate of 12dB per octave below the 300Hz cut off frequency. There is unity gain through this stage at frequencies of 300Hz and above.





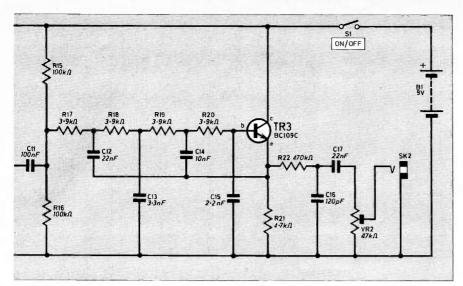
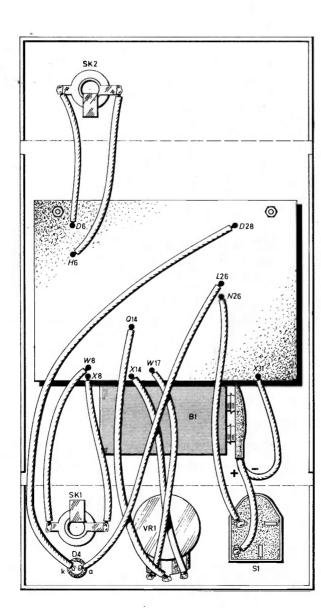


Fig. 2. Complete circuit diagram of Speech Processor.

¶ Fig. 3(a). Topside of stripboard showing component layout.



◆ Fig. 3(b). Underside of stripboard showing breaks in copper strips.

Fig. 4. Exploded view of unit showing interwiring for off-board components.

D1 and D2 are used as the basis of the clipping circuit, D1 processing negative half-cycles and D2 processing positive ones. If the input signal level is less than about ±0.5 volts neither D1 or D2 will conduct and the signal can pass straight through R11 to the next stage of the unit. If the signal level should exceed about

0.5 volts, either D1 or D2 (depending on the polarity of the signal) will be biased past its forward threshold voltage and will conduct heavily. This produces a voltage drop through R11 which tends to hold the signal voltage at little more than 0.5 volts even if the input level should be much more than this.

R12 introduces the softening of the clipping action since a current flowing through D1 or D2 must also flow through R12 as well, producing a small voltage across R12, which is proportional to the current flow. Thus the output signal at the junction of R11 and R12 can go slightly above the clipping threshold and the soft clipping is obtained.

CLIPPING INDICATOR

IC2 is used as the l.e.d. driver, and in this application IC2 is a comparator rather than an operational amplifier. R13 and D3 form a simple voltage regulator circuit which biases the inverting input to about ± 0.5 volts. The non-inverting input will normally be at a lower potential than this so that the output will be at zero volts and l.e.d. indicator D4 will be switched off.

During positive signal peaks if the clipping level is exceeded, the non-inverting input will be taken above 0.5 volts so that the output of IC2 switches to virtually the full positive supply voltage and D4 is pulsed on to indicate that clipping is occurring.

The output filter is another Sallen and Key active filter, but this time a four section circuit has been used so that a nominal attenuation rate of 24dB per octave is achieved. It is of course a low-pass filter that is used here, and the cut off frequency is about 3kHz. Further low-pass filtering is provided by R22 and C16.

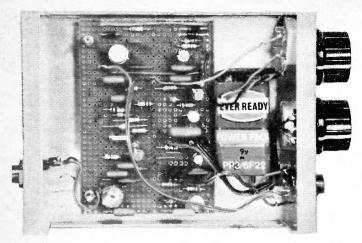
C17 is the output d.c. blocking capacitor and VR2 is the pre-set output

attenuator.

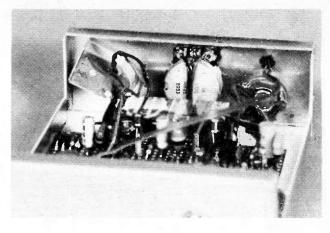
As the circuit has a current consumption of only about 4.5mA a small (PP3 size) 9-volt battery will give many hours of use before needing replacement.

CASE

An aluminium box having approximate outside dimensions of 133 x 102 x 38mm makes a suitable housing for the processor, and this is about the smallest case that will comfortably accommodate all the components. SK1, D4, VR1 and S1 are fitted on the front panel, and SK2 is mounted on the rear panel. SK1 and SK2 are both standard ($\frac{1}{4}$ inch jack sockets).



Plan view showing component layout inside case.



Rear view showing front panel component wiring.

CIRCUIT BOARD

The component panel is a 0.1 inch matrix stripboard having 24 strips by 36 holes, and this can conveniently be a standard 37 holes by 24 strips board with one row of holes trimmed off or just ignored. Drill the two 3.3mm diameter mounting holes (which accept M3 or 6BA fixings) and make the numerous breaks in the copper strips before fitting the components on to the board. There are also six link wires to be soldered in place on the board. Use pins at the points where connections are to be made to the off-board components. Fig. 3 gives full details of the component board.

The completed component panel is mounted on the base panel of the case leaving sufficient space for the battery between the board and the components on the front panel. Use 6mm spacers over

the mounting screws to keep the connections on the underside of the board well separated from the metal case. The remaining wiring is then completed using ordinary p.v.c. insulated connecting wire, and finally the battery clip is wired in place. All this wiring is illustrated in Fig. 4.

ADJUSTMENT

Only one internal adjustment is necessary before the unit is ready for use, and that is to set VR2 to give an output level which is comparable to the output of the microphone used with the unit.

If suitable measuring equipment to assist with this is not available, then it is really a matter of connecting the processor to the main unit using a suitable lead, plugging the microphone into the processor, and then adjusting VR2 by trial and error to a setting which

gives results similar to those obtained without using the processor. While doing this VR1 should have a setting that is just high enough to cause clipping, which will be indicated by D4 just flashing briefly on signal peaks.

In normal use VR1 would be advanced somewhat further than this so that D4 lights up quite brightly whenever an input signal is present. It is probably best to monitor the output signal using an amplifier and headphones, or using a taperecorder perhaps, when initially experimenting with various settings of VR1. This soon shows the benefit in apparent volume increase as VR1 is taken above the clipping threshold, and how excessive clipping simply gives increased distortion and no further increase in volume. Best results are obtained with VR1 advanced, just far enough to give a well clipped signal.

BOOK REVIEWS

PRACTICAL ELECTRONICS FOR **RAILWAY MODELLERS**

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Roger Amos

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Topics covered include controllers, train detection systems, automatic signalling and points controls, and lighting systems, including high-frequency supplies. The only conspicuous lack is of detailed constructional information.

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A. P. Stephenson

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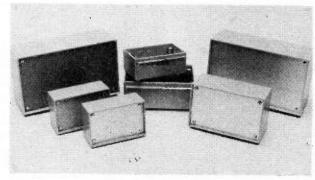
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PROGRAMMER FOR THE AGENT ATOM



BY D. C. GRINDROD

THE device to be described here will provide a simple, fast and relatively inexpensive method of programming and reading 2K and 4K byte single supply rail (+5V) EPROMS (Erasable and Programmable Read Only Memory) using the Acorn Atom computer.

24 Vcc A7. 1 23 A8 A6 2 22 A9 A5 3 21 Vpp A4 4 CE/PGM A3 5 20 19 A10 2532 A2 6 18 CE / PGM A1 7 17 D7 A0 8 D0 9 16 D6 15 D5 D1 10 14 D4 D2 11 13 D3 GND 12 TOP VIEW

Fig. 1. Pin-out diagram for the 2716 and 2532 EPROMS. Note the different functions for pins 18 and 20 for the two types.

It is able to program the following EPROMS (1) TMS2516, 2716 and other manufacturers' pin compatible types; (2) TMS2532 and other manufacturers' pin compatible types. It is not suitable for use with Intel 2732 type EPROMS or other manufacturers' compatible types.

Although specifically designed for use with the Acorn Atom (for which it was developed to overcome loading of frequently used long programs from cassette), it could also be used with other computers with modifications to connecting cables and suitably developed software

The software controls all pulses and addressing, leaving the user to select only MODE (READ/PROG) and SIZE (2K/4K).

To be able to address up to 4096 bytes (4K) address lines A0 to A11 are required. A0 to A7 are provided by port A (£B801) of the VIA and A8 to A11 by the four lowest bits of port C (£B002) of the 8255. This break is not as awkward as would first be thought, as it occurs on a 255 byte boundary. The maximum number held in a single-byte is 255, hence a carry procedure would be needed anyway.

Figure 1 and Table 1 show the pinning, programming and reading requirements of the two types of EPROMS catered for by this programmer from which we can see that they differ in the following respects:

- (1) CE/PGM is a different pin
- (2) 2532 has an extra address line
- (3) The pulse required for programming is —ve going for the 2532 and +ve going for the 2716

The first two differences are overcome by a d.p.d.t. switch, S1, while the third is dealt with by the software.

PROGRAMS

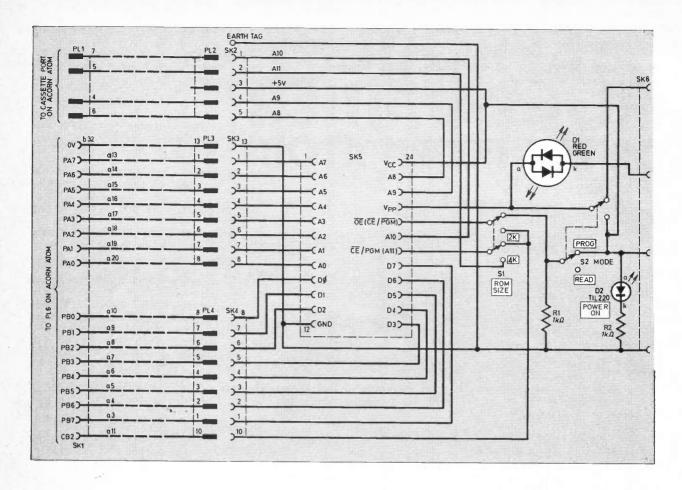
The original idea was to use EPROMS as a semi-permanent storage medium for long programs that were used often and hence reduce the time needed for loading from cassette.

Since each byte programmed needs a 50ms pulse the programming time is not greatly affected by the language used, thus, it was decided to use a Basic progam to control programming of the EPROM.

Table 1: Mode selection for the 2716 and 2532 Eproms

TYPE		2716	(2K)		4 X*		2532	(4K)	
Pin Mode	CE/PGM (18)	OE (20)	V _{PP} (21)	V _{CC} (24)	Outputs (9—11, 13—17)	CE/PGM (20)	V _{PP} (21)	V _{CC} (24)	Outputs (9-11, 13-17)
Read	V _{IL}	V _{IL}	+5	+5	D _{OUT}	V _{IL}	+ 5V	+5	D _{OUT}
Standby	V _{IH}	Don't Care	+5	+5	High Z	V _{IH}	+5V	+5	High Z
Program	Pulsed V _{IL} to V _{IH}	V _{IH}	+25	+5	D _{IN}	Pulsed V _{IH} to V _{IL}	+25V	+5	D _{IN}
Program Venty	V _{IL}	V _{IL}	+25	+5	D _{OUT}		_		_
Program Inhibit	V _{IL}	V _{IH}	+25	+5	High Z	V _{IH}	+25V	+5	High Z

VIL logic low, VIH logic high



When reading the EPROM however, the time taken is determined by the language, therefore a fully re-locatable machine code program was developed which is small enough (79 bytes) to fit above the floating point variables at £2890. EPROM/RAM start and EPROM end addresses are stored in the two lowest bytes of the integer variables A, B and C, respectively.

CIRCUIT DESCRIPTION

The complete circuit diagram for the Eprom Programmer is shown in Fig. 2. Nearly all the circuitry is for the generation from the mains supply of the +5 and +25 volt supply lines. We shall discuss this section first.

A.c. mains voltage enters the unit and reaches T1 primary winding via the onoff switch S3 and fuse FS1. T1 is a stepdown transformer having two independent secondaries, each developing 20V a.c. across their windings.

The diode bridge D3-D6 provides full-wave rectification of the upper secondary voltage. The resulting pulsed d.c. is smoothed by C3 to reach the input of IC1. The latter is a monolithic voltage regulator i.c. which normally provides a stabilised 15V output. However, in this circuit the output voltage is stepped-up to 25V by the action of VR2 in series with the common connection.

COMPONENT Resistors See SK3.4 16-pin d.i.l. (2 off) R1 $1k\Omega$ SK5 24-pin d.i.l. + 24-pin R2 390Ω zero insertion force $4.7k\Omega$ socket All ½W carbon ±5% SK6 4-way inter-p.c.b. connector page 85 Capacitors PL1 7-way DIN 5-way 270° DIN PI2 C1,4 0.047µF 35V tantalum PL3,4 16-pin d.i.l. header (2 off) (2 off) C2.5 0.022µF 35V tantalum PL₆ 4-way inter-p.c.b. (2 off) connector C3,6 100µF 35V elect. VR1 5kΩ multiturn preset (2 off) VR2 1kΩ miniature horizontal skeleton **Semiconductors** preset D1 bi-coloured (red/green) FS₁ 1A 20mm with panel mounting fuseholder D2 TIL220 red l.e.d. T1 mains primary/0–20V, 0–20V 6VA D3-6 VM18 1A 50V bridge D7-10) rectifier d.i.l. (2 off) secondaries p.c.b. 7815 15V 1A voltage mounting-see text regulator (TO-220) 7805 5V 1A voltage Stripboard, 0.1 inch matrix; 17 IC2 strips x 51 holes, 25 strips x 25 regulator (TO-220) holes; 6BA mounting hardware for circuit boards; Speedbloc cable Miscellaneous or other; case, plastic, size 190 x \$1,2,3 d.p.d.t. miniature 110 × 60mm (Tandy 270-224); toggle (3 off) 2-core mains cable; clips and SK1 64-way (a+b) in-line bushes for l.e.d.s.; self-adhesive indirect connector rubber feet for case (4 off); sleev-SK2 5-way 270° DIN Guidance only. Appox. cost

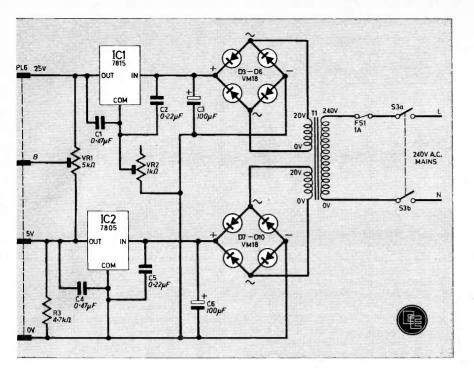


Fig. 2. Complete circuit diagram of the EPROM Programmer. The power supply section is shown above and the circuit (left) gives the interconnections from the 24-pin d.i.l. socket to the Acorn Atom computer.

C1 and C2 are included for reasons of stability. Similarly, a smooth voltage level of about 28V reaches the input of IC2 to provide a +5V output stabilised. C4 and C5 are included for reasons of stability. R3 acts as a bleed resistor for IC2.

The power supply is more than adequate, the maximum requirements being for the 4K EPROM: 5V at 150mA, 25V at 30mA.

The power supply is built on a circuit board separate from the remainder and

connects to it by means of p.c.b. inter sockets/plugs (SK6/PL6).

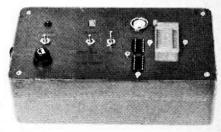
The +5V and 0V rails reach pins 24 and 12, respectively, of the EPROM socket SK 5.

The bi-coloured l.e.d. D1 is an optional extra which was included to show the mode of operation. This was also marked around S2 on the top panel. In the prototype it was orientated so that it lights up red for PROGramming mode, and green in the READ mode. VR1 needs to be set to give equal brightness of the two colours, this will be described later.

The address and data for the EPROM are made available by the software at various pins on *PL6* on the Atom and the cassette port. These signals are under full software control.

The 25V supply is only needed for programming and therefore only reaches pin 21 of the EPROM when S2 is in the PROGramming mode. In the READ mode V_{op} should be at +5V.

The only other control is S1, the ROM SIZE switch. This routes the program signal to the appropriate pin according to Fig. 1, and brings in A11 as required.



Next Month: Construction and Testing

CIRCUIT

IC2 forms a slow-running astable multivibrator the output of which is fed to the reset pin of IC1. The chirping will be switched on and off because IC1 can only oscillate when pin 4 is positive. VR1 alters the time between each burst of

"chirps". The component values are by no means critical and may be experimented with.

Mark Robinson, Winsford, Cheshire.

CHIRPING BIRD

This circuit produces a sound similar to a chirping bird. TR1 and associated components form a sine-wave oscillator which runs fairly quickly. Increasing the values of C1, C2 and C3 would slow down the chirp rate, but they must all be the same value to produce the required phase shift.

The output of this oscillator is fed to the control pin (pin 5) of a 555 astable multivibrator IC1 which produces a high frequency square-wave at pin 3. R7 may be increased to lower the volume but the sum of R7 and the speaker resistance must not be less than 100 ohms, otherwise the current drawn by LS1 could damage IC1.

The output, therefore, is a square-wave modulated by the sine-wave which produces the characteristic "chirp" sound.

SI
ON/OFF

R4.7kn

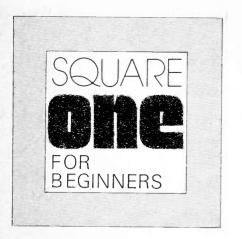
R5.56kn

R7.1000

R8.1000

R8

MORE ON PAGE 118



THE trouble with earphones, headphones and allied devices is that it is only too easy to take them for granted. You make your miniature personal radio, say, and then you think: Ah, yes; it'll need an earphone. The chances are that you try some old earphone salvaged from a defunct radio and plug it in, hoping for the best.

IMPEDANCE

Quite often, it doesn't work, or at least doesn't work well. The most frequent explanation is that its impedance is too low or too high to suit the circuit to which you attach it. Most of the little plastic earphones which come with pocket radios have low impedance, often 8 ohms. This

A common arrangement was to have two earpieces each of whose impedances was 1000 or 2000 ohms. These could be connected in series to give 2000 or 4000 ohms or in parallel for 500 or 1000 ohms. They were extremely sensitive: they had to be to give any volume from a crystal set whose only power was what the aerial picked up. They were usually poor quality sound reproducers with a huge resonance at about 1kHz.

Today's two-earpiece headphones are usually very different. A typical pair of low-cost "stereo phones" contains in its rather large and comfortably padded earpieces a couple of small loudspeakers. These are usually of 8 ohms impedance and they are connected as shown in Fig. 1a. The connections are often brought out to a jack plug with three contact segments.

For stereo listening (b) the "live" sides of the two audio channels are connected to points 1 and 3; point 2 is the common or earthy connection.

For non-stereo use you have the option of using the two in series (c) to give an impedance of 16 ohms or in parallel (d) for 4 ohms. It is possible to obtain stereo phones of other impedances. Sound engineers may use 600-ohm phones for instance. But 2 × 8 ohms is by far the commonest impedance. Actually this too

The low-impedance earphones contain a little coil of fine wire, a magnet, and some sort of diaphragm which is moved either by the magnetic field or by movement of the coil in the field.

The crystal types contain a thin piece of special material (an insulator) metallised on both sides to form a capacitor. The material bends under the influence of an audio voltage, to produce the sound. Being a capacitor, a crystal earphone does not pass d.c. Its resistance is infinite. But it offers an impedance to a.c. which falls as the frequency rises. This tends to make it accentuate treble notes.

In circuits like Fig. 2, where the audio is developed across a fairly high resistance (here $10k\Omega$) a crystal earphone is the natural choice. Since it passes no d.c. it may also be connected as shown dotted, no coupling capacitor being needed.

TRANSFORMER MATCHING

There are times when it is necessary to match a low impedance earphone to a high impedance audio source. This is a job for a transformer (Fig. 3).

Transformers need to be specified with care since many factors affect their performance. But when correctly designed

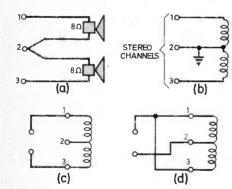


Fig. 1. The wiring configuration for stereo headphones is shown in (a) with the audio channels connected at points 1 and 3 (b) and earth at point 2. For mono use the headphones may be connected in series (c) or parallel (d).

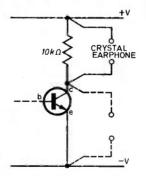


Fig. 2. In this circuit the audio is produced across a high resistance, so a crystal earphone is used since it passes no d.c.

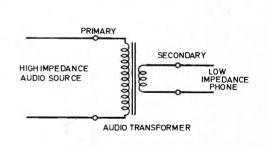


Fig. 3. When matching a low impedance phone to a high impedance audio source a transformer is used to multiply the impedance of the phone.

is fine if they are connected so as to replace an 8-ohm loudspeaker. But many home-built radios are designed to work into earphones of very much higher impedance.

In the old days of radio, the crystal set era, people listened on headphones whose two earphones were magnetic devices with an impedance of as much as 4000 ohms. In fact that was the d.c. resistance; the impedance to audio frequencies was very much higher.

is really the d.c. resistance but the a.c. impedance at most audio frequencies is about the same.

CRYSTAL EARPHONES

For really high impedance nowadays you must use crystal earphones. The single type looks just like one of the low-impedance magnetic earphones, and beginners sometimes come to grief by mistaking one for the other. But inside they are totally different.

and used they multiply the impedance of the earphone by the square of the ratio of turns on the primary to turns on the secondary. Thus a turns ratio of 10 multiplies 8 ohms to 800 ohms.

The current is multiplied by the same number which is why the use of a matching transformer can produce a big increase in volume, though only when conditions allow this. A transformer can't make energy, it can only enable you to make the best use of the available energy.

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CIRCUIT

This is the spot where readers pass on to fellow enthusiasts useful and interesting circuits they have themselves devised.

Payment is made for all circuits published in this feature.

Contributions should be accompanied by a letter stating that the circuit idea offered is wholly or in significant part the original work of the sender and that it has not been offered for publication elsewhere.

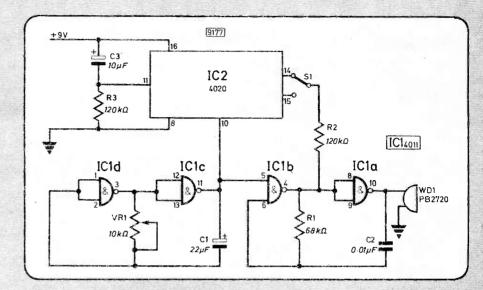
JOGGER'S PACEMAKER

This circuit produces a pulsing tone that can be used to help joggers keep a constant pace and, by increasing the pulse rate, improve their running.

By counting the number of steps the jogger takes it is also possible to calculate the approximate distance that has been run. This is achieved by changing the frequency of the pulsing tone every 500 or 1000 steps.

IC1a and IC1b form an oscillator that generates a tone of about 1000Hz. A second oscillator is formed by IC1c and IC1d and this controls the running pace by switching the first oscillator on and off. The pulse rate is set by potentiometer VR1.

To count the steps, the pulses from the slower oscillator are fed into a 14-stage binary counter IC2. When approximately 500 (exactly 512) pulses have been counted by IC2, pin 14 goes high and changes the frequency of the tone produced by IC1a-IC1b. After a further 512 pulses pin 14 goes low again and the tone changes back to the first frequency. If the



tone is to change every 1024 steps switch S1 should be set to pin 15 of IC2.

To calculate the distance you have run simply multiply the number of steps with

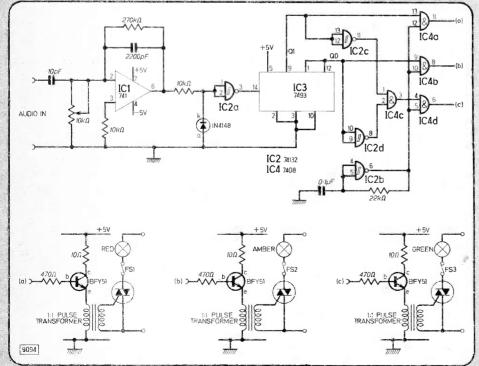
the length of your steps. For example: $1500 \times 1 \text{ metre} = 1.5 \text{km}$.

Joachim Ramkull, Lund, Sweden.

DISCO TRAFFIC LIGHTS

During the past few years many sound-to-light units have become

available. A variation on the normal light sequencer is presented here, which provides a traffic light sequence. The operation of the circuit is as follows:



An audio input of greater than 200mV is amplified, and high frequency components of the signal are reduced by the inclusion of a low-pass filter. The signal then enters IC2a, a Schmitt trigger, which is included to reduce spurious pulses entering IC3. IC3 is a 4-bit binary counter, though this application utilises only two of its four outputs.

IC2b causes high level signals entering IC4a, IC4b and IC4d to be modulated at around 1kHz. This is required for the operation of the transformers. The triacs should be chosen to suit the power rating of the bulbs.

The sequence is as shown below:

Q0	QI	Output
0	0	Green
1	0	Amber
0	1	Red
1	1	Red and Amber

Since this cycle repeats, the standard traffic light sequence (Red, Red and Amber, Green, Amber and back to Red) is followed.

A. Marshall, Old Basford, Nottingham.

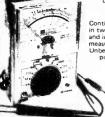
CIRCUIT

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THIS MONTH'S NEW KITS:

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channel receiver kit. Each kit comes with diagrams and notes,

but no circuit boards, the component layout being left to you. The data shows how to drive, reverse and steer two or more motors, With spare channels to perform other functions, Price £9.50 for both kits.

As in December Hobby Electronics, Designed originally for listening to wildlife this could also be used to listen through walls or from long distances. Complete kit including the case at £9.50

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I.I.V. 4 FRE-AWIE
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you have to do is fit this into the TV down lead and plug into the mains. Complete kit including the case at £9.50.

THE HE MICROLOG

This is a biggish project but you build a complete computer! Full constructional details appear in December Hobby Electronics. We will supply the complete kit less the rather expensive case for £18.50. We feel sure you can make a case yourself just as efficiently and save most of the cost.

3 CHANNEL SOUND TO LIGHT KIT



Complete kit of parts for a three channel sound to light unit controlling over 2000 watts of lighting. Use this at home if you wish but it is plenty rugged enough for disco work. The unit is housed in an attractive two tone metal case and has controls for each channel, and a master on/off. The audio input and output are by W's sockets and three panel mounting fuse holders provide thyristor protection. A four-pin plug and socket facilitate ease of connecting lamps. Special price is £14.95 in kit form or £25.00 assembled and tested.

TANGENTIAL BLOW HEATER



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6 pole 2 way

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high-resolution
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You have access to a range of 8 colours for foreground, background and border, together with a sound generator and high-resolution graphics.

You have the facility to support separate data files.

You have a choice of storage capacities (governed by the amount of RAM). 16K of RAM (which you can uprate later to 48K of RAM) or a massive 48K of RAM.

Yet the price of the Spectrum 16K is an amazing £125! Even the popular 48K version costs only £175!

You may decide to begin with the 16K version. If so, you can still return it later for an upgrade. The cost? Around £60.



Ready to use today, easy to expand tomorrow

Your ZX Spectrum comes with a mains adaptor and all the necessary leads to connect to most cassette recorders and TVs (colour or black and white).

Employing Sinclair BASIC (now used in over 500,000 computers worldwide) the ZX Spectrum comes complete with two manuals which together represent a detailed course in BASIC programming. Whether you're a beginner or a competent programmer, you'll find them both of immense help. Depending on your computer experience, you'll quickly be moving into the colourful world of ZX Spectrum professional-level computing.

There's no need to stop there. The ZX Printer – available now – is fully compatible with the ZX Spectrum. And later this year there will be Microdrives for massive amounts of extra on-line storage, plus an RS232/network interface board.



Key features of the Sinclair ZX Spectrum

- Full colour 8 colours each for foreground, background and border, plus flashing and brightness-intensity control.
- Sound BEEP command with variable pitch and duration.
- Massive RAM 16K or 48K.
- Full-size moving-key keyboard all keys at normal typewriter pitch, with repeat facility on each key.
- High-resolution 256 dots horizontally x 192 vertically, each individually addressable for true highresolution graphics.
- ASCII character set with upper- and lower-case characters.
- Teletext-compatible user software can generate 40 characters per line or other settings.
- High speed LOAD & SAVE 16K in 100 seconds via cassette, with VERIFY & MERGE for programs and separate data files.
- Sinclair 16K extended BASIC incorporating unique 'one-touch' keyword entry, syntax check, and report codes.





ZX Spectrum software on cassettes—available now

The Spectrum software library is growing every day. Subjects include games, education, and business/household management. Flight Simulation... Chess... Planetoids... History... Inventions... VU-CALC... VU-3D... Club Record Controller... there is something for everyone. And they all make full use of the Spectrum's colour, sound, and graphics capabilities. You'll receive a detailed catalogue with your

ZX Expansion Module

Spectrum.

This module incorporates the three functions of Microdrive controller, local area network, and RS232 interface.
Connect it to your Spectrum and you can control up to eight Microdrives, communicate with other computers, and drive a wide range of printers.

The potential is enormous, and the module will be available in the early part of 1983 for around £30.

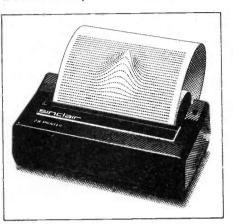
Sinclair Research Ltd, Stanhope Road, Camberley, Surrey GU15 3PS. Tel: Camberley (0276) 685311.

The ZX Printeravailable now

Designed exclusively for use with the Sinclair ZX range of computers, the printer offers ZX Spectrum owners the full ASCII character set—including lower-case characters and high-resolution graphics.

A special feature is COPY which prints out exactly what is on the whole TV screen without the need for further instructions. Printing speed is 50 characters per second, with 32 characters per line and 9 lines per vertical inch.

The ZX Printer connects to the rear of your ZX Spectrum. A roll of paper (65ft long and 4in wide) is supplied, along with full instructions. Further supplies of paper are available in packs of five rolls.



The ZX Microdrivecoming soon

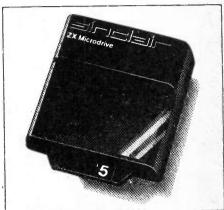
The new Microdrives, designed especially for the ZX Spectrum, are set to change the face of personal computing by providing mass on-line storage.

Each Microdrive can hold up to 100K bytes using a single interchangeable

storage medium.

The transfer rate is 16K bytes per second, with an average access time of 3.5 seconds. And you'll be able to connect up to 8 Microdrives to your Spectrum via the ZX Expansion Module.

A remarkable breakthrough at a remarkable price. The Microdrives will be available in the early part of 1983 for around £50.



How to order your ZX Spectrum

BY PHONE-Access, Barclaycard or Trustcard holders can call 01-200 0200 for personal attention 24 hours a day, every day. BY FREEPOST-use the no-stamp needed coupon below. You can pay by cheque, postal order, Access,

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EITHER WAY-please allow up to 28 days for delivery. And there's a 14-day money-back option, of course. We want you to be satisfied beyond doubt-and we have no doubt that you will be.

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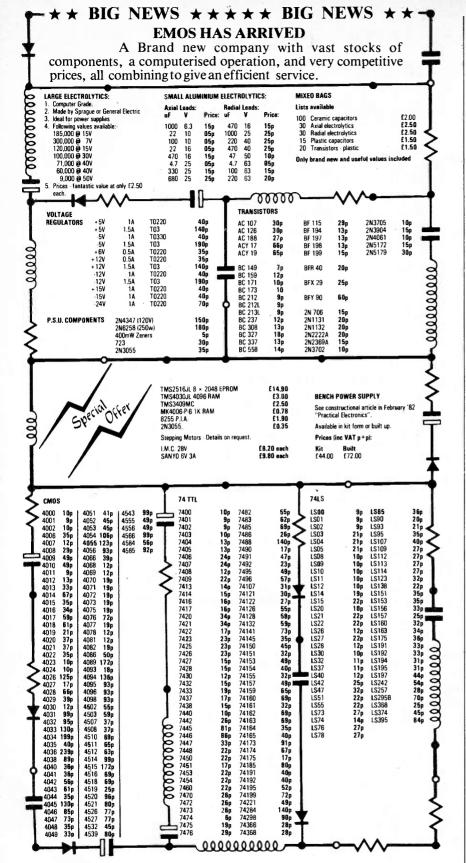
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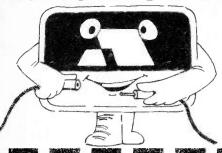
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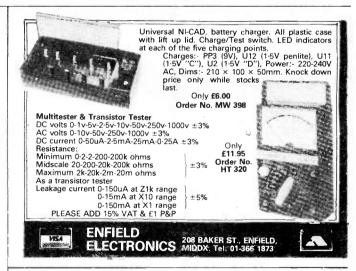
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console of own choice. These units are brand new, ready built and tested.

Features: Three digit tape counter. Autostop. Six piano type keys, record, rewind, fast forward, play, stop and eject. Automatic record level control. Main inputs plus secondary inputs for stereo microphones. Input Sensitivity: 100mV to 2V Input Impedance: 68K. Output level: 400mV to both left and right hand channels. Output Impedance: 10K. Signal to noise ratio: 45dB. Wow and flutter: 0.1%. Power Supply requirements: 18V DC at 300mA. Connections: The left and right hand stereo inputs and outputs are via individual scruened leads, all terminated with phono plugs (phono sockets terminated with phono plugs (phono sockets provided). **Dimensions**: Top panel 5½in × 11¼in. Clearance required under top panel 21/4in. Supplied complete with circuit diagram and connecting diagram. Attractive black and silver finish.

black and silver finish.

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SPEAKERS (15", 12" and 8"). These loudspeakers are ideal for both hi-fi and disco applications. Both the 12" and 15" units have heavy duty die-cast chassis and aluminium centre domes. All three units have white speaker cones and are (ground finish) fixing escutcheons.

Specification and Price:— 100 watt R.M.S. Impedance 80hm

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6 piano type keys

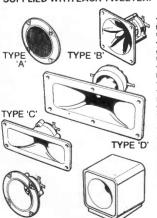
759 oz. magnet, 2" aluminium voice coil. Resonant Frequency 20Hz. Frequency Response to 2.5KHz. Sensitivity 97dB. Price £32 each. £3.00 Packing and Car-

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P232 Turntable * 'S' shaped tone arm

★ Belt driven ★ Aluminium platter

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100 WATT R.M.S. AND 300 WATT R.M.S. MODULES

100 WATT R.M.S. AND 300 WATT R.M.S. MODULES
Power Amplifier Modules with integral toroidal transformer power supply, and heat sink. Supplied as one complete built and tested unit. Can be fitted in minutes. An LED Vu meter is available as an optional extra.

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Sensitivity 1or Max. Output:
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T.H.D.: Less than 0-10
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Purposefully, designed 40 watt R.M.S. and

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