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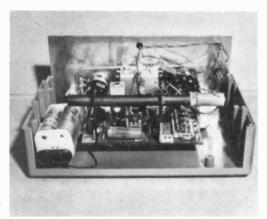
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VOL. 11 NO. 2 FEBRUARY 1982

PROJECTS . . . THEORY . . . NEWS . . . COMMENT . . . POPULAR FEATURES . . .





SCHOOLS

Electronic Design Award

COMPETITION

To allow for the Christmas holiday period and seasonal postal delays, the SEDAC organisers have agreed that Registration Forms (Part A) returned by January 31, 1982 will be accepted.

The final date for submission of Papers is unaltered, i.e. February 16, 1982.

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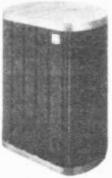


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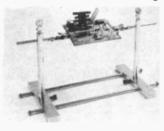
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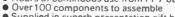


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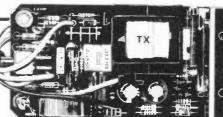
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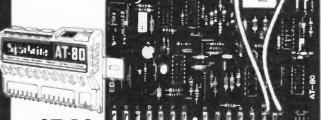
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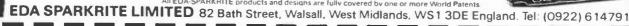
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NOT IN ISOLATION

A stranger to electronics flipping through the pages of this month's EVERYDAY ELECTRONICS might be forgiven for imagining that carpentry and mechanical engineering are essential parts of this hobby. Yet upon reflection, our browser might be nearer the truth than we perhaps

Evidence of the intrusion of "foreign technology" is to be found in two of this month's projects. The second and final part of the Automatic Garage Door article includes details of the electro-mechanical arrangements. No avoiding this-all the electronic wizardry comes to nought if the mechanics are not properly organised.

Carpentry? The Stereo Record Player entails the construction of a wooden cabinet and a pair of loudspeaker enclosures. Although the latter are purely acoustic devices they are integral parts of any sound reproducing system and their design and construction is reckoned to come fairly within the jurisdiction of electronics.

These two projects are different in that most of our designs begin and end with the electronics. In practical terms this usually means simply a circuit board assembly neatly stowed in a readily available plastic case. Rarely are tools other than screwdriver, pliers and soldering iron required. This is fine, generally speaking, and it makes the building of useful electronic gadgets and equipments a beautifully uncomplicated occupation.

Yet the appearance of projects such as the Automatic Garage Doors and the Stereo Record Player is rather important for-apart from their intrinsic functional value-they illustrate the fact that electronics does not have to be separate, self-contained and isolated. Quite the opposite in reality. The interface of electronic circuitry with non-electronic devices and mechanisms is important and upon this progress in technology is largely dependent.

No true enthusiast of electronics should be put out or disturbed by the appearance of any supposedly alien art, craft or technology in our pages, if it makes for overall completeness of the project in question. Obviously there are limits to how far we can or should go in these nonelectronic areas; but it is important to recognise that in its everyday usage, electronc technology has no real bounds. We cannot therefore logically draw lines around our favourite subject and ignore what lies beyond.

To paraphrase John Donne, "No technology is an Island, entire of itself . Most certainly not electronics.

The Bennett

Readers' Enquiries

We cannot undertake to answer readers' letters requesting modifications, designs or information on commercial equipment or subjects not published by us. All letters requiring a personal reply should be accompanied by a stamped self-addressed envelope.

We cannot undertake to engage in discussions on the telephone.

Component Supplies

Readers should note that we do not supply electronic components for building the projects featured in EVERYDAY ELECTRONICS, but these requirements can be met by our advertisers.

All reasonable precautions are taken to ensure that the advice and data given to readers are reliable. We cannot however guarantee it, and we cannot accept legal responsibility for it. Prices quoted are those current as we go to press.

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Stereo Record Player

Going back about 15 years or so "record players" were commonplace in many homes, but since the arrival of the music centre and portable cassette player, the availability and choice one has for this type of separate unit is now limited.

This was highlighted recently when the author was looking for a suitable unit for his son to use at home. The unit needed to be compact and portable so that it could be easily moved from one room to another and on occasions carried to a friend's house.

It was therefore decided to design and build a stereo record player which was inexpensive, simple to construct and yet robust enough to be carried around.

The unit was designed around a dual amplifier i.c. which requires few external components, a simple power supply, and when fed directly from a high output ceramic cartridge produces about 2 watts per channel into 8 ohm speakers.

RECORD DECK

There are many types of "deck" on the market that could be used for this project. The use of a top quality deck is not really justified. In fact a deck from an older type of "mono player" could be used providing a stereo ceramic cartridge could be fitted. Obviously the base board "cutout" for another deck must be made to suit.

The unit described here uses the inexpensive BSRP207 record deck fitted with a SC12H high output ceramic cartridge. It is a single play semi-automatic deck, that is, the arm is placed manually onto the record and after playing returns automatically and switches off. Use is also made of this on/off switch to switch the amplifier on and off, thus saving the need for an extra mains switch.

THE CIRCUIT

The complete circuit diagram of the Stereo Record Player amplifier is shown in Fig. 1. Right channel components have component references of corresponding left channel references preceded by 100.

ICl is a dual 2 watt amplifier for 8 ohm loads. It has internal current limiting short circuit protection and internal thermal shutdown, thus making it a very robust device.

The ceramic cartridge used in the prototype produced an average output when measured on an oscilloscope of approximately 1 volt peak-to-peak.

The cartridge outputs are fed directly across the left- and right-channel volume controls VR1, VR101.

As both channels operate identically only the left channel will be described.

Capacitor C2 couples signals from VR1 to pin 6 of IC1, the input to the amplifier which here is being used in the non-inverting mode. Resistor R1 sets pin 6 at half the supply voltage via pin 1 the internal bias pin. Any mains ripple on this pin is decoupled to earth via C3.

The resulting amplified output signal at pin 2 varies about half supply rail voltage. Part of this output voltage is fed back (via a resistor-capacitor network) to the inverting input, pin 7, to affect the gain of the amplifier as required.

GAIN

The gain of an LM377 operating in the non-inverting mode may be set by two external resistors R1, R2 as shown in Fig. 2, where part of the output is fed back to the inverting input.

This (purely resistive) arrangement should theoretically produce equal gain at all frequencies within the amplifier bandwidth.

In Fig. 1, you will see that the feedback circuitry is much more complex than the simple resistor circuit, but this can be reduced to just two "impedance" elements Z1 and Z2 to replace R1 and R2 in Fig. 2.

To make the gain frequency dependent, capacitors are introduced into the feedback network.

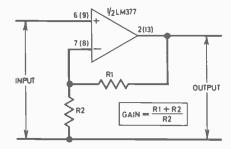


Fig. 2. Calculating the gain of the LM377 wired in the non-inverting mode.

Capacitors have reactance (resistance to a.c., where this resistance is inversely proportional to frequency). By incorporating these in the feedback circuitry, increase or decrease in the gain of the amplifier will occur within a specified frequency range.

In the prototype this method of varying the gain for particular frequencies is provided for by the adjustment of a single potentiometer, VR2. Together with its associated capacitors and resistors it has been designed to provide bass boost and bass cut.

BASS TONE CONTROL

The components R2 to R5, C4, C5 in the feedback network were chosen to give the amplifier a gain of about 23dB (×14) at all frequencies when the TONE control VR2 is set mid-way.

Potentiometer VR2 is a log. type and so when set to the mid-way position, the slider splits the resistive element into two portions equal to $90k\Omega$ and $10k\Omega$ with C4 and C5 respectively across each section.

Full bass boost occurs when the wiper is set at the C4 end of its travel. and full cut occurs when the wiper is set at the C5 end of its travel. Varying degrees of cut and boost occur between these two extremes.

OUTPUT

The two components R6 and C6 form what is known as a Zobel network. This has been included for reasons of stability when feeding reactive loads. Capacitor C8 couples the output signal to the speaker via the headphone socket internal contacts.

When the headphone plug is inserted contacts 2 and 3 open and are themselves isolated from the plug. Contact 4 meets with the tip of the plug while the body connects to earth. Resistor R7 (100 ohms) is thereby placed in series with the headphones (8 ohms) reducing the speaker output level by a factor of about 1/13.

Fig. 1. The complete circuit diagram of the Stereo Record Player.

COMPONENTS

Resistors

 $1 \text{M}\Omega$ (2 off) R4, R104 47k Ω (2 off) R5, R105 R1, R101 $10k\Omega$ (2 off) R7. R107 100Ω ½W (2 off) R2, R102

£30 excluding

10kΩ (2 off) R8 $2 \cdot 2k\Omega$

R3, R103 R6, R106 $1k\Omega$ (2 off) 2.7Ω (2 off)

All 1W carbon ± 5% except where stated otherwise

VR1, VR101 $1M\Omega$ carbon log. law (2 off) VR2 100k carbon log, law dual ganged

Capacitors

C1, C101 47μF 16V elect. radial leads (2 off)

C2, C102

0·1μF polyester (2 off) 47μF elect. 16 V radial leads C3

C4, C104 0.33µF polyester (2 off)

0.033µF polyester (2 off)

C5, C105 C6, C106 C7, C107 $0.1\mu\text{F}$ polyester (2 off) $470\mu\text{F}$ 25 V elect. (2 off) $2200\mu\text{F}$ 25 V elect.

C8

Semiconductors

IC1 LM377 dual 2 watt audio amplifier i.c.

D₁ TIL220 I.e.d. panel mounting bezel type

D2-D5 1N4001 1A 50V silicon rectifier diodes (4 off)

Miscellaneous

mains primary/15 V 1 A secondary transformer

FS₁ 1 A 20mm

FS2 315m A 20mm

SK₁ enclosed 6.3mm stereo switched jack socket (Tandy 274-277)

SK2, SK102 DIN chassis mounted loudspeaker sockets (and plugs) (2 off) Stripboard: 0.1 inch matrix size 24 strips x 37 holes; Veropins, single-sided (16 off); control knobs (3 off); 7-way 1A screw terminal block; self-adhesive rubber feet (16 off); 3-core mains cable, length to suit; screened cable, approx. 50cm; speaker cable, length to suit; 14-pin d.i.l. socket for IC1; cable sleeving; 4B A fixings for T1; 20mm chassis mounting fuseholders for FS1 and FS2 (2 off); heatsink for 14-pin d.i.l. i.c. (Tandy 276-9180)

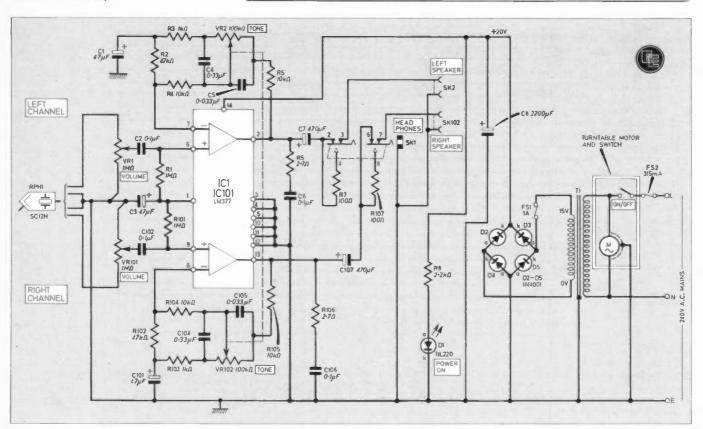
case: cabinet catches (2 off), attache case type hinges (2 off), chrome screw cups, see cutting list for wood sizes; speaker cabinets: speaker grille cloth; metal-

lized self-adhesive angle, see cutting list for wood sizes.

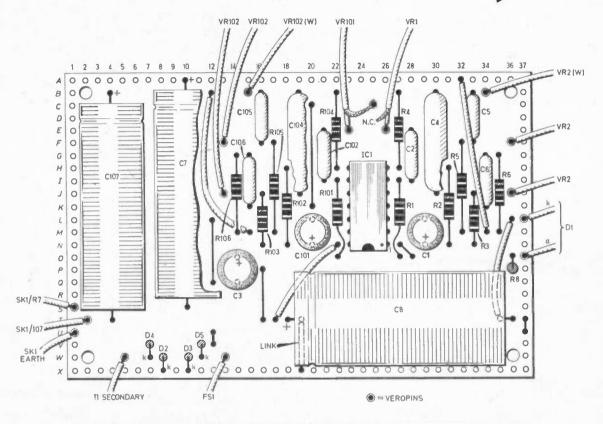
Record Deck **BSR P207**

Cartridge SC12H stereo ceramic

8 ohm 8 × 5 inch elliptical type rated at 2 watts or greater (2 off) Loudspeakers



Stereo Record Player



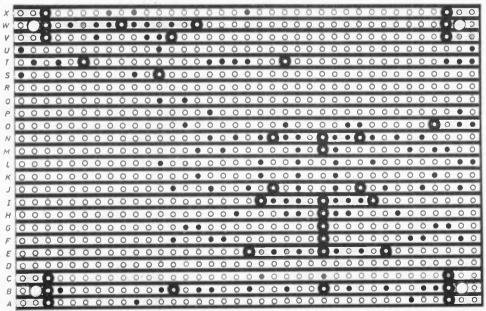


Fig. 3. The layout of the components on the topside of the stripboard and the breaks to be made on the underside (trackside) of the board. Note that all flying leads from this board are via Veropins. The screen on the cable from VR1 is not connected at the board end.

POWER SUPPLY

The power supply and amplifier are switched "on" and "off" as mentioned previously by the on-board motor switch.

T1 secondary provides 15 volt a.c. at 1 amp. Full-wave rectification is accomplished by diode bridge D2-D5. Its output is smoothed by reservoir capacitor C8 producing a d.c. supply voltage of 20 volts.

Resistor R8 limits the current through l.e.d. D1 which functions as a POWER ON indicator and is located on the control panel.

ennstruction starts here

AMPLIFIER BOARD

The layout of the components on the stripboard (24 strips \times 37 holes \times 0·1 inch matrix) is shown in Fig. 3. All flying leads are soldered on the board topside via Veropins.

Start by drilling the four fixing holes and cut the strips where shown. Next fit and solder the Veropins and link wires into position. After locating and soldering the i.c. socket into position, the rest of the components can then be soldered in place, starting with the smallest components first.

Glue the heatsink to the i.c. using "Araldite" and mark the position of pin 1 on the sink for future reference, and then leave to one side to set. All lead-out wires are to be connected when the board is in position. The next stage recommended is to prepare the deck baseboard.

DECK BASEBOARD

The baseboard was made from 8mm thick plywood. A scaled template for this is shown in Fig. 4. The cut-out

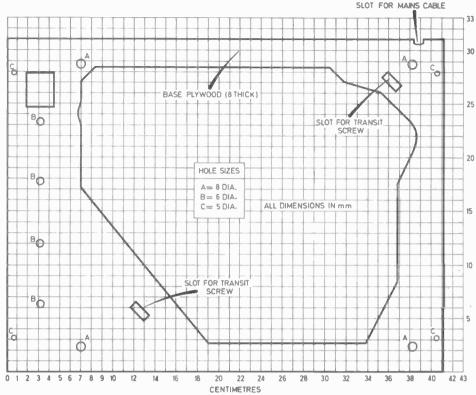


Fig. 4. Scaled template for the cut-out in the deck baseboard to suit the BSRP207 and chassis mounted transformer. Viewed from underside.

was made using a Stanley trimming knife fitted with a type 1275B saw blade. If using a deck other than the BSRP207, a different cut-out contour will need to be originated.

Simply drill a hole to clear the blade and saw around the line finishing off with a woodfile and sandpaper. Next drill all holes, cut the three slots with a small chisel and fit the two

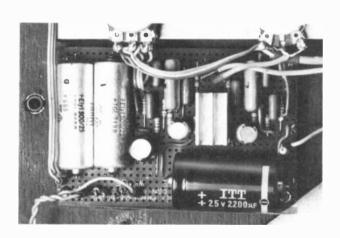
wooden blocks that support the control panel. Now press into place the deck "spring cups", and check that the deck sits in these correctly.

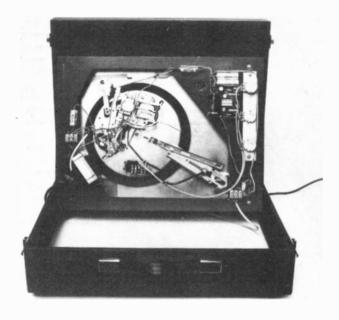
SECURING TRANSFORMER

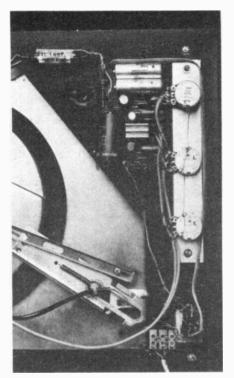
With the deck fitted to the baseboard unscrew the transit screws to tighten the deck against the base.

The fully wired prototype with deck baseboard raised to show component assembly on underside.

Lower left: Close up view of the prototype component board.





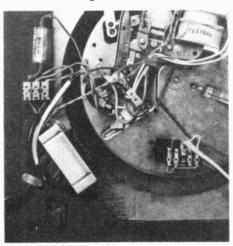


Shows the fully wired up component board and control panel. You can also see the headphone socket and speaker terminal block.

Then position the transformer so that its fixing bolts (when fitted) will be hidden by the turntable. Remove the turntable by easing out the spring clip attached to the centre spindle and lift off. Drill 4BA clearance holes, secure the transformer and replace the turntable.

The deck support spring nearest the transformer may have to be stretched a little to compensate for the extra weight, thereby keeping the deck level. The baseboard is now ready for covering or painting as required.

The mains and motor wiring of the prototype unit. The pick-up tag strip can be seen lower right.



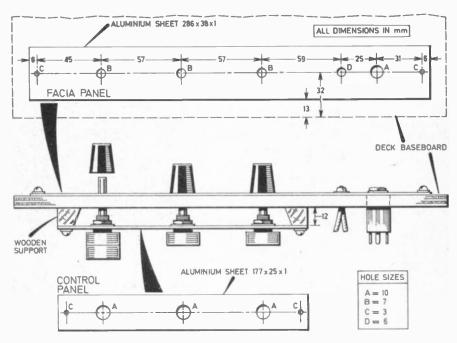


Fig. 5. Drilling and mounting details for the control and facia panels.

Before proceeding to the wiring stage the control and facia panels are cut and drilled as shown in Fig. 5.

WIRING UP

The complete wiring-up details are shown in Fig. 6. Unscrew the transit screws and support the upturned baseboard with two wooden bearers. Locate the mains wire that goes from the microswitch to the motor (lead "X"). Cut this wire and insert the single terminal block into position. Now fit the mains and speaker terminal blocks in place followed by the two fuseholders.

Fix the three potentiometers to the control panel and then screw this assembly into the small wooden supports. Fix the bezelled l.e.d. and headphone socket to the facia panel feeding the l.e.d. wires through the baseboard.

The amplifier board can next be screwed down to the baseboard on spacers. Small lengths (5mm) of plastic sleeving make suitable spacers.

In the prototype there were two additional (white) leads from the motor. These are not required and should be adequately insulated after cutting back close to the fitted motor wiring clamp. Wiring between pick-up and 5-way tag-strip is carried out by the deck manufacturers.

With reference to Fig. 6, carry out the wiring up, keeping all leads as short as possible. Note that the contact numbers for SK1 are printed on the socket body. Use screened lead where appropriate.

INITIAL TESTING

The first test is to make sure the deck functions correctly. Insert the mains fuse FS2 leaving FS1 out for the moment.

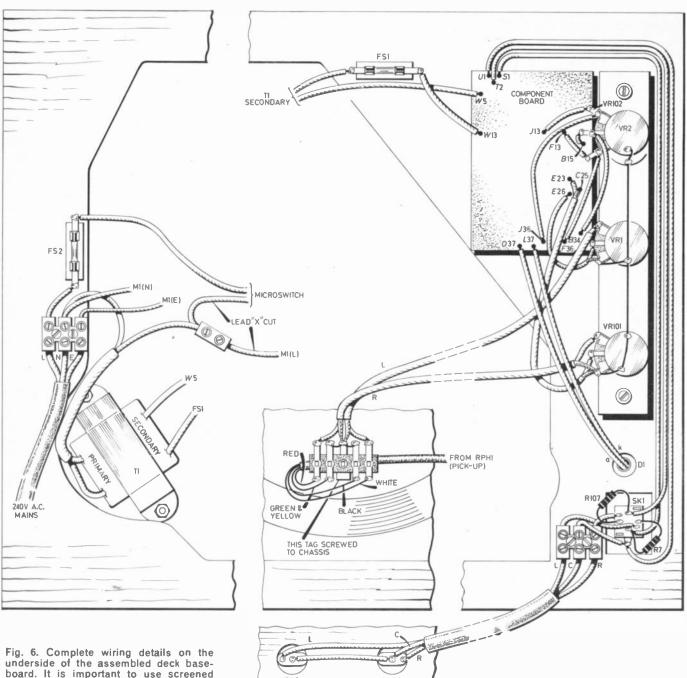
Now with the deck and baseboard supported by the wooden bearers, release the transit screws and switch on. Put a record on and check that the deck works satisfactorily.

Next test the power supply. Lock the arm and tighten the transit screws and turn the deck upside down. In this position the motor must be disconnected from the supply otherwise its bearings may be damaged. Unscrew the motor lead from the single terminal block and insulate it with tape. At this stage do not have the i.c. in its socket. Insert FS1 and switch on. Measure the voltage at pin 14 of the i.c. socket, this should read 20 volts or be close to this. Check also that D1 lights up.

Once this has been confirmed, switch off and insert the i.c. in its socket making sure it is the correct way round, otherwise it will certainly be destroyed. Turn both volume controls fully down and without speakers or headphone connected, switch on and measure the voltage at pins 2 and 13. Both should read 10 volts (or half the measured pin 14 voltage). If any of these tests fail, thoroughly check all the wiring around each stage to locate the fault before proceeding.

FINAL TESTING

Assuming all voltages are correct, solder a short length of cable to each



board. It is important to use screened cable where indicated.

speaker and connect up to the speaker terminal block. With both volume controls fully down and the deck still upside down, switch on. A faint hiss should be heard in both speakers, move each volume control about a quarter of a turn, hold a screwdriver by the blade and touch each amplifier input at the small tag strip on the deck. A loud buzz should be heard. Leave the unit switched on for a while and check that the heatsink stays cool.

Once satisfied switch off and reconnect the motor lead and set the

deck up to play a record. Make sure both channels give equal volume without distortion and that the tone control functions satisfactorily: anticlockwise for cut, clockwise for boost, from centre position. Carry out the same check with the headphones connected.

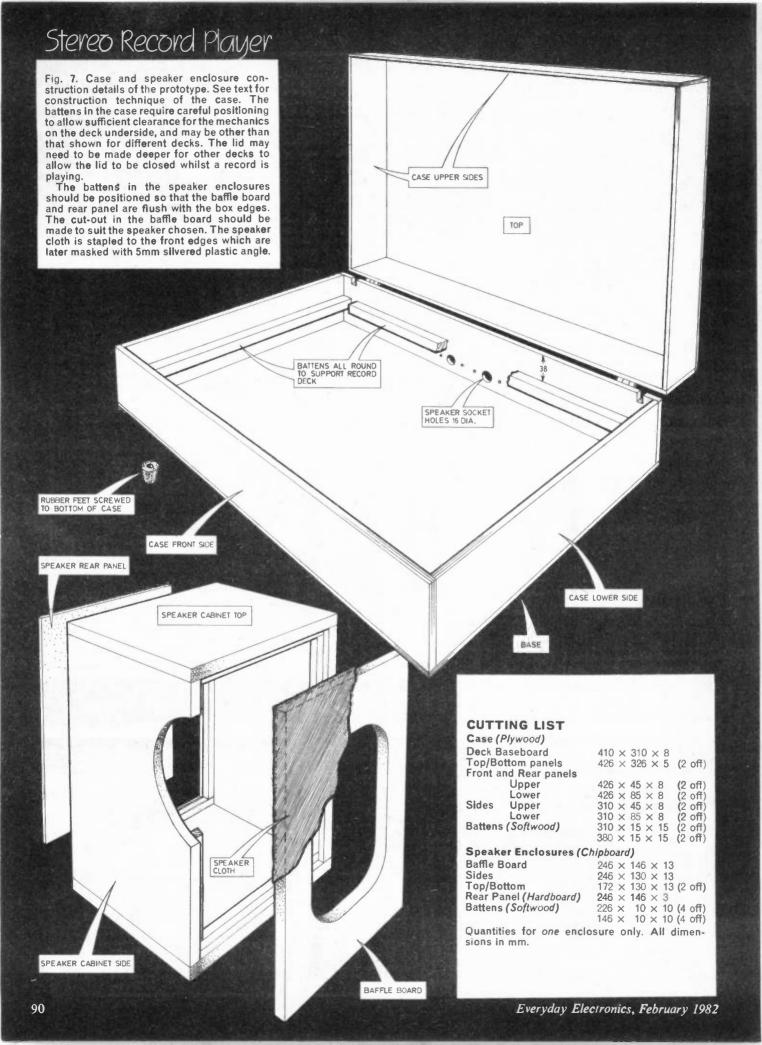
Once again if problems arise check all wiring for possible mistakes and bad joints. If mains hum is present check that there is a good earth to the deck, and the two screened cables to the board. Only one of the screens should go to the board earth pin,

If the unit so far built gives satisfactory performance, the case and speaker enclosures may now be prepared.

THE CASE

Plywood was chosen for making the cabinet, the main structure being 8mm thick with the base and lid 5mm thick. For details see Fig. 7.

The cabinet was first made as a complete box all joints being glued and pinned. Then using a woodworkers marking gauge set to the



correct lid depth mark around the box and cut with a small tenon saw. This method ensures the lid fits the base perfectly.

Next glue and pin the baseboard supporting battens and cut the slots at the rear of the cabinet to accommodate the mains and headphone cables. Drill 15mm holes for the speaker sockets and temporarily screw them into place.

The hinges and catches are fitted with 9mm No. 6 round head screws. Drill the two holes and fit the carrying handle which comes complete with fixing nuts and bolts. These bolts need to be cut flush with the nuts to allow the deck baseboard to slot into place.

All these items must be removed before the final finish is applied to the cabinet.

Finally, the self-adhesive rubber feet are stuck into position on rear and underside. The rear feet are necessary to protect the hinges and speaker sockets should the unit be put down whilst it is being carried around.



The speaker cabinets are made from 13mm chipboard the rear panel being hardboard, see Fig. 7 for details.

The speaker cut-outs accept 8×5in elliptical speakers and are made using the same technique as described for the deck baseboard. All joints are glued and pinned, the rear panel being pinned only, to allow access. Once the baffle has been fixed, countersink and fill-in all holes.

Because some grille cloth is an open weave its a good idea to paint the front of the enclosure with a similar matt colour as the cloth.

At this stage the outer finish can be applied to the cabinets; the prototype was given several coats of matt black paint.

The grille cloth should be cut about 25mm or so larger than the front of the enclosure, then stretched across and temporarily held down at the edges with drawing pins. The cloth can be permanently fixed down using a small staple gun. To mask the staples and edges of the cloth a length of right angled metallised self-

adhesive edging was fitted all round, mitreing the corners to give a neat appearance.

The speaker is fixed into place with 4×15 mm long No. 8 round head wood screws. Solder the speaker leads on, and fit a cable staple to prevent any stress being applied to the speaker terminals. Cut a small slot at the bottom of the rear panel to allow the cable to come through, and secure the panel with short panel pins.

The enclosures may now be completed by fitting self-adhesive rubber feet.

FINISHING OFF

Prior to cleaning and lettering, the facia panel is removed by unsoldering the panel indicator wires and removing the headphone fixing nut and washer. Once off it can be rubbed down lightly with steel wool to give a brushed aluminium effect, then thoroughly washed with hot soapy water before lettering.

"Magic letters" sold by W. H. Smith were used on the prototype.

They are simply rubbed down onto the aluminium, then the whole panel sprayed with clear lacquer for protection.

Finally, the deck baseboard is secured down with chrome finished screws and cups. H





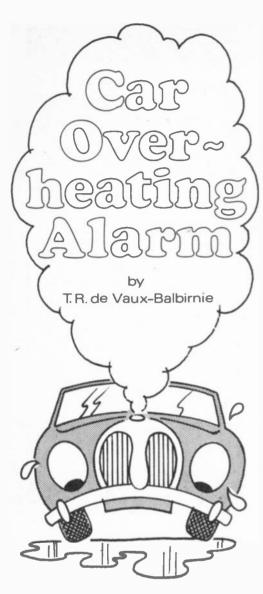
JACK PLUG & FAMILY...

BY DOUG BAKER









THIS circuit has been designed to give an audible warning when a car engine reaches a predetermined temperature. It was originally intended for use as an overheating alarm when towing a caravan but has proven equally useful in a number of other situations where overheating could occur. Trips at high speed, touring in hot countries and high altitude driving can also cause overheating in cars normally quite free from this problem.

If warned in time, the driver may simply ease off a little or stop for a short while to allow the engine to cool. Without such warning, water will be lost and the situation can become more serious.

Although most cars are fitted with a water temperature gauge, they do not give a positive signal in the event of overheating, so it is quite possible to miss the pointer creeping slowly into the red. An audible warning provides the answer; being high pitched rather than loud, it easily "carries" above engine noise.

The driver must always check that there is no actual reason for overheating such as a blocked radiator, leak or broken pump. This project has not been designed to cover-up poor maintenance!

POSITIVE OR NEGATIVE EARTH

The circuit is suitable for both positive and negative earth cars without modification.

The overheating alert is simple to construct, being designed around two cheap and readily available integrated circuits, ICl, a 741 Operational Amplifier and IC2, a 555 Timer. An enginemounted thermistor RTH1, "senses" the temperature.

The circuit is connected to the car's electrical system such that it will only operate with the ignition switched on so there is no drain from the battery when the car is at rest. Sounding or not, the current requirement is very low, around 5mA.

CIRCUIT DESCRIPTION

The circuit works in the following way (see Fig. 1). As the temperature of the engine rises, the resistance of the RTH1 falls. This component, together with R1, from a potential divider across the supply resulting in a voltage appearing at Pin 2 of the opamp IC1. This voltage will therefore fall as the temperature rises. VR1 forms a second potential divider which gives a reference voltage at Pin 3 of the IC1, this voltage dependant upon the setting of VR1. When the voltage at Pin 2 falls below that at Pin 3, IC1 switches 'on' and the supply voltage appears at the output, Pin 6. This enables IC2, the 555 timer, wired as an astable multivibrator, which results in an audible tone from the loudspeaker LS1. The temperature at which this happens is preset by suitable adjustment of VR1. The frequency of the alarm note is determined by the setting of VR2 and this may be adjusted for best effect at the end.

If the thermistor is not as specified, that is with a room temperature (20°C) resistance of about $1M\Omega$ and about $50k\Omega$ at 100°C, then R1 may need changing and the value should be approximately the thermistor resistance at 100°C.

Slight variations in battery voltage will occur but these have no effect on the operation of the circuit. Rapid fluctuations from the charging circuit may result in a "warbling" note from the loudspeaker but this is of no consequence.

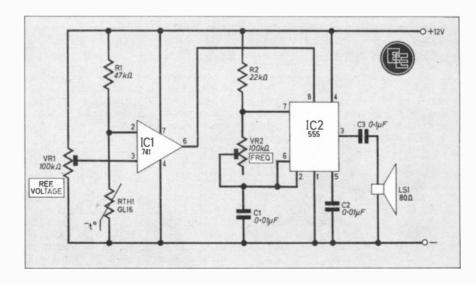
CIRCUIT BOARD

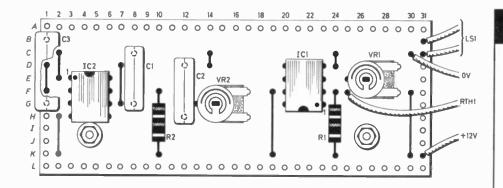
The circuit panel is built on a small piece of 0·1 inch matrix stripboard, 12 strips by 31 holes (see Fig. 2). In the prototype, the i.c.s. were soldered direct to the panel. This is acceptable where the constructor has confidence in the quality and speed of his soldering, but if there is some doubt 8 pin d.i.l. sockets should be used. Note that IC1 is upside down with respect to IC2. This made for simpler construction in the prototype.

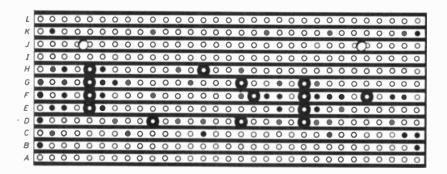
There are several places where the copper strips have to be cut and it is important not to miss the track breaks between the pins of the i.c.s. The best tool to use is a proper stripboard cutter but a small hand held twist drill makes a good substitute. Several wire links are required and these should be on the component side of the panel as shown.

Add the resistors, capacitors and potentiometers and finally wire the loudspeaker in with two flexible wires approximately 100mm long.

Fig. 1. Circuit diagram of the Car Overheating Alarm.







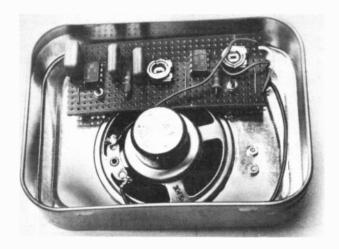


Fig. 2. Layout of components on the stripboard and positions of track breaks on the underside view. Note that the thermistoris connected remotely across RTH1 and 0V wires.

View inside case showing the board assembly and speaker positions.

BENCH TESTING

When the circuit is complete and has been checked visually, it may be tested using a 9 volt battery. To do this, VR1 should be set fully anticlockwise and VR2 to approximately the mid-position. The thermistor and a 9V battery should then be temporarily connected. A note may be heard from the loudspeaker, and if not, VR2 should be adjusted to the position where a shrill note is emitted. VR1 is then backed off to the point where the note is just silenced.

If the thermistor is now warmed slightly by holding it between the fingers or dipping the end in warm water, the tone should be heard once again. When the temperature falls it should switch off. These tests will confirm that all is well with the cir-

cuit and the panel may now be built into a suitable box.

CASE

Although a tobacco tin was used for the prototype, care must be taken when using a metal box to prevent short circuits. However, any suitably sized case of any material can be used.

A series of small holes, drilled in the base, form a "grill" for the loud-speaker which may then be secured in place using epoxy resin adhesive applied thinly around the rim. A hole should be drilled for a grommet through which the connections to the circuit board will pass. A 3-way terminal block should be secured to the underside for these wires (see Fig. 3).

COMPONENTS

Resistors

R1 $47k\Omega$ R2 $22k\Omega$ All $\frac{1}{2}$ W carbon $\pm 5\%$

Shop Talk

Capacitors

C1, 2 0·01µF polyester (2 off)
C3 0·1µF polyester

Semiconductors

IC1 741 operational amplifier IC2 555 timer

Miscellaneous

RTH1 GL16 bead thermistor, negative temperature coefficient, 1MΩ @ 20°C and 50kΩ @ 100°C

VR1, 2 100k miniature horizontal preset (2 off)

LS1 70-80Ω minature speaker, 57mm dia.

0.1 inch matrix stripboard, 12 strips by 31 holes; case to suit; 3 way 5 A terminal block; 2 way 5 A terminal block; 8 pin d.i.l. i.c. holder (2 off); grommet; 7/0.2 connecting wire; thin copper sheet approx 20 × 60mm; epoxy resin adhesive; automotive type spade connectors; mounting hardware; 8B A screws, nuts and insulating stand-offs (2 off); p.t.f.e. sleeving for thermistor leads (approx 100mm)

Approx cost Guidance only £7 excluding

Two holes should be drilled very carefully in the circuit board and corresponding holes drilled in the case. These are to secure the panel using short stand-off insulators and 8BA nuts and screws. The three wires from the circuit board should be passed through the grommet and connected to the terminal block and labelled +12V, 0V and "RTH1" as shown in Fig. 3.

THERMISTOR

Next, the thermistor unit should be prepared. Thermistors are extremely delicate components and could not possibly be used in an unprotected state. Although some form of protection is essential, heat must readily pass to the thermistor.

The method shown in Fig. 4 proved effective in the prototype. A piece of thin copper sheet was cut and bent into the form of a "P" clip, the cavity for the thermistor was formed by rolling the sheet around a thick nail. The thermistor leads must then be sleeved with p.t.f.e. insulation and the thermistor encapsulated in epoxy resin adhesive in the cavity. A 2-way terminal block should be glued on top of the clip and the thermistor connections made to it.

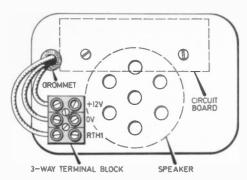


Fig. 3. Underside view of case showing speaker "grill" and terminal block.

The prototype thermistor assembly unit was secured to the engine using epoxy resin adhesive. A flat, horizontal surface is ideal and very thorough cleaning of the contact area is essential and slight roughening of the surfaces helps.

The clip must make good contact with the engine for efficient transfer of heat and it must be situated away from direct sources of heat like the exhaust manifold and also away from cool, moving air. It may require several experiments to find the best position. (A clip attached with adhesive may be removed by careful use of a penknife blade.) Obstinate cases will demand a thermistor unit which is "boxed in" over the top surface with fibreglass insulation. It is also necessary to keep the thermistor unit dry.

INSTALLATION

With the thermistor in position, temporarily at least, two lengths of stranded wire should be run from it to the main unit inside the car. For the time being, the main unit should be placed in a temporary position so that a passenger can adjust VR1 and VR2 whilst driving along. The positive (+12V) connection on the main unit should be taken to a fuse which is "live" only when the ignition is switched on. For cars having a positive earth system this will be the negative (0V) connection.

A 12V test lamp with one side to chassis will locate a suitable fuse. Ensure that the light goes "off" when the fuse is removed thus confirming that the correct side of the fuse is being used. The negative connection on the unit should be run to an "earth" point (the positive connection

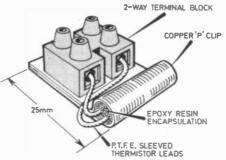


Fig. 4. Thermistor mounting assembly.

on positive earth cars). The leads from the thermistor unit will be connected across RTH1 and the negative (OV) connection regardless of the polarity of the cars electrical system.

IN-CAR TESTING

To test the circuit, the ignition should be switched on and VR1 set fully anticlockwise. A tone should be heard and VR2 adjusted as necessary to give a high-pitched whistle. VR1 should now be adjusted clockwise to silence the alarm and the car driven to bring the engine up to full operating temperature.. VR1 should be advanced at intervals to keep the alarm just "off".

If the sound has a tendency to cut out when the engine is revved up this is usually cured by raising the pitch of the note by means of VR2. If the operation of the circuit is erratic and seems to depend on the speed of the car rather than the temperature this indicates that the siting of the thermistor unit is not good enough and more attention must be given to prevent the unit from being cooled by the car slipstream.

The final settings for VR1 and VR2 can only be found after a proper trial run. VR1 should then be adjusted so that the tone is just "off". Remember, DO NOT attempt to adjust VR1 and VR2 whilst driving along. Get a pas-

senger to assist you.

By PAUL YOUNG

Electroneer

Here we are again at the start of a new year and with our eternal optimism expecting it to be better than the last. I go along with that sentiment, as Robert Louis Stevenson said, "It is better to travel hopefully than to arrive", and although he was talking about a journey, I don't see why it should not equally apply to time.

I would like to thank all those kind readers who wrote to me suggesting a word for an Electronic Engineer. The con-sensus seems to be "Electroneer" but while it seems a logical conclusion, to my rather whimsical way of viewing things it conjures up a vision of a man with ears made of electrons!-They probably are anyway.

I do agree it is difficult to think of something that rolls off the tongue and I

haven't come up with anything better.
Here are some of the suggested words:
From Mrs. Doreen Tomlin of Bracknell,
Berks., "Dynamician", Mr. Stephen J.
Parker of Glossop, Derbyshire, "Electronist", Mr. W. C. Hobson of Humberside and Mr. Albert H. Scott of Jersey, "Electroneer".

Anxious to Please

"He who pays the Piper calls the tune", and never has that been so true as today. I think the day of the vast Electronic store,

run by a Tycoon who changes his Rolls everytime the ashtrays are full, is disappearing, and this has been replaced by a larger number of smaller outfits.

From the customers point of view this may be a step in the right direction. In case of complaints it is easier to get to the man at the top and be certain of getting some speedy action.

Most of us are enthusiasts and anxious to please, but we would ask that you help us to help you, by being clear in your requests. Remember that unless you are ordering a specific item from a catalogue, where it has an identifying number, there are certain data we must know, such as voltage, current, wattage, capacity, inductance, impedance, according to what the item is you are ordering.

You may also required the item not to exceed a certain physical size. If that is the case, please tell us. Only today I had a letter from a customer asking me to send him a mains transformer, without telling me what output current he required, an electrolytic capacitor, without stating the voltage, and would I also send him two general purpose transistors with a medium gain. It must also be said that when a customer sends in a long list of miscellaneous items which he wants us to work out and price, there is always the sneaking suspicion that he is giving us half an hours work, when we are already overloaded, in order to save himself the cost of a catalogue.

Match This

My last little grumble, is an error of which I am sure the true electronic constructor is not guilty. It is the chap who walks in with some odd part of a foreign set, which is broken and asks us if we have one like it. It is almost like going into an ironmongers when you want some spare part for your car. Having said that, I must redress the balance by saying (and I know all my colleagues would agree with me) that our customers are among the very best and always appreciative of any help we are able to give them.

In Brief

Citizen Band Radio is officially with us. and although I haven't dabbled in it yet, I think it will do much good, particularly in bringing companionship and comfort to the old, the sick and the lonely, and more important still, help if they are in trouble.

I was very sorry to have missed the Breadboard Exhibition this year, but I don't think I was entirely to blame. I expected to see some publicity for it a few weeks was scheduled, but none before it appeared.



By Dave Barrington

Circuit Board Holder

We are always on the look out for any "gadgets" or items that help to make the task of construction that much easier.

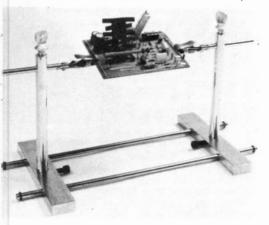
In this respect, we have already drawn readers attention to the excellent Minibench work holder with flexible arms from Absonglen. This is now joined by another form of circuit board holder manufactured by Carlton Nichol.

Two models are available. The CNC6, which is probably more suited to our kind of work, and the CNC9, able to accept Euroboard size boards. Both are constructed in aluminium and plated steel and allow rotation of circuit boards through 360 degrees, with locking in any position.

The CNC6 will take boards up to $254 \text{mm} \times 178 \text{mm}$ (10 \times 7in). The boards are held by spring loaded "jaws" and width adjustment is by lockable metal rods. The CNC9 will take boards up to 203mm × 203mm (8in × 8in) and are held in position by sliding Vee clamps.

An anti-static foam pad is also available as an optional extra to allow the insertion of a number of components before rotating the p.c.b. for soldering. The pad, which is on a backing plate, clips onto the rotating arms of the board holder.

The CNC6 costs £13.80 and the CNC9 £15.95. Whilst the anti-static foam pad



The CNC6 Circuit board holder from Carlton Nichol

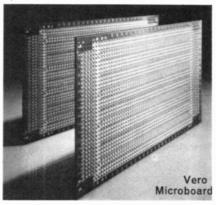
costs £9·20. For more details contact Carlton Nichol & Co. Ltd., Dept EE, Goldkey Industrial Estate, Kelvedon, Essex.

Microboard

We are sure that no one doubts the versatility of stripboard, but there are times when another form of component mounting board would be most welcome. This is particularly evident when dealing with multi-pin i.c. devices.

Recognising this possible gap, Vero Electronics have taken a leaf out of the industrial market and decided to launch their first Euroboard size Microboard onto the retail and amateur constructor markets.

Made from s.r.b.p. material, the Microboards are available in two sizes, 160mm × 100mm and 160mm × 233mm; priced £3.47 and £5.47 respectively. The boards are also compatible with indirect connectors and international card frame systems.



It is claimed that the boards will accept any integrated circuit package, and allow high component packing density. On the component side, each board is printed with an "island" pattern for ease of wiring and is suitable for both solder and wire wrap applications.

According to Mike Humphrey, Vero's Retail Sales Manager, "the development of a board specifically for the hobby market is an important step for the company. More and more, industry has been moving towards the use of Eurocard sized microboards.

"For the hobbyist, the availability of this microboard pattern in Eurocard format increases still further access to our wide range of industrial quality products."

The Microboards are available from Vero's usual retail outlets, or direct from Vero Electronics Ltd., Dept EE, Industrial Estate, Chandler's Ford, Hants, SO5 3ZR.

CONSTRUCTIONAL PROJECTS

Stereo Record Player

The record deck chosen for the Stereo Record Player is the BSR P207 with a SC12H ceramic cartridge. This was selected because of its high output, This was typically 1V peak-to-peak in the authors model.

The system has been designed around this cartridge and to obtain similar performance readers are advised to keep to the components specified. To date, the only source we have been able to locate for the deck and cartridge is from Radio

Components Specialists. We understand that some local hi fi shops may be able to help.

The stereo jack socket for the headphone input is a multi-contact type, with internal switch, and may be difficult to locate. The one used in our unit was obtained from a Tandy store and is listed as order number 274-277.

Practically any general purpose 8 ohm speakers will suffice for the enclosures. The cut-out for the speaker used will have to be tailored to suit. Of course, one of the better quality speaker kits and readymade enclosures, from such advertisers as RT-VC and B.K. Electronics would enhance the final result.

For those constructing their own enclosures, the speaker grille cloth may be obtained from Home Radio. The plastic angled edging for the cabinet should be available from the local hardware or DIY

Cine Interval Timer & Frame Counter

Only a couple of items need further comment when purchasing components for the Cine Interval Timer and Frame Counter.

The prototype unit uses 500m A general purpose plastic encapsulated transistors. However, we feel the BFY50's would be

better suited here.

The designation "E" in the type number for the integrated circuits refers to a "buffered" output and this type should be used. It is quite possible that other equivalent i.c.s will work in this circuit, but we have not tested them in the prototype and cannot vouch that they will work.

Provision for a relay option has been included for a remote single-frame opera-tion. Almost any 9V relay with a coil resistance from approximately 185 ohms and a set of normal open contacts will be suitable. The author used a 9V 410ohm

Medium Wave Radio

The coils called for in the Medium Wave Radio are Denco types and these are available direct from Denco Ltd, Dept EE, 355 Old Road, Clacton-on-Sea, Essex CO15 3RH.

The Jackson 00 gang 208/176pF tuning capacitor is available from Home Radio for the sum of £4.40 plus VAT. They also supply suitable trimmer capacitors and the MT31A/3 or the MT31A/12 can be

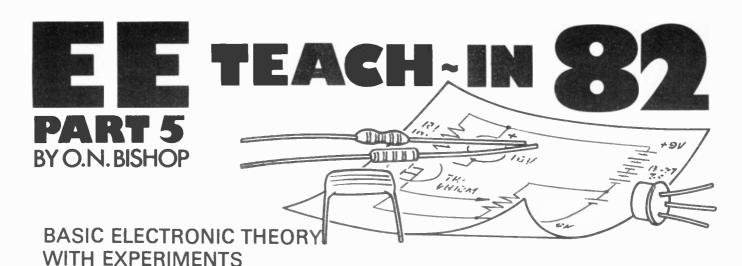
We understand that Watford Electronics are able to supply all the Denco coils, compression trimmers and the Jackson tuning capacitor.

Automatic Garage Door

The final article for the Automatic Garage Door deals with the mechanical arrangements and setting-up.

The motor used in the designer's set-up was a Fracmo type, currently stocked by Service Trading Co, (Dept EE, 57, Bridgman Road, Chiswick, London W4 5BB), with an output shaft running at approximately 56 r.p.m. and a more than adequate torque of 50lbs/in. We have been informed that they have only a limited supply, but are able to supply a near equivalent type in their Parvalux range with 42 or 30 r.p.m.

at 50lbs/in. torque. Most of the hardware for the door gear should be available from local builders merchants or hardware stores.



BIPOLAR TRANSISTORS

When people speak of "a transistor", the bipolar junction transistor is the sort they usually mean. This is the kind which is used in most circuits, except in integrated circuits where the MOSFET is the dominant type.

It is easy to understand how f.e.t.s work, which is why we began this course by studying them. Bipolar transistors are harder to understand, but they are much easier to manufacture, so they have been the most widely used type until recently.

BASE BASE POP

Fig. 5.1. Circuit symbols for the two types of bipolar transistors.

Field effect transistors belong to the group of transistors known as unipolar transistors. This name means that only one kind of charge carrier occurs in it. We have *n*-channel f.e.t.s in which charge is carried by electrons. In *p*-channel f.e.t.s charge is carried by holes.

Bipolar transistors use both holes and electrons in the same transistor. The action of a f.e.t. depends on the effect of an electrical field. The bipolar junction transistor depends on the action of a

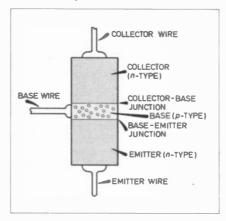
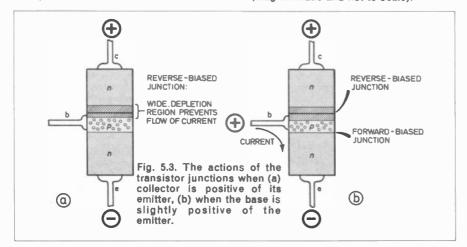


Fig. 5.2. Structure of an npn transistor (diagrammatic and not to scale).



pn-junction. The pn-junction was described in Part 3.

From now on we will use the word "transistor" to mean "bipolar junction transistor", so saving a lot of space. You will already have noticed the symbol for the transistor appearing in circuit diagrams on the pages of this magazine. You may have noticed that there are two types, npn and pnp, see Fig. 5.1.

Transistors can be made from silicon or germanium, though the silicon ones are far more popular today.

A transistor is a three-layered device. An *npn* silicon transistor consists of a layer of *p*-type silicon sandwiched between two layers of *n*-type silicon. The layers are named as shown in Fig. 5.2.

HOW AN NPN TRANSISTOR WORKS

The transistor has two pn-junctions. If the transistor is connected as in Fig. 5.3a, the collector-base (c/b) junction is reverse-biased, so we would expect that it is impossible for a current to flow from collector to base and on to the emitter. If we apply a slight positive potential to the base, the base-emitter junction (b/e) becomes forward-biased (Fig. 5.3b). If the potential is greater than the forward voltage drop (0.6V), a current flows from base to emitter.

Let us look at this current more closely, for it is the b/e junction which is the important one. In Fig. 5.4 we see that the current consists of:

- (1) electrons flowing out of the base, creating . . .
- (2) holes which flow to the b/e junction, where they are filled by . . .
- (3) electrons arriving from the external circuit and flowing through the emitter to the b/e junction.

Now we can see the reason for the calling them bipolar transistors.

The second important point is that the base layer is exceedingly thin (much thinner than shown in the diagrams. As electrons flow through the emitter to the base they come very close to the collector.

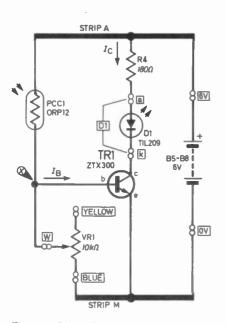
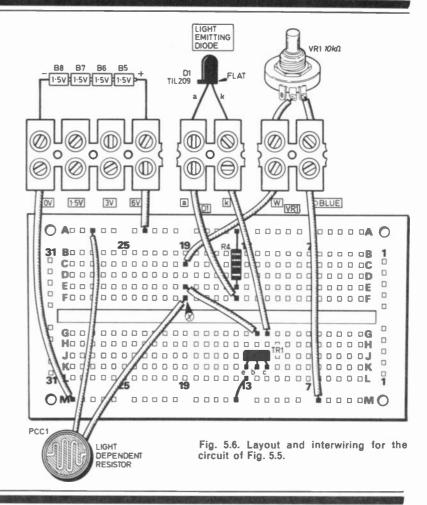


Fig. 5.5. Circuit for demonstrating a light operated transistor switch.



The potential of the collector region is more strongly positive than that of the base. It attracts electrons so strongly that they are pulled across the depletion region at the c/b junction, in spite of the fact that it is reverse-biased.

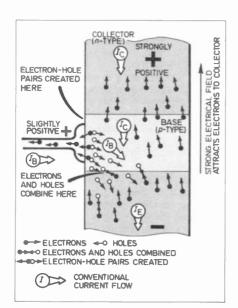


Fig. 5.4. The happenings in the base region of a transistor.

Only about 1 per cent of the electrons are used to combine with holes in the base. The remaining 99 per cent pass into the collector and then flow out of the transistor. In other words, the collector current $(I_{\rm c})$ is around 99 times greater than the base current $(I_{\rm n})$

 $I_{\rm C}$ depends on $I_{\rm B}$ bringing electrons to the base layer, so $I_{\rm C}$ flows only when $I_{\rm B}$ is flowing. By turning the small $I_{\rm B}$ on or off we can control the much larger $I_{\rm C}$. The transistor can be used as a switch.

Secondly, by varying the amount of the small I_B we can vary the amount of the large I_C . The transistor can be used as an amplifier.

A pnp transistor works in a similar way except that polarities are reversed.

EXPERIMENT 5.1 A transistor switch

In Fig. 5.5, PCC1 is a light-dependent resistor (l.d.r.). This is a piece of semiconducting material such as cadmium sulphide. When light falls on it, electrons in the atoms become excited and some escape from their orbits. They become charge carriers, allowing current to flow more easily through the material. The resistance of PCC1 decreases when the amount of light falling on it is increased.

Here, PCC1 and VR1 act as a potential divider. In the dark, the resistance of PCC1 is high, so the potential at X (and at the base of TR1) is very low. In the light, the resistance of PCC1 is reduced so the potential at X rises. We can adjust VR1 so that the potential at X is just less than 0.6V. This is just not enough to switch TR1 on.

Set up the components on the Verobloc as shown in Fig. 5.6. Make sure that PCC1 is not shaded. Turn VR1 until D1 goes out; then turn VR1 slowly back until D1 just comes on again. The potential at X is now a little over 0.6V. Place your hand or a thick cloth over the l.d.r. What happens to D1? TR1 switches D1 on or off according to the potential applied to its base.

EXPERIMENT 5.2 A current amplifier

This experiment illustrates the action of the transistor as a current amplifier, see the circuit in Fig. 5.7. $I_{\rm B}$ flows through the high-value resistor, R4. We use a meter to measure the corresponding $I_{\rm C}$.

The meter belonging to *Minilab* has a f.s.d. of 100μ A. To let it measure larger currents we put a resistor in parallel with

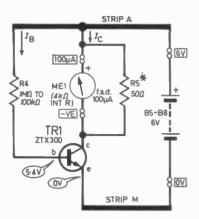


Fig. 5.7. The circuit for investigating the action of a current amplifier.

Fig. 5.8. Layout and interwiring details for the circuit in Fig. 5.7.

B7 B6 ME 1-50 1-50 1-50 0 0 0 1-5V 3V 6V -VE A00000 \Box 0 0 00 000 000000 R5 TWO 10002
R5 TWO 10002
RESISTORS IN PARALLEL R5* 7 O OF 0 • 0 0 0 0 0 0 0 0 0 0 0 F FDD000000000000

it (R5 in Fig. 5.7). The majority of the current by-passes the meter. If the resistance of the coil of the meter is $4k\Omega$ (as in the recommended meter), the p.d. across the coil is $(4 \times 10\Omega^3) \times (100 \times 10^{-6} \ \mu\text{A}) = 0.4V$, when the current is $100\mu\text{A}$.

When a 50 Ω resistor has a p.d. of 0.4V put across it, the current through it is $0.4V/50\Omega = 8 \times 10^{-3}A = 8000\mu A$. So the current flowing through both meter and resistor is $8000 + 100 = 8100\mu A$. (To round it down to $8000\mu A$ is near enough.)

This means that a reading of $100\mu A$ on the meter indicates a total current of $8000\mu A$. To measure the total current, take the scale reading and multiply it by 80.

For those readers having a meter other than the $4k\Omega$ type, a suitable value needs to be derived for R5.

To give approximately the same multiplication factor as above, divide the internal resistance of your meter by 80. Choose the nearest preferred value to this. Next divide the meter internal resistance by the preferred value to give the required factor.

For example, if the internal resistance is 1100 ohms, dividing by 80 gives 13.75. Nearest preferred value is 15. Divide 1100 by 15 to give the factor which here is 73.3. Thus all meter readings must be multiplied by 73.3 to give I_G .

In the experimental layout shown in Fig. 5.8 we use two 100Ω resistors in parallel to give the equivalent of 50Ω . Start with a $1M\Omega$ resistor for R4.

The base of TR1 is at 0.6V (= forward voltage drop), so there is a p.d. of 5.4V across R4. We can calculate I_B as $5.4V/(1 \times 10^4 \Omega) = 5.4 \mu A$. What is the collector current? Maybe it is too small to read.

Try various resistor values for R4, calculate I_B and measure I_C . To help you, Table 5.1 has two sets of figures filled in, though yours may be different depending on the transistor used.

Table 5.1: Results of Experiment 5.2

R4 (ohms)	/ ₈ /(ΩA)	/c/(ΩA)	/ _C // _E
1M			
470k			
270k			
220k 150k	25	3500	140
100k	54	7700	143

We work out the figures in the last column to tell us how many times $I_{\rm C}$ is greater than $I_{\rm B}$. This is called the d.c. current gain of the transistor, symbol $h_{\rm FE}$.

d.c. current gain,
$$h_{
m FE}=rac{I_{
m C}}{I_{
m R}}$$

The d.c. current gain of the author's transistor was about 140. What is the gain of yours?

Gain can be measured in another way which is slightly different. As I_B is increased from 25 μ A to 54 μ A, I_C increases from 3500 μ A to 7700 μ A. Increasing I_B by 29 μ A brings about an increase of 4200 μ A in I_C . We calculate:

Small signal current gain, $h_{te} = \frac{\text{change in } I_{c}}{\text{change in } I_{B}}$

For the author's transistor, $h_{te} = 4200/29 = 145$. What was h_{te} for your transistor.

Transistors of the same type may vary in gain. Transistors of different types vary even more widely (from 15 to 1000 or more), depending on their design.

EXPERIMENT 5.3 A simple audio amplifier

In Fig. 5.9, $I_{\rm C}$ flows through R5. As $I_{\rm C}$ varies, the p.d. across R5 varies in proportion. Variations in $I_{\rm B}$ result in large variation of voltage at point X.

A steady base current of $20\mu A$ flows through R4. This is the bias current. If h_{FE} is 100, I_C is $2000\mu A$. The potential across R5 is $1500\Omega \times 2000\mu A = 3000000\mu V = 3V$. The potential at X is exactly midway between 0V and 6V, allowing it to swing by equal amounts in either direction.

The crystal microphone generates varying small voltages when energised by sound. This causes a varying potential to appear at capacitor C2 (more about capacitors next month).

A small additional base current flows to or from C2. It varies in size and direction according to the waveform of sound received by the microphone. It adds to or subtracts varying amounts from the bias current.

The resulting I_B models the waveform of the original sound. Consequently, the changes of voltage at X also model the original sound, but the changes are large enough to power the loudspeaker, making it emit a sound loud enough to be heard.

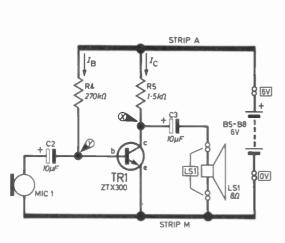


Fig. 5.9. A simple audio amplifier.

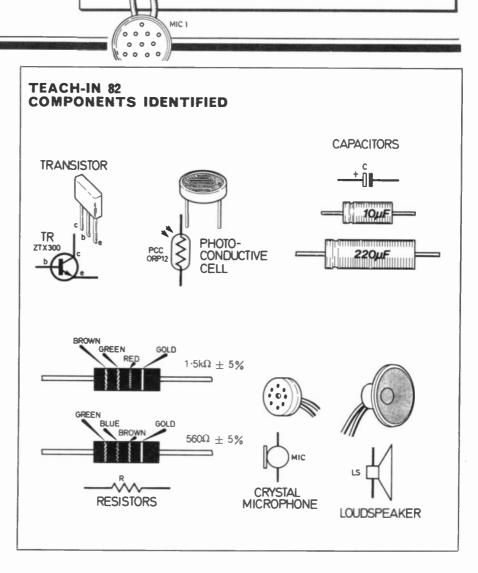
Fig. 5.10. Layout and interwiring for the circuit in Fig. 5.9.

Set up the circuit on the Verobloc as in Fig. 5.10, but do not put in the capacitors yet. Use the meter to measure the voltage at X (location L15). If necessary change R4 for a resistor of lower or higher value to make the voltage as close as possible to 3V, for if the voltage is too high or too low, it will not be able to swing equally in either direction and the resulting waveform will be distorted.

Now insert the capacitors to couple the microphone, MIC1 and loudspeaker, LS1 to the amplifier. Put your ear close to LS1 and tap the microphone with a pencil. You should be able to hear the sound distinctly.

To get a better effect, you can use a pair of long wires (ordinary lighting flex will do) and place the microphone in another room. Ask a friend to speak into the microphone while you listen at the loudspeaker. The amount of sound energy actually striking the microphone is not large so, even with a gain of 100, we would not expect a large output. A more powerful amplifier will be described later in the series.

The amplification of this circuit depends entirely on the value of h_{fe}, which varies a lot from transistor to transistor. If h_{te} is very high, I_C will be high, making the voltage at X swing far too close to 0V, causing great distortion. A way around this is to connect the bias resistor between point X and point Y. Since X should normally be at 3V, we reduce R4 to half its normal value. Alter your circuit to bias the amplifier in this way. (Remove the $270k\Omega$ resistor and place $150k\Omega$ between X and Y G_{16} to L_{15} .)



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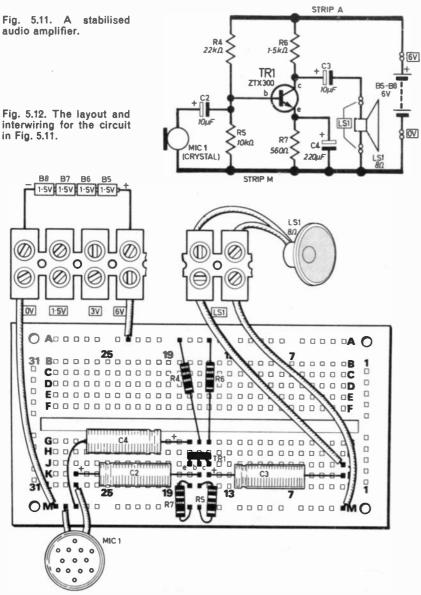
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EXPERIMENT 5.4



EXPERIMENT 5.4 A stabilised amplifier

Fig. 5.11 shows another way of biasing the transistor. R4 and R5 form a potential divider, providing a fixed potential of 1.875V at the base of TR1. The emitter must therefore be at (1.875-0.6)V=1.275V. The emitter current $(I_{\rm E})$ through R7 must therefore be $1.275V/560\Omega=0.0023A=2300\mu{\rm A}$.

Since $I_{\rm B}$ is so small compared with $I_{\rm C}$ we can say that $I_{\rm C}=I_{\rm B}=2300\mu{\rm A}$. This makes the voltage drop across R5 equal to 3·45V, which is much the same as in the previous amplifier. We could make it 3V by varying the ratio between R4 and R5.

Note that we have calculated $I_{\rm C}$ without using $h_{\rm FE}$. Gain does not depend on $h_{\rm FE}$ in this circuit. It depends only on R4, R5 and R7. Amplification is independent of

variations between transistors, or the effects of changes of temperature which also affect h_{FE} considerably.

Build this amplifier up on the Verobloc as shown in Fig. 5.12. Measure the voltages at the three terminals of TR1.

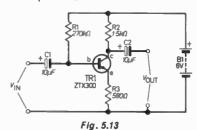
This amplifier has a capacitor (C4) which was not present in the previous one. Remove C4; does the amplifier work now?

Obviously C4 is very important. It is known as the by-pass capacitor. As the voltage from the microphone rises, base potential rises and $I_{\rm B}$ is increased. Increasing $I_{\rm B}$ means increasing $I_{\rm C}$ and $I_{\rm B}$. If $I_{\rm E}$ increases, this increases the p.d. across R7. The emitter potential rises.

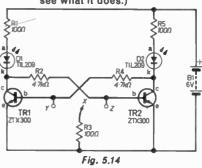
If a rise in base potential causes a rise in emitter potential, the two rises tend to cancel out. There is less increase in I_B than might be expected. The change in I_C is correspondingly reduced, so little sound comes from the speaker.

QUESTION TIME

- 5.1. What do we mean by a bipolar transistor?
- 5.2. What do we mean by a junction transistor?
- 5.3. How would the circuit of Fig. 5.5 act if you interchanged PCC1 with VR1?
- 5.4. If a transistor is receiving base current of 30 μA and its d.c. current gain is 150, what is its collector current?
- 5.5. h_{FE} and h_{fe} measure different things. What is the essential difference between them?
- 5.6. Fig. 5.13 shows a single stage amplifier. If the base potential is increased by 0·1V, what is the change in emitter potential?



- 5.7. Can you think how the circuit above could be useful? (Hint: remember f.e.t.s?)
- 5.8. A transistor in the circuit of Fig. 5.7 has $h_{\rm fe}=50$. When $l_{\rm B}=20\mu{\rm A}$, $l_{\rm C}=1{\rm mA}$. If $l_{\rm B}$ is increased to $24\mu{\rm A}$, what does $l_{\rm C}$ become?
- 5.9. Fig. 5.14 shows a circuit using transistors as switches. Wire X can be touched to points Y or Z. Can you work out what happens if X is touched to Y, to Z, to Y, to Z, and so on, alternately? (Hint: you could try setting it out on Minilab to see what it does.)



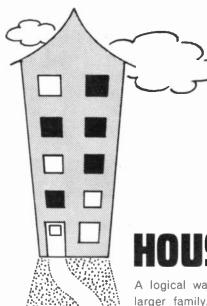
5.10. What is the circuit in Fig. 5.13 called? (Hint: see Part 4.)

C4 prevents rapid changes of potential at the emitter. As emitter potential tries to rise or fall, a large amount of current is required to alter the charge on the capacitor, so the change of potential is much less than it might have been without the capacitor. To put it another way, C4 holds the emitter potential steady. The alternating potentials due to the audio signals are filtered out, or "by-passed" to the 0V line. You could try the effect of using capacitors of other values for C4.

To be continued

FIREGERIPBOARD IN NEXT MONTHS ISSUE

AND FULL DETAILS OF TWO USEFUL INSTRUMENTS TAILOR-MADE FOR THIS BOARD



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MEDIUM

WITH PRESETTUNING **FOR THREE STATIONS**

BY F.G. RAYER

This unit provides pre-set tuning for three stations, plus manual tuning of the MW band, and is for personal headphone reception when used alone, or will feed an audio amplifier for loudspeaker listening. A simplified superhet circuit is employed, and this gives greater selectivity and sensitivity than t.r.f. type circuits. Illuminated l.e.d.s indicate selection of preset or manual tuning.

TUNING ARRANGEMENTS

Signals are picked up by the ferrite rod aerial. This has a coil L1 which is tuned by the selected capacitor.

Switch S1b selects the required means of tuning the ferrite rod winding L1. With manual tuning, C1 is in use. Position 2 of S1b selects the combination C3 and C4. Position 3 selects C5 and C6, while position 4 brings into use C7 and C8.

Mounted over L1 is a smaller coil L2. This is the base coupling winding for TR1, the mixer-oscillator.

L3 is the oscillator coil. Associated with it are two smaller coils which couple L3 to the emitter and collector circuits of TR1, respectively.

Switch S1c selects the required tuning capacitors for L3 as follows: manual tuning: C12 (ganged with C1); Position 2: C14, C15; Position 3: C16, C17; Position 4: C18, C19. Fine tuning or circuit alignment is provided by the trimmer capacitors connected across each main tuning capacitor.

The capacitor values used provide for normal MW coverage with manual tuning, and three pre-set stations: Radio 2 693kHz; Radio 3 1215kHz; and Radio 4 200kHz programmes.

If C1, C12 is left tuned to Radio 1, this allows instant selection of Radio 1, 2, 3 or 4 at the turn of the switch



S1. An l.e.d. above the tuning control indicates manual tuning, with red, yellow and green l.e.d.s for 2, 3 and 4. These l.e.d.s are selected by Sla.

Other frequencies can be provided for as described later.

I.F. AMPLIFIER

The intermediate frequency (i.f.) amplifier uses the double-tuned transformer IFT1, and integrated circuit IC1. The latter incorporates amplifier, automatic volume control and detec-

COMPONENTS

Resistors

1/1	0000	
R2	18kΩ	See
R3	15kΩ	
R4	2.7kΩ	Sno
K4	5. \KT1	211

 $100k\Omega$ R6 470Ω R7 680Ω

R8 1.8kΩ R9 1-8MΩ R10 5-6kΩ

All ‡W carbon ±5%

Potentiometers

VR1 22k Ω log. with switch (S2) VR2 1kΩ miniature vertical pre-

page 95

Capacitors

208pF air spaced variable Č1 C12 176pF air spaced variable

30pF trimmer

30pF trimmer

Jackson 00 gang., 208/176 slow motion with trimmers

160pF silvered micro

C3 C4 C5 C6 C7 C8 60pF trimmer*

22pF silvered micro

2nF silvered micro

60pF trimmer*

C9 10µF ceramic or polyester C10 20µF ceramic or polyester

180pF polystyrene

47pF polystyrene 60pF trimmer* 60pF trimmer* C14

C15

C16

C17 12pF polystyrene C18 220pF polystyrene

60pF trimmer*

10µF ceramic or polyester

C21 0·1μF polyester

0·47μF polyester 0·47μF polyester C23

10µF 10V elect. C24

1µF 10V elect. C25 C26 100µF 10 V elect.

* compression type; 30pF, 40pF,

or 50pF (max.) will suit.

Inductors

L1, L2 5in medium wave ferrite

rod (Denco MW5 FR)

M.W. oscillator coil

(Denco TOC1)

i.f. transformer (Denco IFT18/465)

Semiconductors

BF195 silicon n.p.n.

TR2 BC147 silicon n.p.n.

ZN414 i.f. amplifier and

detector

D1-4 l.e.d. 5mm dia. (4 off-red, green, yellow)

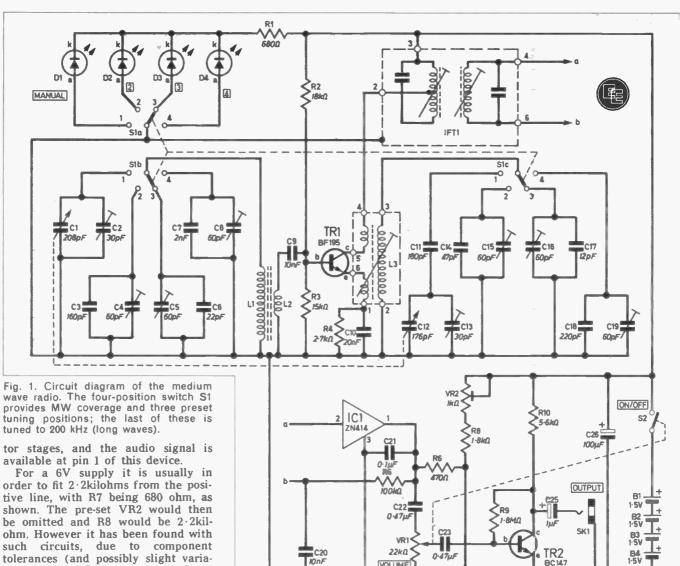
Miscellaneous

B1-4 1.5V cell HP7 (4 off)

3 pole 4 way rotary switch single pole on/off switch (part of VR1) S1 S₂

jack socket, 3.5mm Verobon 75-3009D or similar case, aluminium 5 x Perforated s.r.p.b. 13 × 26 holes 0-15in matrix. 3 knobs. Battery holder (for 4 × HP7) and

connector.



shown. The pre-set VR2 would then be omitted and R8 would be 2.2kilohm. However it has been found with such circuits, due to component tolerances (and possibly slight variations in IC1), that a means of adjust-

ment is often worthwhile. For this reason R8 is reduced in value, and VR2 added. Though this i.f. amplifier needs few

discrete components, it provides very

A.F. SECTION

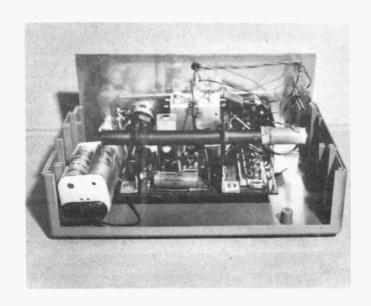
good results.

Audio signals are fed via C22 to the audio gain control VR1, and so to the single transistor amplifier TR2. The collector of TR2 is coupled via C25 to an output socket SK1. This stage provides more than adequate volume for headphones. Alternatively the output may be fed to an external audio amplifier via a screened lead.

POWER SUPPLY

Total current drain of the receiver is about 7mA from a 6V supply. This is readily provided by four HP7 type cells series connected in a holder. These will have a very long life.

If the receiver is always going to be used as a tuner, the supply may if wished be drawn from the amplifier. It is then necessary to connect a



0-47µF

≷R7 680Ω

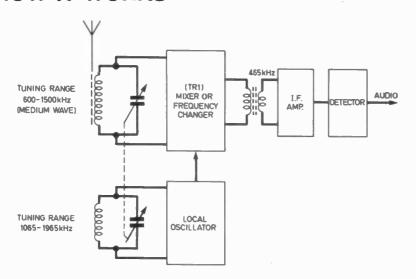
C24 ΙθμΕ

VOLUME

6.1V Zener diode across C26 and to place a series resistor in the positive supply line. With a 400mW Zener, this resistor may be 150 ohm for 9V, 270 ohm for 12V, and 390 ohm for 15V to 16V. A negative power connection must also be provided of course.



HOW IT WORKS



The principle of superhet reception is explained here in simple terms. Reference should be made to the block diagram above, and also to the main circuit diagram given in Fig. 1.

Transistor TR1 is a self-oscillating mixer, or frequency changer. Its purpose is to change the carrier frequency of any station selected by the aerial circuit, to a fixed frequency. The latter frequency can then pass to the section of the receiver having fixed tuning, which is called the intermediate frequency (i.f.) amplifier.

When two different frequencies are present at a mixer or frequency changer, other frequencies are produced. Here, one frequency is that of the wanted signal, obtained from the ferrite aerial L1. The other frequency is produced by the oscillator coil L3, coupled to TR1 emitter and collector by windings provided for this purpose. In the present receiver, the important new frequency obtained from TR1 collector corresponds to the difference in frequency between aerial and oscillator. Examples will make this clear.

Suppose the aerial circuit is tuned to 200kHz, and the oscillator circuit to 665kHz. The difference is 465kHz. So output at 465kHz is obtained from TR1 collector. If now the aerial circuit were tuned to, say, 1,000kHz, and the oscillator circuit tuned to 1,465kHz, the difference is still 465kHz, and the output from TR1 is still at 465kHz. In the same way, the aerial circuit can be tuned to any wanted frequency, and provided the oscillator is always tuned 465kHz higher, the output of TR1 will always be at 465kHz.

With manual tuning, using a two-gang variable capacitor, if the aerial circuit tunes 1,500kHz to 600kHz, then the oscillator circuit needs to tune 1,965kHz to 1,065kHz. That is, 465kHz higher, through the band. This is arranged as accurately as possible by a suitable choice of inductance and capacitor values.

So when all signals tuned in are changed to 465kHz, they can pass to the intermediate frequency amplifier. This does not need to have variable tuning, as it is permanently adjusted to operate at 465kHz. This allows additional tuned circuits, for better selectivity, without the difficulties which would be present if several variably tuned circuits were to be used instead.

All stations received are amplified at 465kHz, and the detector or demodulator follows, to make available the audio signal or programme.

HONSTRUCTION starts here

CIRCUIT BOARD

Components are assembled on a piece of perforated s.r.p.b. On the prototype this has 13×26 holes, $0 \cdot 15$ in matrix, see Fig. 2. Make sure the board is large enough for the trimmers and other parts actually used.

Drill holes for L3 and IFT1. Note that pins 1-2 and 5-6 are slightly closer to each other. This arrangement identifies the connections. Also drill holes for the can tags, and under the middle of IFT1, to give access to the bottom core.

Fig. 3 shows point to point wiring under the board. Two tags MC are fitted with ¹2in 6BA bolts at the corners of the board. These bolts eventually secure the board to the flange of the panel bracket.

Earth the metal cans, top plate tags of trimmers, and other components as shown, to a wire run between the tags. In most places the wire ends of components will be long enough to reach the various points. Elsewhere, use thin wire, with sleeving where needed.

The ferrite rod fits about 1½in above the board. It passes through holes in Paxolin (s.r.b.p.) strips, which are bolted to small angle brackets, see Fig. 2.

Position VR2 so that it can be adjusted from the back.

BEHIND THE PANEL

The front panel is metal, $6^{3}_{4} \times 3^{1}_{8}$ in (172×80mm) for the case used.

A bracket has to be made from a piece of aluminium about 5×3^3 4in. This is bent at 1^3 4in along one long edge, to take VR1, C1, C12 and S1, as in Fig. 4. The bush nuts and screws for C1, C12 hold the front panel and bracket together, and the circuit board is mounted on the horizontal flange, as previously described, clear of C1, C12.

The remaining wiring can then be carried out. Take a lead from Cl, Cl2 frame tag down to the board MC line wire. This lead supports Cl4, Cl7 and Cl8, see Fig. 4. The aerial section of the switch SlA can be wired, and then the oscillator section SlB, followed by the l.e.d. section SlC. Cl is the variable section nearer the panel, with most plates.

Approximate OST £25

Fig. 2. Top view of the circuit board and the ferrite rod aerial.

900

1400

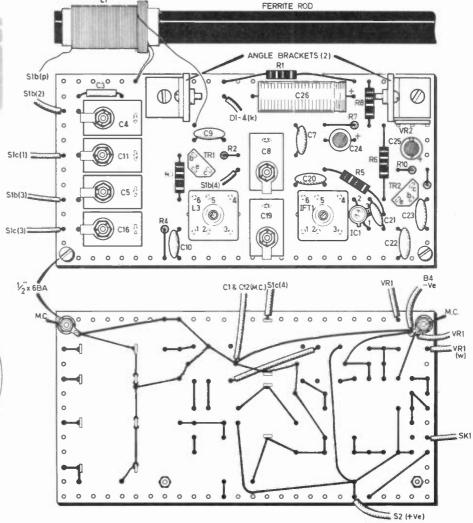


Fig. 5. The calibrated dial of the prototype receiver. Slight differences in the position of these frequency points are to be expected with individual receivers.

600

Fig. 3. Underside of the circuit board. Bare wire may be used for the links, except where wires cross, then insulated wire must be used as illustrated.

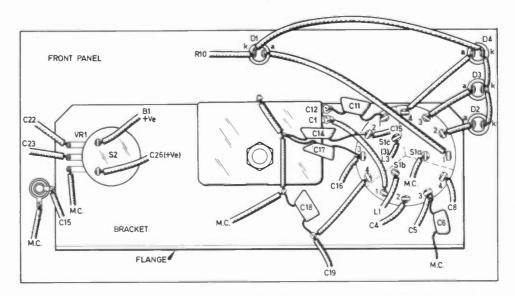
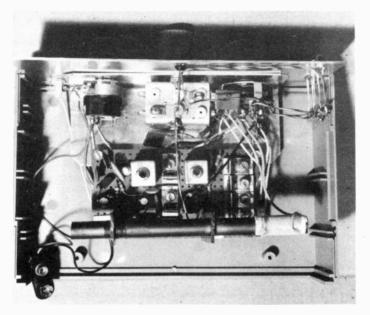


Fig. 4. Rear view of the front panel. The circuit board is secured to the flange of the bracket by two 6BA bolts (these are shown in Fig. 2 and Fig. 3).



The specified ganged tuning capacitor is provided with trimmers. These are designated C2 and C13. If a ganged tuning capacitor without trimmers is used, separate trimmers should be fitted.

Connections to VR1 and the output socket are made as shown. Check the socket is not wired so that output is shorted.

VR2 ADJUSTMENT

Initially, VR2 can be set so that a high resistance meter shows about 1.3V at the side of positive C24. Subsequently, set VR2 for best results. Too low a voltage will reduce sensitivity, while too high a voltage will cause whistles or oscillation. The setting is not very critical.

LETTERS

Something for Everyone

Looking back in previous issues of EVERYDAY ELECTRONICS, I happened to read Letters on page 130 of February 1981 issue.

The first two letters were from fellow Aussies. I am ashamed for them and hope you "Poms" don't think we are all as helpless as those two. They will be wanting printed circuit boards for crystal sets next.

The trouble is they are "spoon fed" over here by some Australian electronic journals and, of course, the trade, both of which seems to cater for the lowest common denominator in intellect. I must assure you that there are many over here who can construct without p.c. boards, matrix boards or Veroboard. All of which are pretty useless at h.f. or v.h.f. anyway.

Also interesting was the letter from F. W. Blakeley of Wrexham. I am not quite as old as your Mr. Blakeley, being a mere

I.F. ALIGNMENT

IFT1 cores are pre-aligned, and should not be touched until a station has been correctly tuned in. Choose a weak signal, correctly tuned in, or rotate the unit for minimum signal pick-up, and adjust top and bottom cores of IFT1, very slightly, for maximum volume.

MANUAL TUNING

Initially position the coil L1 about level with the end of the rod, with C2 and C13 about half closed. Tune in a weak signal correctly with the ganged capacitor C1, C12 nearly open, and adjust trimmers C2 and C13 for best volume.

73 years of age, and my experience only goes back, in radio, to 1923. I am probably a "Johnny come lately" to him.

I consider EVERYDAY ELECTRONICS an

I consider EVERYDAY ELECTRONICS an excellent paper for the electronic enthusiast, and although it does dwell a little too much on "gadgets" sometimes; on the whole there seems to be something for everyone. Your "Home Sentinel Unit" in Volume 1 No. 1 was constructed and worked first time. Happily however, apart from testing when first installed, it was never called upon to deter burglars, we did not have any of these light fingered gentlemen around at the time.

On my retirement from the "hassle" of

On my retirement from the "hassle" of a working life some years ago, the business, and of course all the instruments, was sold, however as is usually the case, a need was found for a few items of test

The first item to be constructed was Mr. Rayner's "Signal Generator" as featured in your September '78 issue. As usual, with most of your projects, it worked first time off.

However, due to small wiring differences the calibrated scale was out, and the only way to correct this fault was by fitting trimmers to the coils. Adjusting the coil slugs at the l.f. ends of each band, and the trimmers at the h.f. ends, brought the

Now tune in a weak signal with C1, C12 nearly closed, and modify the position of L1 on the rod, for best volume.

Repeat these adjustments a few times, to obtain full sensitivity over the whole manual tuning range.

If it is necessary to change frequency coverage by altering the core of the oscillator coil L3, rotate this only slightly. Use a properly shaped tool here and for IFT1, to avoid breaking the cores.

PRE-SET TUNING

After correct operation and alignment for manual tuning, the pre-set tuning adjustments can be made. Proceed as follows: adjust C4 for aerial tuning and C15 for oscillator tuning, at switch position 2, to receive Radio 2. At position 3, adjust C5 and C16 for Radio 3. At position 4, set C8 and C19 for 200kHz long wave, Radio 4.

In some parts of the country LW is not required for Radio 4. The large values at C7 and C18 are then unnecessary, and values up to about 200pF may be used in the aerial position, and up to 100pF in the oscillator circuit a further MW station

tor circuit, a further MW station. With the capacitor values given, coverage is around the frequencies listed earlier. There is considerable latitude in both trimmer values and fixed capacitor values. Small values move tuning towards the high frequency end of the band, and larger values allow tuning towards lower frequencies.

For personal listening, medium or high impedance phones (say 500 ohm to 2 kilohm) will allow excellent results.

readings somewhere near. A crystal calibrator was used, together with an allband radio for setting up. A switch was also fitted to cut out the a.f. modulator when not needed.

Many "old timers" seem to complain about over sophistication of some articles which are published in electronic journals. They are right in this opinion in some respects, as many modern projects are updated versions of previously featured projects. They don't work any better, have more parts which can develop faults, and, their only advantage seems to be in reduced physical dimensions.

A typical example being your "Continuity Tester" featured in your June '78 issue. A better continuity tester was constructed 9 years ago, the components being two surplus transistors and one $6.3\,V$, 0.04 amp flashlight bulb. There are 3 ranges and useful indications can be obtained up to $2M\Omega$. Four of these were installed at a local transformer factory and have been in constant use for years.

Presumably careful selection of the transistors would result in a performance equal to the unit described in the June 78 issue.

J. Ratcliffe, Australia.



This month Barry Fox continues his saga of the "commercialisation" of the Post Office and calls for an amnesty for illicit telephones.

The Beginning

For the Post Office the end of an era, and for the British public the beginning of an era, came on September 12, 1979 when Sir Keith Joseph, Secretary of State for Industry, announced the government's intention of splitting the Post Office into separate sections for posts and telecommunications. He promised consultations with a view to early relaxation of the Post Office's telecommunications monopoly"

On July 21, 1980 Sir Keith Joseph carried through his promise and told Parliament that legislation was being enacted to let private firms to compete with the Post Office Telecommunication Division (to be called British Telecom) for the supply of telephone equipment. After a transitional period of about three years the Post Office, he said, would retain a monopoly only on the main telephone network and the installation and maintenance of the first telephone in each home and business. "I am sure that one of the reasons for America's greater success has been the freedom available there to entrepreneurs to develop new services and a wide range of equipment associated with tele-com-munications" Sir Keith told Parliament.

This was the water-shed speech. But even by then the Post Office had been forced to liberalise its attitudes on A certification answering machines. scheme was inaugurated on April 2, 1980. Any person wanting to sell an answering machine on the British market has since that date been able to submit it to the Post Office for official approval. In what some would call a face saving operation the Post Office claimed that this liberalisation followed "in the light of advances in the functioning of the public switched tele-phone network itself and with improved manufacturing standards"!

Announcement of the answering machine certification scheme was one of the last statements made by the Post Office, because British Telecom was born on

May 21, 1980.

The Telecommunications Bill is likely to become law any moment now, a year after it was first introduced in Parliament (November 1980). When this happens

British Telecom will hand over responsibility for certifying all telephone equipment (including the answering machines which it currently certifies) to a neutral third party, almost certainly The British Standards Institution. In the meantime it's rewarding to catalogue some of the statements made by BT since its May 1980 birth.

On the very same day as the name change was announced, Peter Benton, then Managing Director, said that the new name was being adopted to "reinforce staff pride and make clear to our customers that this is a rejuvenated business". This followed a promise made by Mr Benton of what three weeks earlier was still the Post Office that "we are attacking our problems at the roots; we intend to improve the service permanently".

Accepting the Challenge

On July 21, the day Sir Keith Joseph committed the British government to the break-up of the Post Office, Sir William Barlow, Chairman of the PO proclaimed "We are not afraid of the new challenge". Nevertheless on July 31. British Telecom announced they would have to put up the prices of local calls.

This is one area in which we are certainly not being brought in line with the North American Continent. Over there local calls are either free or very cheap. Most long distance calls are also far cheaper than in Britain. Even stripped of their monopoly, the Bell company are still able to offer far cheaper call rates than the British Post Office.

In November 1980 publication of the Telecommunications Bill caused the Post Office to say it "accepted the challenge" offree competition. But within just six days there was the tell tale sign of an escape route being prepared. "Neglect of investment and determination in previous years will inevitably leave us with difficulties in some places for a while yet" said Mr. George Jefferson, then Chairman of British Telecom.

In February 1981 British Telecom was forced, by an article in *The Daily Mail*, to acknowledge that Telecom engineers had been taking cash back-handers to install business telephones ahead of others in the queue. BT asked for information "which could help us stamp out these activities". There was, of course, one very simple answer; speed up legitimate installation to eliminate the queuel

Time Wasting

In April 1981 British Telecom finally acknowledged what everyone who made international calls already knew—it is absurdly time-wasting for anyone making an international directory enquiry to have to go through the international operator. So separate directory service was promised for May 1982.

In May 1981 British Telecom again pledged "to improve service" and claimed that the waiting list for telephones had been halved by speedier connection. In June 1981 the Chairman Sir George Jefferson again acknowledged past errors. "We must be more market responsive than we have been in the past" he said. A few days later he announced the creation of a sales force to do something wholly novel for the British Post Office-sell telephones. "Until now we have sometimes been hard to find" he admitted.

The signs for the future are good. This vear British Telecom will extend its inhouse certification scheme from answering machines to other attachments such facsimile machines and intruder alarms. Within the three years promised by Sir Keith Joseph, the scheme moves on to privately supplied telephones. The only tragedy is that it has taken government pressure to convert our complacent Post Office into something approaching a

business enterprise.

Call for Amnesty

One problem remains. This is the plethora of illicit telephones already In use, many of them incorrectly connected by subscribers who have had no where

to go for engineering advice.

British Telecom is now sending out a leaflet with all quarterly telephone bills which warn of the dangers to the network, to the service and to individuals. This leaflet repeats the previous threat that British Telecom "will disconnect the telephone service of customers who breach the conditions under which telephone service is provided".

So anyone with an illicit telephone will still not dare ask for help from a British Telecom engineer. When a fault develops on the line they will have to rip out all their illicit phones before calling for service. As often as not this will waste the engineer's time because the fault will usually be on the equipment which has been disconnected.

By the time the certification scheme extends to privately supplied telephones, there could be a million unauthorised connections in Britain. Now, more than

ever before, we need an amnesty. "We must be market responsive" said

Sir George Jefferson when addressing the annual conference of the Post Office Engineering Union on June 2, 1981. Competition will provide many opportunities for us but we must adapt the way we work to make the most of these opportunities".

If Sir George is serious about adaptation, British Telecom must offer an amnesty of technical aid to anyone with an illicit phone connected to the British

NTRODUCTION 1

BY J. CROWTH PART 10

USE OF TRUTH TABLES

Truth Tables are the easiest method of conveying information, or the whole truth, about a gate, or Logic Circuit requirement. Suppose we were told to design a Logic Module which would give a "1" output if:

"A, B, and C were 0, or if C was 1, or if A was 1, or if A and C were 1 or if A and B were 1, or if A, B and C were 1.

This statement tends to be confusing and could be easily misunderstood. It would be less liable to be misunderstood if it were written in the form of a truth table, Table 10.1;

Table 101

	Inputs					
A	В	С	S			
0	0	0	1			
0	0	1	1			
0	1	0	0			
0	1	1	0			
1	0	0	1			
1	0	1	1			
1	1	0	1			
1	1	1	1			

Table 10.1 shows at a glance what combinations of the inputs are required to give an output at S. A Boolean Equation can be derived from the information given in the truth table, this equation can then be simplified and converted into gates as shown previously.

Note

When writing Boolean Equations from truth tables, where A=0 in the Truth Table, this is written as A in the equation, and if A=1 in the Truth Table it is written as A in the equation. example

 \overline{ABC} means A AND B AND C=0

 $\bar{A}\bar{B}C$ means A AND B=0 AND C=1.

Two methods can be used:

Deriving a logic module circuit from a truth table—method 1

Taking the condition when an output is present, that is when S=1. The Boolean Equation for Table 9.1 then becomes: $\overline{ABC} + \overline{ABC} + A\overline{BC} + A\overline{BC} + A\overline{BC} + ABC = S$

Simplifying we get:

$$\vec{A}\vec{B}(\vec{C}+C)+A\vec{B}(\vec{C}+C)+AB(\vec{C}+C)=S$$

$$\vec{A}\vec{B}1 + A\vec{B}1 + AB1 = S$$

$$\overrightarrow{AB} + \overrightarrow{AB} + \overrightarrow{AB} = S$$

$$\vec{A}\vec{B} + A(\vec{B} + B) = S$$

$$\vec{A}\vec{B} + A1 = S$$

$$\vec{A}\vec{B} + A = S$$

$$\overline{B} + A = S$$

This represents an OR gate fed with A and B, B being obtained by passing B through a NOT gate.

Therefore the final module is as seen in Fig. 10.1.

This shows that C is not needed to obtain the required result, and that the original requirement was unnecessarily complicated.

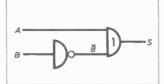


Fig. 10.1. Logic module to realise the truth table above. Shows that C is not required to obtain the required result.

Method 2

Taking the condition when there is no output, that is when S=0, or the output is \tilde{S} .

$$\vec{A}\vec{B}\vec{C} + \vec{A}\vec{B}\vec{C} = \vec{S}$$

$$\overline{A}B(\overline{C}+C)=\overline{S}$$

$$\bar{A}B1 = \bar{S}$$

$$\bar{A}B = \bar{S}$$

In order to make the equation so that the output is S, add a bar to both sides (see Rule 5 in Part 5). $\overline{AB} = \overline{S}$ $\overline{A} + \overline{B} = S$

$$\frac{AB=5}{7}$$

$$\overline{A} + \overline{B} = S$$

$$A + \overline{B} = S$$

This gives the same result as Method I as would be expected. In practice the easiest method would be used, that is the second method in this case, as it involves the least number of terms.

APPLICATION OF LOGIC TECHNIQUES (1) The Equivalent Module, or Exclusive NOR

Symbol

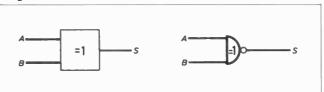


Fig. 10.2. Two common symbols for the EXCLUSIVE NOR gate.

Result

There is an output if A and B are the same.

Truth Table

Inpu	ts	Output
Α	В	S
0	0	1
0	1	0
1	0	0
1	1	1

Boolean Equation

Take the case when S = 1.

$$\overline{A}\overline{B} + AB = S.$$

This equation represents the logic circuit, shown in Fig. 10.3, containing five gates.

By applying Demorgan's Law we can reduce the number of gates to three:

$$\overline{AB} + AB = (\overline{A + B}) + AB = S.$$

This represents the logic circuit shown in Fig. 10.4.

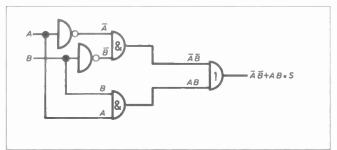


Fig. 10.3. The logic circuit for the above Boolean expression.

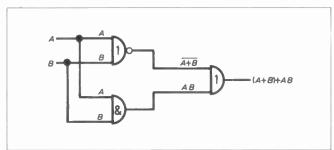


Fig. 10.4. Logic circuit of Fig. 10.3 after applying Demorgan's

(2) The NOT Equivalent Module, or Exclusive OR

Symbol

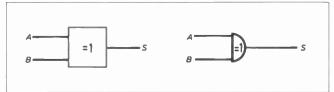


Fig. 10.5. Two common symbols for the OR gate.

Result

There is an output if A and B are not the same.

Truth Table

In	outs	Output
A	В	S
0	0	0
1	0	i
1	1	0

Boolean Equation

Taking the case where S = 0

$$\frac{\overline{A}\overline{B} + AB}{\overline{A}\overline{B} + AB} = \overline{S}$$

$$\overline{AB} + AB = S$$

This represents the logic circuit in Fig. 10.6 which contains five gates.

By applying Demorgan's theorem we can reduce the number

of gates to three:
$$\overline{A}\overline{B} + AB = \overline{A}\overline{B} (\overline{A}\overline{B}) = (\overline{A} + \overline{B}) \overline{AB} = (A + B) \overline{AB}$$

This represents the logic circuit in Fig. 10.7.

An alternative arrangement of logic gates to give the NOT Equivalent Module could be obtained as follows: Boolean Equation

$$\bar{A}\bar{B} + AB = \bar{S}$$

$$\overline{AB} + AB = S$$

$$\overline{(A+B)+AB}=S$$

This represents the logic circuit in Fig. 10.8.

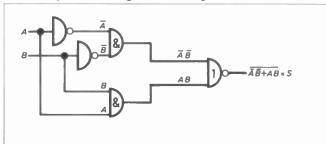


Fig. 10.6. The logic circuit for the above equation and truth

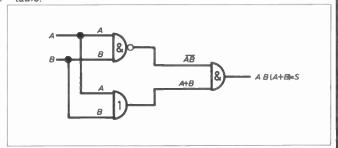


Fig. 10.7. The reduced logic circuit after applying Demorgan's

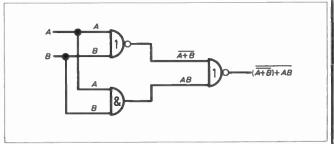


Fig. 10.8. Logic circuit for the above Boolean expression.

It can be seen that this circuit is the same as the Equivalent Module followed by a NOT gate, showing that the NOT Equivalent Module is the inverse of the Equivalent Module.

Because of this there is no need for manufacturers to make both exclusive or and exclusive nor modules. In practice manufacturers rarely made EXCLUSIVE NOR modules, but tend to make EXCLUSIVE OR and follow it by a NOT gate if an EXCLUSIVE NOR is required.

Everyday News

Electronic post - a Royal Mail Service

To underline its position as a world leader in postal communications, The Royal Mail has started "field" trials on a computer-based service which can send mass mailings by wire, to be printed and enveloped in a distant centre for delivery through the mail the next morning.

Anyone who has read some of those old postcards found in Antique shops will have admired the postal service of that era. It was quite common to see comments like "I will be around this afternoon at 4 o'clock."—Such was the confidence in the local delivery service.

The sole reason for mentioning the above is to highlight the latest technological development that brings the 80's system that much nearer the 20's and 30's "same

day" delivery.

Called Electronic Post, this latest advance for Britain's postal service is the first system of its kind to use a combination of computer transmission, laser printing and postal delivery to transmit, print and prepare items for delivery to many thousands of addresses. The service is potentially capable of handling any kind of multiple items, such as statements, bills and invoices.



A London operator keys in Customer's identification to be transmitted to the computer in Manchester. Meanwhile, magnetic tape containing the customer's mailing list is loaded into the computer prior to transmission



At London's Electronic Post Centre at Mount Pleasant, postmen oversee printing, folding and enveloping of letters transmitted from the Manchester computer Centre.

London and Manchester are the two centres chosen for the first trial service and will cover about seven million addresses. This represents about a third of all the 22 million delivery points in the UK.

Mr. Kenneth Baker, Minister of State for Industry and Information Technology, joined with Mr. Ron Dearing, Post Office Chairman, to watch the first transmission of a letter from a major commercial mailer being reproduced in bulk, enveloped and addressed ready for delivery the following morning to addresses in London and Manchester.

The System

The system was developed by the Post Office and GEC Mechanical Handling with help from GEC-Marconi Electronic.

A company produces letters, statements or invoices on a magnetic tape, complete with address and postcode. A computer processes the information and transmits it to the appropriate distant centre.

At the receiving end the items are printed by a laser printer. The letters are then automatically addressed, enveloped and delivered using the normal delivery service.

To counter violence and vandalism the whole fleet of London's 5,500 buses are to be equipped with two-way radio in the driver's cab.

CLASSIC

Covert Local Area Sensor Systems for Intruder Classification (CLASSIC) is a new intruder alarm system for front line use by the British Army. Detection is by seismic or other sensors radiolinked to a monitoring point.

Developed as a private venture by Racal-SES Ltd, CLASSIC is already the subject of negotiations with overseas customers as well as having been purchased by the British Ministry of Defence.

-COAL RADAR-

A novel application of radar is determination of the level of coal in underground vertical storage bunkers. The radar system only suitcase-sized will measure the level of coal within five per cent.

The system has been developed by Marconi Avionics in conjunction with the National Coal Board's Mining Research and Development Establishment.

UK Computers for USA

The United States Air Force and Navy have standardised on a British range of air data computers for re-equipping 27 variants of 10 types of aircraft from jet fighters to large military transports.

Marconi Avionics won the design and prototype contract, worth five million dollars, against fierce competition from US suppliers.

ATE Winner

The newest COMPACT automatic test equipment (ATE) from Marconi Avionics had clocked up over £1 million of orders before its first public demonstration just before Christmas.

It is just half the size of the original COMPACT system developed in 1973.

Government funding of £77 million is promised as the British share in the European Space Agency's Large Satellite due for launch in 1986.

Prime contractor is British Aerospace with Marconi as one of the British subcontractors.

... from the World of Electronics



_ANALYSIS__

NOTHING NEW

Technological advance means better, cheaper or easier ways of doing things, often all three. In the process it is inevitable that some people experience damage to their interests but the majority benefit. This has been true throughout history. It is still true today.

The invention of the power loom put the cottage weavers out of business. Transatlantic flight killed the passenger steamship. Mass-produced cars injured the classical craft of coach building. The home washing machine led to the disappearance of the steam laundry. Radio and TV shut down the music halls. These are just a few of hundreds of examples.

Man is distinguished in his power of reason and invention and every invention has led to change but never more so than in this age of the electronics revolution. In principle this is no different, say, than the innovation of steam power in the last century. What is different, however, is the speed of change. Whereas in the past we have more or less successfully adapted to change over a period of years, today we seem to have been overtaken by it.

Speaking only of the United Kingdom this is partly our own fault. We have been inventive enough but slow to exploit technological advances in general industry. We preferred old work practices and resisted the new.

An exception is the electronics industry itself, the cradle of invention and unafraid in exploiting every means of improving its own performance. The result is that leading companies like GEC-Marconi, Plessey, Ferranti and Racal have all done well in world markets throughout the recession and have full order books.

The industrial robot, the electronic office, computeraided-design, automatic draughting machines are all aids to efficiency and economy in the supply of goods and services. They may not always be welcome but to fail to embrace these and other innovations is to eventually put ourselves out of business altogether.

Brian G. Peck

X-Stream

Preliminary orders worth £750,000 a year in rental have already been received for advanced new digital communication services which British Telecom is to launch during the year, as part of its contribution to promoting information technology in the UK.

The new services, to be marketed under the general title "X-Stream", are made possible by the steady spread digital transmission through Britain's cable. microwave and optical fibre network.

Digital techniques are now being increasingly used instead of traditional analogue transmission as part of Telecom's £2 billion a year modernisation programme.

BBC Radio Birmingham is now to be known as BBC Radio WM, the local radio service for the West Midlands.

WORLD FIRST

Britain achieved a world first recently when the Duke of Kent inaugurated a fully computerised cargo system Harmondsworth Heathrow.

air cargo Called ACP80, processing in the 80s, it serves Heathrow and Gatwick airports and is claimed to be the world's most sophisticated airport system handling both exports and imports.

Robots Make Robots

The Department of Industry is reported as examining a £20 million project for a robot factory with robots building robots.

Participation by private industry is anticipated and the factory, if it comes to fruition, will be a working demonstration of the changing nature of production engineering as well as producing urgently needed robots.

BASICally Sinclair

Coinciding with ICL's announcement of major new developments employing Sinclair technology and Sinclair BASIC, Sinclair Research reports that worldwide sales of the ZX81 computer now exceeds 250,000.

With ZX81 monthly production running at 50,000 units, over 60 per cent for export, Sinclair claims to have more units installed than any other personal computer manufacturer worldwide, firmly establishing Sinclair BASIC as the most widely used microcomputer programming guage.

Commented Clive Sinclair, "With so many units in the field, supported by a large number of new projects using Sinclair BASIC, we believe our language now merits serious consideration by the industry as the 'standard', if confusion, particularly in the educational field, is to be avoided in the future."

ICL has announced a new range of ZX81 software with orders for more than 100,000 packages already received from W. H. Smith's, the only licensed ZX81 retailer, and is to develop further software in support of both the ZX81 and future products using Sinclair BASIC.

ICL is also in advanced discussion with Sinclair to develop an ultra low cost terminal/digital integrated phone work-station employing Sinclairs' flat tube technology and Sinclair BASIC.

Prestel Delivers the Goods

A new information service for freight forwarders and exporters, entitled "Mini Cargo Tariff", has heen launched by British Airways Cargo on Prestel, British Telecom's viewdata service. Radio Rentals Contracts are supplying approximately 150 colour viewdata terminals for the system.

British Airways have invested substantial sums in producing a 2,000 page database which provides freight forwarders with up-to-theminute details of rates, dates, loads, restrictions and class for the shipment of goods. The information is available to BA agents on a "Closed User Group" basis to ensure system security.

Agreement in principle has been reached between Gould Advance Ltd, and Ashcroft Electronics Ltd, for the acquisition of Gould Com-ponents Ltd by Ashcroft.

All-Digital Desk

The BBC is to order an all-digital 48-channel sounds mixing desk from Electronics International of Royston. A prototype desk has been installed at the BBC in Broadcasting House.

When the production version of the desk is delivered it is believed that it will become the world's first comprehensive all-digital sound mixing desk to enter operational service in broadcast-

-STICKY BREAKERS-

Watch out for illegal CB rigs carrying stickers claiming immunity from prosecution.

The Home Office have just issued this advice to the CB trade and prospective buyers.

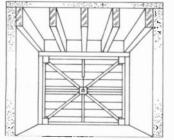
"Don't be misled by unfounded rumours claiming that the use of illicit 27MHz AM sets will be legalised; the Government has no intention of making any changes to the new legal 27MHz FM CB service".

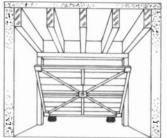
The warning to traders and CB users follows a large number of inquiries to the Home Office concerning rumours of AM legalisation, and reports of AM sets carrying labels stating that the apparatus cannot be used "until April 1982" or similar wording.

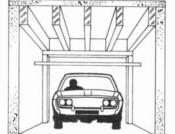
Any such stickers or labels which imply pending

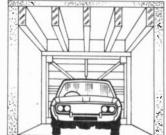
changes in the UK CB service are quite simply hoaxes.

AUTOMATIC









GARAGE

AST month having discussed the suitability of converting an upand-over type garage door and described the Ultrasonic Remote Control system (Transmitter and Receiver), we now move on to the Control Logic/Power Supply unit and installing the mechanics.

POWER SUPPLY

The Control Logic and Receiver units are powered from the same p.s.u., a rectified and smoothed 12V supply, built into the Control Logic case. For the circuit diagram see Fig. 7. The mains supply is connected via switch S1 and fuse F1 to the transformer, T1. The live line is supplied via switch S2 to the relay contacts RLA1 and RLB1 (shown in Fig. 8), the neutral line also being connected to TB2/5, then to the appropriate point on the motor. A neon indicator with integral resistor, LP1, is connected across the transformer primary.

TRANSFORMER

The transformer used in the prototype was a 6VA type, with two 4.5 volt secondary windings wired in series to provide 9 volts. The 600mA output was more than required and 300mA would be sufficient.

The nominal 9V a.c. output from the transformer is rectified by bridge rectifier D1 to D4. The voltage developed across smoothing capacitor C1 is about 12.7 volts, but in practice may be higher according to the regulation of the transformer. The exact working voltage of the circuit is unimportant, hence no regulating components are employed. Capacitor C2 provides high frequency de-coupling, and switches S3 and S4 form the safety cut-out elements, S3 being the microswitch at the base of the door and S4 the tension cut-out switch, discussed later.

CONTROL LOGIC CIRCUIT DESCRIPTION

The logic circuit is based on a cmos integrated circuit type 4081, a quad 2 input AND gate, see circuit diagram Fig. 8.

The input from the receiver circuit is connected via diode D5 to the input of gate ICla, capacitor C3 suppressing unwanted noise. As the voltage at the input of ICla rises, the output will go high, with feedback occurring via R3. Thus a Schmitt trigger is formed, the output from ICla rising very quickly to full potential. R2 will reduce the input to zero, when no signal from the Ultrasonic Receiver is present. ICla may also be triggered by push-button switches S5 and S6. These switches are the MANUAL DOOR OPEN/CLOSE controls.

The output from gate ICla is coupled by C4 to the next stage. This capacitor along with R4 forms an integrator to reduce the output to a single pulse, otherwise a confused state of re-triggering would occur in the next stage of the circuit.

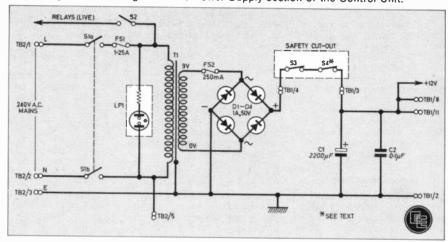


The finished electronic control units. The Power Supply is built into the Control Logic case.

OPEN DOOR

Assuming the garage door is at present closed, the normally closed contacts of S7 are now open, and S8 is closed. Hence pin 6 of gate IClb is at logic 1 (high) and pin 12 of gate ICld is low. Thus gate ICld cannot switch on, and we will not consider its associated circuit at this point.

Fig. 7. Circuit diagram of the Power Supply section of the Control Unit.



IClb produces a high output upon receiving the pulse from D6. Feedback occurs via R9, and gate IClb stays on. This will turn transistor TR1 on thus illuminating the SYSTEM ON indicator, an l.e.d., D14.

Current also passes through R15 to C7. This charges in about two to three seconds, turning on TR3. As will be apparent later, the circuit is arranged so that the motor may be instantly reversed when closing the door. The delay provided by C7 charging prevents any damage to the motor by allowing a "pausing time" before

Transistor TR3 activates the relay RLA with D12 providing protection against the back e.m.f. Relay RLA switches on the "open door" connection to the motor.

When the door is fully open, it operates microswitch S8, which now opens. Thus gate IClb is turned off and diode D10 ensures that the base of TR3 is rapidly returned to zero volts, thus quickly stopping the motor.

CLOSE DOOR

Microswitch S7 is now closed, and S8 open. A pulse from C4 will now trigger gate ICld and this channel operates exactly as described above.

The delay provided by C8 may not seem necessary, but there are certain circumstances where the motor may be required to suddenly change from "open door" to "close door", and C8 provides a suitable margin of safety.

When the door is fully closed, microswitch S7 opens, and gate ICld is turned off thus stopping the motor.

PRIORITY OPEN

It can be seen that whilst the door is in motion, both microswitches are closed and thus both gates IClb and ICld are able to trigger if another pulse is received from C4. Thus both relays could operate together, causing problems (to say the least!) for the motor.

This arrangement is deliberate. since it offers the opportunity to instantly reverse the door whilst closing, and also enables the door to be set in motion from a mid way position, if it has stopped for some reason.

In order to prevent TR3 and TR4 switching on together, IClc has each input connected to the output of IClb and ICld respectively. If both these gates turn on together, IClc goes high, turning on TR2. This reduces the feedback voltage via R13 to zero, thus switching off ICld. Resistor R12 limits the flow of current from ICla into TR2, without affecting the rest of the circuit.

All this happens so quickly that the relay RLB does not operate, leaving relay RLA to safely open the door.

If the door is already closing, and a pulse is received from C4, the same action will follow, with ICld turning off, whilst IClb switches on. Diode D11 will ensure that the

switches off rapidly, whilst C7 will delay the switching-on process. Thus the closing door stops, pauses, then re-opens.

It will be apparent that a pulse received whilst the door is opening will have no effect on the operation of the door

Capacitors C5 and C6 suppress noise picked up from the microswitches

TRIGGERING THE CIRCUIT

The logic circuit may be triggered from the car, with the ultrasonic link as previously described, or alternatively a positive triggering pulse may be generated by S5 or S6 and any number of push button switches may be wired in parallel to these for additional control.

One push switch, S5, should be mounted near the door, inside the garage, and is used to close the door after parking the car inside. Another push switch, S6, may be mounted inside the house to open the door, or if too inconvenient regarding the necessary wiring, it could be mounted outside the garage. However, this may be considered an unacceptable security risk, whereby a key switch electronic combination switch could be used.

Note that both these switches will open or close the door as will the ultrasonic remote control system.

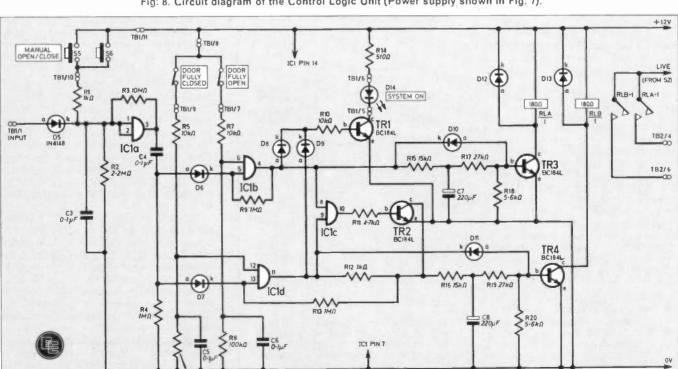


Fig. 8. Circuit diagram of the Control Logic Unit (Power supply shown in Fig. 7).

CONSTRUCTION starts here

CONSTRUCTING THE POWER UNIT AND CONTROL LOGIC

The circuit requires a piece of stripboard measuring a minimum of 46 holes by 22 strips, however a larger piece measuring 50 holes by 34 strips was found to be more convenient in the prototype. It must be securely mounted bearing in mind the various mains connections in close proximity.

A rebate in the stripboard is cut out as shown in Fig. 9 to provide access to one of the case mounting

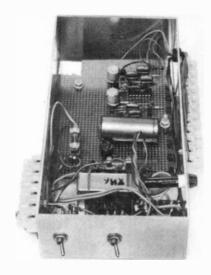
screws. Drill the three stripboard mounting holes where shown, and break the tracks (20 in all) as indicated in the second part of Fig. 9.

Mount the i.c. holder, wire links, resistors and fuse clips where indicated, followed by the non electrolytic capacitors. Take care to fit the diodes, transistors, bridge rectifier and electrolytic capacitors the correct way round. Finally, connect the flexible leads and fit the cmos i.c. taking great care not to touch the pins during fitting, as it is static sen-

THE CASE

An aluminium case measuring 205 ×100×50mm was chosen, and this provided sufficient space for the stripboard, transformer, relays, neon mains indicator LP1 and switches S1 and S2. Carefully arrange these items drilling, to ensure that adequate space is left especially in the region of the neon and switches.

£28.70



View inside the Control Logic case.

The relays need to be securely fixed in position and it may be necessary to fabricate a small bracket to do

Mark the hole positions for the stripboard (3 off), case mounting holes (2 off), transformer (2 off), neon, switches (2 off), leads (at least 4 holes), and terminal blocks. Drill the holes and fit rubber grommets to all holes requiring leads to pass through.

The positions shown for the main components given in Fig. 10 are intended as a guide, in practice great care must be taken in ensuring that there is no danger of any components breaking loose and shorting out. REMEMBER THAT THERE IS LIVE MAINS PRESENT IN THIS EN-CLOSURE.

Mount the transformer, switches, neon, terminal blocks and circuit board, passing the appropriate leads neatly through a grommet. Note that the mains fuse is mounted on an insulated 20mm board mounted fuse holder, secured with a fixing screw. On the prototype, rubber grommets were fitted over the screws securing the circuit board, between the board and the aluminium case, to prevent the circuit board being short circuited by the case.

Fit the relays, and wire up the mains circuit as indicated in Fig. 10. The mains and motor terminal strip may be mounted outside the case as shown, in which case a protective cover should be fitted, or inside for extra security against an electric shock. It is extremely advisable to sleeve all solder joints on the mains side of the circuit.

Finally, complete the low voltage connections. Ensure the mains EARTH lead is securely fixed to the case, using a tag on a transformer mounting screw.

COMPONENTS

Resistors R1, 12 $1k\Omega$ (2 off) R2 2-2MΩ R3 $10M\Omega$ R4, 9, 13 R5, 7, 10 $1M\Omega$ (3 off) 10kΩ (3 off) R6, 8 100kΩ (2 off) R11 4-7kΩ 510Ω ±W 15kΩ (2 off) R14 R15, 16 R17, 19 27kΩ (2 off) R18, 20

5.6Ω (2 off)

Capacitors

C1 C2-C6 C7, 8 2200 µF 25 V elect 0.1μ F polyester (5 off) 220 μ F 25V elect (2 off)

Semiconductors

1A, 50V bridge rectifier D1-D4

1N4148 small signal silicon (9 off) D5-D13 0.2 inch red l.e.d. high brightness, wide angle 4081 CMOS quad 2 input AND gate D14

IC1

TR1-TR4 BC184L silicon npn (4 off)

Switches

S₁ d.p.d.t. miniature toggle S2 s.p.d.t. miniature toggle

S3, 7, 8 s.p.d.t. lever actuated heavy duty microswitch (3 off)

\$4 See text

S5, 6 Push-to-make, non-latching (2 off)

Miscellaneous

Mains transformer, 9V, 0.6A secondary LP1 Mains neon indicator with integral series resistor F1 20mm anti-surge, 1.25 A (for the motor specified) 20mm 250m A

RLA, RLB Relay, 180Ω 12V coil, 2 normally open contracts rated at 240V,

chassis mounting (2 off) Terminal block, 12 way, 2A Terminal block, 6 way, 2A

Stripboard, 0.1 inch matrix, 50 holes by 34 strips; aluminium case size 201 x 100 \times 50mm; case size 72 \times 50 \times 25mm (2 off, for S5, 6); fuse holder, 20mm chassis mounting (for F1); fuse clips, p.c.b. mounting (2 off, for F2); 14 pin d.i.l. i.c. holder; small grommets (5 off); equipment wire; rubber sleeving; solder tags; board mounting hardware (M3 or 6BA); heavy duty wire, 16/0·2 (for mains wiring).

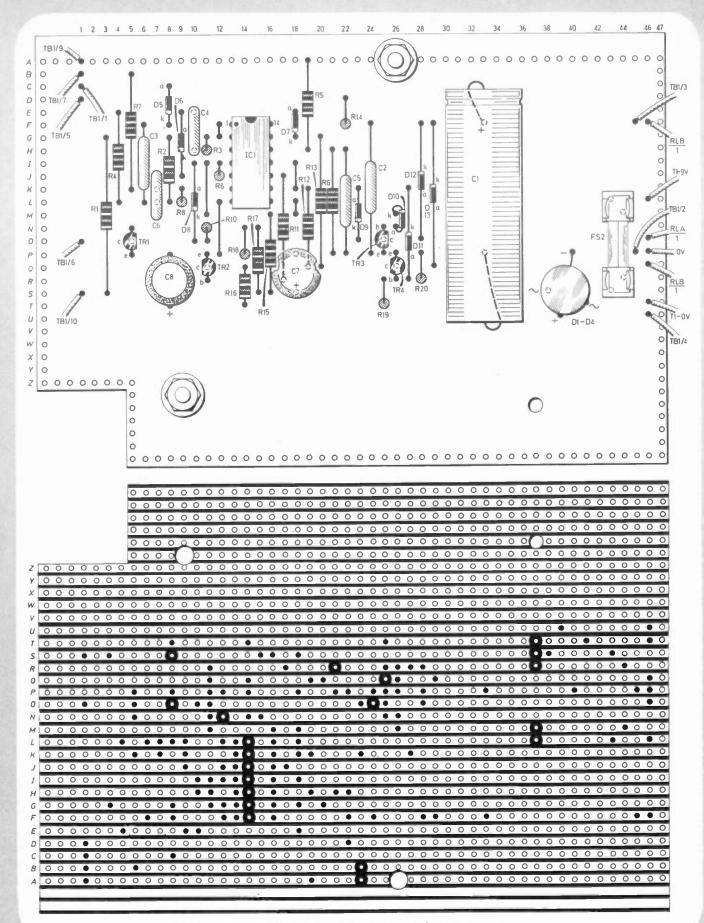


Fig. 9. Component board layout and track breaks for the Control Logic Unit.

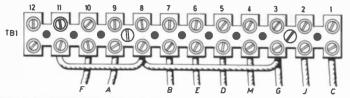
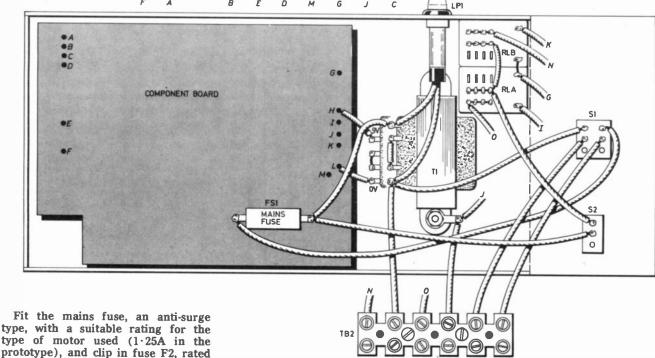


Fig. 10. Interwiring diagram for the Control Logic component board and Power Supply. Note that terminal blocks have been folded up for clarity.

NEON



TESTING THE LOGIC

diagram, Fig. 10.

Connect a voltmeter set to the 20V d.c. range across TB1/4, the positive output from the rectifier and TB1/2, 0V. Connect the mains, and switch on. The meter should read 12 to 16 volts.

at 250mA, as indicated in the layout

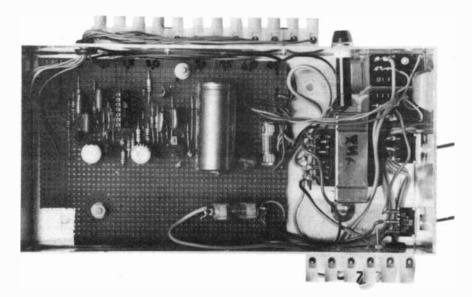
Now wire D14, an l.e.d. across TB1/5 and TB1/6 ensuring correct polarity, the cathode, k, to TB1/5.

Connect a short wire to joint TB1/3 and TB1/4 and then join TB1/7 with TB1/8 (to represent the DOOR FULLY CLOSED switch, S7).

Using a piece of flexible wire, join TB1/10 to TB1/11 for a moment. This represents the MANUAL OPEN push-button switch, S5, being activated and the l.e.d. should light. About two to three seconds later, the OPEN DOOR relay, RLA, should activate

Remove the link from TB1/7. The l.e.d. and relay should instantly

Control Logic Unit with lid removed to show component board, power supply and relays. Note that this prototype has the non-preferred mounting of TB2 on the outside of the case.



switch off. Join block TB1/8 to TB1/9 and re-join TB1/10 to TB1/11 for a moment. The l.e.d. should again light and three seconds later the 'CLOSE DOOR' relay, RLB, should activate.

Remove the link from TB1/7 and the l.e.d. and relay RLB should switch off. Now join TB1/8 with both TB1/7 and TB1/9 (representing the door half open). Again join TB1/10 with TB1/11 for an instant. The l.e.d. should light, but only relay RLA should activate, NOT both relays.

Repeat one or more of these tests, activating the circuit by connecting TB1/11 to TB1/1, input instead of TB1/10 (this will represent the Ultrasonic Control system activating the Control Logic).

If the circuit does not perform correctly, take voltage measurements throughout the circuit, using the circuit description as a guide. Note that a logic one (high) should produce a voltage nearly equal to the supply, and a logic 0 (low) should read nearly zero volts. Check that the components are fitted the correct way where applicable, and that BC184L transistors are not confused with BC184 types, which have a different lead arrangement.

Assuming all is well, connect the ultrasonic receiver positive power lead, 0 volts and output wire to TB1/3, TB1/2 and TB1/1 respectively. It should now be possible to trigger the circuit with the Ultrasonic

Transmitter.

MECHANICAL ASSEMBLY

The system works as follows: (note that all item numbers referred to are given in Fig. 11). With the door closed, the electronic control system receives the command to open from either the Ultrasonic Remote Control (item 4) or the push-button switches (item 9). The solenoid (item 8) will then activate to release the lock mechanism and the motor (item 1) will unwind the cord thus raising the door since it is counterbalanced towards opening.

To achieve this bias towards opening, a system of pulleys and weights (item 10, see also Fig. 2(d) last month) is attached to the door and to overcome the initial inertia (that is, reluctance to move) when raising the door, a cord secured to the roof joists by springs (item 5) "catapults" the door backwards by means of a bracket (item 9) fixed at the top of the door. This is providing a force in direction A shown in Fig. 2(a), discussed last month.

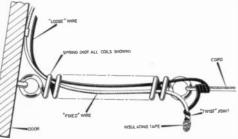


Fig. 12. Method of constructing tension cut-out switch, S4 (item 11, Fig. 11).

A microswitch, S8, not shown in Fig. 11 but mounted outside so as to activate when the door is fully raised, signals for the motor to stop.

CLOSING

Upon receipt of the command to close the door, the motor will reverse, hence winding up the cord and lowering the door.

Microswitch S7 (item 7) will stop the motor when the door is fully shut.

FAIL-SAFE SWITCHES

A large microswitch, S3 (item 12), is fixed in position at the base of the door, such that it activates if it touches any object during the descent of the door. A piece of rubber or plastic sleeving may be slipped over the lever to protect the car, and the switch should be carefully positioned to ensure that it operates correctly.

Should any other fault occur which would prevent the motor switching off when the door is closing or fully closed, the tension cut-out switch, S4 (item 11), comes into being. This is made with a pair of thin insulated wires pushed through the spring which connects the main driving cord with the door. One wire is "loose"

(that is, not fixed firmly at the garage door end) with its other end fixed to the cord. The other wire is secured to the door, allowing enough slack in the wires to enable the spring to expand in normal use. The other end of the secured wire is twisted around the wire attached to the cord, with the two ends stripped, twisted, and then insulated with tape. See Fig. 12.

As the spring stretches in normal use, the join should hold. Under fault conditions the spring will stretch well beyond its normal limit, and the fixed wire will disconnect at the twisted join, thus disabling the system.

MICROSWITCHES

A small lever microswitch, S7, must be fixed to signal the logic control circuit when the door is fully closed. It should be screwed to the wooden frame about 500mm from floor level such that when the door is fully closed, it is activated.

A similar microswitch, S8, is also used to signal the fully open position. It must be mounted outside the garage, near the top of the frame, and be activated when the door reaches the end of its travel. A small piece of plastic may be used to protect the switch from direct rain. or preferably it should be encased with only the lever protruding.

A certain amount of experimentation will be required to find the ideal locations of these switches so this is largely left up to the constructor.

OPENING MECHANISMS

To initially open the door, arrange two large nails or screws in the garage roof joists sufficiently high to clear the door. Fit two springs and a piece of cord tightly between the nails as shown in Fig. 11. Attach a metal bracket to the inside of the door, so that as it closes, it presses against the cord. The tension in the cord should initially open the door sufficiently to allow it to slowly rise. The nails should be fixed at about 250mm back from the door frame, but the exact position must again be located by trial and error.

WEIGHT AND PULLEYS

To raise the door to the fully open position, a system of pulleys are employed whereby the point on the door which runs vertically in a "track" up the side of the frame is linked, via a cord, to a weight. This is illustrated in Fig. 2(d) and Fig. 11.

It is possible that the door may already be fitted with a counterweight or powerful spring in which case no further weights need be added. If not then fix the two pulleys to a roof joist as shown at one edge of the door, feed the cord through and attach a weight (we found 4kg to be sufficient) to one end, the other end fixed to the door at the aforementioned point, as shown.

The length of cord, optimum weight and best suited mounting positions are ideally established by experimentation.

ELECTRIC MOTOR

At the heart of the system is the electric motor, and this will probably be the most expensive item. The motor used in the prototype was a reversible mains motor, complete with gearbox.

COMPONENTS

Mechanical Components

Type 1-Extension, approx. 300mm long, 15mm dia. with looped Springs

ends to attach cords (used for closing the door).

Type 2—Extension, approx. 100mm long, of the type used in spring

balance scales (for the counter balance weights).

Type 3—Extension, approx. 100mm long, about half the strength of type 2 (two required, for the initial opening mechanism).

4kg counterwight (8 to 9lbs) however this may vary and in practise

Weight

will need to be adjusted to suit.

Pullevs 2 required, of the type available at most builders yards. Brackets

Type 1-mild steel strip approx. 200 × 20 × 1.6mm (for initial opening mechanism.

Type 2—approx. $100 \times 100 \times 1.6$ mm (2 off required for mounting

baseboard to garage wall). 400 × 300 × 25mm chipboard or plywood sheet.

Base

Electro-mechanical Components

Fracmo 240 Va.c. reversible motor and gearbox, output speed Motor

56r.p.m., output torque 50lbs/in.

Solenoid 240 Va.c. type, optional if lock is required.

Miscellaneous

Nylon cord, approx. 10kg minimum breaking strain, about 10m required; metal disc to prevent cord slipping off gearbox spindle; 2 large nails or screws; assorted nuts, bolts, washers and screws for installation.

This list is intended primarily as a guide and due to the nature of the system and the enormous variation in design of garage doors, it is quite likely that the project will have to be tailored to suit individual needs.

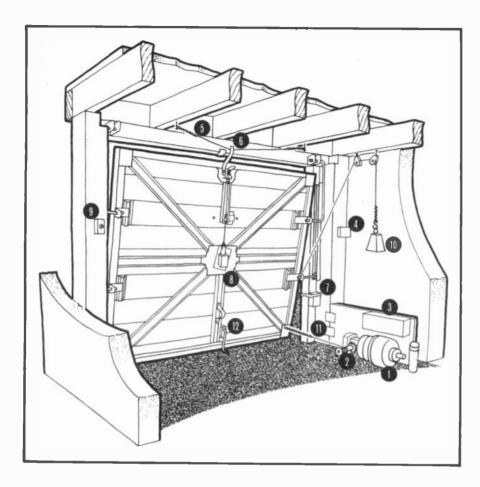


Fig. 11. Cut-away drawing of the garage showing the locations of the major mechanical components and electronic units

ITEMS LIST FOR FIG. 11

- 1 240V mains reversible electric motor
- 2 Gearbox and winch
- 3 Control Logic unit
- 4 Ultrasonic Receiver unit (transducer is mounted in the door frame)
- 5 Initial opening mechanism
- 6 Top bracket
- 7 Door fully closed switch, S7 (the door fully open switch, S8 Is mounted outside the garage at the top of the frame)
- 8 Lock solenoid
- 9 Manual open/close pushbutton switch, S5
- 10 Pulley/weight counterbalance system
- 11 Wind-up cord/spring incorporating the safety cut-out switch, S4
- 12 Obstruction safety cut-out switch, S3

The motor employed was a Fracmo type, with an output shaft running at 56 rpm and a more than adequate torque of 50lb/in.

A piece of wood measuring 400 × 300mm may be used to mount the motor and the Control Logic circuit. The motor is bolted to this wooden base, which can then be fixed vertically to the wall at the correct angle. See photograph for layout.

Ensure that the motor winding shaft has no rough edges to cut the cord and drill a small hole in the end of the shaft to take a self-tapping screw to secure a circular disc of metal. This will prevent the cord slipping off the end of the shaft.

Drill and de-burr a hole in the metal disc, so that the cord may be pushed through and fastened. The other end of the cord should be fixed to the door via a spring to even the tension. Leave the cord slightly slack, in the fully open position to enable a few turns to be wound on the shaft.

Fix the wooden base vertically to the garage wall near floor level, and at the correct angle so as to allow the cord to wind neatly onto the shaft

SOLENOID DOOR LOCK

Lock designs vary even more than garage doors, and again it is impossible to give exact details. A 240V

mains solenoid is used to unlock the door, and ideally it is to be linked to the existing lock, so that the door may still be unlocked by key in the event of a mains or system failure.

The solenoid used in the prototype was wired across the "door open" motor connections. It was bolted to the garage door in the centre below the lock release plunger. A piece of strong, heavy gauge steel wire linked the solenoid plunger with the lock release mechanism, so that the lock was released as the solenoid energised.

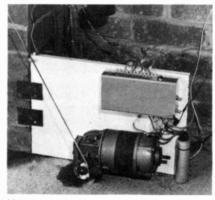
HIGH BRIGHTNESS L.E.D.

A hole should be drilled right through the door frame just below the microswitch, S8, to house the high brightness l.e.d., D14. Insulate the long wires joining the l.e.d. to the Control Logic, before pushing the device fully into the hole, with the wires emerging inside the garage.

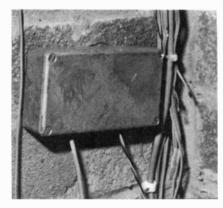
A piece of red transparent gel glued in front of the l.e.d. will greatly improve visibility in strong light, or alternatively a complete l.e.d. indicator with lens etc may be purchased.

RECEIVING TRANSDUCER

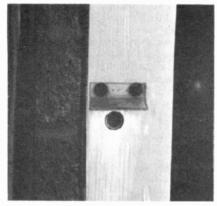
Park the car in front of the garage, and drill a hole into the garage door frame for the receiving transducer,



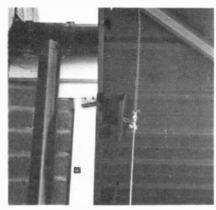
Motor, gearbox and Logic unit (items 1, 2 and 3) on baseboard.



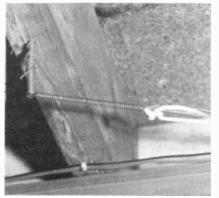
Ultrasonic Receiver (item 4) mounted on inside wall.



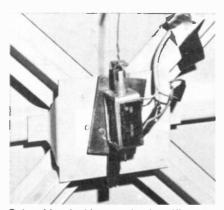
Receiving transducer (part of item 4) mounted outside in the door frame.



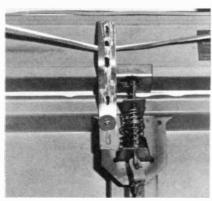
Door fully open switch, S8 (not shown in Fig. 11).



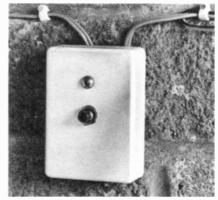
Part of the initial opening mechanism (item 5).



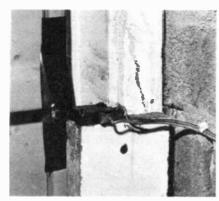
Solenoid unlocking mechanism (item 8).



Top bracket (item 6).



Manual open/close switch, S5 (item 9) mounted on inside wall.



Door fully closed switch, \$7 (item 7).



Obstruction safety cut-out switch, S3 (item 12) shown actuating on bonnet.

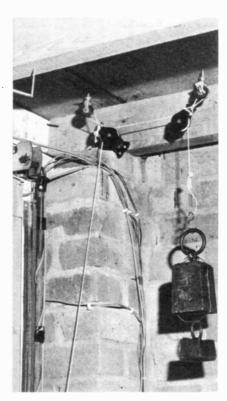
X1, so that it will line up with the transmitter. This hole should be wide and deep enough to flush mount the transducer, and a small hole should be continued into the garage for its lead. A small piece of stiff plastic may be fixed just above the transducer as a rain shield.

Mount the receiver unit on the garage wall near the transducer, and wire it in with the screened cable.

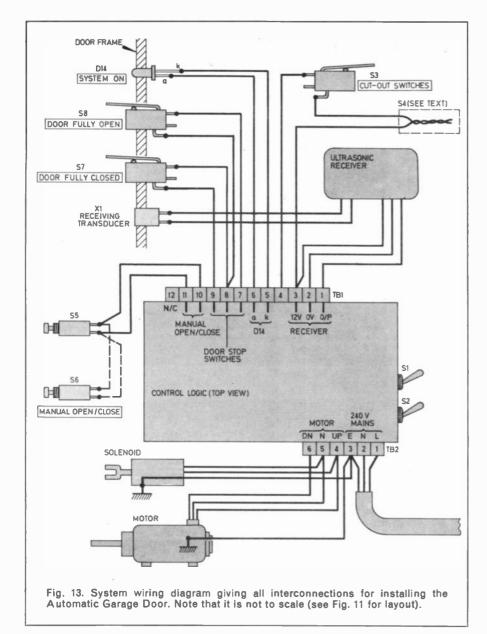
TRANSMITTING TRANSDUCER

The transmitter unit and transducer must be securely mounted inside the engine compartment of the car, such that the transducer has a direct line of sight through the front grille. Note that ultrasonic transducers are fairly directional, and mount the unit at a suitable angle to align with the receiving transducer embedded in the door frame.

A piece of very thin plastic (for example Cling Film) may be used to protect the transmitting transducer from dirt and water. It will reduce the output slightly, but is virtually essential bearing in mind the adverse conditions with which it must contend.



Pulley and weight system (item 10). Note that the front most pulley is mounted in the roof as close as possible to the door frame without fouling the hinge struts, the rear pulley clearing the moving weights away from the corner support.



Having fixed the transducer and transmitter unit, mount the push button inside the car, and connect it with the transmitter.

Alternatively, the transmitter can be built as a hand held unit with the transducer and switch mounted into the same case as the circuit board and therefore can be used from any vehicle.

POWER CUTS

Unless the garage is accessible through a rear door, some provision should be made in case of a mains power failure, which would, to say the least, be inconvenient! Depending upon the type of spring used, it may be possible to simply inch the door open just enough to allow the spring to be unhooked. Alternatively the spring could be fixed to the door by

means of a bolt which may be unscrewed, if necessary, from the outside.

WIRING

Having fitted all the various components, the wiring can commence. The wiring diagram, Fig. 13, used in conjunction with Fig. 11, shows all connections.

Start with the Ultrasonic Receiver, the output, 0V and 12V rails going to terminals 1, 2 and 3 respectively, on TB1, Control Logic.

Continue by wiring the microswitches S7 and S8, the l.e.d. D14 and the manual open/close push-button switches S5 and S6. As these connections are all in pairs, twin cored cable of the type used for doorbell installation is ideal.

Where possible, run all wires along the same route to assist in lacing them together when completed. All joints to be sleeved after soldering.

Now we move on to the emergency cut-out switches S3 and S4, wired from terminals TB1/3 and 4. As both these switches are mounted on the moving door, the wires must be carefully routed via the hinge struts and great care taken to prevent the wires being severed by any moving parts.

MAINS WIRING

When wiring the mains side of the circuit, exercise extreme caution when dealing with the potentially lethal voltages involved. Remember that a garage is a damp environment with a large amount of exposed metal, so on no account at all should any mains terminations be exposed to the elements or enquiring young fingers! The importance of this cannot be overstressed.

For these reasons, TB2 must either be a completely insulated type or mounted inside the Control Logic case with all wires fed through grommeted holes. Also all metalwork, the door, solenoid frame and the motor housing must be properly earthed (by a firmly bolted connection).

For the solenoid wiring, the same precautions must be taken when routing the leads as for the fail-safe switches, as this too is mounted onto the door.

When it comes to the motor connections, follow the manufacturers instructions supplied, remembering that the neutral connection is common, the UP (or "open") will be forward and the DOWN (or "close") will be reverse.

When all wiring is complete, form the wires into a neat bundle and secure them to the door frame or masonry where necessary. Then vigorously check all wires and terminals both visually and with the aid of a continuity tester before applying any power to the circuit. If all is well, the system is ready for a trial run.

IN CONCLUSION

As may have become evident in the description of the automatic garage door, it is extremely difficult to give exact details and dimensions due to the enormous variety of door types and hinge design. However, we hope to have given you enough inspiration to develop the idea to suit your own particular needs, having shown how it can be done.

The system illustrated is very much a prototype and has room for improvement, but it does work and very well too. So come on all you budding Heath-Robinsons, armed with the necessary electronics, I am sure even a shuttered type door could be adapted!

RADIO WORLD

By Pat Hawker, G3VA

News Gathering

In television broadcasting one of the continuing problems is that of providing "links" back to the studios for pictures of news events from places to which there are no permanent broadband circuits. This can be accomplished today by means of transportable satellite terminals, such as the one built by the IBA in 1978 or the more recent BBC unit.

But, at least in Europe, these are still experimental systems, although used from time to time for operational broadcasts from the Channel Islands, a North Sea oil rig and as a temporary link between Madeira and Portugal. There is still no European "domestic" satellite system that includes transponders dedicated to this type of work, and even when this does happen the equipment and satellite costs seem likely to remain quite high.

One of the unanticipated "hazards" of electronic news gathering (ENG) stems directly from the compact size and light weight of the expensive equipment. In the turmoil of news gathering it is sometimes difficult to keep an eye on equipment costing many thousands of pounds. One team recently lost its small colour camera and portable tape recorder and is beginning to believe that ENG stands for "easily nicked gear".

Slow-Scan

It has always seemed surprising to me that more use has not been made by the broadcasters of high-quality "slow-scan" television systems capable of providing a succession of still pictures over narrowband radio links or even the public switched telephone system. In the 1950s, various "cable TV" systems were developed both by the BBC and ITV for sending rather jerky black-and-white film across the atlantic, but with the coming of Intelsat 1 ("Early Bird") these were thankfully abandoned. But even today, with the worldwide satellite system, quite a lot of news reports, for example, consist of a telephone voice set against a "library" photograph of the correspondent or the city concerned.

Yet the development of digital frame stores and digital synchronisers during the past few years could result in much improved slow-scan systems, and this type of approach is now being used in Germany. Brunswick Technical University has developed a process for transmitting colour photographs via the public switched telephone network to allow weather maps to be sent to the Frankfurt television production centre.

The picture is viewed with an ordinary television camera and is displayed on a domestic receiver. The colour image from the camera is stored in a digital memory and then transmitted slowly over the telephone network. Each picture takes 20-100 seconds, depending on the quality required, although only 5 seconds are

needed for a less clear picture for recognition purposes.

British Telecom at their Martlesham research centre are also developing various forms of "narrow-band" television. Radio amateurs have used both black-and-white and colour slow-scan television for a considerable time, but are satisfied with results rather different from the high-quality pictures sought by broadcasters.

Military ingenuity

The very first contact entered in my post-war log was with SVIEC in Athens on February 27, 1946. In the current issue of *Mercury* the journal of the Royal Signals Amateur Radio Society, vice-president Major General (retd) Eric Cole, CB, CBE, former Director of Telecommunications, War Office, SUIEC and G2EC, tells the story of how SVIEC came on the air.

As Chief Signals Officer, Land Forces,

As Chief Signals Officer, Land Forces, Greece in 1945-46 he arranged, after the German evacuation of Crete, for a large consignment of their signals equipment to be sent to Athens. But while looking it over a 30lb generator fell on his foot, resulting in a stay in Hassani Military Hospital

To while away the time he had an HRO communications receiver installed beside his bed and found to his surprise that, although the amateur bands were still nominally closed, a lot of stations were already jumping the gun on 7MHz. The Corps transmitters, some miles from the hospital, included a number of much-prized American BC610 transmitters (the military version of a Hallicrafters pre-war amateur transmitter). An "order-wire" and keving line were soon installed.

In his capacity as a national telecommunication authority he allotted himself

On the Beam

on the amateur bands.

A few years later, General Cole, while living in a black of flats in Mayfair showed similar ingenuity and determination when objection was taken to his roof-top aerials. He devised a system of three aerials, including a 28MHz rotary beam, that could all be erected and dismantled at short notice, during darkness or when they were unlikely to be observed by censorous eyes.

the call SVIEC and was soon working

stations all over the world. British amateurs were re-licensed for 1-8 and 28MHz only from January 1, 1946 and soon afterwards British Service personnel in the Mediterranean area were permitted to use four letter callsigns, beginning XA,

At other times all poles and wires were dismantled and could not be seen from the ground 80ft below.

New Emissions

For many years the various types of radio emissions have been classified internationally as part of ITU's Radio Regulations: A1 for CW, A3 for a.m telephony; A3J for s.s.b. with fully suppressed carrier; A5 for television, and so on. But now the old order changeth and is giving place to new.

As a result of the 1979 World Administrative Radio Conference a revised list officially came into operation on January 1. What was A1 now becomes A1A; A3 becomes A3E; F1 for r.t.t.y. becomes F1B. In a rather more radical change s.s.b. becomes J3E; f.m. telephony becomes F3A; vestigial-sideband, amplitude-modulated television is C3F.

In front of these basic classifications a further set of hieroglyphics can, when required, be used to indicate the bandwidth of the emission. This is now written according to a code in which the frequency designation is used as a "decimal point": thus 0·1Hz becomes H100; 400Hz is written 400H; 2·4kHz becomes 2K4; 1·25MHz is 1M25; while 5·5GHz would be written 5G5.

■ Satellite TV

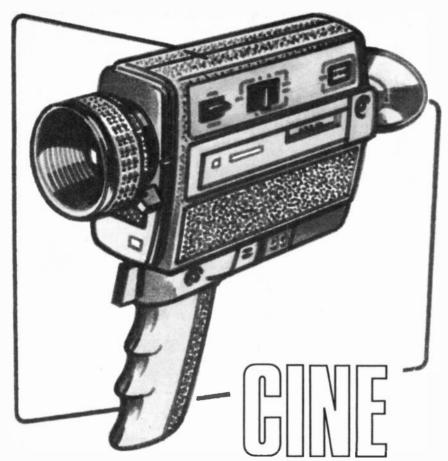
Despite much talk, the outlook for UK direct-broadcast satellites (DBS) is still confused. Britain is to pay a one-third share (£77 million) towards the £230 million cost of the European Space Agency's L-Sat project which plans to put a large satellite capable of carrying a high-power DBS transponder into geostationary orbit in 1986. But this will be at 19 degrees West, and not the allotted position for a fully operational UK DBS satellite at 31 degrees West.

The L-Sat transponder may in fact be used for a Pan-European system under the aegis of the European Broadcasting Union which has been offered experimental use of a DBS transponder on L-Sat free of charge.

With the support of broadcasters in the UK, Austria, Belgium, Greece, Ireland, Israel, Italy, Malta, The Netherlands, Spain, Switzerland, Turkey and Yugoslavia, EBU are to make feasibility studies this year (1982) using lower-power distribution satellites. But although these will include transmission of many TV programmes it is not intended that these should be available to the public.

In the meantime there is a growing feeling that before DBS can flourish some way must be found of overcoming at least some of the barriers that make it difficult to provide television programmes across frontiers. This applies particularly to the vital differences in transmission standards and particularly those between the PAL and SECAM colour systems.

One proposal for overcoming this problem has come from IBA engineers, an ingenious system called MAC (Multiplexed Analogue Component). This system, if used for satellite transmission, would provide high-quality pictures, in a universal form more suited to the characteristics of satellite transmission.



Interval Timer & Frame Counter

PRODUCING special effects with a cine camera can be aided by using one or both of the circuits in this article. These are not normally found on medium priced cine cameras.

The obvious use for the Interval Timer would be in time lapse cine, for example, the opening of flowers, fast moving clouds or simulation of early black and white comedy films.

The Frame Counter has rather a

The Frame Counter has rather a different use in as much as mode of operation is shorter but it is still very useful, for example, the inclusion of a Matte, such as a laser shot from a spaceship could be fired on the film where you want it. Using the Frame Counter at the beginning of the first exposure it will tell you the number of frames that have been taken. A backwinder may then be used to rewind the film to the precise position where the second exposure is to take place.

CINE SOCKETS

It should be noted now that not all cine cameras will have the facility to use both circuits. For using the Frame Counter the camera must be equipped with a flash or sync. socket. BY L.A. PRIVETT

This will have a pair of internal contacts that make as each frame is taken.

The Interval Timer can only be used with a cine that has a single frame switch, that is for taking one frame at a time. It must also be of the type which uses an electro mechanical solenoid as the trigger and be connected to an output socket for remote control. This will normally be a 2.5mm or 3.5mm jack socket. If in any doubt about either of these sockets, contact your local photographic dealer who should be able to help.

CIRCUIT DESCRIPTION

The complete circuit diagram is shown in Fig. 1. There are two distinct sections, the Interval Timer and the Frame Counter, and could if required be built separately. It must be remembered, however, that IC2 is shared by both sections and so if only building one section, the gates not used should have their inputs connected to the positive supply rail.

ICl forms the basis of the Interval Timer, the 555 device being wired in the astable mode, to provide a very low frequency square wave oscillator. The speed of this is controlled by VR1.

The pulse from the output, pin 3, is inverted by IC2a and fed to the base of TR1 via R5. Pulses from IC1 cause TR1 collector/emitter resistance to rise and fall in sympathy; when low it acts like a switch being closed across SK1 to reach the REMOTE socket on the camera.

The frequency of the output pulses ranges from approximately 1 a second to 1 a minute. To indicate that the timer is working, for setting up purposes, an l.e.d. is connected across IC1 output. The l.e.d. in fact is the d.p. pin on X1.

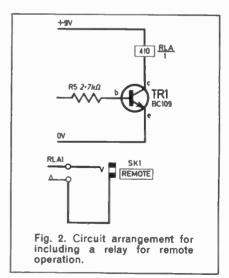
RELAY OPTION

TR1 can be connected to the camera using the collector and emitter as described above, or alternatively if this method is not favourable then a relay could be incorporated as shown in Fig. 2. This would of course consume more current so a larger battery or a small 9 volt power pack would be advisable for prolonged use. Each pulse is fed to transistor TR1 which activates the relay and closes its contacts to complete the circuit to the solenoid and so take one frame of film.

IC3 and IC4 are decade counters with 7-segment display outputs. These are wired together to form a counter from 0 to 99.

The pulses to the clock input to IC4 are derived from the SYNC. contacts on the cine camera via SK2, switching positive to pin one IC3. So the number of pulses received at SK2 are counted by IC3 and IC4 and displayed on X1 and X2.

R6 and C2 eliminate contact bounce from camera sync. output. R7 holds pin 15 on both IC3 and IC4 to ground (count condition) until S1 is depressed when pin 15 is taken "high" and resets both i.c.s to 00.





DISPLAYS

The outputs from IC3 and IC4 directly feed displays X1 and X2 whose cathodes are connected to the collectors on TR2 and TR3. Gates IC2a and IC2b and their surrounding components form a square wave oscillator whose output is fed via R11 to the base of TR3 and also to the input of gate IC2d wired as an inverter feeding the base of TR2 via R10. This circuit grounds each display alternately to function as a simple multiplexer. This keeps the current consumption down, as does the use of cmos i.c.s.

Once the counter reaches 99 the next pulse will reset the display to zero and continue counting.

COMPONENTS TO THE

Resistors

R1	470Ω	R4	680Ω	R7	$10k\Omega$	R10	6-8kΩ
R2	$18k\Omega$	R5	$2 \cdot 7k\Omega$	R8	$16k\Omega$	R11	6.8kΩ
R3	$1k\Omega$	R6	470kΩ	R9	$16k\Omega$		

All 1W carbon ± 5%

Capacitors

47µF 10V tantalum bead

C2, C3 0.047 µF miniature sleeved polyester (2 off)

Semiconductors

TC1

555 timer i.c. CD4011 CMOS Quad 2-input NAND gates CD4033 CMOS decade counter/7-segment outputs IC2 IC3, 4

TR1 BC109 silicon npn

TR2, 3 BFY50 silicon npn (2 off)

X1, X2 DL704 common cathode 7-segment displays

Miscellaneous

VR1 2.2MΩ log, law carbon potentiometer S1 momentary action push-to-make switch

S₂ miniature single pole on/off switch

2.5mm jack socket 3.5mm jack socket 9V type PP3 battery SK₁ SK2

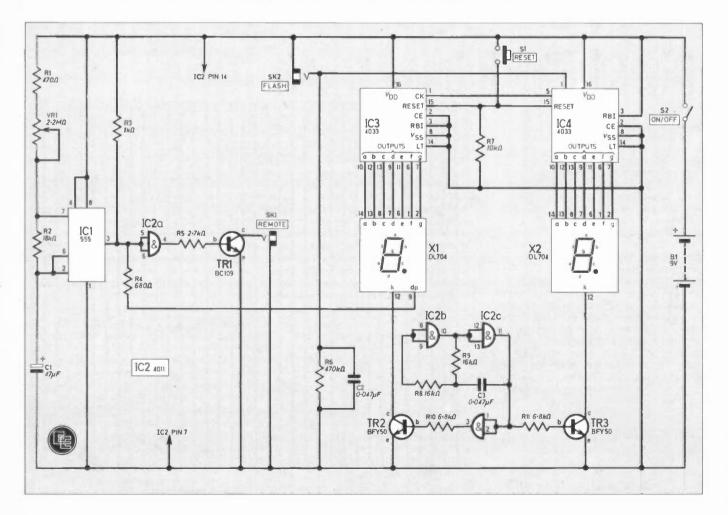
B1

Stripboard: 0.1 inch matrix size 34 strips \times 29 holes; d.i.l. i.c. sockets: 16-pin (2 off), 14-pin (1 off), 8-pin (1 off); PP3 battery clip; 22 s.w.g. tinned copper wire (for links); general purpose miniature stranded connecting wire (or ribbon cable—see text); control knob; case, sloping front panel size 160 x 100 x 60mm approx., West Hyde TEK362 or similar.

REMOTE Option—see text, Fig. 2.

9V 410 ohm coil with 1 set of normally open contacts.

Fig. 1. Complete circuit diagram for the Cine Interval Timer and Frame Counter.



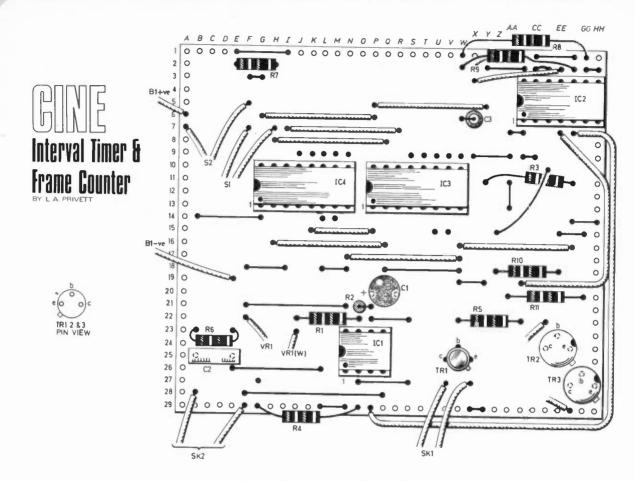


Fig. 3. Component layout on the topside of the stripboard.

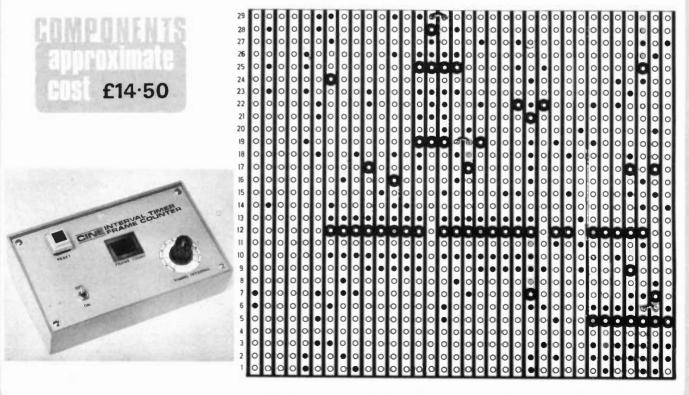


Fig. 4. Underside details for the circuit board showing breaks in the copper strips. Also note the small links across strips.

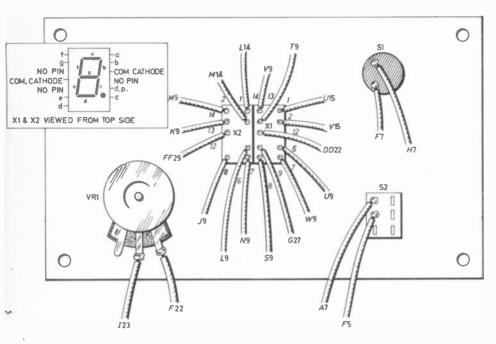


Fig. 5. Wiring to the rear of the front panel. The letters and numbers at the ends of the wires refer to locations on the topside of the circuit board.



ASSEMBLY

The prototype was constructed on a single piece of stripboard, size 87 x 75mm as shown in Fig. 3. First, make all the breaks required on the underside and then mount the i.c. sockets followed by the resistors, link

wires, capacitors and transistors. Do not insert the i.c.s. at this stage. Note that these are cmos devices and the usual precautions should be taken.

In the prototype the flying leads from the board to other case mounted components used ribbon cable for this to keep wiring together and neat. General purpose "separate" hook-up wire can be used here if preferred.

The two displays were glued to a red diffuser (Perspex) using Araldite and this assembly when set glued to the underside of the front panel behind a previously made rectangular cut-out.

If preferred a separate l.e.d. (TIL 209 for example) could be used in place of the d.p. l.e.d. on X1

as was used on the prototype. The power supply is a 9V PP3 battery, but if desired, a socket could be wired in place of the battery clip to allow an external power source to be used.

Finally check all wiring, especially to the CMOS i.c. power lines and the displays. If satisfied, insert the i.c.s into their sockets paying special attention to their orientation.

TESTING

Connect up to your camera using suitable leads. With the interval timer first check the polarity of the REMOTE socket and connect SK1 so that the more positive lead goes to the collector of TR1. Operate the single frame switch on the cine camera, turn VR1 to minimum resistance (fully anticlockwise), switch on and the "monitor" l.e.d. (d.p. or X1) should flash on and off at approximately one second intervals and if connected correctly to your cine camera, the camera should take 1 frame per second.

A stop-watch or other means will be needed to calibrate the scale around VR1.

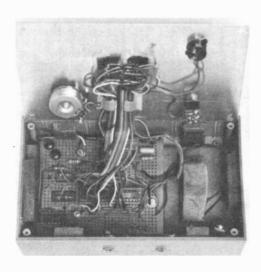
Connect the Frame Counter to the SYNC. or flash socket, press RESET and manually take 1 frame at a time to check that the counter is counting properly and not jumping. If contact bounce is affecting counting then some experimenting with the values R6 and C2 will be necessary to cure this.

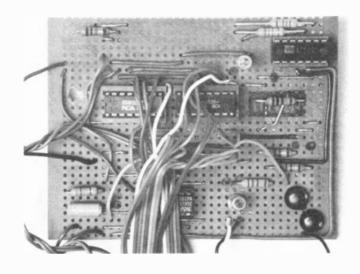
Using both facilities of this unit will be advantageous in time lapse cine as the Frame Counter will show the number of shots taken by the Interval Timer and also allow the "run time" of the film to be calculated.

For example: 96 frames at 24 frame per second=4 seconds. 90 frames at 18 frames per second=5 seconds.

Components mounted inside the case and interwired with ribbon cable.

Completed prototype board. The two transistors in the bottom right have been replaced with more common types.







THE transformer is often regarded in awe as a seemingly complicated and confusing device but in reality, it is a very useful and efficient component.

The first encounter newcomers to electronics are likely to have with transformers will probably be in power supply circuits, but in fact they have applications in many aspects of electronics, including radio, high frequency work and audio circuits. However, here we shall concern ourselves only with the mains power supply.

STEP-UP AND STEP-DOWN

Basically, a transformer is a device for converting one a.c. voltage to another a.c. voltage (note that they will not work with d.c. voltages) and and it can be either converting a low voltage to a high voltage (step-up) or converting a high voltage to a lower one (step-down). It is this latter type which feature mainly in EVERYDAY ELECTRONICS as it includes transformers used for stepping down the mains to low voltages.

PRIMARY AND SECONDARY

The input of the transformer is known as the primary winding and the output as the secondary winding ("winding" because the transformer consists of coils of wire wound onto a bobbin).

It is not uncommon for a transformer to have two or more primary or secondary windings and this is best illustrated on mains transformers with two 120V primary windings. The reason for this being that the transformer can be used on 120V mains (in Europe for example) or by connecting the two primaries in series, the transformer will work on 240V (the UK standard).

It is also permissible to join secondary windings together in series, so for example, two 4.5V secondaries in series will result in a 9V output.

It is not correct, however, to connect the primary to the secondary, except under exceptional circumstances.

PARALLEL WINDINGS

When considering connecting primary or secondary windings in parallel, you have to look at the current rating of the transformer and this is usually governed by the size of the component and the gauge (thickness) of the wire used.

So for example, if a circuit requiries 12V at 2A and the available transformer has two 12V secondaries rated at 1A each, these can be wired in parallel to supply the 2A.

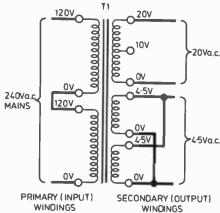


Fig. 1. Circuit diagram of a step-down transformer with two 120V primary windings connected in series for 240V mains operation, a centre-tapped 20V secondary and two 4.5V secondaries wired in parallel.

Note the two lines down the centre of the symbol representing the metalwork or "laminations" of the transformer. It is also customary to draw fewer "coils" on the low voltage windings.

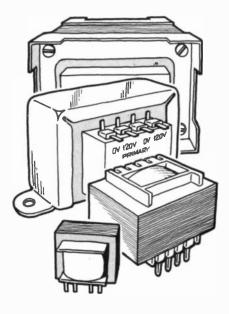


Fig. 2. Various types of mains transformer. From the front: a 1·2VA sub-miniature with p.c.b. mounting pins; 12VA p.c.b. mounting; 20VA chassis mounting with solder tag connections; 100VA chassis mounting with screw terminals.

Two very important points must be remembered when connecting windings in parallel: (1) they must be of the same voltage, that is, you cannot wire a 9V winding in parallel with a 12V winding, and; (2) the starts of the windings must be joined together and likewise the two finishes must be connected. If this rule is not observed and the windings are reversed, there will be no output, as the voltages cancel each other, and one very hot transformer!

The same applies to primary windings in parallel.

CENTRE TAPPING

Another form of winding is the centre tapped winding, and this simply means that an extra output voltage can be achieved by literally "tapping" off from a winding without breaking the continuity of the coils. For instance, if a tapping is made half-way up a 20V winding, 10V will result at this point.

VA RATING

The power capacity of a transformer is often quoted as a VA rating. This is the total output voltage multiplied by the maximum output current. For example, a transformer with a single 12V secondary rated at 0.5A is a 6VA transformer.

$12V \times 0.5A = 6VA$

Equally, by knowing the VA rating of a transformer, the maximum current can be calculated by dividing the VA rating by the output voltage. For example, a 20VA transformer with a single 10V secondary can supply 2A.

20VA+10V=2A
Incidentally, the VA rating is also true of the primary windings, so to calculate the current drawn from the supply on maximum load, simply divide the VA rating for the transformer by the primary (input) voltage. For example, a 12A, 240V mains transformer will draw 50mA from the supply.

 $12VA \div 240V = 0.05A$ or 50mA

TERMINATIONS

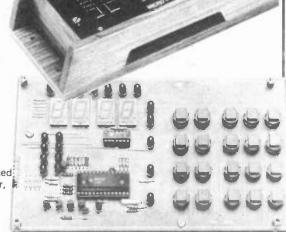
To a large extent, the VA rating will determine the physical size of a transformer, the higher the rating, the bigger the component.

A variety of shapes and methods are available, some of which are illustrated in Fig. 2. Transformers up to about 12VA can be supplied with solder tags or p.c.b. mounting, from 20VA to 100VA will almost certainly be too heavy for p.c.b. mounting so therefore they are only available with solder tags and integral fixing bracket. Greater than 100VA will be quite a substantial assembly with mounting frame and terminal block connections.

Micro-processor universal Timer

This incredibly versatile programmable timer can control up to 20 functions at accurately timed intervals over a period of a week. Originally developed for industrial and laboratory use it offers many interesting and exciting possibilities for the amateur constructor. Based on a pre-programmed TMS 1000 Microprocessor, the unit provides a 24 hour clock with four independent relay

controlled outputs with



a programmable period of one week. Up to 20 daily or weekly programmable functions can be set via a keyboard. Any of the timer functions can be assigned to control any one of the four relay outputs thus providing almost unlimited programming possibilities.

No previous experience of microprocessor programming is necessary since the manual explains all the possible operations, clearly and simply, enabling the inexperienced user to be fully conversant within one hour. Completed programme steps are indicated by LED's

The kit comes complete with printed panel and may be installed either as a 'built-in' or a 'free-standing' unit. A stabilised power supply mounted on a separate printed circuit board is supplied with the unit. It requires the addition of a 12V, 1A transformer. There is space on the board for up to four output control relays. One is supplied with the kit. Further relays maybe ordered separately as required. Price: (excluding wooden housing as illustrated) £48,37 inclusive of VAT and DELIVERED FREE on U.K. mainland.

APPLICATIONS

The programmable timer can provide central control of domestic electrical cooking, heating and entertainment equipment. The possibilities are limited only by the imagination of the user Control of house lighting to discourage intruders; control of TV or audio equipment; sound or video recording control; automatic plant watering; automatic pet doors or feeding—are a few simple examples. For the professional or industrial user many uses in this area of process control will be found.

TECHNICAL DATA

ower supply:

Mounted on separate pcb with space for up to four output control relays. Requires 12V/1A transformer.

CONTROL SWITCHING:

Standard relays (one supplied with kit) will switch 2A. Additional relays may be ordered separately. National relay, order no. HT 12V. Siemans relay, order no. R1 INV12

MICROPROCESSOR:

TMS 1000 DISPLAYS:

12mm 7 segment LED numerical display, LED programme function indicators.

DIFFICULTY GRADE: 3

KIT NUMBER: K1682

Box 30, St. Leonards-on-Sea, East Sussex TN37 7NL Tel: Hastings (0424) 753246

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Power supply 1 Amp Power supply for stereo 60 Watt amplifier Running light

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FM stereo decoder High quality FM tuner
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Digital thermometer FM stereo receiver (19 in. rack-mounting) 2 channel infra-red remote control light

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ape/slide synchronizer 3 channel coloured light organ 20 cm display (common anode) 20 cm display (common cathode)

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Microprocessor-controlled EPROM programmer (built and tested) Universal start/stop time

Repair Service available (for a nominal charge) if your soldering technique is not quite what it should be!

Any technical enquiries welcomed -in writing—and will be answered promptly by letter.

This easy to build 3 band stereo AM/FM tuner kit is designed in conjunction with Practical Electronics (July issue). For ease of construction and alignment it incorporates three Mullard modules and an I.C. IF. System. FEATURES: VHF, MW, LW Bands, interstation

muting and AFC on VHF. Tuning meter. Two back printed PCB's. Ready made chassis and scale. Aerial: AM - ferrite rod, FM - 75 or 300 ohms. Stabilised power supply with 'C' core mains transformer. All components supplied are to P.E. strict specification. Front scale size 10½"x 2½" approx. Complete with diagrams and instructions.

SPECIAL OFFER! TUNER KIT PLUS

- amplifier kit (usually £3.95 + £1.15 p&p) Mullard LP1183 built preamp, suitable for ceramic and auxiliary inputs (usually £1.95 + 70p p&p)
- · Matching power supply-kit with trans former (usually £3.00 + £1.95 p&p)
- Matching set of 4 slider controls complete with knobs for bass, treble and volumes (usually £1.70 + 80p p&p)

£3.80 p&p.





- Featuring latest SGS/ATES TDA 2006 10 watt output IC's with in-built thermal and short circuit protection.

 • Mullard Stereo Preamplifier Module.
- Attractive black vinyl finish cabinet, 9"x8%"x3%"(approx)

 10+10 Stereo converts to a 20 watt Disco amplifier.
To complete you just supply connecting wire and solder Features include din input sockets for ceramic cartridge, microphone, tape or tuner. Outputs - tape, speakers and head-phones. By the press of a button it transforms into a 20 watt mono disco amplifier with twin deck mixing. The kit incorporates a Mullard LP1183 pre-amp module, plus power amp assembly kit and mains power supply. Also features 4 slider

level controls, rotary bass and treble controls and 6 push button switches. Silver finish fascia with matching knobs and contrasting cabinet. Instructions available, price 50p. Supplied FREE with the klt. £14.95

Frequency response Input sensitivity

Tone controls

Distortion Mains supply Plus £2.90 p&p.

SPECIFICATIONS: Suitable for 4 to 8 ohm speakers. Suitable for 4 to 8 unin spec 40Hz - 20KHz, P.U. 150mV, Aux, 200mV

Mic. 1.5mV, Bass ±12db @ 60Hz

Treble ±12db @ 10KHz 0.1% typically @ 8 watts 220 - 250 volts 50Hz.

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2 WAVE BAND MW - LW

- Easy to build5 push button
- tuning . Modern design

6 watt output • Ready etched
and punched PCB • Incorporates suppression circuits.

All the electronic components to build the radio, you supply only the wire and the solder, featured in Practical Electronics March issue. Features: pre-set tuning with 5 push button options, black illuminated tuning scale. The P.E. Traveller has a 6 watt output neg. ground and incorporates an integrated circuit output stage, a Mullard IF Module LP1181 ceramic filter type pre-aligned and

assembled, and a Bird pre-aligned push button tuning unit.

£10.50 Plus £2.00 p&p

Sultable stainless steel fully retractable aerial (locking) and

speaker (6"x 4"app.).

£1.95/pack. Plus £1.15 p&p available as a kit complete

HIGH POWER

PRACTICAL ELECTRONICS - STEREO

READY BUILT OR IN KIT FORM KIT

125 WATT MODEL

SPECIFICATIONS:

200 WATT MODEL £14.95 Plus £1.15 p&p

£10.50

Plus £1.15 p&p

125 W Model

25Hz - 20KHz

400mV @ 47K

200 W Model

200 watts 70 · 95 max 4 · 16 ohms

25Hz - 20KHz

Max, output power (RMS) 125 watts Operating voltage (DC) 50 - 80 max 4 - 16 ohms Frequency response

measured @ 100 watts Sensitivity for 100 watts Typical T.H.D. @

50 watts 4 ohms Dimensions (both models) 205 x 90 and 190 x 36mm

The power amp kit is a module for high power applications — disco units, guitar amplifiers, public address systems and even high power domestic systems. The unit is protected against short circuiting of the load and is safe in an opcircuit condition. A large safety margin exists by use of

BUILT £14.25 Plus £1,15 p&p £18.95 Plus £1.15 p&p

> generously rated components, result, a high powered rugged unit. The PC Board is back printed, etched and ready to drill for ease of construction and the aluminium chassis is preformed and ready to use. Supplied with all parts, circuit diagrams and instructions

ACCESSORIES:

Suitable LS coupling electrolytic for 125W model

Suitable LS coupling electrolytic for 200W model

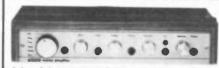
Suitable mains power supply unit for 125W model Suitable Twin transformer power supply for 200W model

£1.00 plus 25p p&p.

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30+30 WATT STEREO AMPLIFIER

Viscount IV unit in teak simulate cabinet, silver finished rotary controls and pushbuttons with matching fascia. mains indicator and stereo jack socket. Functions switch mains indicator and stereo jack socket, runctions symbol for mic magnetic and crystal pickups, tape and auxiliary. Rear panel features fuse holder. DIN speaker and input socket 30+30 watts RMS, 60+60 watts peak. For use with

4 to 8 ohm speakers. Size 14%"x 10" appr BUILT AND TESTED.

£32.90 Plus £3.80 p&p

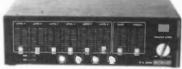
TV SOUND TUNER KIT

as featured in E.T.I. December '81 issue. Kit of parts including PCB, UHF tuner, I.C.'s, all components excluding case, and selector switch. £11.45 + £1.50 p&p.



Transformer £1.50 + £1.50 p&p (p&p free on trans former if ordered with kit). Ready built LP1183 Module for simulated stereo operation £1.95 + 75p p&p.

MONO MIXER AMPLIFIERS



50 WATT Six individually mixed inputs for two pick ups (Cer. or Mag.), two moving coil microphones and two auxiliary for tape, tuner, organs, etc. Eight slider controls - six for level and two for master bass and treble, four extra treble controls for mic and aux inputs, Size: 13%"x6%"x3%"app., Power butput 50 watts R.M.S. (continuous) for use with 4 to 8 ohm speakers, Attractive

black vinyl case with matching fascia ahd knobs. Ready to use

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ive vertical sliber controls, master volume, tape level, mic vel, deck level, PLUS INTERDECK FADER for perfect graduated change from record deck No. 1 to No. 2, or vice versa. Pre fade level controls (PFL) lets YOU hear the next disc before fading it in.

Note: Goods despatched to UK postal addresses only.

For further information send for instructions 20p

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PRODUCTS

ZX PRINTER

Sinclair Research has introduced a new Printer to complement its existing ZX range of personal computers and software. The ZX Printer is only available direct from Sinclair, mail order, for the sum of £49.95 including VAT.

Designed by Sinclair exclusively for use with the ZX81

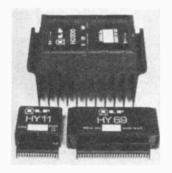
The second secon

POWER MODULES

One of the leading British designers and manufacturers of encapsulated audio amplifiers, pre-amplifiers and power supplies, ILP Electronics of Canterbury, is in the process of launching a major range of new modular products for home hi-fi and disco constructors. The new audio modules, like the existing range, are all totally compatible with each other and can be combined to create a "tailored" system.

The HD Power Amp modules are a new range of heavy duty Bipolar power amplifiers, specially designed to withstand the heavy usage and potential misuse of disco and guitar amplifier work. Available in a choice of three outputs; 60W, 120W or 240W per channel (each with or without heatsink), the prices range from just under £20 plus VAT to approximately £39 plus VAT.

The HY7 mono mixer is an encapsulated unit capable of mixing up to eight signals into one. The HY7 cost £5·15 plus VAT.



Other new mixer modules are the HY11 mono mixer, which mixes five signals into one and has provision for bass and treble controls, and the HY69 mono preamplifier with two input channels for magnetic cartridge or microphone with mixing volume, treble and bass control facilities. Price £10.45 plus VAT.

ILP Electronics Ltd, Dept EE, Graham Bell House, Roper Close, Canterbury, Kent CT2 7EP.

DIGITAL MULTIMETERS



Two new liquid crystal display (LCD) multimeters just introduced onto the market by Lascar Electronics are claimed to be nearly half the price of competing instruments.

The six function, including a diode check facility, instruments have 21 ranges

and can resolve voltages to lmV, current to $1\mu A$ and resistance to 1 ohm. Autopolarity and auto-zero are standard. The 4mm input terminals are protected against overloads and transients.

Housed in identical moulded cases, the DP200 model claims a 0.5 per cent accuracy and is available at £27.95 plus VAT, while the DP2010 model costs £23.95 plus VAT and claims an accuracy of 1 per cent.

Both types feature a 0.5in LCD readout with a 200 hour battery life and an indication on the display when the battery needs replacing.

Lascar Electronics Ltd., Dept EE, Unit 1, Thomasin Road, Burnt Mills, Basildon, Essex SS13 1LH. personal computer, and the ZX80 with retrofit 8K ROM, the new printer features full alphanumerics and high resolution graphics. A special feature is a Copy command, which prints out exactly what is on the TV screen without further instructions.

Sinclair Research Ltd, Dept EE, 6 King's Parade, Cambridge CB2

CALCULATOR WITH PUNCH

A novel design pocket calculator, type BG15, combining the extra functions of a quartz clock with alarm and a realistic electronic boxing game is the latest offering from Casio and available through Tempus.

When switched to the boxing game, the display shows a couple of pugilists in action, throwing and blocking punches to the head and body, swaying to avoid blows, reeling when struck, retreating to the ropes, and so on in imitation of a real bout.

One boxer is under control of the player, the other a programmed hitter and defender. As the player's skill progresses, successive opponents become more robust and competent, as "weight" divisions and handicaps increase.

The calculator automatically keeps score of rounds, points and verdicts.

The Casio BG15 is available from Tempus and costs

Tempus, Dept EE, 164/167 East Road, Cambridge CB1 1DB.



CHARGE IT

Appearing in the shops now, under the brand name of X-Cell, is a new rechargeable battery designed to replace the existing, fairly expensive 6V, type 996, lantern battery. Developed by NiTech Ltd.,

Developed by NiTech Ltd., a British Company, the battery features a "built-in" charging unit, which means that you need no other adaptor or recharging device; you simply plug into the nearest power supply with the leads provided. The units can be directly recharged from almost any mains supply worldwide, or from vehicle, boat or aircraft batteries.

The manufacturers claim that tests have proved that the X-Cell will give a minimum of 300 and as many as 3,000 charge/discharge cycles at a cost of less than 1/10 of a penny per charge. It is claimed that they cannot be overcharged and connection polarity makes no difference.

Two versions are available, the X-Cell Plus with continuous operation times between charges of 3^{1}_{2} hours

and a retail price of £19.95, and the X-Cell Regular, a light duty household version, at 112 hours and a retail price of £14.95. The X-Cell is guaranteed for three years.

NiTech Ltd, Dept EE, 4 Castle Street, Hastings, East Sussex TN34 3DY.



CIRCUIT

MICROPROCESSOR-CONTROLLED MUSIC BOXES

This is a modification that can be applied to the *Microchime* (February 1979) and the *Micro Music Box* (February 1980).

The TM1000N, MP0027A musical integrated circuit has two switching functions: one to select the tune and the other to select the bank of tunes to be played. Instead of using a three-

position switch for the former I designed a push button unit which gives the project a little more style!

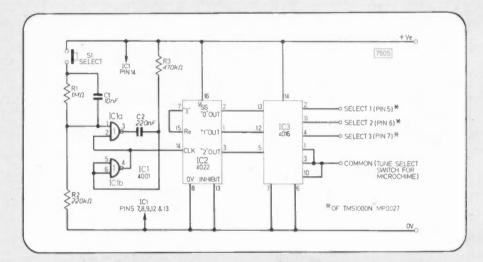
The operation of the circuit can be considered as follows:

ICl is used to "clean up" the pulses made available from S1 to feed IC2. IC2 is a decade counter wired to count from 0 to 2 giving three outputs which are used to control IC3. The bilateral switches in this integrated circuit are operated by these outputs to cause either select 1, 2 or 3 lines to be activated. Therefore, together with the common line this part of the circuit functions as a three-way switch.

The unit can be fitted into the existing case taking its power from one or both batteries.

D. Butler,

Lexden, Colchester, Essex.



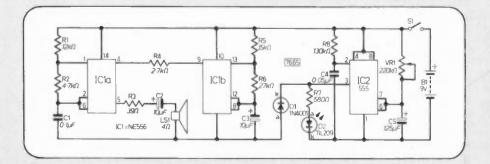
DARKROOM TIMER

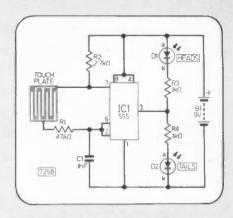
This circuit was designed around two timer i.c.s. The 555 is wired as a monostable and the 556 is wired as two astables. The timing period is determined by adjusting VR1. The 556 forms the audio alarm section of the timer, and the 555 does all the timing.

On closing S1, the output (pin 3) of IC2 goes high, thus preventing the 556 from turning on. The l.e.d. will be on, indicating that timing has begun.

When the timing period is over, the l.e.d. will go off as pin 3 (IC2) goes low. At the same time that the l.e.d. goes off, ICla, b switch on and the alarm sounds. The alarm will be short bleeps caused by the switching on and off high frequency oscillator ICla. This is achieved by connecting the output of IClb to the reset pin (4) of ICla. IClb is oscillating at about 1Hz. VR1 is 220 kilohm linear.

R. T. Nkambule, Mbabane, Swaziland

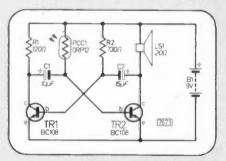




COIN FLIP

This is a cheap and novel way of producing an electronic "heads and tails". Not only does it use only one i.c., it is touch sensitive as well. The resistance of one's finger forms the timing resistance and, in fact, if only a very light touch is applied the l.e.d.s can be seen to flip between states. Taking the finger of the touch plate (which can be Veroboard with alternate stips joined) will end the "throw". One l.e.d. has to be labelled "Heads" and the other "Tails"

A. J. Boulton, Stoney Stanton, Leics.



ELECTRONIC CHICKENS

Now I know there are kits on the market for electronic chickens, but this one is different! The circuit is a simple multivibrator using common components, is simple to build, and is unique! It is light controlled. So what, you say. The ORP12 used has black tape over the window, the circuit is that sensitive!

How to use it: Put it in a box with the PCC1 on the outside of the lid, and black tape on the window. Carry the box about in the room to obtain random clucking as it goes in and out of the shadows. It also does the long baaachackle in brightish daylight, near a window. It clucks slowly in the dark, faster in medium dark.

Brian S. Craigie, Edinburgh, Scotland

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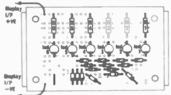
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The suspense and excitement of the casino in your own home. Just press the button, the circle of lights go round and there is the sound of the roulette wheel as well, both gradually slowing down to reveal the winning number.

No. 12 EGG TIMER

How do you like your eggs done, hard or soft, just set the timer and it will sound when the egg is done to your liking, Long battery life because it switches itself off automatically. So get cracking now!

Want to get started on building exciting projects, but don't know how? Now using **EXPERIMENTOR BREADBOARDS and** following the instructions in our FREE 'Electronics By Numbers' leaflets, ANYBODY can build electronic projects. For example, take one of our earlier projects, a L.E.D. Bar Graph;



One EXP 300 or EXP 350 You will need. breadboard 15 silicon diodes 6 resistors 6 Light Emitting Diodes Just look at the diagram, Select R1, plug it into the lettered and numbered holes on the EXPERIMENTOR BREADBOARD, do the same with all the other components, connect to the battery, and your project's finished. All you have to do is follow the large, clear layouts on the 'Electronics by Numbers' leaflets, and ANYBODY can build a perfect working project.

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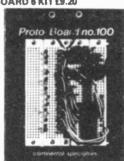
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1980 saw a genuine breakthrough – the Sinclair ZX80, world's first complete personal computer for under £100. Not surprisingly, over 50,000 were sold.

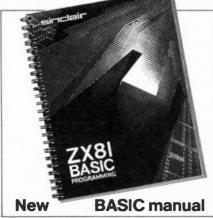
In March 1981, the Sinclair lead increased dramatically. For just £69.95 the Sinclair ZX81 offers even more advanced facilities at an even lower price. Initially, even we were surprised by the demand – over 50,000 in the first 3 months!

Today, the Sinclair ZX81 is the heart of a computer system. You can add 16-times more memory with the ZX RAM pack. The ZX Printer offers an unbeatable combination of performance and price. And the ZX Software library is growing every day.

Lower price: higher capability
With the ZX81, it's still very simple to
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It uses the same micro-processor, but incorporates a new, more powerful 8K BASIC ROM – the 'trained intelligence' of the computer. This chip works in decimals, handles logs and trig, allows you to plot graphs, and builds up animated displays.

And the ZX81 incorporates other operation refinements – the facility to load and save named programs on cassette, for example, and to drive the new ZX Printer.



Every ZX81 comes with a comprehensive, specially-written manual – a complete course in BASIC programming, from first principles to complex programs.

Kit: £49.95

Higher specification, lower price - how's it done?

Quite simply, by design. The ZX80 reduced the chips in a working computer from 40 or so, to 21. The ZX81 reduces the 21 to 4!

The secret lies in a totally new master chip. Designed by Sinclair and custom-built in Britain, this unique chip replaces 18 chips from the ZX80!

New, improved specification

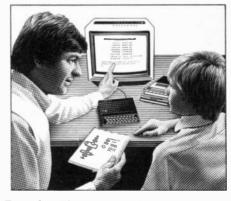
- Z80A micro-processor new faster version of the famous Z80 chip, widely recognised as the best ever made.
- Unique 'one-touch' key word entry: the ZX81 eliminates a great deal of tiresome typing. Key words (RUN, LIST, PRINT, etc.) have their own single-key entry.
- Unique syntax-check and report codes identify programming errors immediately.
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- Graph-drawing and animateddisplay facilities.
- Multi-dimensional string and numerical arrays.
- Up to 26 FOR/NEXT loops.
- Randomise function useful for games as well as serious applications.
- Cassette LOAD and SAVE with named programs.
- 1K-byte RAM expandable to 16K bytes with Sinclair RAM pack.
- Able to drive the new Sinclair printer.
- Advanced 4-chip design: microprocessor, ROM, RAM, plus master chip – unique, custom-built chip replacing 18 ZX80 chips.

Built: £69.95

Kit or built - it's up to you!

You'll be surprised how easy the ZX81 kit is to build: just four chips to assemble (plus, of course the other discrete components) – a few hours' work with a fine-tipped soldering iron. And you may already have a suitable mains adaptor – 600 mA at 9 V DC nominal unregulated (supplied with built version).

Kit and built versions come complete with all leads to connect to your TV (colour or black and white) and cassette recorder.



Everyday Electronics, February 1982



16K-byte RAM pack for massive add-on memory.

Designed as a complete module to fit your Sinclair ZX80 or ZX81, the RAM pack simply plugs into the existing expansion port at the rear of the computer to multiply your data/program storage by 16!

Use it for long and complex programs or as a personal database. Yet it costs as little as half the price of competitive additional memory.

With the RAM pack, you can also run some of the more sophisticated ZX Software - the Business & Household management systems for example.

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How to order your ZX81

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And of course you can print out your results for permanent records or sending to a friend.

Printing speed is 50 characters per second, with 32 characters per line and 9 lines per vertical inch.

The ZX Printer connects to the rear of your computer - using a stackable connector so you can plug in a RAM pack as well. A roll of paper (65 ft long x 4 in wide) is supplied, along with full instructions.

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	Mains Adaptor(s) (600 mA at 9 V DC nominal unregulated).	10	8.95	
	16K-BYTE RAM pack.	18	49.95	
	Sinclair ZX Printer.	27	49.95	
	8K BASIC ROM to fit ZX80.	17	19.95	
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High-Pass Filter .										
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Ideal for use with ZX80/81. Has 50 keys and many other parts for your spares box. Probably cost in excess of £100. In very good used condition $= (13.50 \pm 2.00 \, \text{post})$. Diagram showing how to connect to ZX80/81 $= £2.00 \, \text{extra}$.

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Size approx. 4'x 2' x 2'6" high. These were made for hard work, the top being formica covered. Suitable for housing instruments or for use as office desks. Beautifully made, these cost over £100 each, our price only £11.50 each, however, you must arrange to collect.

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Mains operated - ex. Computer. 5" Woods extractor 5" Woods extractor £5.75 Post £1.00. 6" Woods extractor £6.90 Post £1.25 6" Plannair extractor £7.50 Post £1.00 4" x 4" Muffin 115w, £4.50 Post 50p, 4" x 4" Muffin 230v,

" Muffin 230v. £5.75 Post 50p.



INTERRUPTED BEAM

This kit enables you to make a switch that will trig steady beam of infra red or ordinary light is broke. Main com ponents - relay, photo transistor, resisto Circult diagram but no case. Price £2.30 tors and caps, etc.

INSTRUMENT BOX WITH KEY

rey strongly made (ply-wood sides with hard board top and pttom). This is black grained effect, vinyl covered, very pleasi poperance, Internal dimensions 12% 1000, 4% wide, 6" deel keal for carrying your multi range meter and small tools and feecing them in a safe place, £2.30. Post paid if ordered with ther goods, otherwise £1.00.

ROPE LIGHT

4 sets of coloured lamps in translucent plastic tube arranged to give the appearance of a running or travelling light. With variable speed control box, ideal for disco or shop window display. Complete, made up, ready to plug into mains. £36.00 + £2 post.

COMPUTER KEY SWITCHES (make your own keyboard)



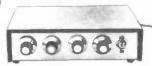
OUR CAR STARTER AND CHARGER KIT has no doubt saved OUR CAR STARTER AND CHARGER KIT has no doubt saved many motorists from embarrassment in an emergency you can start car off mains or bring your battery up to full charge in a couple of hours. The kit comprises: 250w mains transformer, two 10 amp bridge rectifiers, start/charge switch and full instructions. You can assemble this in the evening, box it up or leave it on the shelf in the garage, whichever suits you best. Price £11.50 + £2.50 post.

GPO HIGH GAIN AMP/SIGNAL TRACER. In case measuring only 5% in x 3% in x 1% in is an extremely high gain (70dB) solid state amplifier designed for use as a signal tracer on GPO cobles, etc. With a radio it functions very well as a signal tracer. By connecting a simple coil to the input socket a useful mains cable tracer can be made. Runs on standard 4% battery and has input, output sockets and on-off volume control, mounted flush on the top. Many other uses include general purpose amp, cueing amp, etc. An absolute bargain at only £1.85. Suitable 80ohm earpiece 69p.

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3 CHANNEL SOUND TO LIGHT KIT

Complete kit of parts for a three-channel sound to light unit contro ing over 2000 watts of lighting. Use this



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make a wonderful girt for almost anyone. In easy to assemble modular form this should sell at about £30 — but due to a special bulk buy and as an incentive for you to buy this month we offer the system complete at only £16.75 including VAT and post. FREE GIFT — buy this month and you will receive a pair of Goodman's eliptical B"x5" speakers to match this amplifier.

THIS MONTH'S SNIP POCKET AUDIO COMPONENT TESTER



With it you can quickly test diodes, rectifiers, transistors, capacitors, check wiring and p.c. boards for open circuits, find the anote and cathode of a diode or rectifier and whether a transistor is PNP or NPN, which are the base collector and emitter connections. Concensers, if bad, give a continuous signal, but if good, give intermittent signals of varying length depending on their value. The test current is very low (2uA) and the voltage only 1.4v, so it is also possible to check MOS devices, as well as sensitive transistors without fear of damaging them. The unit is supplied complete with internal battery, which should last many months.

Price £3.45p

THERMOSTAT ASSORTMENT

10 different thermostats, 7 bi-metal types and 3 liquid types. There are the current stats which will open the switch to protect There are the current stats which will open the switch to protect devices against overload, short circuits, sett., or when fitted say in front of the element of a blow heater, the heat would trip the stat if the blower fuses; appliance stats, one for high temperatures, others adjustable over a range of temperatures which could include $0-100^\circ C$. There is also a thermostatic pod which can be immersed, an oven stat, a calibrated boiler stat, finally an ice stat which, fitted to our waterproof heater element, up in the loft could protect your pipes from freezing. Separately, these thermostats could cost around £15.00 - however, you can have the parcel for £2.50.

6 WAVEBAND SHORTWAVE RADIO KIT

D WAVEDAND SHORTWAVE RADIO KIT Bandspread covering 13.6 to 32 metres. Based on circuit which appeared in a recent issue of Radio Constructor. Complete kit in-cludes case materials, six transistors and diodes, condensers, resist-ors, inductors, switches, etc. Nothing else to buy if you have an amplifier to connect it to or a pair of high resistance headphones, Price £1136.

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n-note bleeper, push latching switch, plastic case and battery ctor. Will scare away any villain and bring help. £2.50 com-

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should not miss!
When ordering please add £2.50 post and £1.27 VAT.

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Contains labelled connection block, latching relay, test switch and removable key control switch. Simplifies the whole installation, all you have to do is to take wires to pressure pads and to alarm bell. Price £7.95, with complete diagram.

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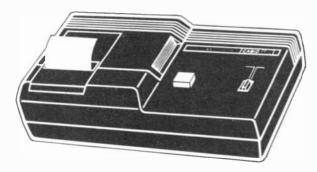
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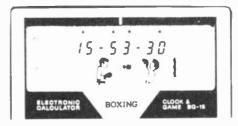
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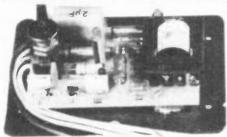
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Pic of Kraftwerk dummy by Anton Corbijn.

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ignition gives all the well known advantages of the best capacitive discharge systems.

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IMPROVED ECONOMY - no loss of ignition performance between services.

FIRES FOULED SPARK PLUGS no other system can better the capacitive discharge system's ability to fire fouled plugs

ACCURATE TIMING prevents contact wear and arcing by reducing load to a few volts and a fraction of an amp.

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SUPER POWER SPARK -- 3½ times the energy of ordinary capacitive systems - 31/2 times the power of inductive systems.

OPTIMUM SPARK DURATION 3 times the duration of ordinary capacitive systems - essential for use on modern cars with weak fuel

- full spark power even with low

CORRECT SPARK POLARITY unlike most ordinary C.D. systems the correct output polarity is maintained to avoid increased stress on the H.T. system and operate all voltage triggered tachometers.

L.E.D. STATIC TIMING LIGHT for accurate setting of the engine's most important adjustment.

LOW RADIO INTERFERENCE fully suppressed supply and absence of inverter 'spikes' on the output reduces interference to a minimal level.

DESIGNED IN RELIABILITY an Inherently more reliable circuit combined with top quality components - plus the 'ultimate insurance' of a changeover switch to revert Instantly back to standard ignition.

IN KIT FORM

it provides a top performance electronic ignition system at less than half the price of competing readybuilt systems. The kit includes everything needed, even a length of solder and a tiny tube of heatsink compound. Detailed easy-to-follow instructions, complete with circult diagram, are provided - all you need is a small soldering iron and a few basic tools.

AS REVIEWED IN ELECTRONICS TODAY MAGAZINE

JUNE '81 ISSUE

Quote "the kit is very impressive"

"well written instructions and a good performance"

"Exellent value for money. Highly recommended"

FITS ALL VEHICLES, 6 or 12 volt, with or without ballast NEGATIVE EARTH ONLY

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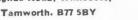
STANDARD CAR KIT £14.85 PLUS £1 TWIN OUTPUT KIT £22.94

For MOTOR CYCLES and CARS with twin ignition systems Prices include V.A.T.



ELECTRONIZE DESIGN

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Phone 0827-281000





DIMENSIONS

Length 12.5 cm Width 8.9 cm Height 4.3 cm Lead length 100.0 cm

TECHNICAL DETAILS

The basic function of a spark ignition system is often lost among claims for longer 'burn times' and other marketing fantasies. It is only necessary to consider that, even in a small engine, the burning fuel releases over 5000 times the energy of the spark, to realise that the spark is only a trigger for the combustion. Once the fuel is ignited the spark is insignificant and has no effect on the rate of combustion. The essential function of the spark is to start that combustion as quickly as possible and that requires

The traditional capacitive discharge system has this high power spark but, due to it's very short spark duration and consequential low spark energy, is incompatible with the weak air/fuel mixtures used in modern cars. Because of this most manufacturers have abandoned capacitive discharge in favour of the cheaper inductive system with it's low power but very long duration spark which guarantees that sooner or later the fuel will ignite. However, a spark lasting 2000µS at 2000 rev/min. spans 24 degrees and 'later' could mean the actual fuel ignition point is retarded by this amount.

The solution is a very high power, medium duration, spark generated by the TOTAL ENERGY DISCHARGE system. This gives ignition of the weakest mixtures with the minimum of timing delay and variation for a smooth efficient engine.

SUPER POWER DISCHARGE CIRCUIT A brand new technique prevents energy being reflected back to the storage capacitor, giving 31/2 times the spark energy and 3 times the spark duration of ordinary C.D. systems, generating a spark powerful enough to cause rapid ignition of even the weakest fuel mixtures without the ignition delay associated with lower power 'long burn' inductive systems.

HIGH EFFICIENCY INVERTER A high power, regulated inverter provides a 370 volt energy source - powerful enough to store twice the energy of other designs and regulated to provide sufficient output even with a battery down to 4 volts.

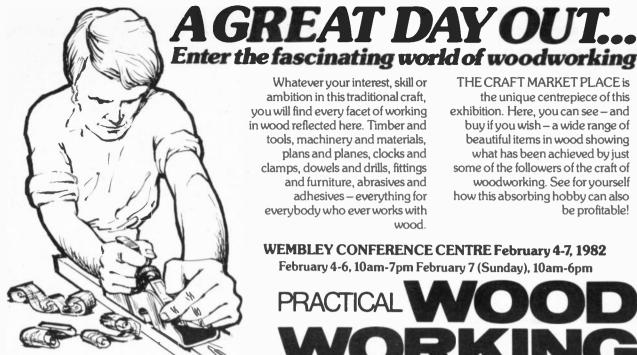
PRECISION SPARK TIMING CIRCUIT This circuit removes all unwanted signals caused by contact volt drop, contact shuffle, contact bounce, and external transients which, in many designs, can cause timing errors or damaging un-timed sparks. Only at the correct and precise contact opening is a spark produced. Contact wear is almost eliminated by reducing the contact breaker current to a low level - just sufficient to keep the contacts clean.

TYPICAL SPECIFICATION

	TOTAL ENERGY DISCHARGE	ORDINARY CAPACITIVE DISCHARGE
SPARK POWER (PEAK)	140 W	90 W
SPARK ENERGY (STORED ENERGY)	36 mJ 135 mJ	10 mJ 65 mJ
SPARK DURATION	500 µS	160 µS
OUTPUT VOLTAGE (LOAD 50pF EQUIVALENT TO CLEAN PLUGS)	38 KV	26 KV
OUTPUT VOLTAGE (LOAD 50pF + 500 KΩ EQUIVALENT TO DIRTY PLUGS)	26 KV	17 KV
VOLTAGE RISE TIME TO 20 KV (Load 50pF)	25 µS	30 µS

TOTAL ENERGY DISCHARGE should not be confused with low power inductive systems or hybrid so called reactive systems.

Goods normally despatched within 7 days



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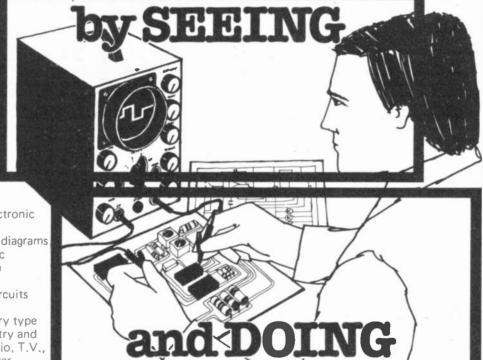
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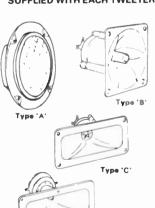
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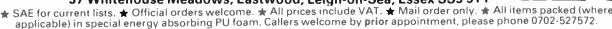
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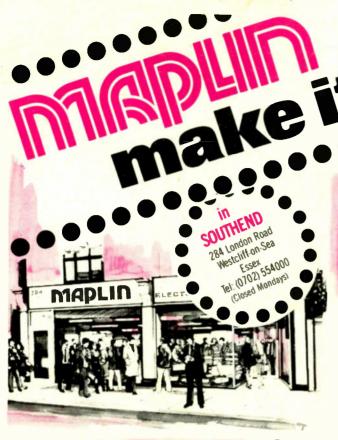




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