

PROJECTS—HI-FI—COMPUTERS

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Radio-Electronics

THE MAGAZINE FOR NEW IDEAS IN ELECTRONICS

part 2: phone accessory

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carbon, alkaline, mercury

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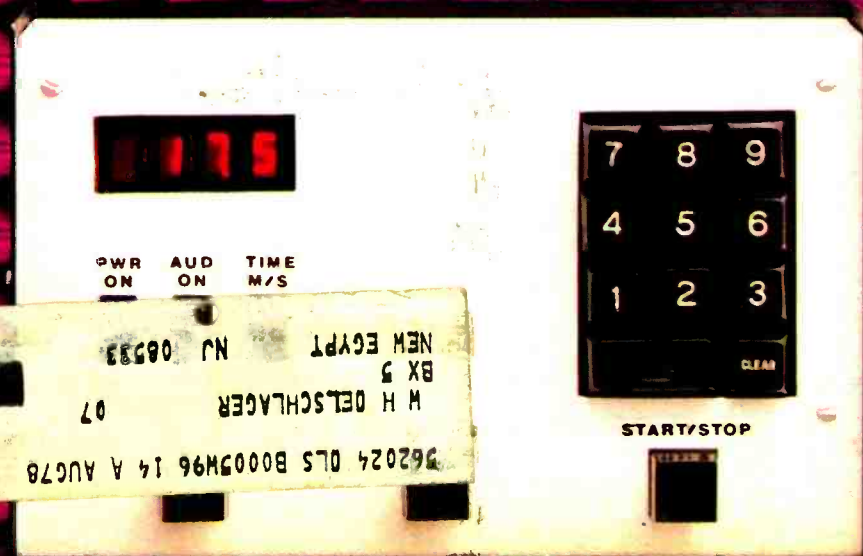
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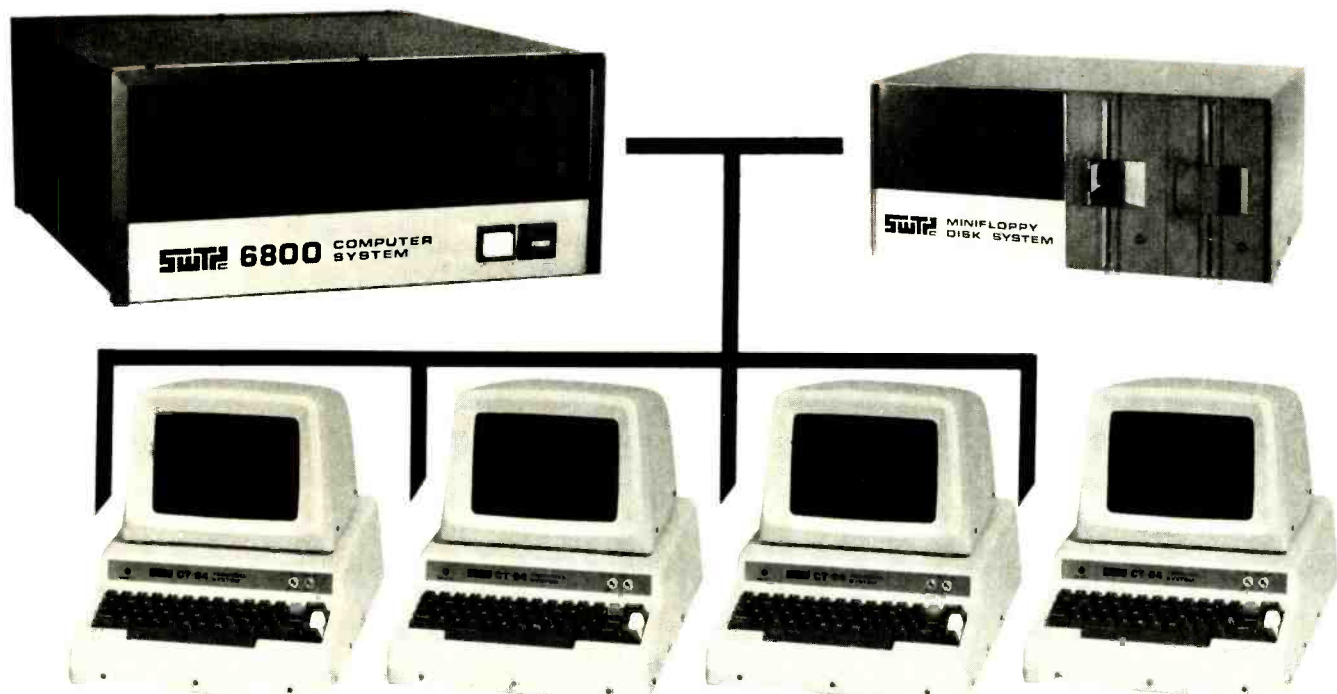
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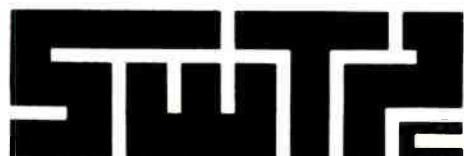
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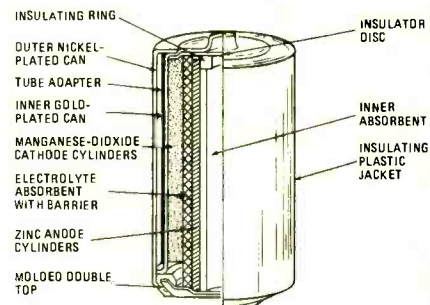
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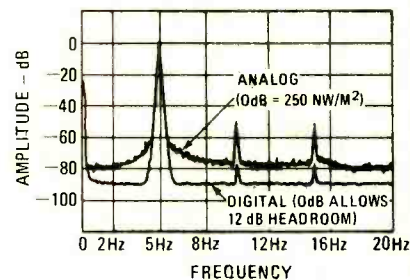
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ON THE COVER

Unique darkroom timer counts down in either seconds or minutes and seconds. Precise digital readout and audible pulses too. See page 33.



TYPICAL ZINC-CARBON CELL is one of the three basic disposable cells described in this story. Turn to page 44.



SPECTRAL PURITY of analog recorder vs digital recorder. For complete details on recording sound as a digital signal, turn to page 57.

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looking ahead

TV-audio wedding: For years, audiophiles have been irked by the quality of television sound, and for years the TV networks have been blaming the television set manufacturers, while the manufacturers have been blaming the public for being too stingy to pay extra for good sound systems in their sets. And for years, almost everybody has been blaming AT&T, whose TV audio lines have been limited to the same frequency bandwidth as AM—so that while TV sound channels may theoretically be capable of passing frequencies as high as 15 kHz, most network broadcasts have had an upper limit of 5 kHz.

The logjam may finally be breaking. The limited substitution of satellites for land lines in long-distance TV transmission as well as some special musical simulcasts, mainly by the Public Broadcast System, have led to an increased awareness that good TV sound is both possible and desirable. But the biggest boost probably came from the AT&T's recent switchover of its intercity television relay system to a new diplexing system that boosts the sound bandwidth to a full 15 kHz (see **Radio-Electronics**, April, 1978). This widely publicized move, perhaps triggered by the new competition from satellites, began to spur public demand for receiving systems that can take advantage of this better sound. And, as the FCC's current inquiry into the desirability of stereophonic sound for TV progresses, it's expected to stir up more interest in TV sound.

Short of waiting for TV makers to improve their sound systems, there hasn't been much the public could do. Until now. For years, stereo component manufacturers have been experimenting with TV tuner components for audio systems, but haven't moved them to market because of the general inadequacy of the transmitted sound. Now they're beginning to dust them off, geared to the networks' new sound capability. The first to appear probably will be a component from Pioneer, to list at around \$250, which permits the substitution of the home stereo system for the sound circuits in the TV set. Complete details weren't available at presstime.

More than a year ago, Sony introduced in Japan its *Audioscope TV Tuner*, a similar component which also included a tiny black-and-white monitor that could be switched to display either the incoming picture or the audio waveform. It was shown in the U.S., but never actively marketed here, presumably because of the generally poor quality of transmitted audio. Panasonic has a TV sound component waiting for introduction. Now other audio manufacturers can be expected to go forward with long-delayed plans to make TV sound easily available to home audio systems.

GE's home projector: The concept of home projection television got a huge boost when General Electric introduced its highly engineered rear-screen unit. The one-piece furniture-styled console, called "Widescreen 1000," has a picture area of 1,003 square inches, with a diagonal of 45.7 inches—or three times the size of a 25-inch set. The relatively compact cabinet is 50 inches high, 70 inches long and 24 inches deep.

What makes GE's projector different from many others is that it was designed from the ground up as a projection set, that it is backed by one of the major domestic manu-

facturers and that it is the first rear-screen unit with a truly bright picture. The heart of the new set is a specially designed extremely bright 13-inch picture tube with the electron gun and shadow mask made especially for projection. Although GE didn't give any figures, one knowledgeable observer said its brightness was five times that of a conventional direct-view 13-inch tube. It uses a three-element coated plastic lens and two float-glass mirrors to focus and direct the picture to the one-piece acrylic viewing screen with a fresnel lens pattern molded into the rear surface. The optical components, including the screen cabinet, are sealed against dust.

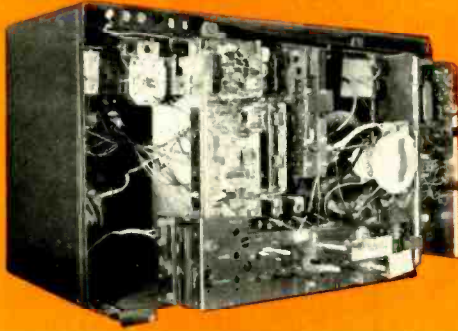
How well does it work? The picture is surprisingly bright—brighter than any other single-tube projection set on the market. The picture is clear and undistorted from corner to corner. The viewing angle is restricted, to perhaps 30 degrees. The set is tuned by digital wireless remote control and contains VIR color-adjusting circuitry. The suggested retail price, which hadn't been determined at presstime, was expected to be around \$2,800. GE estimated industrywide sales of about 125,000 projection TV sets in 1978, 220,000 in 1979 and 500,000 by 1983.

New TV highlights: Microprocessor tuning will be coming to American TV in a big way starting this year. Although the first programmed TV set in this country has been available for some time from Heathkit, the floodgates may now be opening in mass-market television. Toshiba has introduced a 21-inch model with a programmable 12-channel tuner (list price about \$950). The set can be pre-programmed to tune in 16 different shows during the course of a day—turning on and off and changing channels. Up to eight selections can be stored permanently and repeated daily, so you don't have to miss your favorite soap opera just because you forgot to turn on the set. A similar set is on its way from Hitachi. Other microprocessor-based tuners are expected this year. Not all of them are programmable—some merely use the microprocessor circuitry for extremely accurate tuning.

Radio's future: British Broadcasting Corporation is looking into several new radio broadcast services, which were recently outlined to American broadcasters by its engineering director, James Redmond. Among them: (1) BBC has proposed to the government an information network of low-power stations, all operating on the same frequency and using time-division multiplexing to prevent mutual interference. These stations would give motorists detailed information about traffic in the area in which they are driving. A special low-cost fixed-tuned receiver in the car is arranged so it interrupts the regular car radio when there is a message for the specific area. (2) The addition of inaudible signaling to radio broadcasts to provide a station identification or other information on a liquid-crystal display built into radio receivers. (3) Quadriphonic sound on FM, stereo on AM. (4) Data and facsimile transmission multiplexed into audio broadcasting.

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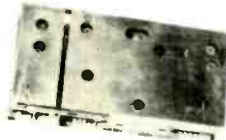
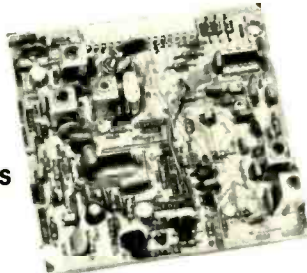
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CB'ers are heroes in storms of '78

The first two months of 1978 provided blizzards, ice storms and severe weather conditions to much of the U.S.; it was also a time when CB operators performed many acts of heroism and neighborliness by helping those stranded by snow, providing emergency fuel deliveries and aiding local authorities in coordinating rescue and relief operations.

"For many of those stranded in the recent blizzard," said John Sodolski, vice president of EIA and head of its communications division, "CB radio was the only communications device available for emergency rescue."

It is certain that without the volunteer efforts of CB'ers, snowmobile owners and four-wheel-drive organizations, rescue efforts would have taken longer and many lives would have been lost. CB radio provided the essential communications link during rescue efforts, since in many cases police vehicles were out of commission due to the storms.

Among the thousands of reports of CB'ers' heroism, some stand out:

—In Massachusetts, CB'ers directed oil deliveries into stricken areas, facilitated emergency transportation to and from hospitals, delivered food and rescued stranded motorists and people whose homes had lost heat.

—In Ohio, when the power failed, CB operators helped bring over 2000 people through freezing temperatures to shelter in churches and other homes.

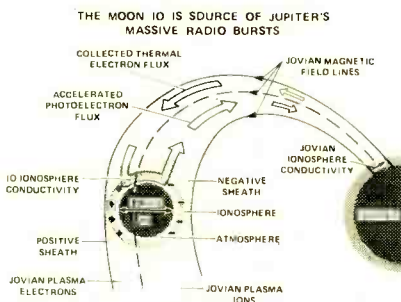
—In West Virginia, React Club members routinely delivered groceries and performed other neighborly acts for elderly people trapped by the snow in remote areas.

—On the West Coast, during recent rainstorms and floods, CB'ers took an active part in aiding the local Red Cross with emergency communications at two flood shelters housing 25 homeless families. Manned vehicles with CB radios shuttled flood victims and volunteers to Red Cross headquarters and the shelters. And once the immediate emergency was over, several members of the CB emergency team volunteered to stay on to help the victims and assess flood damage.

Planet Jupiter emits surf-like radio signals

It has been discovered that radio signal bursts emitted from the planet Jupiter and one of its moons, IO, resemble the rushing sound of waves on the shore—a phenomenon well-known for its soothing qualities. Shield Products, Inc., of Cleveland has used highly sophisticated receiving equipment to record these signals (originating from a distance of over 390 million miles)

on magnetic tape, and have dubbed the project *Jupiter Tranquil*. These recordings are available to the general public.



PLANET JUPITER'S MOON IO is source of massive radio signal bursts recorded for *Jupiter Tranquil*. Diagram shows signal path.

The signals are produced when satellite moon IO passes through Jupiter's intensely charged magnetic field, at which point the moon's ionosphere becomes both positively and negatively charged. When IO is in the optimum orbital position, an electrical current of more than 100 megawatts flows along Jupiter's magnetic field down to the planet's own ionosphere and back to IO. This tremendous flow of energy is responsible for the signal bursts that so resemble the sounds of the sea. For further information, write to Shield Products, Inc., Space Technology Div., 1104 Prospect Ave., Cleveland, OH 44155.

Telesat Canada Anik C spacecraft to help link south Canada cities

Telesat Canada has awarded a contract to Hughes Aircraft, El Segundo, CA, to build three new communications satellites, the Anik C series (*anik* is Eskimo for brother). The satellites will open up interurban telecommunications in the vast, 3000-mile area stretching from coast to coast in the southern half of Canada. This area also lies within 1000 miles of the United States border.

The spacecraft, scheduled to be sent into orbit in 1981 via NASA's Space Shuttle, will operate over the Equator between 105 and 120 degrees west longitude, south of Alberta, Canada. It will operate in the super-high-frequency range of 12 to 14 GHz-per-second. The use of such high frequencies allows the antenna to produce the narrow beam necessary for such concentrated radiated power. This narrow-beam capability will allow earth stations receiving the frequencies to be located in high-density areas without interfering with terrestrial or lower-frequency space signals.

Other Anik C features include 16 communications channels to supply audio, video and data telecommunications services;



ANIK C SPACECRAFT, being built by Hughes Aircraft for Telesat Canada, will link southern Canada's cities from coast to coast (shown in artist's conceptual above). The satellite will provide telecommunications services in super-high-frequency range of 12 to 14 GHz.

and polarization diversity that permits using the same frequency twice. Its solar array consists of two concentric cylindrical panels; when the satellite is in orbit, the outer solar panel extends downward exposing both panels to the sun and boosting power generation to more than 900 watts.

U.S. investigative agency reports dangers of microwave radiation

The General Accounting Office, which is the investigative arm of the U.S. Congress, has released a report that warns the public may be exposed to dangerous microwave-radiation levels. This represents the first time any Government agency has recognized such hazards. The report further states that microwave transmission facilities are being built at a rate of 15% a year, and there is as yet no Government agency to monitor them or set safety standards.

Prior Environmental Agency Protection lab tests are described in the report as having concluded that microwave radiation may "affect the immune system" in laboratory animals as well as produce a "trend toward lowered behavioral performance."

Studies conducted in the Soviet Union and Eastern European countries reported that individuals exposed to radiation levels much greater than the U.S. Government's standard of 10/1000ths of a watt complained of a host of symptoms, including

continued on page 12

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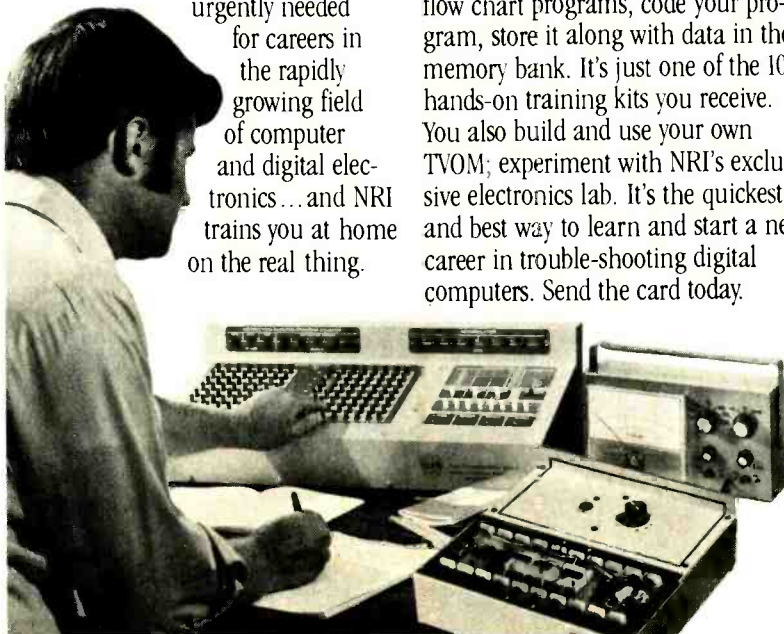
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William Mumford, a retired Bell Labs scientist who helped devise the U.S. standard for microwave radiation, asserted that this standard was safe under normal environmental conditions, and that of the 4000 medical cases of alleged microwave radiation reported since the 1950's not one fatality had been reported.

World's most powerful vacuum tube switch used in nuclear-fusion tests

A 20-man engineering team from RCA's Department of Electro-Optics and Devices (Lancaster, PA) has designed and developed what they term "the world's most powerful vacuum switch tube." The RCA tube switches 25 megawatts (MW) or enough energy to turn on 300,000 TV sets simultaneously. It will be one of 12 similar tubes that are to be installed in Princeton University's Plasma Physics Lab for use in a fusion-test reactor. This reactor will be used to determine the feasibility of generating electrical power via nuclear fusion.



320-POUND VACUUM TUBE designed by RCA switches 25 MW—enough power to turn on 300,000 TV sets—and will be used in Princeton University nuclear fusion tests. Computers were used in designing the tube, constructed by engineers of RCA's department of Electro-Optics and Devices in Lancaster, PA.

The reactor, which is destined to become operational in 1981, will use the vacuum switch tubes to regulate power systems that will provide the major heating of the plasma used in the fusion reaction. Dr. Ralph E. Simon, division vice president,

explained that the tube's function "is to switch and control ion beams as they are injected into the fusion test reactor . . . The tube will supply 25 million watts of energy to the ion beams in pulses varying in duration from several millionths of a second to several seconds."

Made of copper, stainless steel, tungsten and ceramic, the RCA tube weighs 320 pounds and measures 42 inches long and 22 inches in diameter. Tube design was facilitated by the use of computers that were able to analyze the tube's electrical characteristics and help determine the shape of the tube's anode. The power dissipated by the anode is 2 MW. Sixteen kilowatts is required just to heat the cathode. The tube is completely water cooled with a flow rate of 250 gallons-per-minute through the tube's anode section.

New FCC legislation cracks down on unlicensed CB operators

In March, 1978, President Carter signed legislation that at long last empowers the Federal Communications Commission to enforce its rules against *unlicensed* as well as licensed CB violators.

Until recently, a Federal court order was necessary to act against unlicensed violators of the Commission's CB regulations; these regulations were really effective only against licensed violators. James C. McKinney, the FCC deputy chief of field operations, stated, "We can now make it economically unfeasible to be unlicensed. I suspect fines . . . will be greater than for licensed violators."

The new law increases the maximum fine for violations from \$500 to \$5000, and extends the time period from 90 days to a year during which the FCC can issue a citation for a violation.

Present-day color TV set prices lower despite inflation

According to Roy Pollack, vice president of RCA's Consumer Electronics Division, the cost of a color TV set has declined during the past 25 years. In 1954, the average price for such a set was \$1000; today, it is \$575, representing a 43% drop. This price drop occurred despite an inflationary period that saw skyrocketing prices on other household items such as washing machines, refrigerators and the like.

Pollack attributed this counter-inflationary price trend to be a reflection of "improved production techniques, new materials, new technologies, bigger markets and severe competition." He added that during the past 25 years, the average price of a new car rose 134% and the median cost of a new one-family home mounted a whopping 290%.

Earth satellite uses infrared band for 24-hour monitoring

The Landsat 3 earth satellite developed



MULTISPECTRAL SCANNER (MSS) can take thermal photos of the earth at night. Here, Hughes Aircraft technician examines instrument's aperture, housing reflective mirror and telescope to record earth-surface images.

by General Electric has had an infrared (IR) channel added to its multispectral scanner (MSS). This IR channel, designed by Hughes Aircraft, will enable round-the-clock photography of terrestrial resources such as water, snow, glaciers, etc. The satellite (scheduled for a West Coast NASA launch) is planned to act in tandem with its sister satellite, Landsat 2, already in orbit.

NASA spokesman asserted that the earth resolution of the IR band of the MSS will allow a 24-hour monitoring of water temperature variations, water pollution, ocean currents and geothermal sources and enable scientists to determine more precisely the ecological impact on earth resources of such large heat-producing facilities as nuclear power plants and industries as well as expanding urban areas.

The vehicle's MSS system traps the earth's radiation with a flip-flop mirror that moves from side to side during the orbit. A precision telescope inside the scanner channels the radiation to detectors in five spectral bands, including the IR band. The detectors then convert the light into voltages, which are, in turn, translated into numerical values. This data is stored on tape recorders for later transmission to a ground station, or in real time at a 15 million bits-per-second rate.

The satellite will circle the globe 14 times daily as the scanner photographs strips of land below. Each time the satellite passes over the Equator it will do so at a distance about 1800 miles west of the last trip.

continued on page 15

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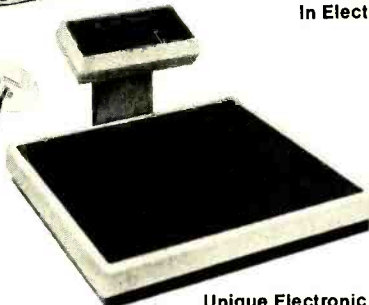
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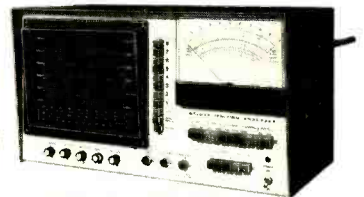


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Contest Of Contests

Our readers deserve a really great contest—one that will test their skills to the limit—one that will let them show the world just how great they are. So, we here in the editorial offices of **Radio-Electronics** have given much thought to running a really terrific contest for our readers.

We first thought of a design contest—like build the best circuit. The problem here is how can anyone pick a winner. Is a mini flat-screen color TV better than a wristwatch computer or a portable telephone?

So, we need your help. We need some reader to dream up the very best contest that has ever been held. We need a contest that will put before the electronics community a real challenge and yet permit us to select a definitive clear-cut winner. Like who built the frequency counter that could read to the highest frequency; or who built a CB radio with the least number of parts; or who built the fastest electronic car.

In addition, we need a handicap factor since the cost of the item built must also be considered. And this handicap must increase as the cost goes up. As we all know, if we spend enough dollars we can reach the moon. But how do you get to England for \$2.98?

Send us your ideas. We'll reward the contributor of the one that we do use with two items—first, a 10-year subscription to **Radio-Electronics**; second, a check for \$250. The editors are the judges and our decision is final. If two similar winning ideas come along, the one with the earliest postmark wins. No prizes for second place, and all entries become the property of Gernsback Publications Inc. Now, start filling our mailbox.



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Geostationary satellites aid federal antimuggling efforts

Recently, General Electric communications experts ran a six-week-long series of communications and position-fixing tests involving the use of two NASA experimental geostationary satellites. The purpose: to demonstrate that the satellites could help the federal government reduce significantly the present rate of contraband drug and illegal alien smuggling occurring across the U.S.-Mexico border. These tests, conducted together with the U.S. Drug Enforcement Administration (DEA) and the Immigration and Naturalization Service (INS), showed that a geostationary satellite, orbiting about 23,000 miles above a fixed area of the earth, can help keep government field agents in contact not only with their colleagues but with their base station, even in very remote locations.

A geostationary satellite stays in a permanent location, its line of sight coinciding with about 43% of the earth's surface. It receives the signal from the agent's mobile radio transmitter, amplifies it, then sends it

back to earth, either to another agent's two-way radio or to a base station. The base station uses a system called tone-code ranging to pinpoint the location of the agent's car to within 600 feet—a distinct advantage when much of the tracking must be done through mountainous and often roadless territory.

The two NASA space vehicles, the ATS-3 (positioned above the Amazon River) and the ATS-1 (over the Equator south of Hawaii) plus a station wagon were used in the tests. The station wagon contained a special antenna and radio equipment consisting of a GE two-way mobile radio, with a special tone-code data responder that reacts automatically when addressed by a coded signal, plus a low-noise preamplifier to increase sensitivity and a power amplifier to boost signal output to 250 watts. General Electric's network of transponders located around the globe were used to determine precisely the location of the two NASA satellites, at which point range measurements were transmitted via satellite to the station wagon, whose automatic re-

sponder returned the signal to the satellite. A computer was then able to use the travel time of the signal to calculate the location of the station wagon in real time.

RCA American satellites beam pay TV programming

A pay TV program supplier, Showtime Entertainment Inc. has started using RCA's American Communications satellites to transmit quality programming to cable TV subscribers throughout the U.S. on a daily basis. The service provides an average of nine hours of programming, with most of the shows planned for the popular evening hour and weekend time slots.

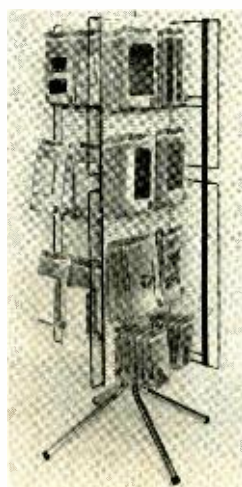
The *Showtime* programs are transmitted via two RCA American satellites (in orbit 22,300 miles above the Equator), and are beamed on two separate feeds—one for the Eastern and Central time zones, the other for the Mountain and Pacific time zones. It is also planned to eventually extend this programming service to Hawaii. **R-E**



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letters

MONEYMAKING IDEAS

Here are two moneymaking ideas that might interest your readers:

Games and money: Doesn't almost every TV sales and service organization have old TV sets (trade-ins, etc.) that just "sorta work"? These sets may be almost unsaleable as receivers . . . but, with minimum repair, they might be ideal as *second* sets for video games only.

We had an old set in the garage with a torn speaker that would receive only one channel. With *no* work, the set became just fine for use with a TV game . . . and the set was 14 years old! Some advantages of this old TV "retirement job" are:

1. No conflict between kids wanting to play the video game vs. adult viewing time.

2. No worry about "burning in" the game image on your good set through accidentally leaving it on overnight, etc.

3. The set can have many faults—poor contrast, poor AGC, nonworking channels, poor audio tone—and yet be just fine for video game use.

4. A customer can try his game on the used set in just minutes to determine whether it's usable . . . and there will be no questions or variable performance due to location, type of antenna, etc.

Program stores: in the *near* future, tape programs for video games and home computers will become a rapidly growing business, especially if standardization is such that tape readers, video displays and keyboards are used in both systems. This market may also include plug-in PROM's and, of course, video entertainment tapes.

They could even become combined. A tape of a well-known novel could (for students) even have a question-and-answer section at the end to emphasize comprehension and interpretation.

The point is—those stores that are now just hi-fi and record stores should be learning just as much as they can about the new technology. It's all interrelated . . . tapes, records and the machine to play them. Five years from now I predict such businesses will be known as "Program Stores," featuring game programs, video programs, music

programs (both live and computer-produced), computer programs, etc., etc.

PETER LEFFERTS

San Martin, CA

NO AM DATA

The logic shown by Len Feldman in the Letters column of the December 1977 issue escapes me completely. Apparently he is saying that since most AM sections of AM-FM tuners are poor, they shouldn't be mentioned.

Good FM performance is easy to find; one hardly needs a product report to locate a tuner with FM that is as good as the transmitting station's characteristics. The difference in tuner performance will show up in the AM section, which varies from terrible to fairly good. Isn't that where the emphasis should be?

Mr. Feldman states that most AM stations use "standard" 5000-Hz lines. I don't know the source of his data, but there isn't one single audible AM station in the Washington, DC, area that isn't audibly better

continued on page 22

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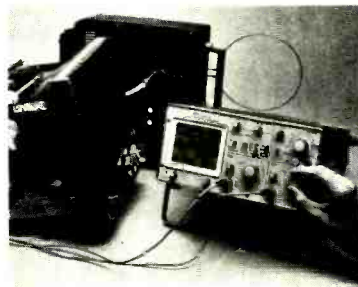
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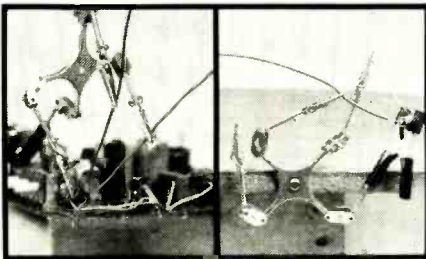
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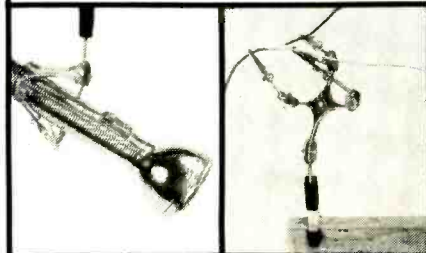
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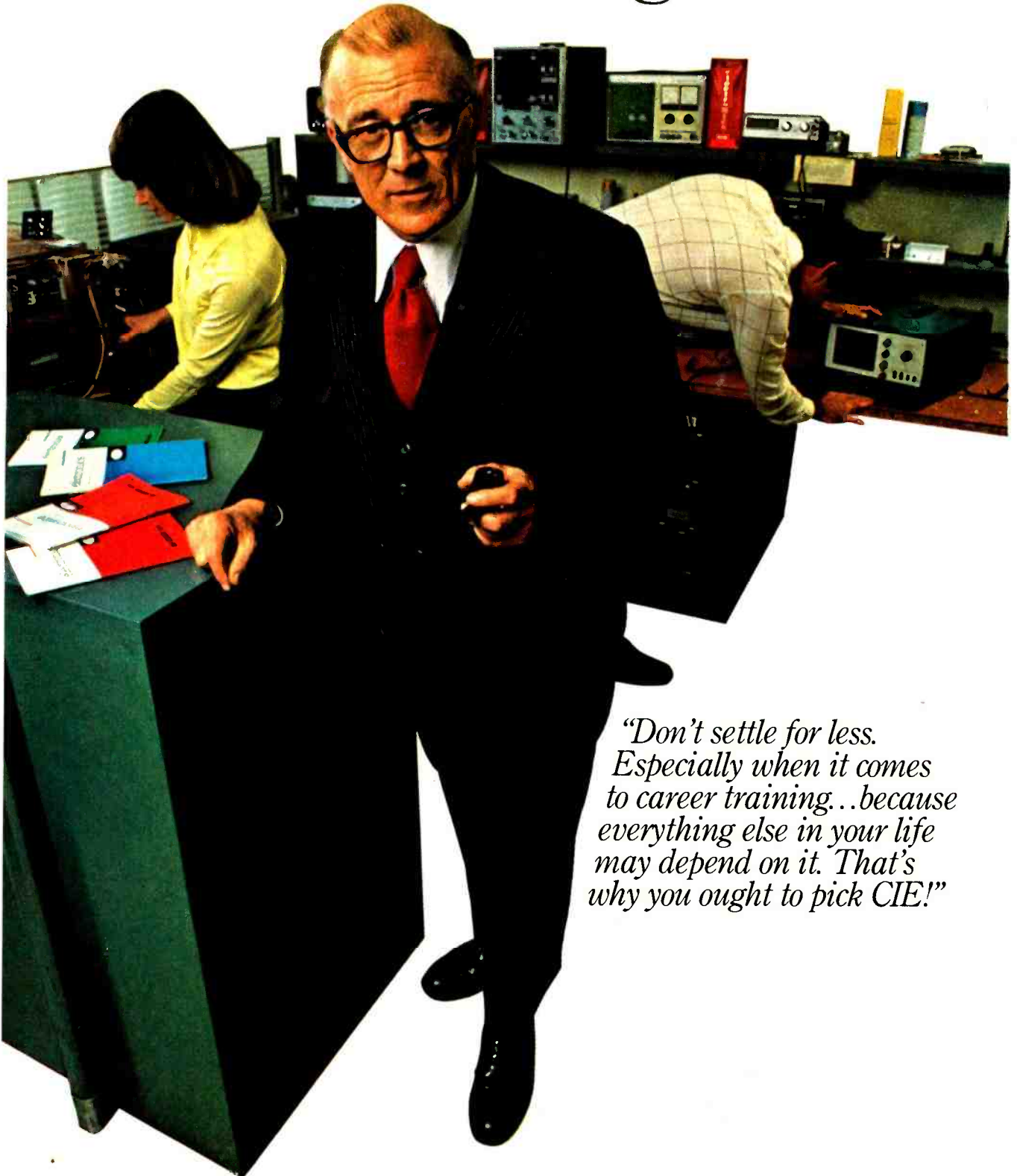
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LETTERS

continued from page 16

than that on a wideband receiver. Broadcast lines are nearly always "unloaded" and stations do equalize them to much better performance than 5 kHz.
CDR. HAROLD W. CORNELIUS
Arlington, VA

On the contrary, Commander Cornelius, it is precisely because I do know how good AM can be made to sound that I seldom bother to discuss the AM sections of the tuners and receivers that I measure. There is one station in the New York area that, in fact, boasts a "proof of performance" from the FCC showing that it is essentially flat to

15,000 Hz (just like FM stations). The only time I was ever able to take advantage of that excellent capability was when I was testing a little-known AM-only tuner known as McKay-Dymek. That tuner had virtually flat frequency response to beyond 9 kHz and an excellent 10-kHz whistle filter as well (a necessity if the response of an AM tuner is to be extended to that degree).

I see little point in discussing the fact that one AM tuner rolls off at around 3 kHz (typical of the breed, by the way), while another goes "all the way" to 4 kHz. If and when I run into a tuner that does measurably better than that, I will certainly devote more space to that fact in my product test reports.

I would also like to take issue with your

statement that "one hardly needs a product report to locate a tuner with FM that is as good as the transmitting station's characteristics." There are still several FM stations that do an exemplary job of transmitting a clean signal, and such tuner parameters as selectivity, distortion, separation, 50-dB quieting and signal-to-noise ratio (not to mention capture ratio and AM suppression in multipath-laden reception areas) are very important if one wants to hear the best that these stations have to offer.

I would hope that if and when the FCC provides us with standards for stereo AM, stations (and receiver manufacturers) may find an additional incentive for creating, broadcasting (and receiving) higher-fidelity AM programming than has been the case up to now. If and when that happens, I shall be happy to demonstrate that I do know a little bit about AM too. —Len Feldman

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CIRCLE 14 ON FREE INFORMATION CARD

FREQUENCY/TIME DISPLAY CORRECTIONS

With reference to my article, "Build AM/FM Frequency Display," in the January 1978 issue, there are a few corrections that should be made in it.

First, in the Parts List, R7 was left off—it should be 15K, as per the Fig. 2 schematic. In Fig. 9, there is a jumper shown going to Q3; this is really R21, or 3.9K. Also in Fig. 9, the "C11" shown just above the crystal, XTAL1, is really R11, or 15 megohms. Finally, the photograph entitled "View of a Partially Constructed Board" shows Q3 through Q6 turned the wrong way. Turn 'em around and everything will be fine.

GARY McCLELLAN

PET COMPUTER NEWSLETTER

Some of your readers may be interested to know of a new coast-to-coast newsletter we are publishing called *The PET Paper*, which is intended to provide "something for everyone" who owns a Commodore PET computer.

Both beginners and experts should find articles in *The PET Paper* of interest, as well as the news it contains of User Groups in their communities, reviews of known and available software, plus hardware how-to's, sources and costs of peripherals as they become available. We even plan to print sections from local User Group newsletters. The cost of a year's subscription to the newsletter is \$15.

For further information, contact Terry L. Laudereau and/or Rick S. Simpson, *The PET Paper*, P.O. Box 43, Audubon, PA 19407.

TERRY L. LAUDEREAU
RICK S. SIMPSON
Audubon, PA

PINK-NOISE GENERATOR

I encountered one difficulty with the pink-noise generator described in the January 1978 issue.

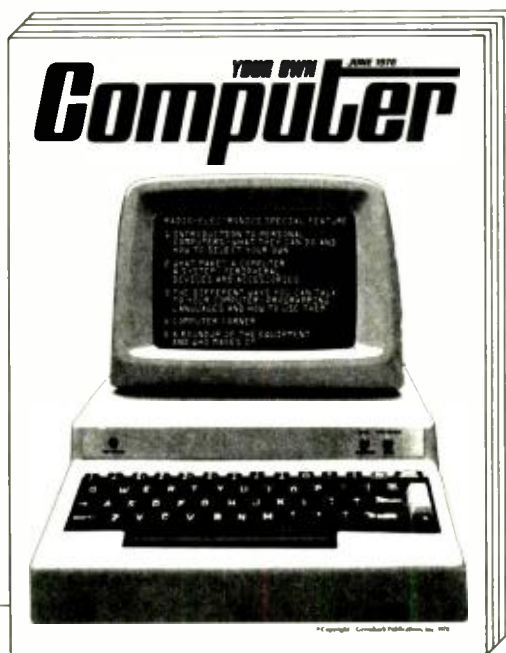
As mentioned in the article, pink or white noise is used to produce the sounds of wind, rain, or surf, and one of the uses of this sound is to provide a steady background for masking out unwanted noises. The pink-noise generator as described in the article is unsuitable for these purposes for one major reason:

continued on page 24

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LETTERS

continued from page 22

A pseudorandom noise generator such as the MM5837 has a periodicity of about $1\frac{1}{2}$ seconds. This is because the noise is produced by an internal shift register and at the internal clock frequency, the register is reset when the random sequence is exhausted—at the periodicity frequency.

The sound the noise source makes is a combination of noise and sounds associated with this cycling. Without the pink-noise filter, the sound resembles metallic machinery with a rotating gear that has a "tinking" sound. With the filter, the sound is much like that of a slow heartbeat.

When I contacted National Semiconductor, manufacturers of the MM5837, the company confirmed this condition. For the purpose for which the IC was intended, namely, for *short-duration* white noise, the cycling effect is unnoticeable. For those interested in avoiding this, a much longer shift register can be constructed, say 28 bits, and the periodicity now becomes over 12 hours, virtually unnoticeable.

W. CURTISS PRIEST

Lexington, MA

Your pink-noise generator is a simple, no frills gem. It is the greatest, cheapest piece of audio gear that I ever bought, and has done more to balance my system to the room than anything.

I have an SL meter, and the results of using the $\frac{1}{3}$ -octave pink-noise record never seemed right. I know my Dynaco 10-

band equalizer does its job so the results were puzzling. A five-minute adjustment using your generator and it sounded better than ever.

However, nowhere do you specify the reasonable lower and upper frequency limits of the unit. I am also sure that the 3-dB octave filter must be a trade-off, and I wish you could specify \pm dB limits for the slope—or supply a frequency plot showing the notches.

If as one of your projects you could offer a reasonably priced sweeper or 10-octave filter set, it would be great to have a variable source of pink noise. Of course, then you get into the question of whether it should be $\frac{1}{3}$ octave.

I fear most equalizers are of the op-amp cut-and-boost type and there is somewhat of a problem at the frequency extremes. Since one can only cut all frequencies about -12 dB, the noise floor approaches the normal output at very low and high frequencies, and an accurate plot is questionable. The noise floor through my system seems to be only about 6 to 8 dB down from a 0-70 dB or 80 dB middle (I mean residual pink noise from the equalizer, not system noise) and trying to find the plot in a system that is only -5 dB at 26 cycles gets sticky.

It seems one can test and experiment cheaply at mid-audio band but playing with the top and bottom costs plenty of money! Thanks for an excellent economical product.

FRANK B. HORNER

Allenhurst, NJ

In response to Mr. Horner, let me assure you that the PNG covers the entire audio spectrum (20 Hz–20 kHz). In fact, an actual spectrum analysis shows the noise to extend from below 10 Hz to beyond 40 kHz. As for the -3 dB/octave filter, it is accurate to $\pm \frac{1}{2}$ dB. There is a low frequency beat that may cause the meter to swing back and forth. The correct setting is when the needle bounces equally on either side of the reference. As Mr. Priest points out, this beat may make the PNG unsuitable for psychological testing and sound effects generation. For these uses, a more conventional semiconductor junction noise source may be required.

JEFFREY G. MAZUR

ENGLISH VS. AMERICAN TV'S

I agree with E. M. Kubilus' letter in the January 1978 issue of *Radio-Electronics*. I have been repairing English TV's for $5\frac{1}{2}$ years in my spare time. The English sets are a dream to repair compared with American models. I think the American manufacturers could benefit from the English design. Also the English picture quality is better than the American quality.

ALFRED R. WATERS
APO, NY

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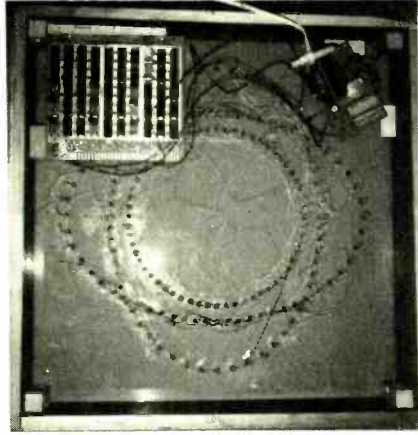
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DIGITAL CLOCK VARIATIONS

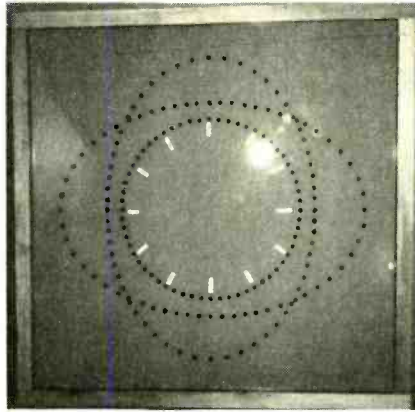
I thought you might be interested in my version of the No-Digit Digital Wall Clock featured in the June 1977 issue of **Radio-Electronics**. I added a couple of features that I think make the clock more interesting:

1. I added a ring of LED's that indicate the seconds and give a sense of motion.
2. I made the minutes and seconds rings into ovals to add interest.



3. I used 60 LED's for the hours instead of 12 to more closely approximate the motion of a mechanical clock.

I wired the clock to a surplus wire-wrap board and modified the author's circuit to allow for the ring of seconds. Also I allowed for the extra hours LED's, plus the minutes



circuits had to be modified to allow moving the hours every 12 minutes. In order to synchronize the minutes and hours I added a power-up reset circuit. The time-setting circuits are more complex to permit setting the seconds and to keep the minutes and hours in sync.

In the fast-set mode, the clock is fascinating to watch, with the minutes and seconds whizzing around and the hours ticking along. I'm thinking of building a similar project that doesn't keep time at all (or can be read as a clock only with great difficulty) but is just visually interesting.

Thanks a lot for the informative article.
MICHAEL MORSE
Garrett Park, MD

R-E



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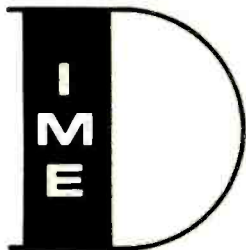
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CIRCLE 101 ON FREE INFORMATION CARD

CHEMTRONICS INC. (45 HOFFMAN AVENUE, Hauppauge, NY 11787) has developed a new handy spray-can product called *Electro-Wash*. This compound is a freon degreaser/cleaner that is very effective for loosening and flushing away grease and dirt that can cause problems in sensitive electronic circuits. It's nonflammable, has a fast wetting action, and will not affect plastics or any electronic component.

The manufacturers claim *Electro-Wash* can dissolve oxidation. I had an excellent test subject in my own antique TV set, which is a jewel but has some problems with dirty tube-socket contacts. So I pulled a few strategically located tubes and sprayed *Electro-Wash* on the sockets and tube bases. This seems to have cleaned up a cute little intermittent that had been bugging me for some time. It also helped the tuner on Channel 13 as well as a couple of controls!

My test sample was a 24-ounce can, a good size for bench use. The can comes with a long plastic needle for getting into tuners and similar places while the TV set is on. Along with the can came a *Vibra-Jet* (model VJ-1) which has a 26-inch-long plastic hose with a fitting that plugs right into the top of the spray can. The nozzle is a 7-inch-long plastic tube, and the handle is a slightly bent piece of larger tubing. This design makes it exceedingly handy for getting into those tight places the big can has problems reaching.

The *Vibra-Jet* delivers the spray in a fast

series of pulses (similar to a "flutter" action) for better cleaning. I cleaned up a couple of filthy PC boards with this and they look like new. This device is also very handy for getting at color codes on resistors that are covered with a 1/8-inch layer of gook, and for identifying resistors that have been overheated and discolored.

The *Vibra-Jet* tubing is made of polyurethane, which is safe for use with any other Chemtronics products: coolant spray, moisture displacers, lubricants, etc. (If you worry about the environment, *Electro-Wash* uses CO₂ propellant, which is harmless.)

Electro-Wash can be used on any kind of electronic equipment from computers to little transistor radios. I also used it to clean up a sticking "U" key on my nonelectronic typewriter with the greatest of ease! *Electro-Wash* sells for \$4; the *Vibra-Jet* costs \$1.98. R-E

Advanced Video Model FSII Video Camera



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continued on page 28

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EQUIPMENT REPORTS

continued from page 26

usually required since the camera "sees" well in normal room lighting. This camera has an $f/1.9$ zoom lens with a focal range of 9 to 30 mm (over a 3:1 zoom ratio) and an optical through-the-lens viewfinder with a built-in sliding lens protector and adjustable-focus eyepiece. The dual-range automatic light control gives usable pictures from 3.72 candlepower to 9293 candlepower, with only one bright-dim control—there is no iris to adjust!

The camera weighs only 2.7 pounds and comes with a quick-release pistol-grip handle. Dimensions (without the pistol-grip handle) are 9 inches long by 4¼ inches high by 2¼ inches wide. Snap off the pistol grip, remove a large screw and you have a standard ⅜-inch tripod socket. A PAUSE-CONTROL pushbutton on the top of the camera lets you remotely start or stop a video recorder at the camera, if the videocassette recorder has a compatible pause circuit. (With the RCA *SelectaVision* videocassette recorder, remove the camera PAUSE pushbutton cap, unscrew and remove the miniature "on" bulb and replace the cap. This remote control works perfectly when connected to the *SelectaVision* pause-control jack.)

All connections are made through a 15-foot five-conductor cable that screws onto the side of the camera. A female connector is provided to mate with the plug at the other end. The five conductors carry the camera video signal, the sync pulses from the power supply (60 Hz), the positive supply voltage, ground return and pause control. The *model FSII* output has standard composite video specifications: 1 volt, 75 ohm, negative sync, with 525-line scan and full 1:2 interlace. The *model FSII* is all solid-state construction, except for the 8844 vidicon tube and the incandescent "on" bulb.

A wide-angle lens is available for \$55, and a close-up lens set, which gives up to $\times 6$ magnification, costs \$10.

The camera can be used with a video recorder—for amateur TV film production, for home or business security purposes, or for any other closed-circuit use. Since the camera does not have an RF output, a *Video Modulator Kit* (\$12) is available from the manufacturer. This kit converts the camera video output to RF on Channels 2 to 6 (slug-tunable) so that the camera can be "played" into any regular TV set. This modulator can also be operated from the same 18-volt power supply as the camera.

Using the *model FSII* is simple. Plug one end of the conductor cable into the camera and the other end into the power supply. The *model PSVI* power supply has a BNC connector for the camera video output that connects to the video monitor or recorder (or to a video modulator to use with a TV set). After about a 15-second warmup, a picture appears on the TV screen (unless you've forgotten to lift the sliding lens protector or turn on the power supply!). Depending on the lighting, set the switch on the side to OVERCAST or BRIGHT. As you watch the TV set, zoom in on some subject, using the lever on either side of the lens mount. Adjust the focus for the clearest picture. Now you can zoom in or out, and the picture will not need refocusing. Incidentally, the focus ring has two convenient detents for average and distant subjects.

Full technical information, including a schematic and alignment data, plus a full guarantee come with the camera.

continued on page 30

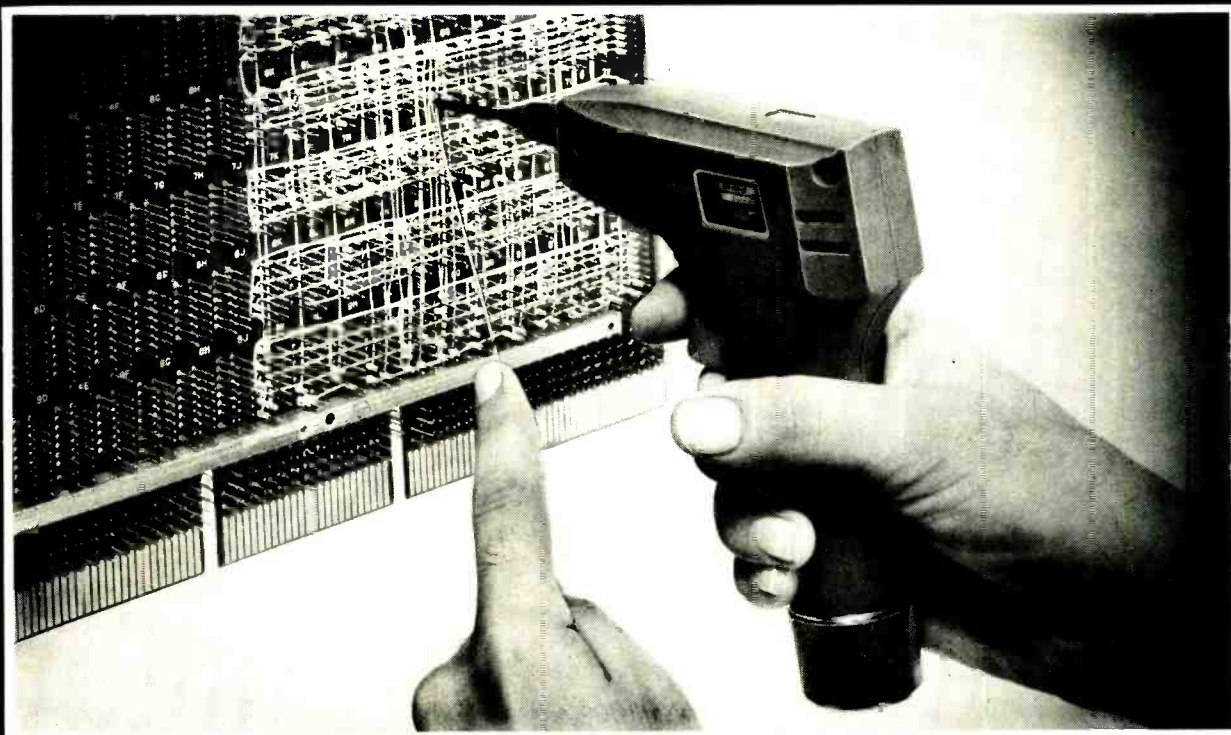
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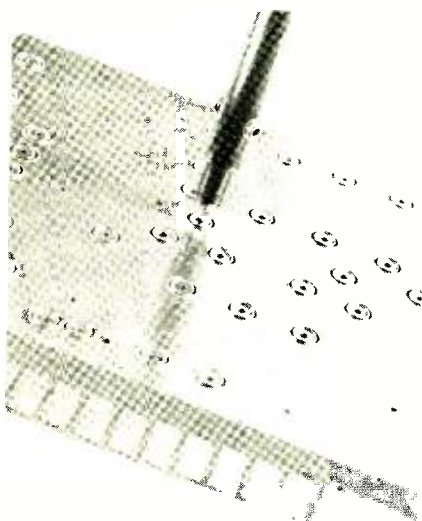


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A. F. Stahler Tools



CIRCLE 103 ON FREE INFORMATION CARD

SEVERAL BREADBOARDING SYSTEMS USE PC boards with etched or stamped conductors that you cut and jumper according to your schematic. Other systems use socket buses into which you plug components and wires. While these are very convenient, timesaving alternatives to conventional PC board layout and etching, sometimes less-disciplined methods

are preferable. Improved layouts often result when IC's and other components can be positioned at nonparallel angles. Radio-frequency circuit optimization requires painstaking attention to lead length, positioning and shielding. A. F. Stahler Company (P.O. Box 354, Cupertino, CA 95014) manufactures tools and accessories that can be used to create a flexible breadboarding technique that overcomes these shortcomings for small-quantity development work.

The basic concept is to create terminal islands within a larger sea of a single- or double-sided PC ground plane. Two tool types in three sizes each are used either to insulate or isolate terminal areas on the PC layout: The first, a *series IS6000* isolated-pad-drill tool (shown), drills a pilot hole and a larger-diameter coaxial ring, leaving a component hole surrounded by an isolated terminal. Surface tension tends to prevent the melted solder from bridging the gap between the isolated terminal and the surrounding foil. A replaceable No. 60 or No. 69 bit is used to drill the component hole; and, depending on tool size, the outer concentric milling edges form a metal circle with a diameter of 0.01, 0.15 or 0.20 inch. The smallest-diameter tool is just right for drilling IC patterns with 100-mil pin spacing. The recommended drilling speed is 600 RPM.

The second type, the *series IS6900* tool, is an insulated-spot-drill that removes all the conductive material within the outer diameter of the tool. If, for example, you want to mount an IC or feedthrough terminal on a double-sided PC board, this tool removes the metal on the component side to prevent the terminals

from shorting against the metal. A *model IS6010* tool could be used for additional mechanical support on the component side, but soldering IC pins on the top of the board makes removal of the IC more difficult and increases capacitance between terminals. The *series IS6900* is also convenient for wire-wrapping terminals and sockets. In this case, the tool is used to insulate the terminals from the surrounding metal so that wire-wrapped leads can be used to complete the circuit.

After initial layout, the hole center pattern is transferred to the PC board directly. For single-quantity designs, the layout can be done right on the PC board itself. Isolated and insulated holes are drilled and the metal is polished with emery cloth to remove burrs and prepare the surface for soldering. Components are inserted, with their leads extending a fraction of an inch from the board to act as wire terminations. The final steps are wiring, soldering and cleaning the board with a flux remover.

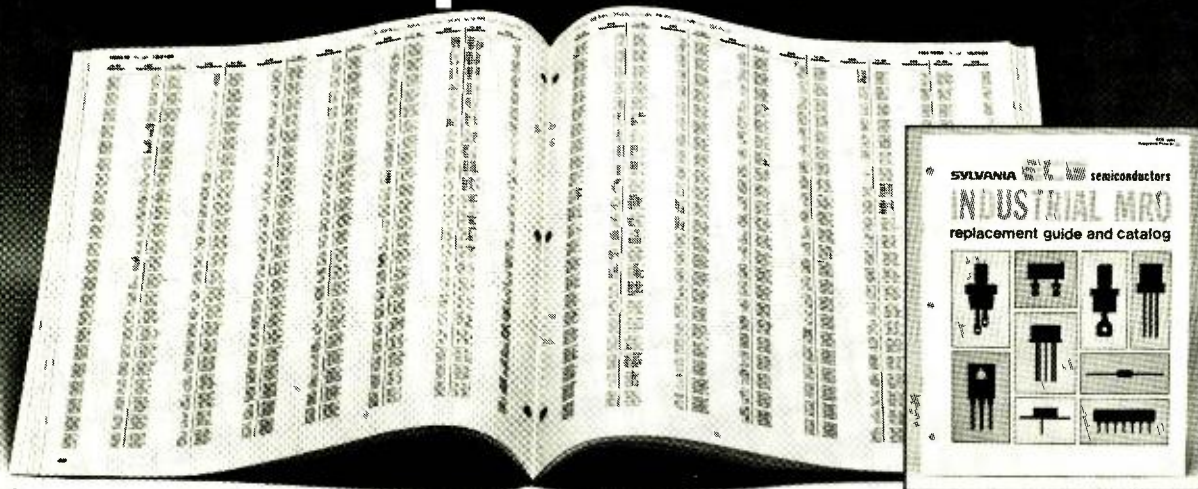
These drilling tools are also very handy if you want to modify existing PC boards; you can easily add pads to unetched areas.

The isolated and insulated drills are available in both high-speed steel at \$10.50 each, or carbide-treated, costing \$12.50 each. Sets of the three sizes are priced at \$25 and \$30 for steel and carbide, respectively.

A. F. Stahler also markets the *model RSDT-DIP16* template set that prints the drill pattern for 16-pin (or fewer) DIP IC's. This \$12.50 set includes a rubber stamp, a 1-ounce bottle of fast-drying ink and a stamp pad.

For additional information, write A. F. Stahler Company, P.O. Box 354, Cupertino, CA 95014. **R-E**

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*The ability to receive short wave broadcasts will vary with antenna size, time of day, operator's geographic location and other factors. You may need an optional outside antenna to receive distant short wave broadcasts.

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CIRCLE 21 ON FREE INFORMATION CARD

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BUILD

Digital Timer for your Darkroom

Modern photographic processes require precise timing for the best overall results. Mechanical timers are going out and electronics are in. This digital timer is the last word.

RAYMOND G. KOSTANTY

HERE'S A MODERATELY PRICED, HIGH-ACCURACY, high-repeatability digital timer with a displayed range from 0.1 seconds to 99 minutes and 99 seconds, and an internal range up to 694.45 days (when 8 nines are entered in the minutes/seconds range). Although originally intended for photo darkroom use, it can be used in any application that requires switching on up to 150 watts for a precise time period.

The desired time is entered via a keyboard and displayed on four 7-segment displays. As the timing interval progresses, the display counts down and shows the remaining time, either in minutes and seconds or in seconds and tenths of seconds. An audio signal consisting of 0.1-second bursts of a 1000-Hz sinewave, once a second, is available as an aid in dodging and burning-in operations, when looking at the display would divert too much of the printer's attention. At the end of the timing interval, the last selected time will automatically be entered into the display and be ready for repeat use. This allows multiple prints from the same negative to be made without the nuisance of entering the same time over and over.

Calculator IC controls operation

The heart of the timer is a General Instruments calculator IC. In conjunction with peripheral logic IC's, it functions as shown in Fig. 1. Initially, the desired time is entered with the mechani-

cal keyboard, and the START/STOP button pressed. Internal control circuits now come into play and activate an *electronic* keyboard. First, the selected time is entered into the calculator's memory and the "minus" function is activated. If the TIME switch is in the SECONDS position, the number 1 is next entered. Then, the "equals" key is electronically pressed at a 10-Hz rate, and .2 seconds later the enlarger and then the audio are turned on. Each operation of the "equals" key subtracts one (decrements) from the display until the contents of the display are -1. Then the enlarger is shut off and the audio is terminated. Next the contents of the calculator's memory are recalled, and finally the memory is cleared. At this point we are back to where we originally started, with the display showing the last selected time, and the memory cleared. If the same timing interval is still required, merely press the START button to start the sequence again. If a new time is desired, enter it via the keyboard. The first key pressed automatically clears the previous time out and enters the new number in the extreme right position. Additional entries always appear in the rightmost position and shift previous entries to the left.

If the TIME switch is in the MIN/SEC position, the selected time is stored in memory and then a loop whose sequence is -, delay, 4 and 1 is executed once a

second until the contents of the D8 display become minus. Although the loop's sequence is -, delay, 4 and 1, the entry of the 4 is inhibited unless the two rightmost digits are both zero as determined by the zero-zero detector. So for 59 cycles out of each 60, the sequence is effectively -, delay, blank and 1. When a whole number of minutes is displayed, such as 3:00, 41 is subtracted so the display next shows 2:59. After the display shows minus in the D8 position, the same ending sequence described above is executed.

Time intervals up to 9999999.9 seconds or 999999 minutes and 99 seconds may be entered, but only the last four entries will be displayed. If more than 4 digits are entered, the OVERFLOW LED between the first and second digits of the display lights to remind you that the interval will be longer than the display contents. It extinguishes when the remaining time becomes 99:59 minutes or 999.9 seconds.

When the TIME switch is in the MIN/SEC position, the last two entries are always interpreted to be seconds (even if they are greater than 60), and earlier entries become minutes. Thus, an entry sequence of 4, 8, 5 will give an interval of 4 minutes and 85 seconds. The same interval could also be obtained by entering 5, 2, 5 (5 minutes and 25 seconds).

The CLEAR button sets the display to zero and allows you to correct mistaken time entries. Since it does not clear the memory or reset the audio latch, it should not be pressed during a timing interval.

The timing cycle can be halted by either pressing the START/STOP button, which will shut off the audio and the enlarger, reset the display to the last selected time, and clear the memory, or by pressing the PAUSE button, an alternate-action switch, which will halt the timing cycle and shut off the enlarger and audio. A second tap on the PAUSE button allows the interval to continue from where it was interrupted.

The FOCUS button, also an alternate-action switch, turns the enlarger on for focusing the first time it is pressed. Its next operation turns the enlarger off and restores control to the timer.

How it works

Figure 2 shows the block diagram of the timer, and Fig. 3 its schematic. The calculator portion consists of IC7, the four displays, four of IC2's buffers, and the mechanical keyboard. The display segment outputs a-g of IC7 contain the numerical information to be displayed in seven-segment format. Each segment output is connected to its corresponding anode of all four displays. The digit outputs, D1-D9, go high, one at a time, starting from D1 and going through D9 (and two additional internal states), and complete the cathode circuits of each display sequentially, thus determining which readout will display the information on the segment lines. The segment outputs are open-drain current sources to

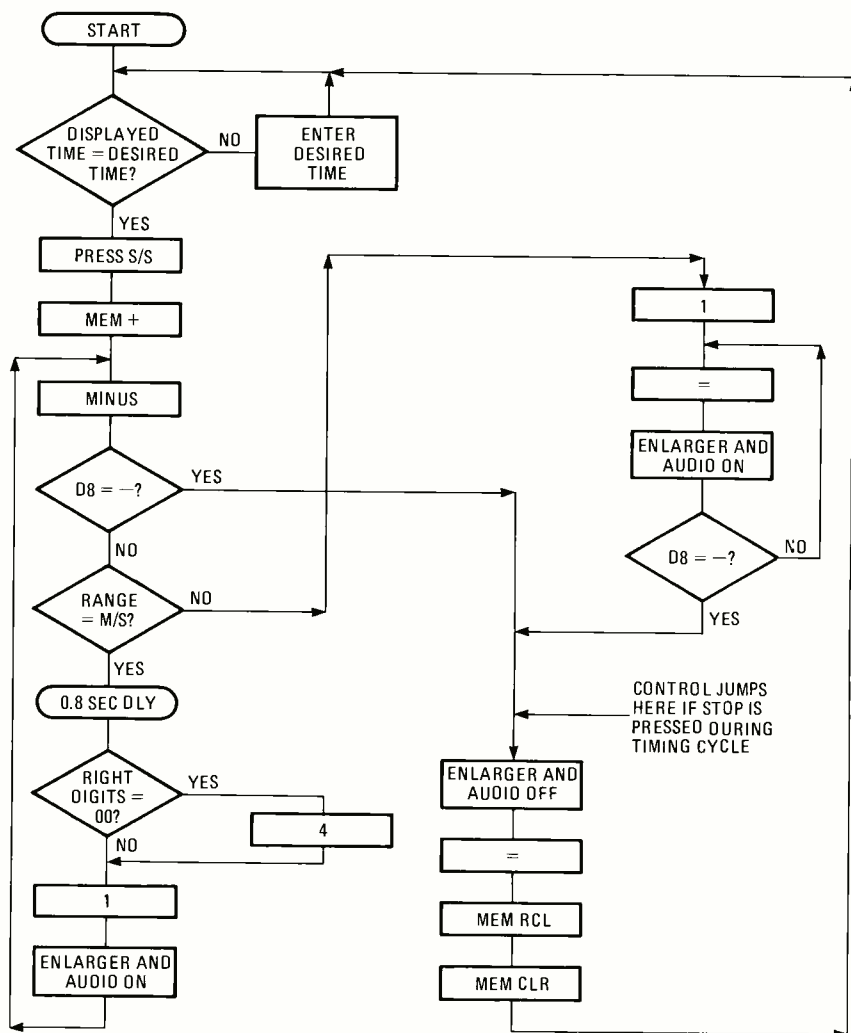


FIG. 1—OPERATIONAL DIAGRAM of the digital darkroom timer. An electronic calculator IC and its memory are vital to the timer operation.

MAIN BOARD PARTS LIST

Resistors 1/4 watt, 10% unless otherwise noted

- R1—10,000 ohms
- R2—22,000 ohms
- R3—R5, R10—R13, R15, R17, R19, R20—100,000 ohms
- R6—R8—2700 ohms
- R9—100 ohms
- R14, R16, R18—5600 ohms
- R21—75,000, 5%
- R22—510,000 ohms

Capacitors 10%, 15 VDC ceramic discs

- C1—100 pF
- C2, C6—C9—0.1 μF
- C3—.02 μF
- C4—.05 μF
- C5—.01 μF
- D1—D3—1N4001, 1N459 or equal silicon diode
- LED1—LED3—light-emitting diode, Monsanto MV-50 red or equal
- DIS1—4—7-segment LED digits, Fairchild FND357 (0.362 in.) or FND70 (0.250 in.)
- IC1—4018 resettable divide-by-N counter
- IC2—75492 6-digit MOS to LED driver
- IC3, IC15—4001 quad 2-input NOR gate
- IC4, IC16—4011 quad 2-input NAND gate
- IC5—4071—quad 2-input OR gate

IC6, IC10—4017 decade-counter/divider, 10-line output

IC7—C-685 calculator (General Instruments)

IC8, IC11—4016 quad bilateral switch

IC9, IC12—4081 quad 2-input AND gate

IC13—4019 quad AND-OR select gate

IC14—4027 dual J-K master-slave flip-flop

PB0—PB10—SPST momentary push-button switches. Part No. DC-61-05 (Datanetics, 18065 Euclid St., Fountain Valley, CA 92708)

Key caps for PB0 to PB10. Datanetics part numbers:

40-3091-03	40-3097-03
40-3092-03	40-3098-03
40-3093-03	40-3099-03
40-3094-03	40-3100-03
40-3095-03	40-7170-03
40-3096-03	

S1—S3—SPDT rocker switch, UID type RSW-06-12-SD-BB-S-B1-BK or RSW-06-12-SD-B1-S-B1-BK (UID Electronics Div., 4105 Pembroke Rd., Hollywood, FL 33021)

S4—S5—SPST alternate (push-to-pock, push-to-release) action switch,

Switchcraft UP-501L with PB-11-02 caps

S6—SPST momentary switch, Switchcraft UP-101M with PB-11-02 cap.

J1—RCA-type phono jack

Miscellaneous—Four rubber or plastic feet, 1/4 × 2 1/2-inch red filter for display cutout, duplex receptacle, IC sockets, nuts, bolts, wire and solder.

Note that the following parts are available:

IC7 (C-685) \$6.50; case \$11.00; main PC board, glass-epoxy etched and drilled with plated-through holes \$15.00. The power supply PC board, etched and drilled \$6.00. Switches S1—S3 \$2.25; S4—S6 \$2.50; PB0—10 \$7.00. Relay \$2.30. Complete kit of all parts except solder \$90.00. Assembled, tested, ready-to-use timer \$120.00. All prices include postage. California residents add sales tax as applicable.

Order from R. Kostanty, PO Box 1042, Gardena, CA 90249. Allow 3-4 weeks for delivery; add 2 weeks for non-California checks to clear. Foreign orders same as above, U.S. funds. Power transformer not shipped to foreign countries.

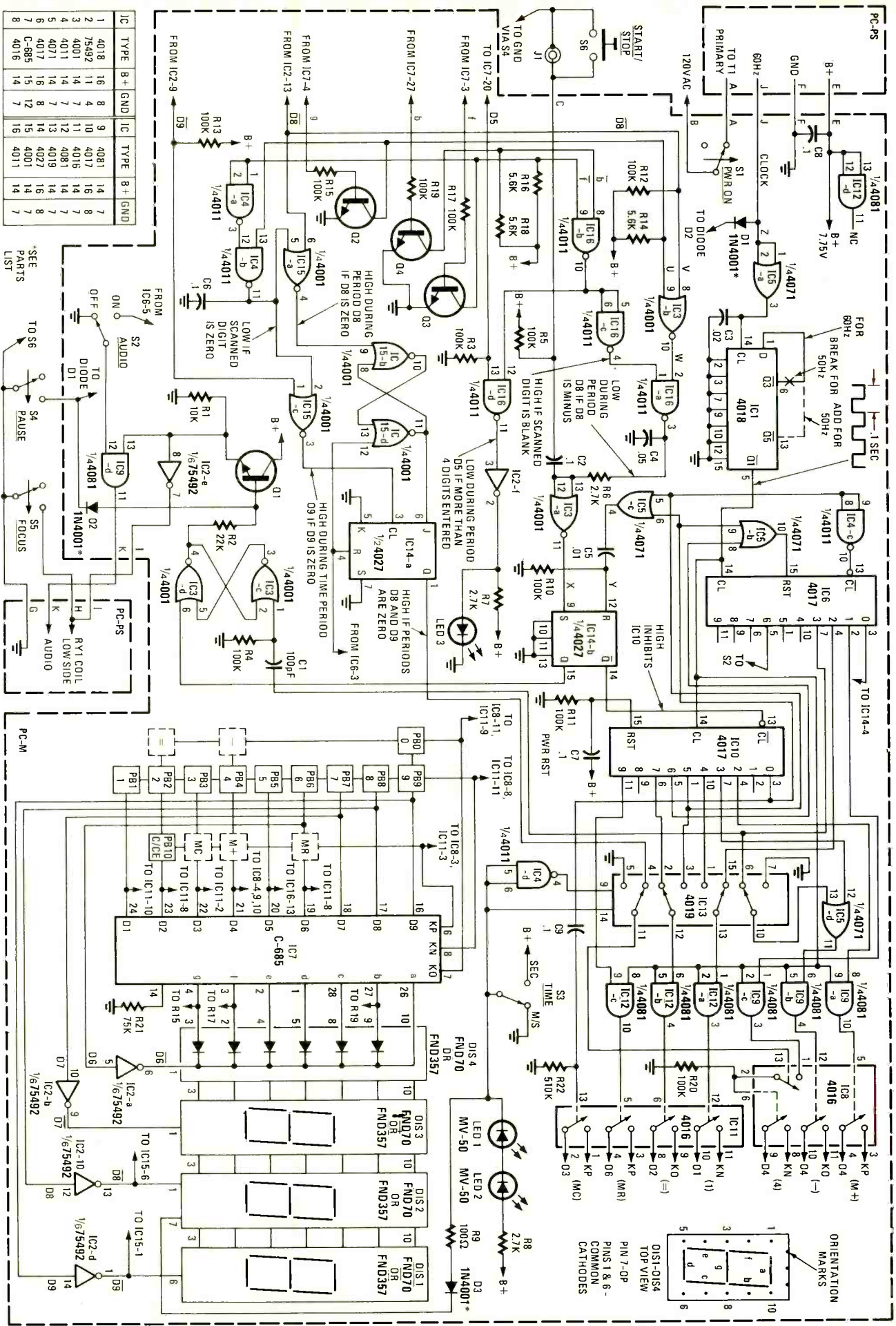


FIG. 3—THE HEART of the darkroom timer. The power supply, audible counter and safelight-enlarger switching will be covered next month.

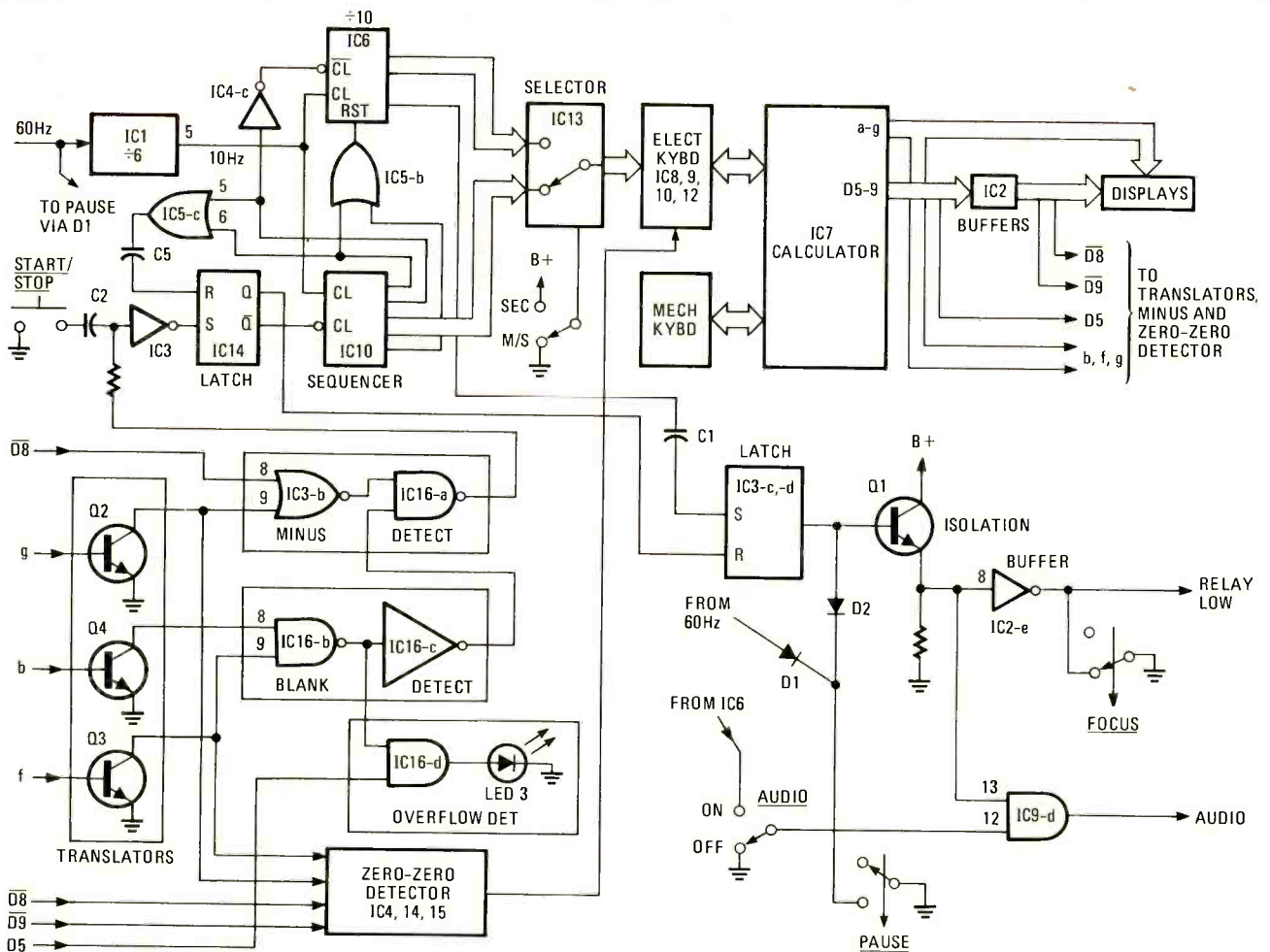


FIG. 2—FUNCTIONAL BLOCK DIAGRAM of the timer. Timing input is via a mechanical keyboard. The electronic keyboard operates from internal commands.

B+ which can directly drive one segment of a LED readout. The digit outputs, also open-drain current sources to B+, do not have the capability of handling up to seven times the current of each segment, so they are inverted through the high-power buffers in IC2 before being connected to the common cathodes of the displays. Each digit output from IC7 is high $\frac{1}{10}$ of the time, at a frequency of about 130 Hz.

The calculator IC (IC7) also uses the digit outputs to determine which functions or numbers are being entered. In Fig. 3, PBO to PB10 (the solid squares) are the mechanical keyswitches which, when pressed, short one of the 9 digit lines to one of the 3 "K" inputs. (Functions electronically activated are shown in dotted squares.) The calculator decides what to do based on a combination of the voltage present at KP, KN or KO, and the timing of this voltage with respect to the digit outputs. If, for example, KN is high simultaneously with D5 (as it will be if PB5 is pressed), the number 5 is entered as input data. Similarly, if pins 3 and 4 of IC11 are shorted by the electronic switch inside IC11, a high will appear on KP during interval D6 and the memory recall function will be activated.

(Yes, a complete calculator could have been included as a function of the timer.

But in view of the \$8.00 cost of ready-to-use commercial units, the additional cost of 5 extra mechanical keyswitches and the 4 extra displays required to display the answer to noneven division could not be justified.)

Each time the START/STOP button is pressed, differentiator C2-R6 produces a negative pulse which lasts about 250 microseconds at IC3 pin 12. This pulse is latched by half of IC14. When pin 14 of IC14 goes low, sequencer IC10 is enabled.

Device IC10 is a decade divider whose 10 outputs sequentially go high, non-overlapping, in the direction from 0 to 9, each time its clock input on pin 14 goes high if pins 13 and 15 are both low. Just prior to the pressing of the START button, the divider is inhibited because pin 13 of IC10 is high, and is resting with only its 0 output high. When IC14's pin 14 is latched low, divider IC10 outputs 1, 2, 3 and 4 sequentially go high at a rate of 0.1 second per step, and in conjunction with IC8, 9, 11, 12 and 13 perform the Memory Add, Minus and, if in the seconds range, 1 functions as shown in the flow chart. When the 5 output of IC10 goes high, the latch which feeds pin 13 of IC10 is reset, causing pin 1 IC10 to remain high for the duration of the timing cycle, IC6 to be enabled, and pin 6 IC12 to be

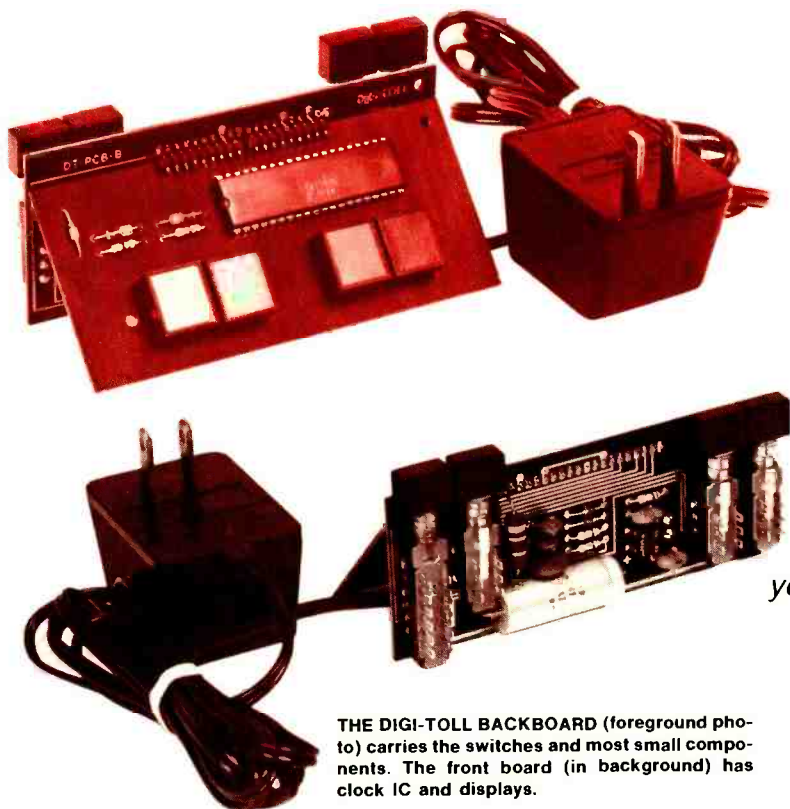
high. Succeeding clock pulses advance decade divider IC6 (which is identical to IC10) and activate the Equals function, which decrements the display once each time it is activated.

The outputs from IC6 reset the zero-zero detector, provide a signal to turn on the audio oscillator and set the latch formed by 2 of IC3's gates 0.2 second after the Equals function is first performed. (The zero-zero detector, IC4-a and b, IC15 and IC14-a, produces a high when zeroes are in the rightmost position on whole-minute intervals.) This delay compensates for a one-time delay in the calculator IC the first time the Equal function is performed. (The fact that latch IC3 is set once a second by IC6-7 is incidental.) IC3's output, after being buffered by Q1, drives a relay on the power supply board which turns on the triac that energizes the enlarger. The relay simplifies gating the triac from low-level logic.

Countdown continues until a minus is detected in the D8 position or until the STOP button is pressed. Either event causes a positive pulse at IC14-9. Integrated circuit IC14 changes state and resets IC3 to turn off the enlarger and allows IC10 to continue sequencing. Output 7 from IC10, which is only required when in the minutes/seconds range, ter-

(continued on page 86)

BUILD DIGI-TOLL



THE DIGI-TOLL BACKBOARD (foreground photo) carries the switches and most small components. The front board (in background) has clock IC and displays.

Save On Long Distance Phone Calls

Part II—Special-purpose electronic timer helps you effect considerable savings in telephone toll charges as you time calls the same way Ma Bell does

FRED BLECHMAN, K6UGT

BEFORE STARTING TO BUILD ANY OF THE *Digi:Toll* versions, take care in component selection and substitution. Certain electrical parts are critical for maximum performance. For example, components T1, D5 and R6 are carefully balanced to provide maximum within-specification segment current to the displays. Since the 3817 IC can only drive 8 mA-per-segment, specially graded displays (brightness code 08 or 09) are a *must*. Also, D3 and D4 are special low-voltage-drop germanium diodes needed to equalize DIS1 segment brightness.

The *Digi:Toll* circuitry can be hand-wired on perforated board, if desired, using “junk-box” components, but the results will reflect this approach. You can make your own PC boards, as shown in Figs. 3, 4 and 5; however, note that the back board, shown in the foil patterns of Figs. 3 and 4, is two-sided with plated-through holes; this is not exactly a beginner’s PC board project. In addition, both boards must be precision-cut when using the *Digi:Toll* enclosure.

Also, several special mechanical components, such as the header strips used to join the PC boards, the specified switches, the extruded aluminum case, the display lens and the screened data plate, are necessary to reproduce a high-quality unit. Since some components may be hard to obtain in single quantities, a complete kit is available.

Assembly of the standard *Digi:Toll* from the complete kit is very simple, and shouldn’t take more than an hour. The instructions provided are detailed and complete. However, since you may want to purchase partial kits, or use some of your own components, the construction described in this article will assume you’ve decided to “roll your own,” using the PC boards and critical components specified.

The standard *Digi:Toll* 12-hour display/24-minute elapsed timer does not require any foil breaks or special jumpers. If you plan to construct any of the optional versions, you should first make the specified foil breaks shown on the schematic table of Fig. 2 and in Figs. 6 and 7. You can make the breaks easily using an *X-Acto* knife or a razor blade, being careful not to cut *other* foil traces. All foil breaks, except for F and D, are on the back board, and all breaks on the back board, except for C, are on the *component* side of the board. Do *not* add any jumper wires yet.

The back board

Start by assembling the back two-sided PC board, oriented as shown in Fig. 6. Install a few components at a time into the board in the locations shown. Solder on the bottom side, using a 25- to 50-watt soldering iron and 60/40 resin-core solder, then clip off the excess leads. Save

the cutoff leads for jumper wires if you make a modified unit. Install resistors R1, R2 and R3 first, then diodes D1 and D2, observing diode polarity. Next, install the bridge rectifier, making sure the marked pins (+ and -) face capacitor C1. Remove the excess coating from the leads of the small capacitors before installation so that the coating doesn’t interfere with proper soldering; twist the coating off with pliers, even with the bottom of the capacitor body. Install capacitors C2, C3, C4, C5 and C6 in the positions shown. Insert resistor R6 into the board and then raise it about 1/8 inch above the board; this resistor runs hot in normal operation and should be given some “breathing” space. Next, install large capacitor C1 (be careful of the polarity!), then the switches. When soldering the switches to the board, make sure that all the terminals are seated in the board holes, that the switch bodies are parallel with the board and that the plungers face upward. After soldering the switches to the board, snap the switch pushbuttons onto the plungers—it takes firm pressure. Next, insert the AC-cord leads from the transformer into the *bottom* of the PC board and solder them on the *component* side. Set this subassembly aside while you work on the front PC board.

The front board

Assembling the front PC board is even

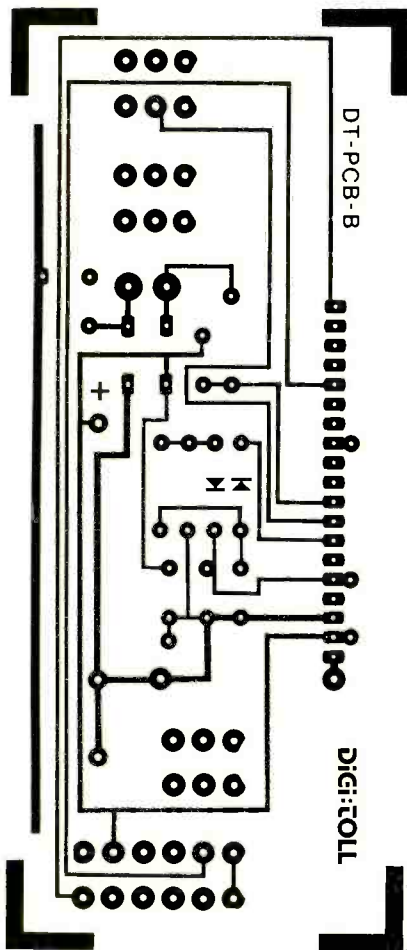


Fig. 3—FOIL PATTERN for rear (solder side) of back board is shown full-size. Switches are on this board.

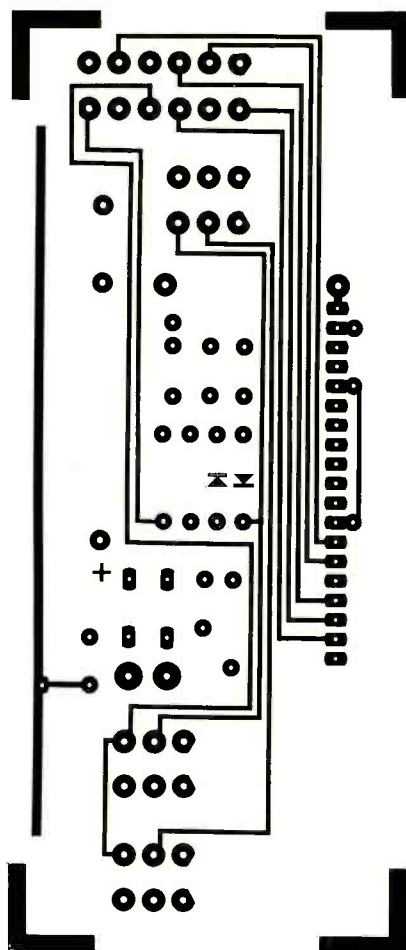


Fig. 4—BACK BOARD component side has this foil pattern (also full-size). Plated-through holes connect front and rear foils.

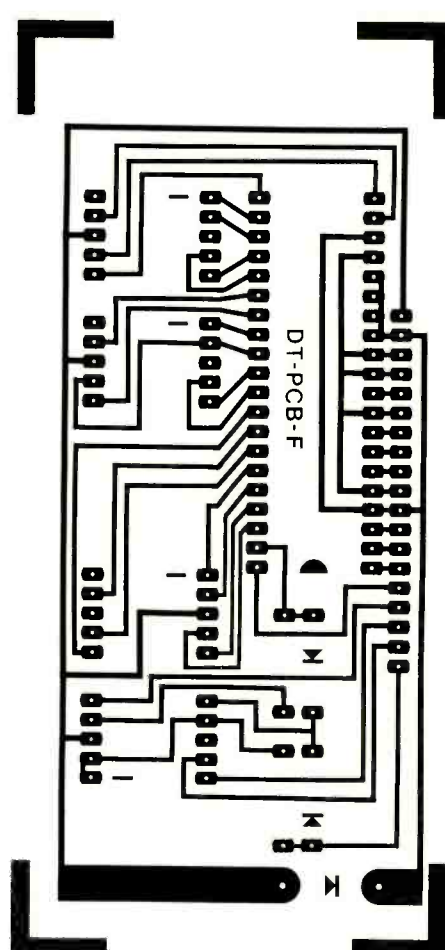


Fig. 5—FULL-SIZE FOIL PATTERN for the front board. This board holds the clock IC and 7-segment displays.

easier. Figure 7 shows the components layout. Insert all components into the blank side of the board, and solder on the foil side of the board. Resistors R4 and R5 are installed first, followed by diodes D3 and D4, and then by diode D5. The colored ends of D3 and D4 are the cathodes (the bars in diode symbols). The cathode of D5 is clearly marked with a black band. Next, solder the wire jumpers into all the 19 holes at the top of the board *except* for the center hole. These jumpers will mate with the back board, so their length and type will depend on whether you intend to place the display a remote distance from the switches that are installed on the back board. You might, for example, want to use some ribbon cable between the boards. You should be aware, however, that this could cause some spurious signals to be generated, and the boards should be as close as your application will allow. The best arrangement is to close-couple the boards using header strips. If you cannot locate header strips, then you must use the more tedious (and structurally inferior) approach of installing individual bare-wire jumpers. The header strips (supplied with the kit) are bent at a right angle at one end, to be used for right-angle panel-mounting (see Fig. 8). If you are using

the *Digi:Toll* cabinet, use a bending tool to form the free ends of the jumpers or header strips to a 45-degree angle. Fig. 9 shows how this is done with the bending tool provided in the complete kit. (The header-strip wires are very tough, and while they can be bent with pliers, multiple flexing to achieve just the right angle can cause them to break—hence, we recommend using the bending tool.)

Next, install the displays, being careful to orient them properly; the scalloped edge is the top of the digit, but DIS1 must be installed *inverted* so that its decimal point can be used as an indicator dot.

Install the clock IC last. Be careful when you handle this device, since it can be damaged by static discharges when not in its conductive carrier. Before installing the IC in the PCB board, it will first be necessary to make the pins perpendicular to the body (they are manufactured bowed-out). Figure 10 shows how to preform the pins on a flat (and non-charged) surface. When you install this IC into the board, be *very sure* the index notch faces the left side, as shown in Fig. 7, unless you enjoy unsoldering 40 pins simultaneously to remove the IC! A socket was not used in the *Digi:Toll* since it would not allow the unit to clear the

extruded custom case; if you use a different enclosure, you could install a low-profile 40-pin socket here.

Joining boards and modifications

The front and back boards are mated by inserting the 18 projecting header-strip pins (or jumper wires) from the front board into the holes at the top of the back board. Do not solder these pins or wires until you align the boards. If you intend to mount the unit behind a panel, as shown in Fig. 8, make sure to provide enough space between the boards so that the display and switches can be properly located behind the panel openings. If you mount the *Digi:Toll* in its custom enclosure, the boards must be properly spaced and aligned at a 45-degree angle, as shown in Fig. 11. The easiest way to insure proper alignment is to use the enclosure itself as an assembly "jig." Insert about 1/2 inch to 1 inch of the boards into one end of the extrusion, which has slots for the boards, and space the boards 1/32 inch apart at the top, using the two plastic strips supplied with the cabinet.

If no modifications are being incorporated in the version you build, solder all the header-strip pins, or jumper wires, on the component side of the back board. If

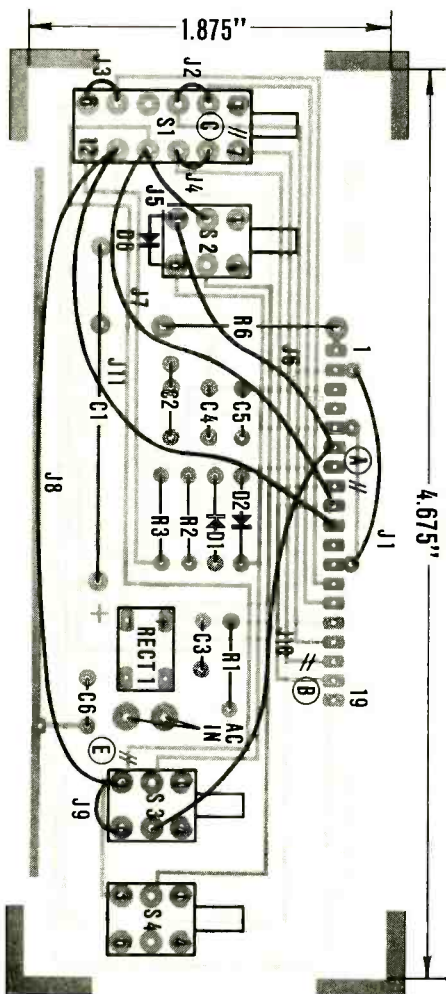


Fig. 6—COMPONENT PLACEMENT for the back board. Check locations of jumpers and foil breaks used in the various *Digi-Toll* options.

modifications *are* being made, do not solder those pins that connect to modification jumpers.

The Modification Table of Fig. 2 defines which jumper wires are required for each version. The longer jumpers are made from an appropriate length of insulated wire, with $\frac{1}{16}$ inch of insulation stripped from each end. The shorter switch jumpers can be bare wire, or salvaged cutoff component lead wires. Figure 6 shows all 11 jumpers; but use only those that are required for the chosen modification! Diode D6 (used only with modification "L") and the jumpers going to the switches should be connected to the solder lugs on the switches. Jumpers going to pins 6 and 9 on the header strips should be wrapped once around the header pin and soldered. The top end of jumper J11 can be soldered in the empty hole (position 10). Keep jumper J1 clear of the top of the board by at least $\frac{1}{8}$ inch so that it does not interfere later on with installation into the custom cabinet.

Testing & final assembly

The unit should be tested before being installed in a cabinet or behind a panel. For the unmodified units, follow this procedure:

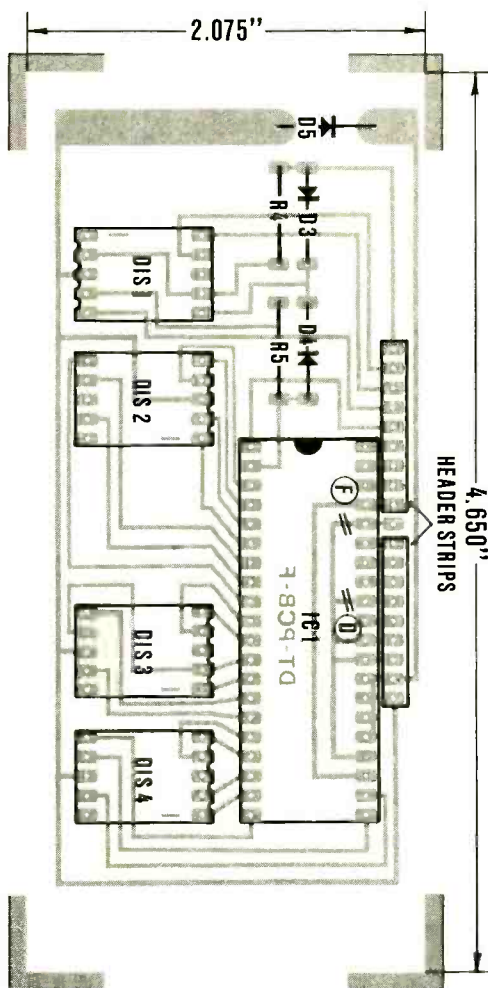


Fig. 7—THE CLOCK IC, four displays and five other components are on the front PC board.

1. Place switch S1 in the up (time-mode) position and plug in the transformer. The display should light up, with the dot indicator blinking to indicate a power interruption. Pressing either switch S3 or switch S4 (SLOW-SET or FAST-SET) will stop the blinking.
2. Set the proper time using switches S3 and S4, checking the segments as the time advances to see that they all light up when they should.

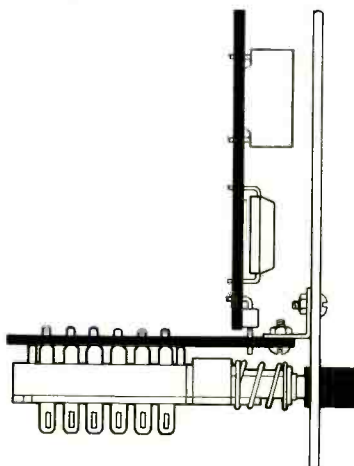


Fig. 8—THE DIGI-TOLL can be panel-mounted if desired. This is a suggested installation scheme.

3. Press switch S1 once so that it stays down (elapsed-time mode). The indicator dot should disappear and the display should count seconds.
4. Press switch S2 to reset the elapsed-time display to 00 00.
5. After several minutes, return switch S1 to the up position by pressing it again. The time-of-day and indicator dot should reappear, and should have advanced to the current time. Pressing S1 down again should display the elapsed time at the point where it was stopped, and seconds-counting should continue from that reading.

For version "T," the test procedure is similar to that for the standard unit, except that there is no dot indicator, and power interruption is signaled by the blinking of one or more segments of DIS1.

Version "L" is the same as modification "T," except that depressing switch S2 (ELAPSED-TIME RESET) while in the *time* mode (switch S1 is up) will reset the *time* display to 00 00.

For modification "W," the testing procedure is somewhat different:

1. Place switch S1 in the up (TIME) position and plug in the transformer. The display should light up, and power interruption is indicated by one or more missing segments in DIS1 (the segments do not blink, since counting is inhibited). The display should *not* change until the elapsed timer is used.
2. Place switch S1 in the down (ELAPSED-TIME) position and the display should start counting seconds.
3. Reset the elapsed-time display by pressing switch S2. The display should go to 00 00 and resume seconds-counting when switch S2 is released. Allow the unit to run for 20 minutes to verify that all segments light up when they should.
4. Return switch S1 to the up position. The display should read the total time that the elapsed timer was on. This display should *not* change (you must watch it for more than a minute to verify this). When the elapsed timer is used again (switch S1 is down) it will "add" to the total-time display (S1 is up), thus accumulating (or totalizing) the individual elapsed times to 23 hours and 59 minutes, and then repeat.
5. When switch S1 is up, pressing switches S3 and S4 *simultaneously* should reset the *total-time* display to 00 00.

Modification "P" does not change any of the operating modes, therefore testing is the same as that for any of the above versions. When operated on a 50-Hz line with the recommended transformer, the display may dim a little, and the transformer will become warmer than when

the unit is operated on a 60-Hz line. Finally, it should be noted that a power failure in the elapsed-time mode (switch S1 is down) is indicated by flashing DIS1 segments on all versions.

If problems develop in testing, the display indicates the symptoms, and Fig. 2 can be used for examination and diagnosis. Be particularly wary of incorrect polarity of RECT1, capacitor C1 and all the diodes. Also, make sure the IC1 index

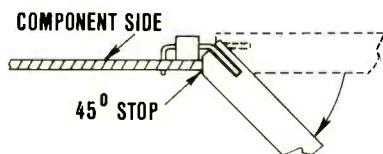


Fig. 9—A SPECIAL BENDING TOOL—supplied in the complete kit—is recommended for forming the header strip lugs.

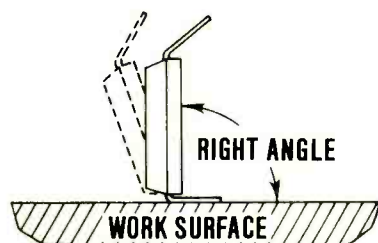


Fig. 10—HOW IC PINS ARE PREFORMED so they are perpendicular to the body before the device is mounted on the PC board.

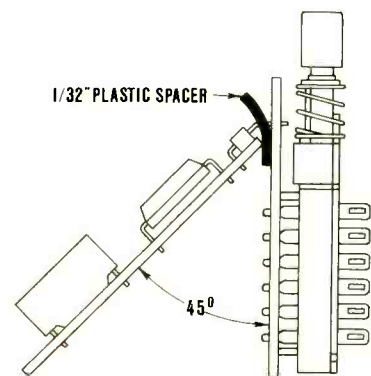


Fig. 11.—HOW THE PC BOARDS ARE ALIGNED prior to mounting in the custom *Digi-Toll* cabinet.

notch and all the display digit notches are properly located. Remember that DIS1 is inverted! If all the components are properly installed, check for bad solder joints—continuity checks should find those. Of course, there is always the possibility of a defective component, but this is less likely than the troubles just described.

When your checkout is complete, you must provide a suitable enclosure or mounting. If you mount the unit behind a panel (Fig. 8) be careful not to short-out or drill through any circuit paths when making mounting holes in the PC boards.

If you use the *Digi-Toll* custom case, slide the completed and tested assembly into the right end of the extrusion with the line cord extending to the right. The display should face you. It is necessary to

slightly depress switches S3 and S4 when they reach the end of the right-hand switch notch so that the assembly can continue to slide into the case. You can use the long plastic strip provided with the case to protect the switches from scratches. Insert the strip from the left end of the extrusion and fit it between the pushbuttons of switches S3 and S4 and the underside of the extrusion as you move the pushbuttons into the case and slide the strip along with the buttons until they pop up into the left-hand case notch.

Using screws, attach the left-end cap loosely. Slide the plastic data plate into the upper slots. Remove the protective covering from the display lens and insert it into the lower front slots, making sure the antireflective dull side of the lens faces out. Now, screw on the right-hand end cap with the two remaining screws, dressing the power cord through the notch in the rear of the right-hand end cap. Before tightening each end cap,

it is only 2:30 PM in California). Conversely, Californians dial New York early (before 8:00 AM) and reach New Yorkers at work at the lowest possible rates!

Figure 12 summarizes the three least known facts about company billing:

1. The initial period rate for *unassisted* calls (no operator), such as direct-distance-dialed calls, is for only *one* minute, not three minutes, as generally thought. Although this first minute costs more, you are *not* charged this rate for additional minutes.
2. On operator-assisted calls, the initial period rate is for *three* minutes, so any amount of time up to three minutes cost the *same*.
3. Additional minutes are timed in full-minute increments.

In other words, you pay for a full minute even if you are on the phone for only one second of that minute! For example, a five-minute, one-second call is billed as six minutes. Hang up a few seconds soon-

FAST SET ◀ TIME ▶ SLOW SET			Digi-TOLL			RESET ◀ ELAPSED TIME ▶ RUN		
UNASSISTED	1 MINUTE	ADDITIONAL MINUTES	DISCOUNT DATA	SAT	SUN	MON-FRI		
ASSISTED	3 MINUTES	TIMED IN FULL MINUTE INCREMENTS	8:00 AM - 5:00 PM			FULL RATE		
Flashing dot ▼ indicates power failure - Reset TIME			5:00 PM - 11:00 PM	35% DISCOUNT				
			11:00 PM - 8:00 AM	60% DISCOUNT				

Fig. 12—DATA PLATE helps you save 15% or more when you use it while making long-distance calls.

make sure the display lens is properly seated in the end-cap pockets. Finally, attach the four adhesive foam feet to the bottom of the cabinet near the corners.

Put it to work

With or without a *Digi-Toll*, the real key to significant telephone savings is for you to understand the phone company timing and discount schedules and use them to the best advantage. Just timing calls for logging purposes, or using a timer to become aware of the seconds and minutes, is not enough! It is how you use this timing data that will *really* add up to dollars saved.

The timing and discount schedules are summarized in the data plate in Fig. 12. It is not only how *long* you talk that's important, but *when* you place the call. For example, if you originate a long-distance call between 8:00 AM and 5:00 PM, local time in your area, on Monday through Friday, you pay the *full* rate. On Saturday and Sunday, however, during the *same* hours, you are only charged 40% of the full rate—a 60% discount. The same applies to calls made from 11:00 PM to 8:00 AM *any* day, and from 5:00 PM to 11:00 PM on Saturdays. On all other days (Sunday through Friday), the 5:00 PM to 11:00 PM calling period costs only 65% of the full rate (35% discount); so, obviously, there is a considerable savings in placing calls at the best time of day, and in taking advantage of East-West time differences (i.e., when it is 5:30 PM in New York (35% discount)

er, and you save the charge for one whole minute. At 15¢ to 40¢ per minute, this can add up!

Now, with this information, you can use your *Digi-Toll* to save yourself a lot of money. When using the standard *Digi-Toll* on a long-distance or timed local call, first select the time-of-day display and check the discount data on the data plate (Fig. 12) to see if you should reschedule your call to a discount period, or to a *higher*-discount period. Then, *immediately* after dialing, push switch S1 down (into the elapsed-time mode) and *hold* down the RESET pushbutton, switch S2, until the other party *answers* (or when conversation starts on person-to-person calls). When you release switch S2, the count starts and you have synchronized the *Digi-Toll's* elapsed timer with phone company timers.

Since the *Digi-Toll* now tells you exactly how long you've been on the phone, you can use several methods to save money: (1) You can limit your call to some pre-established average maximum length. (2) You can eliminate unnecessary additional minute charges by trying to complete calls before a new minute begins. (Telephone company timing ends when *either* party hangs up.) (3) You can use time that you have paid for but, without a seconds timer, would normally lose to the telephone company. If you must begin a new minute or do so accidentally, take advantage of the unused portion of the minute by making sure you've covered everything you want to and avoid

making another call.

When the call is completed, record the elapsed time in a long-distance log if you use one and return switch S1 to the up (time-of-day) position. If you can't log the call immediately, the "memory" feature allows you to recall the elapsed time by just pressing switch S1 down when you want it displayed. It will, of course, continue counting from that point as long as switch S1 is down.

If the elapsed timer of a standard unit is used when the *real* time goes from 12:59 to 1:00, it is possible for the nondisplayed time-of-day to be altered because the internal clock logic doesn't know whether to go to 1:00 or 13:00. To minimize this possibility, return the unit to the time-of-day mode when the elapsed timer is not actually in use. (On any modified unit, such a condition will not occur since all display modes are in the 24-hour format.)

When you use the 24-hour modifica-

tion "T," cost-cutting techniques are similar. However, ham radio operators can set the time-of-day display to Greenwich Mean Time (GMT) for international contacts, and can use the elapsed-time display to remind them to identify their station every 10 minutes, or to time "phone-patch" calls out of the local area. For international phone calls, a time chart can help you determine the best times for the best discount, yet not call someone in the middle of the night at their end!

The modification "L" *Digi:Toll* can be used for long-distance calls just as a standard unit, but you can also use it for timing conferences and consultations, or for time periods longer than 24 minutes. These timed periods can be used for billing or other record-keeping purposes. This modification also makes an excellent research or test-lab 24-hour elapsed timer. At the beginning of the timing period, just press RESET pushbutton S2 while in the time mode, and the display

will go to 00 00 and start counting and displaying hours and minutes. Of course, the time-of-day function is temporarily lost, but can be reset whenever desired by using switches S3 and S4.

Modification "W" provides you with a 24-hour totalizing timer that is valuable on "limited WATS" calls where telephone time is purchased in blocks (soon *all* WATS time will be limited this way). When the 24-minute timer is used in the usual fashion, the 24-hour totalizing timer will keep a continuous record of how much block time has been used or how much remains. At the end of your specified accounting period, just zero the totalizer with switches S3 and S4 and start totalling a new accounting period.

With this minicourse in telephone charges and a *Digi:Toll* to put you in "sync" with Ma Bell's billing equipment, you can now pick up your phone, sit back, put your feet up on the desk and start saving your money!

R-E

Burglar Alarm Switches Made Easy

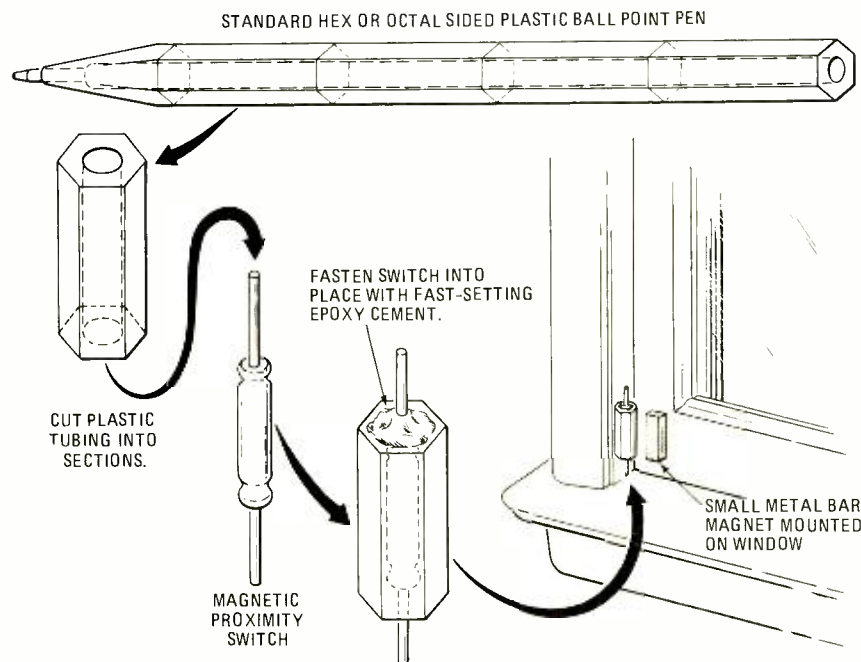
HAROLD PALLATZ

WITH JUST A SMALL SAW, SANDPAPER AND a few parts, a good, inexpensive type of burglar alarm switch can be made. If purchased on the open market, the Magnetic Proximity Switch costs from \$3 to \$5 per unit. You can build it yourself for under 50¢. Each unit takes 10 minutes to make and the materials are readily available. If you have, say, 20 windows and doors in your home or shop, the switches would cost only \$8 instead of \$80.

These magnet-activated switches are good for many reliable operations; they are used by the armed forces in many critical electronic applications. They are most dependable; they have a magnetic reed that flexes to open or close the switch contact when a small magnet is brought near it. Since there is no mechanical linkage or plungers and there is no physical contact with the magnet, there is very little that can go wrong. The switch itself is cemented to the window frame, and the sliding member of the window has a small bar magnet cemented to it. The body of the switch is stationary at all times.

Construction

Take the plastic casing of an old ball point pen—the type with a clear plastic hexagonal or octagonal body. Saw the pen into sections, removing the center ink supply after the first saw cut. It is important to use a fine-tooth saw and to work very slowly to prevent the plastic from cracking or overheating and then melting back together during sawing. File or sandpaper smooth the plastic end sections. Insert the magnetic switches into place and epoxy the ends of the plastic



INEXPENSIVE MAGNETIC INTRUDER SWITCH is made from a tiny magnetic reed switch encased in sturdy plastic housing made from a ball point pen. Mounting in window or door frame is simple.

tube with a fast-setting epoxy cement. (Do not get any cement on your fingers, as some types may react with the skin.)

The wire ends of the switch should protrude from the plastic sections. When the cement is hard, the body of the switch can be glued (or fastened with a small metal or plastic clamp) to the stationary side of the window. A small 1-inch-long bar magnet is cemented, stapled or taped to the door or moving side of the window so that it comes within 1/8 inch or so from the switch. For connection to the switch

ends, the intruder alarm wire is simply soldered to the switch leads. However, solder quickly so as not to overheat the glass or the epoxy.

The uncased magnetic proximity switches can be bought new from most of the larger electronic supply stores. They are also available from electronic surplus outlets for 15¢ to 25¢ each. The small 1- \times 3/16-inch bar magnets can usually be obtained from the same sources at about the same prices, as well as from local hobby supply stores.

R-E

TEST EQUIPMENT

All About Audio Oscillators

Part 2—The audio oscillator most-often-used today is a sophisticated instrument that is a far cry from the simple AF signal source.

This is a continuation of the story on the latest types.

CHARLES M. GILMORE*

THIS IS ONE OF A CONTINUING SERIES OF articles on test instruments, their theory of operation and applications. This story on audio oscillators began in the April issue.

Synchronization output signal

Frequently audio oscillators not offering a squarewave output provide a squarewave of very limited capability. The purpose is to obtain a high-level waveform that can be used for triggering oscilloscopes or synchronizing other equipment. The difference between a synchronizing output and a true squarewave output may be subtle. However, a synchronizing output cannot be expected to have an output attenuator, a constant or controlled output impedance, controlled output amplitude or control symmetry.

Output metering circuits

A metering circuit connected at the power amplifier output permits precise output-level settings with the variable attenuator. This feature eliminates connecting an AC vacuum-tube voltmeter to the audio oscillator for level setting, as is often necessary with oscillators without internal metering circuits where the absolute value of the output signal amplitude is not known. These metering circuits are generally peak-reading and RMS-calibrated. However, such a meter is completely adequate on low-distortion sinewave oscillators. It may begin to jitter as the output frequency is reduced from 20 Hz. In the 1-Hz or 2-Hz range, the meter may actually follow the output waveform, swinging from zero to a peak value. When this occurs, the output amplitude of the generator should be established at a frequency greater than the jitter point.

Some audio oscillators offer a switch-selected internal load. With the internal

load connected, the output amplitude is reduced by a factor of two, compared to the no-load voltage. This internal load is necessary because metering circuit calibration depends on a proper load at the attenuator's output terminals. Often the load connected to an audio oscillator is essentially an infinite impedance. Such a load can be metered correctly by simply switching the internal load in parallel with the output terminals.

Frequency synchronization

A number of higher-priced audio oscillators offer a feature that permits synchronization of the audio oscillator frequency to an external standard. This is useful when a high degree of frequency or phase relationship control is necessary, but when the reference signal is not pure enough to use in the particular experiment. Frequency synchronization specifications indicate the synchronization lock range—that is, how closely the oscillator must be tuned to the reference frequency to establish synchronization. The amplitude and waveshape required of the synchronizing signal is also specified.

Such locking schemes frequently result in a phase shift between the reference signal and the phase of the audio oscillator output. If the two frequencies are identical—that is, if the audio oscillator has been tuned to the reference frequency—the phase shift is a constant. However, it is common for the phase shift to vary when tuning the audio oscillator frequency away from the signal. Of course, the instrument must remain within its lock range to achieve any synchronization at all.

Low-distortion modes

Some audio oscillators have a special low-distortion feature, in which an increased time constant for amplitude feedback control is used to reduce distortion. The effect is primarily intended to improve the distortion specification at low

frequencies. The increased time constant increases the settling time of the oscillator. This means that amplitude variations, caused by changes in frequency, last longer than under normal conditions.

Battery operation

Battery operation, which is offered by a number of oscillator manufacturers, should be used in two cases: When oscillator portability is important, such as in an area where power is difficult to obtain; or where power connections limit operational flexibility.

Battery operation can serve another purpose in the laboratory. Battery power means that the oscillator can be operated with no true connections to earth ground. In other words, the oscillator can be floated at any desired potential. Certain basic precautions must be taken to insure that no contact with an audio oscillator case floating at a high potential is made by either the operator or some object at earth ground; such a contact could cause circuit damage or personal injury.

High output levels

A less commonly advertised feature of the oscillator is its ability to create much higher output levels than a generator. For special applications, audio oscillators can be obtained with levels approaching 100 volts. Such instruments are not considered conventional, but only serve a specialized purpose.

Attenuator range

As noted earlier, the attenuator can vary from a very simple, limited-range device to a complex circuit with a great number of step ranges. Depending upon the intended use of the oscillator, one feature may be more desirable than another. The Hewlett-Packard *model 204C* and *model 204D* audio oscillators are examples of the alternatives available. The *model 204C* offers a single continu-

* Manager, Design Engineering, Heath Co., Benton Harbor, MI.



ously adjustable control with a 40-dB range; the *model 204D* offers a single continuously adjustable control with a 10-dB range, combined with an 80-dB step attenuator (in 10-dB steps).

Where precise amplitude control is necessary, attenuator accuracy must be considered. Even on metered audio oscillators, the output setting accuracy is influenced by the attenuator, since the metering circuits do not read attenuated signals but simply the power amplifier output before attenuation.

As indicated previously, audio oscillators are available in special models with extra high output levels. They can also have extended or specialized frequency ranges for certain kinds of work. Oscillators of this type may vary to an upper frequency of 10 MHz, achieved at the sacrifice of some lower frequencies. Some oscillators also offer a few selected discrete frequencies for production, testing and other limited applications.

Transformer output

Transformer-coupled output, available in a few oscillators, offers complete DC isolation from the rest of the generator. This is particularly useful in driving a bridge. If supplied with multiple secondary taps, the oscillator with a transformer-coupled output can drive extremely low impedance or provide high-voltage output. Usually such oscillators have a limited output frequency range, due to the transformer's limitations.

Applications

It is difficult to enumerate all the applications for an audio oscillator. Briefly, the instrument can be used in any situation requiring a source of relatively pure sinewaves within the frequency range covered. A large majority of applications are found in audio measurements because of the high degree of concentration on reproduction fidelity. However, there are enough applications for a good audio

oscillator that a general laboratory or service shop should place it on the priority list for low-frequency sources (right after a function generator). Of course, if the shop specializes in audio measurements and repairs, the audio oscillator is a *must*.

Total harmonic distortion

The THD measurement determines the amount of harmonic energy added to the output signal of a device by the device itself. With the audio amplifier—or any other amplifier for that matter—the test is performed by supplying a distortionless signal to the input and then measuring the harmonic content at the output. Most THD analyzers reject the fundamental signal with a notch filter and measure all other energy within the audio spectrum. The THD then becomes:

$$\% \text{ THD} = \frac{\text{Harmonics} \times 100}{\sqrt{(\text{Fundamental})^2 + (\text{Harmonics})^2}}$$

The THD measurements for audio equipment are usually made at a number of input and output levels, since the harmonic distortion introduced by the amplifier depends on the gain and the power output at which it is operated. Harmonic distortion is a direct result of nonlinearities within the amplifier.

Note that the described technique for measuring THD includes noise, hum or other nonharmonically related signals not rejected by the notch filter tuned to the fundamental frequency. Two points therefore must be considered. First, the amplifier itself must be thoroughly analyzed for its own hum and noise prior to using it for THD measurements. Any such components should be eliminated before the measurement. Second, the oscillator supplying the distortion-free signal must also be free of hum and noise.

A method of measuring THD that eliminates this problem uses a wave analyzer. This consists of an extremely sharply tuned filter followed by an AC

voltmeter. Although measuring THD with a wave analyzer is more tedious, the result is more accurate. The ultra-low distortion characteristics of the generator are still required, and, of course, there must be no hum and noise. However, levels of hum are permitted with this system of measurement.

The wave analyzer is first tuned to the fundamental frequency and an amplitude measurement is taken. It is then tuned to the second, third, fourth, fifth, etc., harmonics, and amplitude measurements are taken. THD is then calculated using the following formula:

$$\% \text{ THD} = \frac{\sqrt{(2\text{nd})^2 + (3\text{rd})^2}}{\text{Fundamental}} \times 100\%$$

Although truly broadband noise contributes throughout to this measurement, hum and other nonharmonically related spurious signals are eliminated.

When making THD measurements with an analyzer that uses the notch filter, a few basic testing precautions are necessary. Since hum contributes significantly to the harmonic distortion measurement, ground loops must be carefully avoided. A ground loop consists of a second path through which the ground or return signals for the test may pass. Frequently, this second path also contains significant line-frequency currents, which could enter into the measurement and contribute an undesired signal to the output. This signal is generated by neither the amplifier nor the oscillator, but affects the measurement.

Two forms of THD analyzers use the notch technique: With the first and simplest form, the notch filter is manually tuned. The second technique uses manual tuning to within a few percent of the desired center frequency, when an automatic nulling circuit takes over to center the analyzer on the fundamental frequency. Errors can occur in the analysis if the generator drifts from the fundamental frequency to which the analyzer was tuned. An analysis with automatic null compensates for minor drifts. However, drifting introduces errors to measurements made with the manually tuned analyzer.

A note of caution regarding THD measurements: It seems logical that one could measure both the THD of the generator and the THD at the output of the unit under test, and subtract, thus arriving at the THD contributed by the device under test. Unfortunately, because of the complex qualities of the signals, this technique does not lead to a proper measurement. The THD of the audio oscillator must be substantially less than that expected from the amplifier being tested.

In a later issue, we will continue this story on audio oscillators and will show how they are used in measuring audio amplifier parameters such as response, distortion and impedances. R-E

Which Battery Type Is Best

How to choose between Carbon-Zinc, Alkaline and Mercury

There are several types of batteries and dry cells available for use in radios, calculators, electric and electronic toys, and numerous other appliances and gadgets in the home. This is a rundown on how they differ and which to use for different applications.

ALEX F. BURR

THIS ARTICLE GIVES YOU ALL THE INFORMATION NECESSARY TO intelligently select one of the common small nonrechargeable batteries for any routine application. This is not an exhaustive study, but just an article giving the highlights of pertinent characteristics. Some information on the most popular sizes (AA, C, D and rectangular 9-volt) and the most popular types (carbon-zinc, alkaline and mercury) are given.

Battery sizes

The smallest of the popular sizes is the AA penlight cell. It is a cylindrical cell nominally $\frac{17}{32}$ of an inch in diameter and $1\frac{1}{8}$ inches long. It is a very popular size for equipment that must be kept small.

The C-cell is also cylindrical. It is $\frac{15}{16}$ of an inch in diameter and $1\frac{13}{16}$ inches long. It is used where considerable current is required yet size is also important.

The standard flashlight cell is the D-cell. It is a cylinder $1\frac{1}{4}$ inches in diameter and $2\frac{1}{4}$ inches long, and is used where current capacity is more important than size.

For those applications that require a relatively high voltage at a small current, the rectangular 9-volt transistor radio battery is ideal. It is $\frac{21}{32}$ of an inch thick, $1\frac{1}{32}$ inches wide and $1\frac{29}{32}$ inches high, including the two snap terminals on top. Figure 1 shows these four most popular sizes drawn to scale.

There are several other sizes that are not as popular but are easily available. These include the AAA-size, slightly smaller than the AA-cell and often used in pocket calculators, small photoflash guns and electric novelties, and the N-cell, which is a little larger than one-half of an AA-cell and used in certain calculators. Some equipment will use a $\frac{1}{2}$ C- or $\frac{1}{2}$ D-size, but these sizes are rare in a disposable cell. Then, there are the cells designed for electric watches. They come in a bewildering variety of sizes, all small. If the current demands are small and if the size must be small, they can be considered although their cost is great.

Zinc carbon

The most popular type of disposable battery is based on the zinc-carbon cell (see Fig. 2). Despite its name, the positive carbon electrode does not play any part in the chemical reaction that supplies the energy of the battery. The negative zinc case does, however. The positive electrode in the electrochemical sense is the manganese dioxide that is packed around the central porous carbon electrode. This electrode serves as a conductor for the current and for some of the gases generated during the current producing chemical reaction.

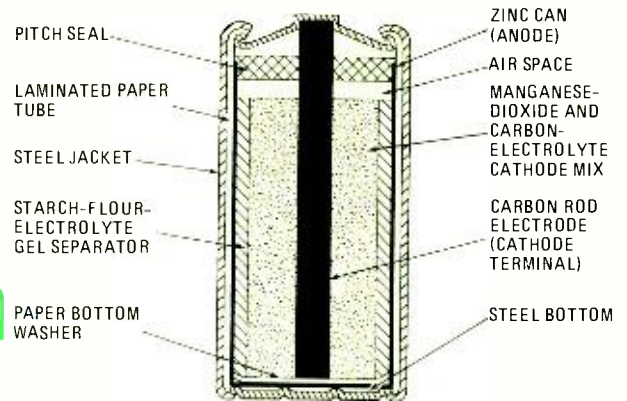


FIG. 2—CROSS-SECTION VIEW shows the construction of a typical cylindrical zinc-carbon dry cell.

The third essential part (after the positive and negative electrodes) of any battery is the electrolyte, a solution serving as an ion-transfer medium between the positive and negative electrodes. In dry cells this solution is in the form of a paste, and in the zinc-carbon cells this paste is a gelatinous mixture of corn starch and flour containing ammonium chloride, zinc chloride and water.

In the modern zinc-carbon cell the positive electrode (manga-

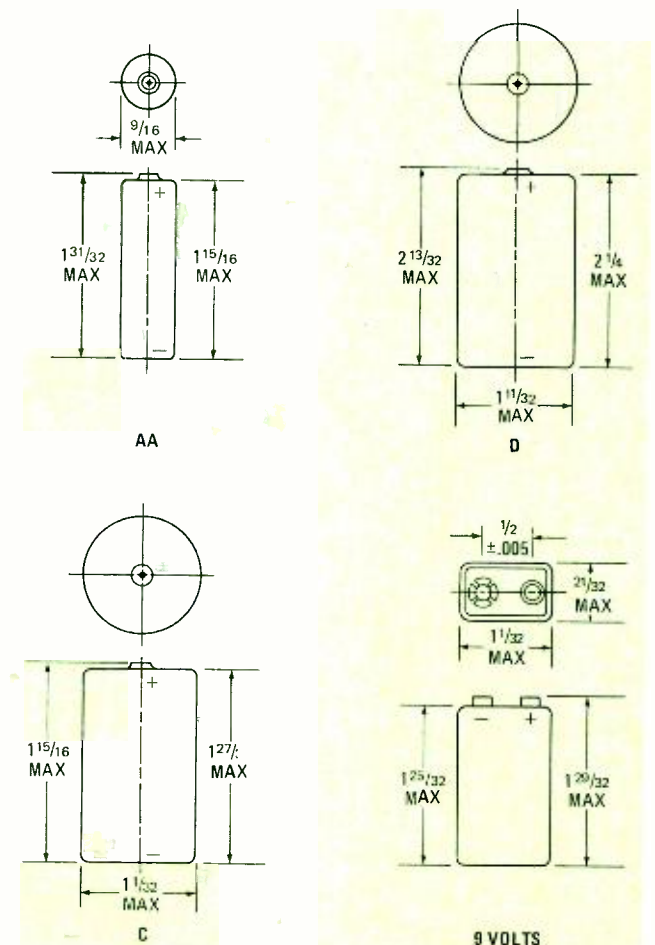


FIG. 1—THE DIMENSIONS of the popular sizes of disposable (not rechargeable) dry cells and batteries.

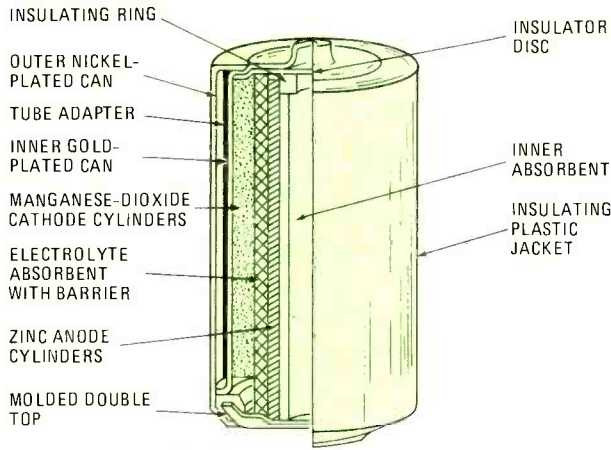


FIG. 3—THE ALKALINE CELL has the familiar zinc and manganese-dioxide electrodes, but the electrolyte is a potassium-hydroxide solution.

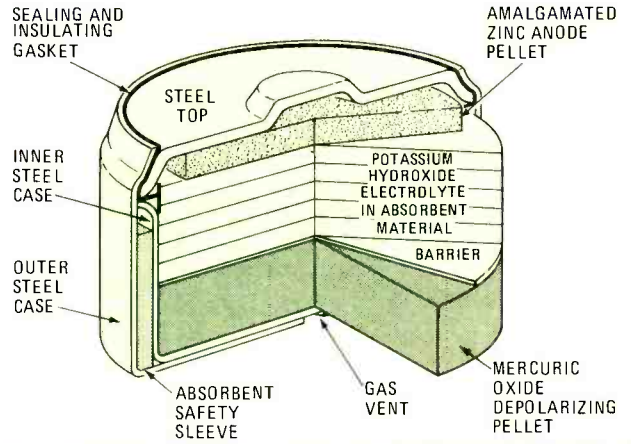


FIG. 4—SECTION THROUGH a typical mercury cell. The construction of this tablet-type cell differs from that of the cylindrical cell or battery.

nese dioxide) also serves as a depolarizer or absorber of gases created during the generation of electricity. The whole cell is encased in a steel jacket to retard the escape of the corrosive materials contained in the cell. During the chemical reaction, the zinc inner case is eaten up as the electricity is produced.

All zinc-carbon cells of a given size are basically the same. There are relatively minor variations in such quantities as the electrolyte and depolarizer-to-electrode ratio that change some characteristics. For example, some batteries are constructed to deliver a *large current for a short time*, while others deliver a *smaller current for a longer time*. Some are optimized for *continuous operation*; others for *intermittent operation*.

There is quite a difference in quality of construction and the purity of materials used. In general, the higher the price paid, the better the cell or battery and the longer it will last in any application. However, a battery twice as expensive as a similar one in many cases will not quite give twice the life.

Alkaline

A popular premium type of disposable battery is the alkaline cell. Its use is often called for when a high current must be supplied. Alkaline cells have the same zinc-manganese dioxide electrodes as the zinc-carbon cells; but the electrode structure is different (see Fig. 3) and the electrolyte is a solution of potassium hydroxide. The result is a cell with the same 1.5-volt open-circuit voltage with a relatively constant energy capacity over a wide range of current drains. Thus, alkaline cells do not have any particular advantage over regular zinc-carbon cells at lower current drains but do have a higher efficiency at high current drains.

Mercury

Another widely used primary battery is the mercury cell. The construction of a mercury cell is shown in Fig. 4. This cell, more correctly called the zinc-mercuric oxide battery, uses red mercuric oxide as the positive electrode and depolarizer and a zinc-mercury amalgam as the negative electrode. The electrolyte is a solution of potassium hydroxide and zinc oxide. The different electrodes give these batteries an open-circuit voltage of 1.4 volts. Mercury cells have a greater energy capacity per unit volume and weight and a better shelf life than zinc-carbon cells. A very important feature is their flat voltage-time discharge curve. That is, the voltage produced by the cell does not decrease until the battery is completely discharged. In fact, these cells are sometimes used as precise voltage-reference sources. The various advantages and disadvantages of these

TABLE I—ADVANTAGES AND DISADVANTAGES of different types of primary cells

Type	Advantages	Disadvantages
Zinc-carbon	Inexpensive	Limited energy capacity
Alkaline	High current efficiency	More expensive
Mercury	Constant output voltage High energy density Good shelf life	Expensive Not for cold weather use

different types of cells are summarized in Table I.

There are other types of primary cells such as small zinc-silver cells often used in watches and the newly developed lithium cell, which has a very high energy density. However, the price of these cells permit them to be used only in special applications.

Current drain

The single characteristic that most are interested in is the amount of current a given battery will supply. Unfortunately, there is no simple answer to that often-asked question. However, some general indications can be given. The larger the individual cell, the lower the internal resistance and the larger the maximum current that can be supplied. Table II gives some

TABLE II—TYPICAL VALUES of short-circuit current and internal resistance found in zinc-carbon cells

Cell Size	Maximum Current (amps)	Internal Resistance (ohms)
AA	4.6	0.31
C	5.4	0.28
D	6.6	0.23
9 volts	0.5	19

typical values of the internal resistance and the short-circuit current of various sizes of zinc-carbon cells.

Of course, don't expect the battery to supply the short-circuit current for very long. What you really need is some idea of the amount of current a given cell will supply for a reasonable time. Unfortunately, what is reasonable in one situation may not be reasonable in another. However, it is possible to give a rough idea of the current that a given type of cell can supply. This data

TABLE III—CURRENT SUPPLIED by different types of batteries

Battery Size	AA			C		D		9 Volts		
	zinc-carbon	alkaline	mercury	zinc-carbon	alkaline	zinc-carbon	alkaline	zinc-carbon	alkaline	mercury
10-hr current (mA)	60	150	240	110	230	250	500	15	30	30
100-hr current (mA)	10	20	25	20	40	50	100	4	8	4

is given in Table III. The first line of the table lists the current that a typical cell of the type named can supply intermittently to a load for 10 hours before the terminal voltage is reduced to two-thirds of the initial value. (This end-point voltage is 1.0 volt for the zinc-carbon cells.) The second line in Table III lists the maximum current a typical cell will supply for 100 hours under the same conditions. These figures should be used only as a rough estimate of the current to be expected because all sorts of factors will enter to decrease or increase the current that the battery can supply. Some of the factors are temperature, the lowest voltage the specific circuit will operate on, the length and frequency of any rest periods, the quality of construction and the age of the battery.

We must emphasize that the current listed in Table III is only a rough estimate—the best that can be made without exactly specifying a great many conditions in addition to the discharge time. We can easily see, especially by considering the data on the C-cell, that the amount of current drawn has a great effect on the total energy that can be extracted from a cell. The table implies that a zinc-carbon C-cell can supply 1100 mA-hours of energy (current drain multiplied by time) in 10 hours, but that 2200 mA-hours of energy can be obtained if the cell is discharged over a 100-hour period.

Table IV gives the mA-hours available from a typical C-cell at

TABLE IV—ENERGY AVAILABLE from a C-cell vs. current drain

Continuous current drain (mA)	5	10	20	50	100
Hours to 1 volt	540	220	90	23	8
Energy supplied (mA-hours)	2700	2200	1800	1200	800

CONVERGENCE PROBLEM

This Heathkit model GR-269 has been working quite well. Now there is a convergence problem at both the top and bottom of the screen. The convergence panel was returned to Heath and checked out. The potentiometers that should affect this area don't seem to have any effect at all.—J. W., Nova Scotia.

Since the convergence board was already checked, there's another likely cause: Check the convergence plug, socket, and all the wiring going to the controls that don't operate. Two things can cause these controls not to work: first, bad components on the board, and second, the loss of the pulses that do the work! Check for an open circuit, a bad solder joint, or poor socket contact. Every control on that board should do something.

VERTICAL PROBLEM

Without a signal, I get a full vertical scan in this Philco model 16M91; however, with a signal it drops to about 80% full. I tried new 6GF7 tubes, changed the

50-μF electrolytic capacitor in the 6GF7 cathode, etc., but got nowhere. The stability and sync are good. Can you help? I've been in poor health and reading your column always makes me feel better!—C. G., Derry, NH.

That's nice, and much appreciated!

Looking at the *model 16M91* schematic, I see it resembles my own RCA *model 15*. Check those 150K resistors in the height-control circuit, as well as all the resistors, particularly those in the input grid, etc. Also, quick-check that VDR (Voltage-Dependent Resistor) in the feedback loop; just disconnect it; it's used as a clamp to keep the vertical scan from being too high. Finally, try unhooking the lead from the pin 2 (output) grid of the 6GF7 tube to the service switch; there might be some leakage.

FUSE IN CATHODE

I found a burned-out fuse in the cathode of the 6LB6 tube in this Zenith model 14A9C50. And there's no high voltage. The Sams schematic does not show this

a variety of continuous current drains. It is clear that the higher the current drain, the less energy is obtainable from the cell. A similar reduction in energy is found with the AA- and D-cells. The energy from an alkaline cell also decreases with increased current but not by as large a factor.

Other factors enter, too. For example, if a C-cell subjected to a 50-mA drain will be discarded after its voltage drops to 1 volt, it will last for about 23 hours. If the end point is lowered to 0.8 volts, then it will last for about 40 hours. At moderate current drains, intermittent use instead of continuous use will increase the number of useful hours by 25% to 50%. Temperature also greatly affects the amount of energy supplied. Table V shows the effect of temperature on zinc-carbon and mercury cells.

TABLE V—EFFECT OF TEMPERATURE on the capacity of zinc-carbon and mercury cells

Temperature (°F)	Zinc-Carbon Capacity (%)	Mercury Capacity (%)
-20	6	—
0	27	4
20	48	10
40	69	58
60	90	93
70	100	100
80	115	103
100	140	106

We can go into all these factors that affect the lifetime of a battery in great detail but most of you will not use the information developed. For most purposes, the simplified current capacities given in Table III will allow you to balance the three main factors of size, current capability and cost to fit the particular application you have in mind.

R-E

fuse. What's it for?—O. K., Oceanside, NY.

Zenith's service data for late runs of the *model 14A9C50* chassis does show this fuse: it's a 0.4-amp Bel-Fuse and is used to protect the flyback, etc., against damage. Since normal current is about 220 mA, a 400-mA fuse should be fine. Be sure to use a fast-blow type of fuse. You can obtain an easy cathode-current reading in TV sets like this. Just pull the fuse and connect your DC milliammeter right across the terminals.

FLOATING BARS, BAD COLOR

I've two floating bars on the screen, bad color, but a fair picture on an Admiral model 3G11. The DC power supply has been checked. Where do I look now?—S. C., Lansing, MI.

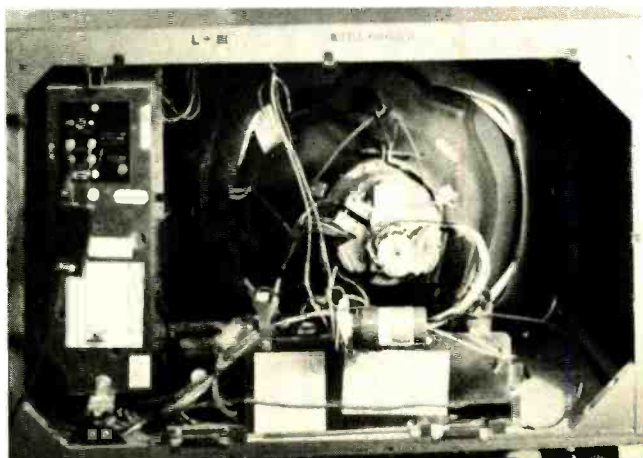
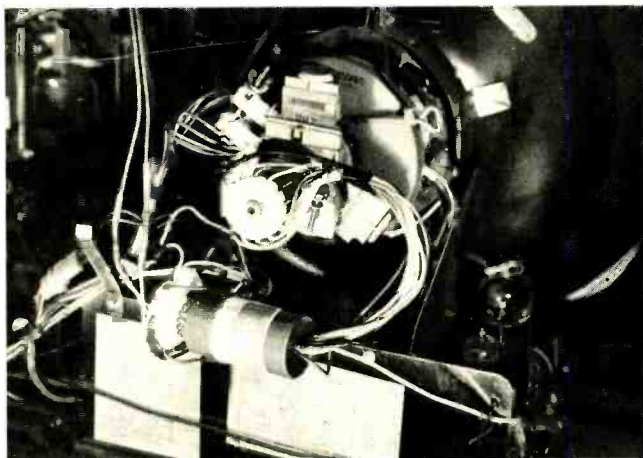
You're close to it. Look at that thermal switch in the autodegaussing circuit. These have caused some funny problems.

(Feedback: "That did it! It took me only two minutes. I'm amazed.") R-E

HOW AND WHY

Picture Tube Brighteners Can Save You Dollars

Television picture tube brighteners have been around for quite a few years. Here is a fresh look at what they are all about.



NORMAN A. ACKERMAN*

THE USEFUL LIFE OF A TV PICTURE TUBE IS usually determined by the weakening of the electron beam. Yet the phenomenon of electron emission in the beam from the oxide-coated cathode is largely unexplained. However, exhausted cathode emitters do have remarkable recuperative power. When certain restoring devices are used, picture tube life can be extended. This article will attempt to explain the process that makes this regeneration possible.

The picture tube

The TV picture tube is a special kind of vacuum tube in which a narrow beam of electrons behaves like an electronic pencil and draws a visible trace or pattern on a specially prepared screen. In many respects, its operation parallels that of an ordinary receiving or transmitting vacuum tube. In an ordinary vacuum tube as well as in a CRT or picture tube, electrons are created by a heated cathode located at the base of the tube inside the glass envelope. Thermionic emission in a

* Vice President, Perma Power Division of Chamberlain Mfg. Corp.

vacuum tube occurs when the electrons in the cathode material contain enough thermal energy to overcome the forces at the emitter surface and escape.

In an ordinary vacuum tube, the emission occurs over the entire outside surface of the cathode (emitter). (See Fig. 1.) In

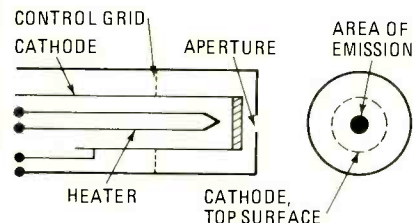


FIG. 1—THE CATHODE STRUCTURE in a TV picture tube. The black dot in the center shows the cathode area that is actually used.

the picture tube, the emission comes from only the top surface of the cathode. More precisely, emission occurs only under the area projected by the aperture in the control grid. To fully comprehend why this happens requires some information about the effect of the control grid and the first anode on emission.

Thermal agitation of the cathode coating material causes the cathode to emit electrons. Since this cathode coating is located in one particular place—on the cathode face opposite the control-grid aperture—principal emission moves in a forward direction from this surface. Because of the shape of the cathode, the emission does not come from a point source. So, while the emission is generally in a forward direction, it takes place randomly. Some electrons leave the cathode in a direction that is parallel to the axis but off to one side, and some leave at various angles to the axis.

Moreover, the velocity of the electrons varies. Some electrons leave the cathode at high speed, but the majority are low-velocity charges. For simplicity, assume that the initial velocity of the emitted electrons is substantially zero and any forward motion they exhibit after leaving the cathode is attributable to acceleration by the voltages applied to the first anode, in opposition to the negative voltage at the control grid.

An electrostatic field (called the first lens) exists between these elements and varies with the control-grid voltage, thus controlling the emission from the cathode. For any fixed voltage applied to the first anode, the control-grid voltage controls the number of electrons that can pass through the aperture.

Figure 2 shows the field distribution for two assumed grid-bias values, 0 (Fig. 2-a) and -30 volts (Fig. 2-b), and a fixed voltage value on the first anode. With a bias of 0 volt, the area adjacent to the cathode (between the cathode and control-grid aperture) has a comparatively high positive potential as a result of the field.

“Under such conditions of 0 grid volt-

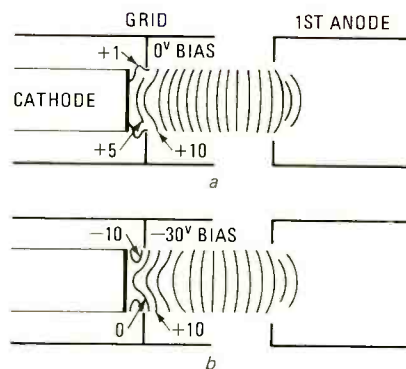


FIG. 2—FIELD DISTRIBUTION FOR two grid-bias values; (a) 0 bias. (b) 30 volts bias.

age, it has been found that the area of the cathode which is emitting corresponds approximately to a projection of the area of the grid aperture; the maximum number of electrons are passing through the grid opening and the beam current density is high.

“When the control grid is made negative by an increase in the bias (-30 volts in the figure), the field distribution in the vicinity of the cathode is altered so that only the center of the emitter surface is behaving as an emitter. The other areas are influenced by the space charge and effectively are not emitting. The result is a reduction in beam density and several other related effects.”^{1,2}

It should be emphasized that showing the electric field and lines of force schematically is strictly a device to help you visualize certain phenomena. The lines of force having certain physical properties are convenient working tools to explain what happens.

The thermionic emission that comes from under the aperture opening of the picture tube emitter continues until there

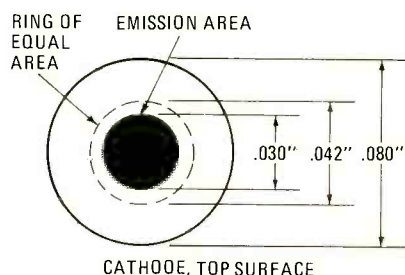


FIG. 3—WHEN BRIGHTENER IS USED, diameter of emitting area is increased slightly as shown.

is a depletion of electrons. Even after emission from the electron gun *appears too low* to provide enough density for a good picture, the cathode emitter material is almost *90% unused*. At this point the useful end-life of the tube has apparently been reached, because the picture is not bright enough for proper viewing.

This apparent contradiction can be better understood by examining the cathode. It is typically a metallic cylinder that is about $.080$ ($\frac{2}{64}$) inch in diameter and $.340$ ($\frac{11}{32}$) inch long, and is closed at one end (see Fig. 3). The heater is inside this

cylinder with its leads through the open end, and it is coated to provide electrical insulation. The outside closed end of the cylinder is coated with an emitter material (i.e., magnesium oxide) that looks pure white when it is new. When the tube is depleted, there is a metallic dot near the middle of the white surface of about $.030$ inch in diameter. This diameter corresponds to the aperture opening of the control grid.

The cathode is held permanently in place by a metallic ring molded into the ceramic supports of the electron gun. The control grid is an inverted dish or cup that almost completely covers the cathode, and also is held in place by the ceramic supports. There is a space between the grid cup and top of the cathode on the order of $.055$ inch. Because of this spacing and type of construction, the cathode cylinder is barely visible through the glass neck of the tube, and the top of the cathode is virtually impossible to inspect once the electron gun has been manufactured.

Picture tube manufacturers rarely run a control sample picture tube to normal emission depletion and then inspect the cathode surface. Quality control samples are generally tested only for about 4000 hours, equivalent to two years of normal use, and then they are inspected. (Normal viewing is considered to be six hours a day.) When a picture tube is operated beyond that 4000-hour period to electron beam depletion and apparent end-life, it seems unlikely that the tube engineer would break the tube apart to check the gun parts let alone saw or cut open the metal control grid to inspect the used-up cathode surface. Yet, if he did he would find new emitter material surrounding the tiny metallic dot of depleted surface.

The size of the dot and the new material surrounding it suggests that the picture tube life could be prolonged if the new electrons could be moved to the area under the control-grid aperture. Again, take a look at the dimensions involved. The aperture opening produces a narrow beam and an electrostatic lens that can project an image on the picture tube screen. This opening is about $.030$ inch in diameter ($.015$ -inch radius). The unused emitter area is about $.0043$ square inch compared with a $.0007$ -square-inch depleted emitter area, yet the picture tube can no longer produce a picture that is bright enough for normal viewing. Contrast is lacking because there is not enough electron beam density to fully activate the picture-tube-screen phosphor dots. Warm-up time is unusually long since the cathode temperature slowly increases to give the last electrons sufficient energy to escape.

TV tube brighteners

In the late 1940's, a transformer device was designed by the Perma-Power Company to increase the power applied to the

heater and raise the temperature of the cathode in the emission-depleted picture tube. (A patent was filed in 1952 and granted in 1956.) This device (called a TV tube brightener) is permanently installed in the TV set, and the increased power it feeds to the older heater is applied whenever the set is on. Thus, the kinetic energy of the electrons in the cathode emitter surface is increased. As a result, the electrons in the ring surrounding the depleted area can migrate to the area under the control-grid aperture opening, and this area of new material is thus used.

It is interesting to observe how small a ring of cathode material of equal area to the depleted area is. This ring has an outside diameter of .042 inch and a radius of a .021 inch. You can easily understand that if only three-thousands of an inch of material surrounding the depleted emitter area emits electrons that migrate under the aperture opening, then this emission includes almost 50% new material.

Early studies of black-and-white picture tubes were made in which cathodes from tubes that had reached end-life were compared with some that had reached a second end-life after using a brightener. The depleted areas in the tubes using brighteners were on the order of 1/4 to 1/2 times larger. The increase in viewing time was often more than 50 percent (see Fig. 4).

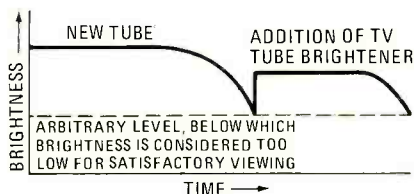


FIG. 4—INCREASED LIFE EXPECTANCY of picture tube with brightener attached is shown here.

While these early studies involved black-and-white tubes, the same observations are true of color picture tubes. In the latter, however, one of the three guns, usually the red gun, loses its electron beam density before the others. Although a separate heater is used for each gun in a color picture tube, the three filaments are tied together and only two leads are available at the picture tube socket. To raise the (heater) power on one gun, the power on all three guns must be raised.

Fortunately, no ill effects result from this procedure. Since the electron gun is a controlled emission device, you use the CRT bias controls on the set to readjust the electrostatic fields in order to keep beam-current densities balanced. This is a function of the electrostatic field distribution. The fact that more electrons are available than are needed does not affect how many are drawn off and used.

In the manufacturing process, picture tubes are processed through exhaust-and-

aging cycles. This process includes procedures that decontaminate the tube elements and remove gases. The aging process also initiates emission from the cathode surface, to assure consistent emission from tube to tube. Procedures vary, but elevated temperatures and heater power are always used. The control grid and first anode areas can be heated to 800 °C for 10 to 20 minutes, and twice the normal heater voltage can be applied for these intervals. One and one-half times the normal heater voltage is generally applied for an extended period of time with normal voltage on the first anode.

Since the cathode heaters and other tube elements are designed to withstand this exhaust-and-aging process, it is easy to understand why a TV tube brightener can be used without damaging the picture tube.

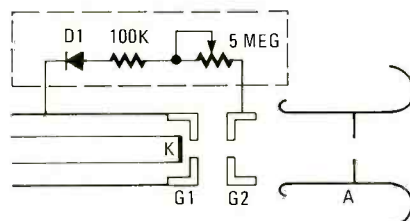


FIG. 5—ADDING RESISTANCE between G1 and G2 reduces the field and tends to open the beam.

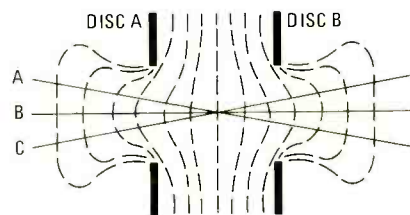


FIG. 6—THE EQUIPOTENTIAL SURFACES between two discs, each having an aperture at its center.

In addition to applying more heater power to regain emission, the electrostatic field can be readjusted as follows: The leads for control grid G1 and first anode G2 are accessible outside the picture tube so that a potentiometer can be installed between them. Since the voltages on the control grid and first anode are essentially fixed with reference to cathode K, applying resistance between the G1 and G2 leads reduces the field and tends to open the beam. The accelerating voltage is constant, resulting in an increase in beam current (see Fig. 5).

However, any condition that alters the

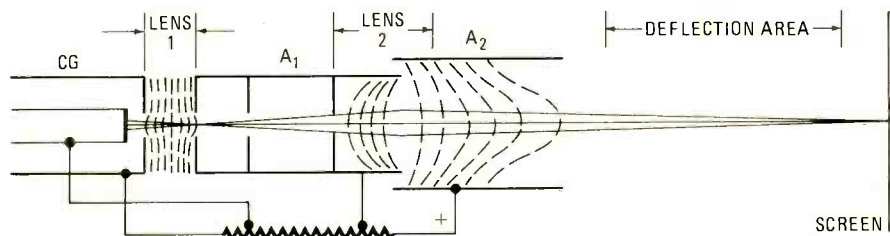
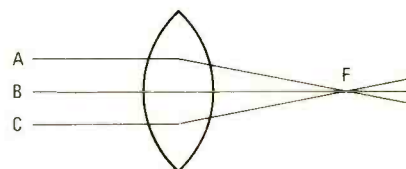


FIG. 8—ELECTROSTATIC FOCUSING SYSTEM as it is used in a TV picture tube.

distribution of the field also affects the focus. Figure 6 shows the equipotential surfaces between two discs, each having an aperture at its center. The equipotential surfaces bulge through the holes, causing some electrostatic field to leak outside the normal region. If electrons a, b and c are placed to the left of plate A, electrons a and c will converge as they cross the convex equipotential lines that protrude through the opening in plate A. Electrons a and c will tend to cross each other within the field between the two apertures, and then continue along a diverging path as they cross the concave equipotential lines that protrude through the opening in plate B. This procedure is



SIMPLE OPTICAL LENS

FIG. 7—SIMPLE OPTICAL LENS is similar to electron lens described in the text.

similar to what occurs in an optical lens having three incident rays (see Fig. 7).

Although it is true that any set of curved equipotential lines resemble a simple electron lens because they refract the electron path, the complete electron gun field is more complex (see Fig. 8).

However, it should not be difficult to visualize how the charged cylinders in an electron gun form an electrostatic field that controls the electron beam and focuses it. The dimensions of the crossover point, as determined by the control-grid and first-anode voltages, are also the dimensions of the beam when it is properly focused by the second electron gun lens and strikes the screen.

However, any device that changes the designed field distribution can produce some undesirable effects. It can affect the dimensions of the first crossover point and the angle of divergence at which the electron beam leaves the crossover point and, to a limited extent, it can affect the position of the first crossover point.

Emission slump

One of the least understood and most prevalent picture tube problem is known as emission slump. It is generally accepted that the cathode slowly becomes con-

continued on page 88

Radio-Electronics Tests JVC Model RC-828 Radio-Cassette



CIRCLE 104 ON FREE INFORMATION CARD

1

LEN FELDMAN
CONTRIBUTING HI-FI EDITOR

WHEN YOU LOOK AT THE FRONT PANEL OF the JVC model RC-828JW radio cassette (Fig. 1), you may wonder why we decided to test and discuss what looks like a "low-fi" portable radio cassette unit. Reading on, however, you will discover that this is not an ordinary entertainment portable. Its tuner sections and cassette record/play sections equal the quality of those found in low- to medium-cost separate receivers and stereo cassette decks. Additionally, although its amplifier has minimum power, this fact is offset by the model RC-828's ability to drive an external amplifier of any wattage rating by its unique biphonic and stereo-expansion circuitry.

The lower section of the front panel shows a central cassette compartment, below which are the usual transport pushbuttons (fast rewind (REVIEW), RECOrd, PLAY, fast forward (CUE), STOP/EJECT and PAUSE). You can use the fast rewind and fast forward buttons to review (i.e., rewind in measured sections by holding down both pushbuttons while also pressing the PLAY pushbutton) and to cue forward. Above the cassette compartment are a three-digit tape counter and its associated reset button. To either side of the cassette mechanism are 6 1/2-inch-diameter woofer drivers. The decorative areas above the woofers, near the outer edges of the unit, are actually built-in microphones that enable you to make stereo live recordings without having to connect separate mikes.

The upper section of the front panel contains a tuning knob for any of the six available RF bands (AM, FM, medium wave and three shortwave) whose frequencies are calibrated in the dial area. A stereo indicator light is located to the left of the frequency scales, while two meters are on the right. These meters are used to indicate reproduced level, recording input

level, signal strength during tuning and even battery condition. A pair of 2-inch tweeters flank the dial-scale area, and just below the left-channel tweeter is a momentary switch that provides dial-scale lighting. The dial light is normally off (unless this switch is depressed) to conserve battery life when the unit is being powered from D-cells.

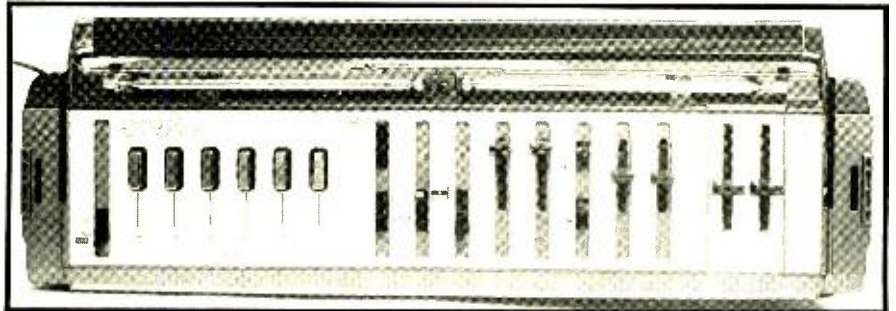
When the carrying handle is swung out of the way, a second panel on the top surface of the unit becomes accessible (see Fig. 2). This panel contains a power ON/OFF switch, six

line-out jacks located on the side of the unit.

The side panel, shown in Fig. 3, also contains line input jacks, twin external microphone input jacks, external speaker jacks and a switch labeled BEAT CUT to help eliminate beat tones that might occur when recording short-wave broadcasts. The alternate side panel (not shown) has an AC receptacle as well as a connector for a car-battery adapter that can also be used to power the model RC-828.

Binaural and biphonic

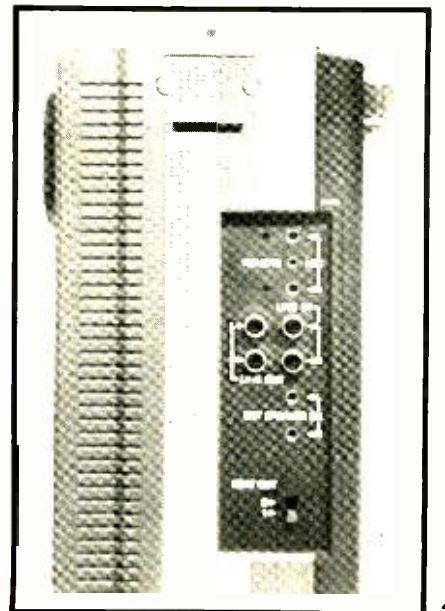
Binaural recording involves recording sound



2

pushbuttons that determine the desired frequency band, a three-position function switch (for selecting LINE-IN, RADIO or TAPE), a Binaural Equalizer/Meter switch (that determines the meter function and applies special equalization when you listen to binaurally recorded tapes), two mode switches (with settings for biphonic processing, expanded stereo listening, or ordinary stereo listening), individual channel recording-level slide controls, tape-selection switches (for standard ferric or CrO₂ tape) a record-selection switch (for activating an automatic level control or ALC circuit, if desired), separate bass and treble boost and cut controls, and separate master volume controls for each channel. The volume controls do not affect levels available at the

using two microphones that are spaced as far apart as two human ears, and often a dummy head is used for making such recordings. This dummy head is equipped with tiny microphones where ears are normally located. Until the advent of JVC's biphonic processor, such recordings needed to be heard via headphones.



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When headphones are used, the left ear of the listener hears only what was received by the left "ear" of the dummy head, and the same

MANUFACTURER'S PUBLISHED SPECIFICATIONS:

RADIO FREQUENCY BANDS:

FM: 88-108 MHz; **AM:** 540-1600 kHz; **MB:** 1.6-4.3 MHz; **SW-1:** 4.3-11.0 MHz; **SW-2:** 11-18.5 MHz; **SW-3:** 18.5-26.0 MHz.

TAPE RECORDER SECTION:

Frequency Response: (Standard tape), 50 Hz-11 kHz; CrO₂ tape, 40 Hz-12 kHz.
Wow and Flutter: 0.09% WRMS. **Signal-to-Noise Ratio:** 50 dB. **Fast Wind Time:** 110 seconds.

AMPLIFIER SECTION:

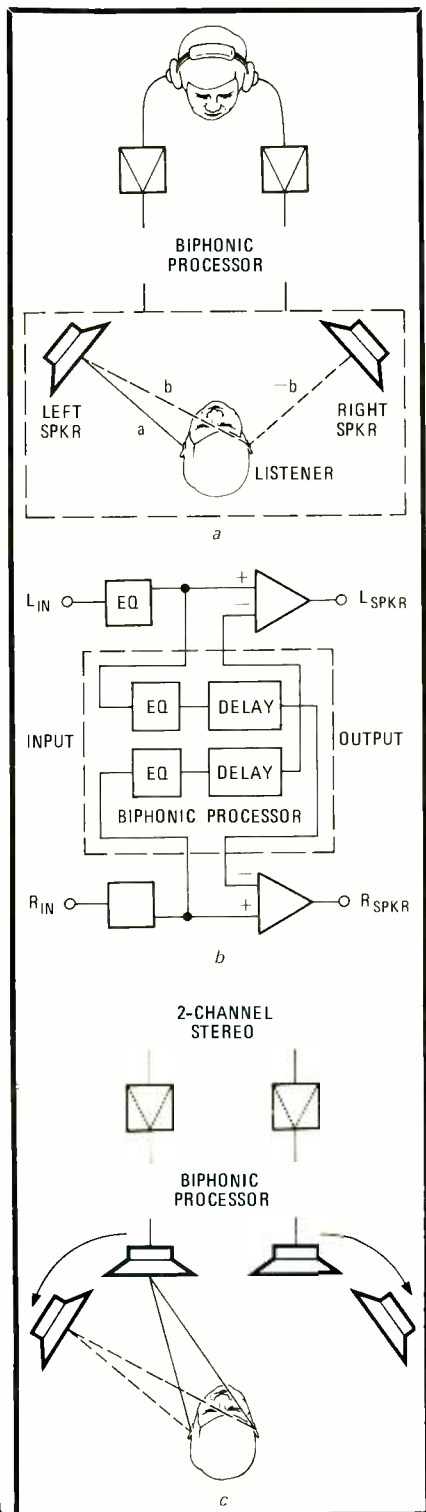
Power Output: 2.3 watts-per-channel (FTC). **Line Input Level:** 13 mV.
Line Output Level: 300 mV.

GENERAL SPECIFICATIONS:

Power Requirements: 12 VDC (8 D Cells) or 120 VAC, 60 Hz. **Power Consumption:** 19 watts. **Dimensions:** 1 1/2" W x 10 3/4" H x 5 inches D. **Weight (Including Batteries):** 14.1 lbs. **Suggested Retail Price:** \$250 (approx.).

holds true for the right ear-microphone. If binaural recordings are played back on a stereo speaker system, *both* the listener's ears hear sounds that were intended for only one ear, and the subtle timing and amplitude differences resulting from such closely spaced microphone recordings are all but lost.

Figure 4-a shows that the biphonic processor circuit contained in the *model RC-828* can reproduce binaural recordings via speakers



successfully. Crosstalk signals that reach the right ear of the listener from the left speaker (shown as line *b* in Fig. 4-a) is cancelled by introducing a time-delayed, out-of-phase quantity of signal *b* (shown as line *-b*). Fig. 4-b is the block diagram of the action of the biphonic

TABLE I
RADIO-ELECTRONICS PRODUCT TEST REPORT

Manufacturer: JVC

Model: RC-828JW

FM PERFORMANCE MEASUREMENTS

	R-E Measurement	R-E Evaluation
SENSITIVITY, NOISE AND FREEDOM FROM INTERFERENCE		
IHF sensitivity, mono: (μ V) (dBf)	3.5 (16.1)	Good
Sensitivity, stereo (μ V) (dBf)	5.0 (19.2)	Very good
50-dB quieting signal, mono (μ V) (dBf)	5.5 (20.0)	Very good
50-dB quieting signal, stereo (μ V) (dBf)	65.0 (41.5)	Fair
Maximum S/N ratio, mono (dB)	65	Very good
Maximum S/N ratio, stereo (dB)	63	Very good
Capture ratio (dB)	2.0	Good
AM suppression (dB)	50	Good
Image rejection (dB)	75	Good
IF rejection (dB)	80	Very good
Spurious rejection (dB)	85	Very good
Alternate channel selectivity (dB)	60	Very good
FIDELITY AND DISTORTION MEASUREMENTS		
Frequency response, 50 Hz to 15 kHz (\pm dB)	1.5	Very good
Harmonic distortion, 1 kHz, mono (%)	0.4	Excellent
Harmonic distortion, 1 kHz, stereo (%)	0.5	Excellent
Harmonic distortion, 100 Hz, mono (%)	0.4	Excellent
Harmonic distortion, 100 Hz, stereo (%)	0.63	Very good
Harmonic distortion, 6 kHz, mono (%)	0.45	Very good
Harmonic distortion, 6 kHz, stereo (%)	0.5	Excellent
Distortion at 50-dB quieting, mono (%)	2.5	Fair
Distortion at 50-dB quieting, stereo (%)	0.75	Good
STEREO PERFORMANCE MEASUREMENTS		
Stereo threshold (μ V) (dBf)	10 (25.2)	Very good
Separation, 1 kHz (dB)	39	Excellent
Separation, 100 Hz (dB)	28	Good
Separation, 10 kHz (dB)	31	Very good
MISCELLANEOUS MEASUREMENTS		
Muting threshold (μ V)	N/A	N/A
Dial calibration accuracy (\pm kHz at MHz)	200	Good

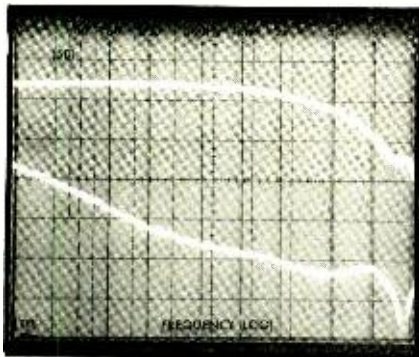
TABLE II
AMPLIFIER PERFORMANCE MEASUREMENTS

	R-E Measurement	R-E Evaluation
POWER OUTPUT CAPABILITY		
RMS power/channel, 8-ohms, 1 kHz (watts)	1.3	See text
RMS power/channel, 8-ohms, 20 Hz (watts)	0.4	See text
RMS power/channel, 8-ohms, 20 kHz (watts)	0.32	See text
RMS power/channel, 4-ohms, 1 kHz (watts)	2.0	See text
RMS power/channel, 4-ohms, 20 Hz (watts)	0.25	See text
RMS power/channel, 4-ohms, 20 kHz (watts)	0.35	See text
Frequency limits for rated output (Hz-kHz)	100-10	See text
DISTORTION MEASUREMENTS		
Harmonic distortion at rated output, 1 kHz (%)	0.3	Good
Intermodulation distortion, rated output (%)	1.5	Fair
Harmonic distortion at 1-watt output, 1 kHz (%)	0.3	Good
Intermodulation distortion at 1-watt output (%)	1.5	Fair
DAMPING FACTOR, AT 8 OHMS	25	
PHONO PREAMPLIFIER MEASUREMENTS		
Frequency response (RIAA \pm dB)	N/A	N/A
Maximum input before overload (mV)	N/A	N/A
Hum/noise referred to full output (dB) (at rated input sensitivity)	N/A	N/A
HIGH LEVEL INPUT MEASUREMENTS		
Frequency response (Hz-kHz, \pm dB)	See text-16 kHz	Good
Hum/noise referred to full output (dB)	70	Very good
Residual hum/noise (minimum volume) (dB)	75	Very good
TONAL COMPENSATION MEASUREMENTS		
Action of bass and treble controls	See Fig. 6	Excellent
Action of secondary tone controls	N/A	N/A
Action of low-frequency filter(s)	N/A	N/A
Action of high-frequency filter(s)	N/A	N/A
COMPONENT MATCHING MEASUREMENTS		
Input sensitivity, phono 1/phono 2 (mV)	N/A	
Input sensitivity, auxiliary input(s) (mV)	120	
Input sensitivity, tape input(s) (mV)	120	
Output level, tape output(s) (mV)	200	
Output level, headphone jack(s) (V or mW)	1.33 V/8 ohms	
OVERALL AMPLIFIER PERFORMANCE RATING		See text

processor. Because the equalization, crossfeed and delay constants are specifically calculated, it is necessary to sit at a distance of between 60 cm and 80 cm (24–32 inches) from the front surface of the *model RC-828*, but once you are positioned properly, the effect is uncanny. Sounds seem to emanate from points well beyond the extent of the receiver itself and you would swear that there are additional speakers located several feet apart. Similarly, ordinary stereo program material played back via the two speakers in the *model RC-828* can be “expanded” via the biphonic processor, as shown in Fig. 4-c. JVC has developed a single IC that performs all these functions for each channel.

FM measurements

Table I summarizes our measurements of the FM tuner section. Obviously, these measurements fall short of what one could expect from a state-of-the-art stereo FM tuner, but not by a significant margin. The 50-dB quieting exhibited in mono at a signal input of only 5.5 μV (20 dBf) is quite satisfactory (even in terms of high-fidelity components) as are the mono and stereo FM distortion figures, each well below 1.0%. Stereo separation (shown by the two sweep traces in Fig. 5) is also quite impressive, approaching 40 dB at midfrequen-



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cies and remaining above 30 dB at the more difficult 10-kHz test frequency. The automatic switchover from mono to stereo FM occurs at a signal input strength of 10 μV (25.2 dBf), and the maximum signal-to-noise ratio in both the mono modes is well above 60 dB. Tuning is relatively noncritical as a result of a moderately applied amount of nondefeatable AFC.

We did not measure the performance of the various medium-wave and shortwave bands of this receiver, but experienced excellent nighttime signal reception from such diverse locations as Great Britain, Japan, Israel, the Scandinavian countries and Germany using nothing more than the dipole antenna components on the receiver.

Amplifier measurements

The less said about the amplifier section of the *model RC-828* the better. The rating of 2.3 watts-per-channel applies to a 6-ohm load for which the output circuitry was optimized. Since our lab is equipped only with precision 8-ohm and 4-ohm loads, we could not obtain the rated power output level, but read 2.0 watts (with 4-ohm speaker loads) at a 1.0% distortion level. The efficient built-in two-way speaker system makes this nominal amount of audio power produce surprisingly loud sound levels before audible clipping or distortion is perceived. The saving grace is the fact that line output terminals are provided, and the signals available at these terminals are taken ahead of the power amplifier section and are, therefore,

TABLE III
RADIO-ELECTRONICS PRODUCT TEST REPORT

Manufacturer: JVC

Model: RC-828JW

CASSETTE TAPE DECK MEASUREMENTS

	R-E Measurements	R-E Evaluation
FREQUENCY RESPONSE MEASUREMENTS		
Frequency response, standard tape (Hz-kHz \pm dB)	40-12.0	Very good
Frequency response, CrO ₂ tape (Hz-kHz \pm dB)	40-13.0	Very good
	N/A	
Frequency response, other (see text) (Hz-kHz \pm dB)	See Fig. 7	N/A
DISTORTION MEASUREMENTS (RECORD/PLAY)		
Harmonic distortion at -10 VU (1 kHz) (%)	Std N/A CrO ₂ N/A	N/A
Harmonic distortion at -3 VU (1 kHz) (%)	1.5 6.5	Very good/poor
Harmonic distortion at 0 VU (1 kHz) (%)	2.3 6.5	Good/poor
Harmonic distortion at +3 VU (1 kHz) (%)	3.0 N/A	Good
SIGNAL-TO-NOISE RATIO MEASUREMENTS		
Standard tape, "Dolby" off (dB)	53	Excellent
Standard tape, "Dolby" on (dB) (dB)	N/A	N/A
CrO ₂ tape, Dolby off (dB)	51	Fair
CrO ₂ tape, Dolby on (dB)	N/A	N/A
MECHANICAL PERFORMANCE MEASUREMENTS		
Wow and flutter (% WRMS)	0.06	Superb
Fast wind and rewind time, C-60 tape (seconds)	100	Fair
COMPONENT MATCHING CHARACTERISTICS		
Microphone input sensitivity (mV)	0.5	
Line input sensitivity (mV)	120	
Line output level (mV)	200	
Phone output level (mV) (V)	1.33 γ	
Bias frequency (kHz)	100	
TRANSPORT MECHANISM EVALUATION		
Action of transport controls		Excellent
Absence of mechanical noise		Superb
Tape head accessibility		Excellent
Construction and internal layout		Very good
Evaluation of extra features, if any		Superb
CONTROL EVALUATION		
Level indicator(s)		Good
Level control action		Excellent
Adequacy of controls		Excellent
Evaluation of extra controls		Excellent
OVERALL TAPE DECK PERFORMANCE RATING		
		Very good

TABLE IV

RADIO-ELECTRONICS PRODUCT TEST REPORT

Manufacturer: JVC

Model: RC-828JW

OVERALL PRODUCT ANALYSIS

Retail price	\$250 (approximately)
Price category	Low
Price/performance ratio	Superb
Styling and appearance	Very good
Sound quality	Good
Mechanical performance	Excellent

Comments: The *model JVC RC-828JW* radio-cassette unit is one of the most unusual products ever tested in our labs. Far from being a typical low-fi portable, this all-in-one entertainment center lets you enjoy good FM and AM reception (not to mention shortwave reception) as well as cassette tapes wherever you are. Its amplifier section is not qualified to drive good external speakers, but thanks to the available line outputs, you can connect the system's program sources to your home-based hi-fi system when you are not traveling. With that sort of hookup, you can avail yourself of both the biphonic processing features and the expanded stereo features. The expanded stereo feature will be especially useful (even if you own a regular components system) if your speakers are too closely spaced in a small listening room. In addition, when you are away from your main hi-fi system, this feature lets you enjoy wide-separation stereo reception from FM sources and from cassettes even when you must monitor such results from the *model RC-828* itself.

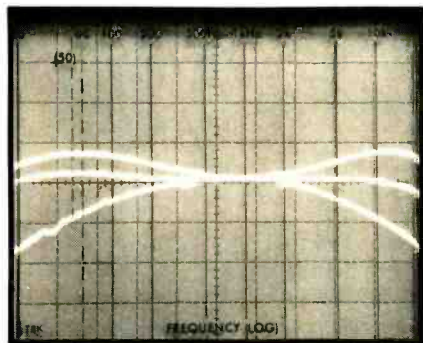
The relatively high quality of the FM section was a surprise and, if your present hi-fi system lacks a tuner, connecting the line outputs to the auxiliary inputs of your main amplifier will provide you with adequate FM and stereo FM reception when you are not using this portable away from home.

Despite the low power output of the self-contained amplifier, listening to the unit via its own speakers offers surprisingly faithful reproduction—thanks in part to the built-in twin two-way speaker systems. The twin-element built-in antenna was satisfactory for FM reception (the elements are expandable in length and can be oriented for any directional pattern) and even better for shortwave reception. By offering four separate shortwave bands, it is possible to adequately separate closely spaced shortwave signals.

distortion-free and suitable for a high-fidelity amplifier of your choice, or even to the auxiliary inputs of a home-based stereo high-fidelity receiver.

Line input jacks also enable you to copy program sources (such as records) from your hi-fi component system to the unit's built-in cassette deck. It is our experience with the *model RC-828* that the available external speaker jacks are not really necessary since few speakers could produce enough sound when fed with the low power levels provided by the power amplifier. Since the stereo-enhance feature is so effective, the use of external speakers fed directly from the built-in amplifier is not justified even on the basis of increased stereo separation.

Tone control range for bass and treble (as shown in the scope photo of Fig. 6) has been kept to a relatively low amount of boost capa-



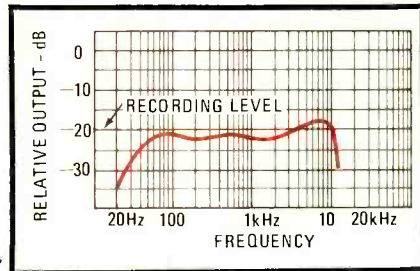
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bility so that you will not tend to overdrive the limited power capability of the built-in amplifier. A moderate amount of bass boost is permanently in the circuit (even with the BASS

control set to its flat or mid-position) to add a bit of depth to music reproduced over the relatively small built-in low-frequency speakers.

Cassette deck measurements

Table III summarizes results of tests made on the cassette recorder section. Record/play frequency response, using TDK-type AD ferric tape, extended from 40 Hz to 12.0 kHz, somewhat better than that claimed. Figure 7 shows a point-by-point plot of the response using this tape at a -20-dB recording level.



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While CrO₂ tape yielded somewhat better results (with response out to 13 kHz), we do not recommend using this type of tape with the *model RC-828* (at least not with our sample). For some reason, when CrO₂ tape was used, distortion at a 0-dB record level was excessive (we checked the switch setting several times to make sure that it was the correct one) and since the signal-to-noise ratio obtained using this tape (even referred to the higher distortion point) was not as good as that obtained using the high-quality ferric tape from TDK, there is no reason to use the CrO₂ variety.

Because of the extremely low price tag of the *model RC-828*, JVC undoubtedly could

not incorporate either Dolby noise-reduction circuitry or their own ANRS hiss-reducing scheme. We almost wish that they had added \$50 or so to the cost of the instrument and used either the Dolby or ANRS circuitry. For many listening applications, however, the signal-to-noise values greater than 50 dB are satisfactory, and the reasonably good frequency response and the superbly low wow-and-flutter characteristics of this cassette tape deck mechanism make it a true high-fidelity tape deck in every sense of the word. We have tested many \$200 cassette deck components that performed no better than the cassette section of the *model RC-828*, which, after all, contains six radio bands, a complete stereo amplifier/speaker system and that amazing biphonic circuitry.

Summary

An overall evaluation, together with our summary comments, is contained in Table IV. We do not suggest that the *model RC-828* is a suitable substitute for a true high-fidelity component system in your home. But there may be occasions when you want reasonably good reproduction such as when traveling, residing in a dormitory or in a hotel, or at other times when you are away from your primary hi-fi system. To ears accustomed to good clean sound, the *model RC-828* offers sound that comes amazingly close to the real thing and permits you to record cassette tapes to be played back later on your primary system or cassette stereo system in your car. JVC also offers an unusual headphone/microphone combination (*model HM-200E*) that allows you to make your own binaural recordings. That product, plus the *model RC-828*, makes an unbeatable combination. **R-E**

CONVERGENCE PROBLEM

After changing the picture tube in this Quasar TS-934, I have a convergence problem. I don't understand this.—A. R., East Haven, CT.

Sometimes, a new picture tube will have slightly different tolerances than the original tube. However, this doesn't seem to be too common. The most frequent cause appears to be something that was *disturbed* during the changing operation, such as loose contacts, broken wires, etc. Check out all the convergence pulses on the plug and make sure they're all normal.

(Feedback: "Right! I had a couple of problems in the vertical circuitry that were upsetting the pulses going to the convergence board.")

NO HORIZONTAL SWEEP

All I can get on this RCA CTC-71J is a thin, bright vertical line. The high voltage seems to be OK. The horizontal-deflection yoke winding checks out. What is this?—J. R., Detroit, MI.

You have several things that are normal (high voltage, etc.). So, the horizontal-output stage is working. Your yoke is OK, but you're not getting *through* it! Check the horizontal yoke return circuit, which goes through the pincushion transformers, a 0.55- μ F capacitor and the plug-socket contacts. Something in this circuit is probably open.

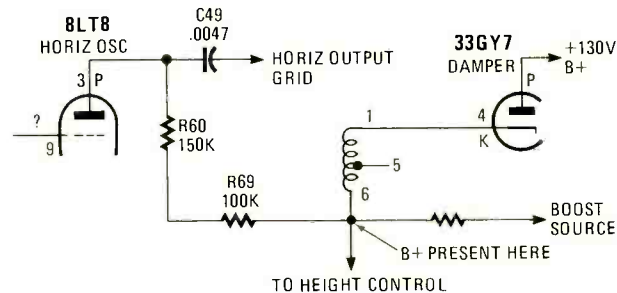
(Feedback: "Bingo! The PC-board conductor between the yoke return contact and the first pin transformer was open.")

VOLTAGE PROBLEMS

This Admiral T12H4-1A has no high voltage, no boost or plate voltage on the horizontal oscillator. I can get the B+ voltage as far as the height control. I tried changing tubes, etc.—no luck. No shorts that I can find. Where am I losing it?—B. T., Delta, IA.

Follow the DC path from the horizontal oscillator plate back

to the B+ voltage (see diagram). Note that there's no starter resistor connected directly to the B+ 130-volt line: this goes back to the damper cathode of the 33GY7. So, your damper tube *must* be conducting before you can get anything out of the horizontal-output stage at all. (Found a set once with a broken cathode ribbon on the damper tube! That was weird.)

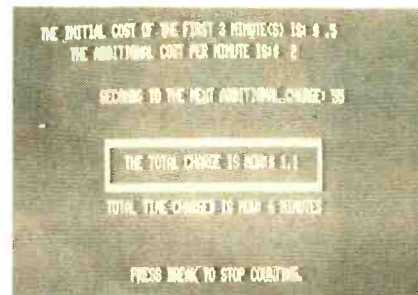


Your path leads from the oscillator plate through R60 (150K) and R69 (100K) to terminal 6 of the flyback, then through two small windings back to the damper cathode. You do have B+ voltage as far as terminal 6, which is connected directly to one end of the height control. So, that much is OK. Resistor R69 may be breaking down under load.

Start checking at terminal 6 on the flyback to see just where you lose the voltage. If the resistor checks good cold, try bridging another resistor across it to see what happens. There is one fairly unlikely possibility that should be checked if nothing else works. Read the control-grid voltage on the horizontal oscillator. If this measures too far *positive*, the tube will draw so much current that the plate voltage will drop to zero. There is no coupling capacitor to a high voltage that could short in this chassis, but you may have a leakage across the socket or something equally weird! **R-E**

Computer Program

Know what your long-distance calls cost as you make them



ACTUAL SCREEN PHOTOS of program as it is running during a real phone call. Display tells you actual cost of call moment by moment.

FRED BLECHMAN, K6UGT

WOULD YOU LIKE TO KNOW EXACTLY WHAT YOUR TOLL AND long-distance calls are costing you, *as you are making them*? With a simple computer program, you can synchronize yourself with the phone company billing machine, and have your computer screen print out your accumulated charge, and the time (in seconds) to the next added charge!

I used a Radio Shack TRS-80 computer, so the program is written in Radio-Shack Level 1 BASIC. If you have a different computer, you should be able to adapt it with no trouble. The program is written so that it allows the operator to request information on rates and discount periods, or bypass these instructions and go right into entering the three variables that determine the cost of the call: the initial time period, the initial time charge and the additional charge per minute. Once this data is entered, the operator places the phone call in the regular manner, and presses the ENTER key on the computer keyboard

when the party at the other end picks up their phone. That synchronizes your computer with the phone company timer—with absolutely **no connection to the phone lines!** From that point on, the computer display shows the initial cost and time period, counts down the seconds to the next added charge and (after the initial time period) displays the total time and cost continuously. To make the total cost more readable, it is enclosed in a graphic rectangle.

You must account for the operating speed of the computer. This is done in lines 130 and 140 by delaying the start of the countdown a few seconds, since it takes time for the computer to get to this point in the program on the first pass (mostly graphic delay). Also, I found my TRS-80 ran a FOR-NEXT loop at a rate of 490 loops-per-second, as shown in line 225. You might have to change this number slightly if your computer runs a bit faster or slower. This program uses 2495 bytes of memory of the 3583 available in the TRS-80 4K memory unit.

Telephone Toll Totalizer Program

© 1978 Fred Blechman

```

5 REM ★ PROGRAMMED FOR TRS-80 LEVEL 1 BASIC ★
10 REM ★ COPYRIGHT 1978 FRED BLECHMAN ★
15 REM ★ THREE LINES AFFECT THE TIMING OF THIS PROGRAM ★
16 REM ★ LINES 130 AND 140 DETERMINE THE STARTING COUNT★
17 REM ★ (THIS LETS YOU CORRECT FOR THE GRAPHIC DELAY) ★
18 REM ★ LINE 225 CONTROLS THE SECONDS COUNTING ACCURACY ★
20 CLS:P.:P.P.
21 P." TELEPHONE TOLL TOTALIZER":P.:P.
25 IN:"DO YOU WANT SPECIFIC INSTRUCTIONS? YES=1,NO=2":A
26 IF A=1 GOTO 300
28 CLS:P.:P.:P.
29 P." TELEPHONE TOLL TOTALIZER":P.:P.
30 IN:"WHAT IS THE INITIAL TIME PERIOD (MINUTES)":P
35 P.:P.
40 IN:"WHAT IS THE INITIAL CHARGE (CENTS)":I
45 P.:P.
50 IN:"WHAT IS THE ADDITIONAL CHARGE PER MINUTE (CENTS)":M
51 CLS:P.:P.:P.
55 P."WHEN THE PARTY AT THE OTHER END PICKS UP THE RECEIVER"
57 IN:"PRESS ENTER TO START TIMING. . . .":A$
60 CLS:P.:P.
70 P."THE INITIAL COST OF THE FIRST":P;"MINUTE(S) IS:$":I/100
80 P." THE ADDITIONAL COST PER MINUTE IS:$":M/100
90 P.AT 847,"PRESS BREAK TO STOP COUNTING."
100 C=I/100:D=1
110 FOR X=22 TO 93
115 SET (X,22):SET(X,28):NEXT X
120 FOR Y=22 TO 28
125 SET(22,Y):SET(93,Y):NEXT Y
130 T=56
140 IF P=3 THEN T=176
150 GOTO 221
200 P.AT 526,"THE TOTAL CHARGE IS NOW:$":C
210 SET(93,24):SET(93,25):SET(93,26)
215 P.AT 651,"TOTAL TIME CHARGED IS NOW:";P+D;"MINUTES"
220 T=59:D=D+1
221 C=C+(M/100)
225 FOR X=1 TO 490: NEXT X
230 P.AT 330,"SECONDS TO THE NEXT ADDITIONAL CHARGE:";T
240 T=T-1
250 IF T=-1 GOTO 200
255 GOTO 225
300 CLS:P:"YOUR CHARGES ARE BASED UPON THREE THINGS:"
310 P." (1) INITIAL TIME PERIOD (1 or 3 MINUTES)"
320 P." (2) INITIAL CHARGE (FOR THE INITIAL PERIOD)"
330 P." (3) ADDITIONAL CHARGE PER MINUTE AFTER INITIAL PERIOD"
340 P.:P:"IF YOU USE AN OPERATOR TO ASSIST YOU, THE INITIAL"
350 P.:P:"TIME PERIOD IS 3 MINUTES. DIRECT DIAL IS 1 MINUTE."
360 P.:P:"THE CHARGES ARE BASED ON THE DESTINATION CALLED. . . ."
370 P:". . . THESE ARE USUALLY LISTED IN THE FRONT OF THE PHONE BOOK"
380 P:". . . OR. . . CALL OPERATOR FOR THE RATES."
390 P."DO YOU WANT INFORMATION ON DISCOUNT PERIODS?"
400 P.:P:"YES=1,NO=2":B
410 IF B=2 GOTO 28
420 CLS:P:"THERE ARE TWO DISCOUNT RATES IN THE CONT. U.S.A.:"
430 P." 35% DISCOUNT: 5PM-11PM SUNDAY-FRIDAY"
440 P." 8AM-11PM HOLIDAYS":P.
450 P." 65% DISCOUNT: 11PM-8AM EVERY NIGHT"
460 P." 8AM-11PM SATURDAY"
470 P." 8AM-5PM SUNDAY"
475 P.:P:"CHARGES ARE BASED ON TIME AT CALLING POINT!"
476 P.:P.:P.
480 IN:"PRESS ENTER TO INPUT TIME AND CHARGE DATA. . . .":A$
490 GOTO 28

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DIGITAL HI-FI RECORDING

Its Time Is Now!



For years, audio engineers have sought practical ways to increase the signal-to-noise ratio and dynamic range in audio recordings. Here's a look at what's coming from the giants in the recording industry.

LEN FELDMAN
CONTRIBUTING HI-FI EDITOR

THREE SEPARATE EVENTS HELD IN THE last several months have convinced me that we are on the brink of a major revolution in recording technology: the High Fidelity Show held in Tokyo, Japan, in late September, 1977; a press conference held by the 3M Company in New York just prior to the Audio Engineering Society (AES) convention in early November, 1977; and the AES convention itself. The revolution, of course, has to do with the coming transition from analog audio recording to digital audio recording.

It is fitting that this transition should be taking place exactly 100 years after Thomas A. Edison inscribed those first recorded words onto a tinfoil cylinder. As crude as that first recording was, its basic concept has remained the same for a century. All presently available recordings are analog representations of sound waves. In the case of discs, the analog is represented by continuous wiggles or undulations in the disc's groove walls. In the case of tape, the analog consists of continuously varying magnetization patterns on a ribbon of tape coated with particles that can store such continuous magnetization patterns.

Analog-to-digital conversion

Any amplitude or number can be rep-

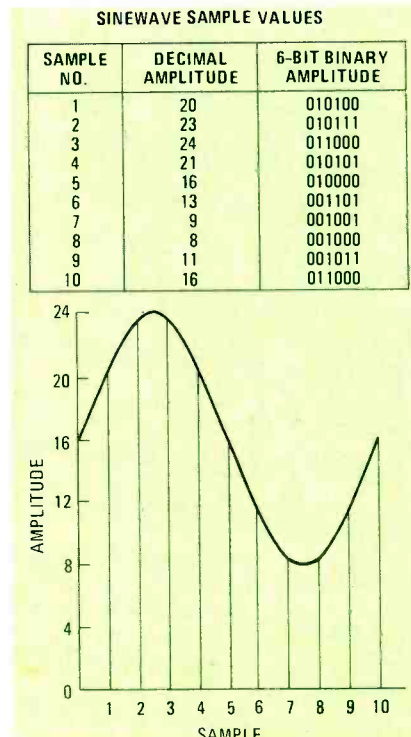


FIG. 1—SINEWAVES can be converted to digital form by sampling at specific points and then converting the amplitude into a binary representation, as shown in the accompanying table.

resented by a binary numbering system in which only "0" and "1" are used. The way to represent a continuous sinewave (see Fig. 1) in binary language would be to sample the amplitude at several points and express the results in binary form. If a 6-bit system is used, for example, you could express any amplitude level from 0 (000000) to 63 (111111). The table in Fig. 1 shows how a waveform is expressed in binary language, using a 6-bit system.

Assume, also, that you are sampling at a rate of 50,000 times-per-second. If the single sinewave shown in Fig. 1 were part of a 5000-Hz tone, you would have ten chances to sample one complete wave, as summarized in the table in Fig. 1. Note, however, that some of the bit values expressed are not totally accurate. In any bit or binary-number expression, only whole numbers can be used (as shown in the accompanying decimal system equivalents of the 6-bit code). Thus, to make the digital equivalent of any waveform *more* accurate, *more* bits must be used. Furthermore, in the example, a 5000-Hz signal has been used. If the same sampling rate (50,000-per-second) were used to dissect a 20,000-Hz waveform, you would only have 2½ chances per cycle to sample the waveform amplitude. Even though it would seem that such a low

sampling would not be enough to represent a pure 20,000-Hz sine wave, it is possible to reconstruct the 20-kHz sine wave as long as it has been sampled at least twice (once for its positive-going peak amplitude, and again for its most negative value).

Advantages of digital recording

Once an analog or continuous waveform signal has been converted to a series of pulses (for example, a positive pulse for "1" and an absence of a pulse for "0"), these pulses can be recorded on tape. For tape recording, the advantages are obvious since the pulses can all be recorded at a constant amplitude. That means that every pulse can be recorded at a level well above the residual noise level of the tape and well below the saturation level. Thus, the system's dynamic range is no longer limited by the parameters of the magnetic tape itself. It can be shown that in a digital recording system, each bit of storage capability offers 6 dB of dynamic range. So, for 96 dB of dynamic range (that's about 30 dB better than the best analog tape recorders can deliver), all you have to do is create a 16-bit digital recording system.

The 3M/BBC system

The 3M Company and the British Broadcasting Company have jointly developed such a recording system. Figure 2 shows their new Digital Audio Mastering System that is designed to replace analog systems in professional recording studios.



FIG. 2—NEW DIGITAL AUDIO MASTERING SYSTEM is manufactured by the 3M Company and contains 32 audio tracks.

It records 32 audio channels on special 1-inch-wide tape. A matching stereo recorder (not shown) is included in the complete digital recording system, comparable in size and style to present analog professional mix-down recorders. Figure 3 shows the operator's panel for the system. This panel can be removed from the cabinet for use at nearby remote locations.

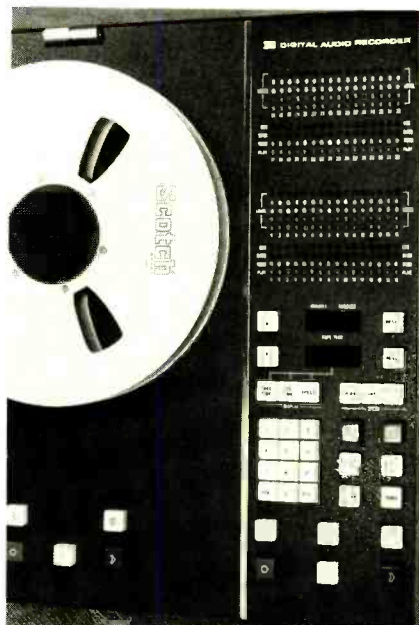


FIG. 3—FRONT PANEL of the new Digital Audio Recorder.

The frequency response is flat within 0.3 dB from 30 Hz to 15 kHz, and within 3 dB out to 20,000 Hz. Signal-to-noise ratio is an incredible 90 dB or better, and harmonic distortion is less than 0.03% at any frequency from 20 Hz to 20 kHz at +15-dBm input levels. Intermodulation distortion is equally low for any two frequencies mixed. By way of comparison, Fig. 4 shows the harmonic distortion in a recorded 5-kHz pure tone for a top-quality analog recording system com-

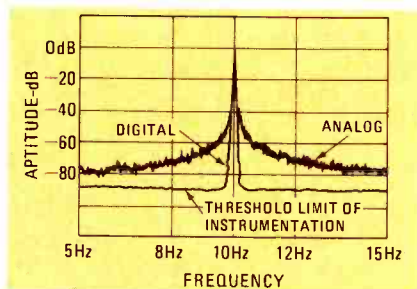


FIG. 4—HARMONIC DISTORTION curves of an analog recorder vs. a digital recorder.

pared with that for the digital mastering system. Second- and third-harmonic components for the analog system are around 55 dB below the 5-kHz fundamental signal, whereas in the case of the digital system, these distortion components are some 80 dB or so below reference record level.

There are other advantages to the system. Because equal-amplitude digital pulses are used to represent all audio signal waveforms, the problem of tape saturation at high frequencies becomes meaningless. The full dynamic range of all types of music (even electronic music, which often has disproportionately high levels of high frequency signals) can be recorded without using compressors, limiters or manual gain riding. In professional recording, it is a common practice

to make several dubbings from the original multitrack master recording before arriving at the final two- or four-channel mix-down recording. In an analog system, each such dubbing adds approximately 3 dB of noise. In a digital system, any number of dubbings can be made with absolutely no degradation of signal-to-noise ratio. Furthermore, using a digital system eliminates such problems as tape print-through, in which high magnetization levels on one layer of tape can become imprinted on adjacent layers to produce pre- or post-echoes.

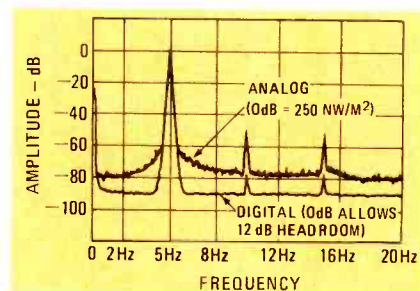


FIG. 5—SPECTRAL PURITY of an analog recorder vs. a digital recorder.

Because of slight motion irregularities in a conventional analog system, recorded tones are often accompanied by modulation noise—that is, the noise tends to rise at frequencies just above and below the frequency of the recorded tone, as shown in the upper trace of Fig. 5. By comparison, the lower trace, representing the results obtained using the 3M digital mastering system, shows no measurable modulation noise surrounding the same 10-kHz recorded test frequency. This spectral purity is a distinctly audible improvement when listening to playback of digitally recorded audio programs.

Wow-and-flutter for the new system is not measurable. Obviously, even the best tape transport is not totally free of such speed variations, but in a digital system it is possible to store the millions of bits corresponding to the audio signal in short-term memory, and then to release them in perfect, uniform sequence determined by a reference quartz-crystal oscillator clock.

Tape dropouts can, of course, pose a serious problem in a digital system if the missing information is not compensated for. In an analog system, a short dropout is often undetectable, since our ears tend to fill in the gaps if they are sufficiently short. In a digital system, dropouts mean an absence of pulses, which the system would interpret not as missing music but as a different bit-code altogether. Accordingly, in the 3M system, each track contains a degree of track redundancy and, with the aid of computer memory, such dropout errors are automatically corrected. This requires 1½ times the normal tape length needed if such a correction scheme were not included, but it is well worth the extra tape. The tape in

the 3M system operates at a speed of 45-inches-per-second; 7200 feet of recommended tape wound on a 12½-inch reel provides 30 minutes of recording time. Of course, the 3M Company has developed a high-storage density tape to go with the machines. At the AES Convention we were also shown Ampex's new Series 460 digital audio recording tape specifically designed for digital-recording applications. Digital dropout for this new tape is less than an average of 1 μ s per 100-feet of tape per track, measured across the entire tape width and length.

Consumer digital systems

The new 3M professional mastering system is expected to cost a recording studio around \$150,000. If this seems extremely high to home recordists, it should be pointed out that a comparable first-quality analog system costs around \$80,000, and provides 24-track capability (as opposed to 32 tracks), using 2-inch-wide tape (as opposed to 1-inch tape). Therefore, in terms of per-track cost and tape usage, the prices are not that disparate. Still, this new system would probably never be bought by a home recordist.

On the consumer front, things seem to be moving along just as quickly. Sony, for example, has developed a PCM add-on tape accessory that can be attached to its Betamax home videotape recorder. For some reason, the Japanese digital audio recording systems are called PCM (*Pulse Code Modulation*), but these systems are, in fact, true digital tape recording systems and operate on the same digital code principle as the more elaborate multitrack 3M system.

At the AES show, we also saw a PCM recorder from Mitsubishi, as well as a new PCM cassette tape deck that features an oversized cassette of specially formulated 2-hour tape, a frequency response from DC to 20 kHz, ± 0.5 dB, 80 dB of dynamic range, distortion of under 0.3% and undetectable wow-and-flutter. This cassette deck uses 13-bit digitization, a 47.52-kHz sampling frequency as well as a sophisticated system of dropout compensation.

Soundstream, Inc. displayed a 4-channel digital mastering system with a digital editing feature that produces inaudible splices with resolution to less than 30 μ s. The editing system can even be used to produce splices at different places on each track and, since the tape is never physically cut, a selection can be repeatedly spliced and auditioned while the all-electronic digital editing system keeps track of acceptable final results.

Direct-to-disc

It can be argued that digital tape recording will mean little if, in the end, these superior master tapes are used to cut only conventional lacquer discs from which vinyl pressings are made. However,

this is not really the case. Readers may be familiar with some of the new limited-edition direct-to-disc records that are now available. In this process, a master lacquer disc is cut during an actual musical performance expressly to eliminate the intermediate analog-tape recorder process. Anyone who has heard these discs can attest to their better sound quality. Making a direct-to-disc record, however, means bypassing all the sophisticated multitrack capabilities that have been developed over the years and are responsible for present-day high-quality recordings, since later editing or mixing is not possible. By introducing digital

Home disc system

The final development in this conversion from analog-to-digital audio technology has already appeared. Several Japanese companies have already perfected a PCM disc and its required player. At the AES convention, two seemingly identical machines were shown—one by Teac, the other by Mitsubishi. These discs contain tiny pits or photoetched depressions beneath a PVC coating. A monochromatic laser beam reads the presence or absence of these tiny pits on the disc; these read-outs are then converted into pulses or digital code that is then amplified and converted back to audio waveforms.

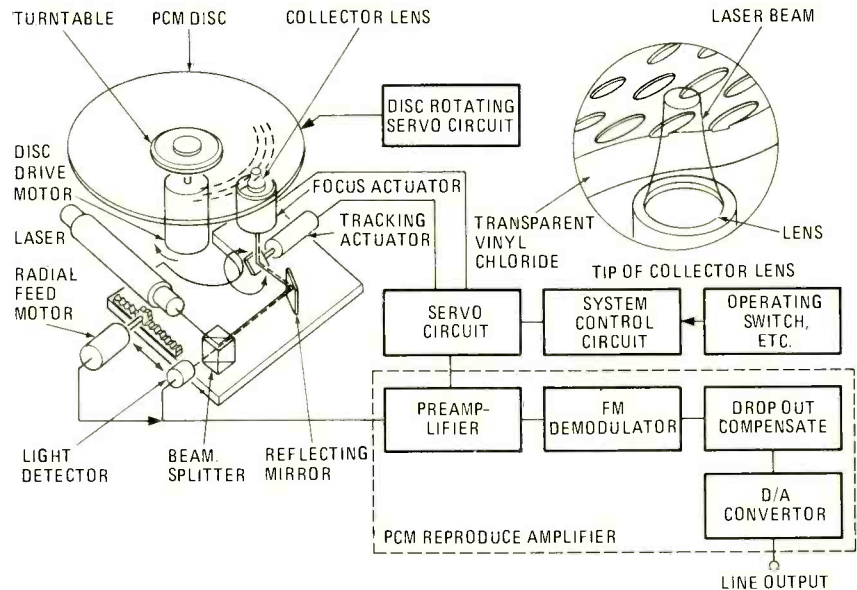


FIG. 6—PCM RECORD PLAYER uses a laser to reproduce music from a transparent vinyl chloride disc.

master tape recording as the initial studio recording technique, the advantages of multitrack recording can be combined with the sound quality of direct-to-disc techniques since the digital master tape system introduces no intermediate degradation of the music signals.

The end of signal processing?

Because of the limitations of analog tape recorders, recording studios have used many kinds of signal processors to try to get around the noise and dynamic-range limitations of the analog tape system, using such schemes as Dolby noise reduction and *dbx* compressors and the like. The widespread use of digital master recording systems could well make such devices obsolete—at least insofar as studio use is concerned. Of course, as long as vinyl long-playing discs are used as the final playback software, some compression of dynamic range will still be necessary before the master disc is cut, since even the best LP's are not capable of much more than 65 dB of dynamic range. However, digital master taping techniques may insure that more discs will reflect that full range along with lower distortion.

Figure 6 shows a simplified diagram of the laser PCM disc system. The turntable platter rotates at a speed of 1800 RPM, controlled by a precise quartz-crystal oscillator locked servo system. For some idea of the system's potential performance, consider the following projected specifications: a dynamic range of 98 dB; a frequency range from 10 Hz to 20 kHz within 0.5 dB; distortion less than 0.1%; and wow-and-flutter (typically) less than 0.001%. Thirty minutes of playing time can be had from a single disc.

While it may be several years before digital PCM discs are as common as LP records, the possibilities exemplified by these outstanding specifications are too good to be ignored. The lack of contact between the disc and the pickup arm means that there should be no performance deterioration no matter how many times the disc is played. There will be no tracking distortion, pinch effect, or tracking error as with mechanical systems. Nor will there be any acoustic feedback (howling) and surface noise in a disc that requires no more delicate handling than present-day LP's. This is a very attractive prospect for the program source of the future.

R-E

Little Pro Beep

Easy to build programmable beeper can be used as the heart of a call system

ISAAC QUEEN

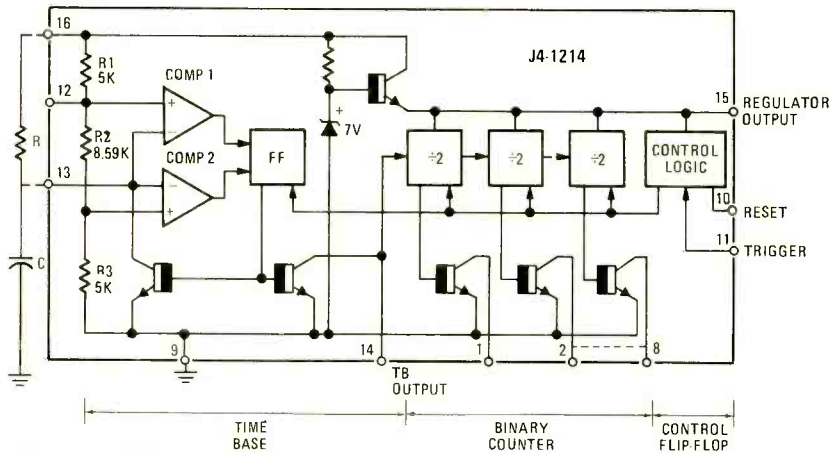


FIG. 1—CALECTRO J4-1214, simplified diagram. (Courtesy Calectro Handbook No. 98-102.)

CALL SYSTEMS ARE COMMONLY USED IN supermarkets and in department stores to summon personnel to the office or the phone. A bell repeats an identification number until the desired person answers. This very simple system may prove ample for many call purposes.

It is a *programmable beeper*. It can be switched to beep once, twice or four times, repeating the signal automatically. If desired, it may be set for a continuous sound. In addition, there is a choice of two audio frequencies—a buzz (about 105 Hz) and a tone (about 210 Hz). A total of eight distinct signals is available.

The beeper uses a programmable counter, Calectro type J4-1214 (an equivalent to the XR-2240), the simplified diagram of the programmable counter is shown in Fig. 1. The timebase section functions somewhat like that of the popular 555 timer. Capacitor C and resistor R are external timing components. The capaci-

tor charges through the resistor. When terminal 13 reaches the critical value, comparator-2 triggers the flip-flop. C is quickly discharged, a negative pulse appears at pin 14, and the sequence repeats. The formula for the time period is $T =$

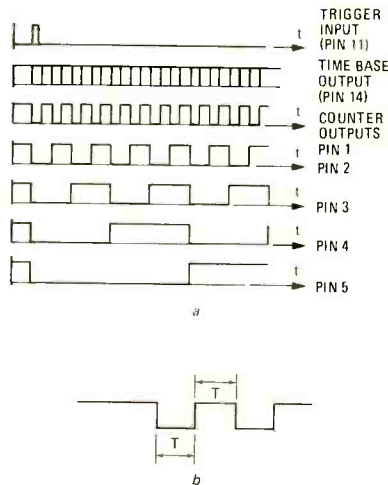


FIG. 2—TIMING DIAGRAM. The counter outputs are shown in a, and the output at pin 1 is shown in b.

RC (seconds = megohms \times microfarads).

The timebase pulses feed a string of eight divider stages (only stages 1, 2, 8 are shown in the diagram). Their outputs are available at pins 1 through 8, respectively. Each pin has half the frequency of the previous one. All eight pins are normally at a logic-1 level. At the trigger instant, all the pins go to a logic-0 level (to ground). Refer to Figs. 2-a and 2-b. For example, pin 1 remains at a logic-0 level for an interval T, then snaps back to a logic-1 level for a time T, and so on. At pin 2, the interval is 2T, at pin 3 it is 4T, etc.

The squarewave outputs from J4-1214 are available from pins 1 through 8, individually or in any desired combination. Figure 3 shows how to tie pins 1 and 8 to

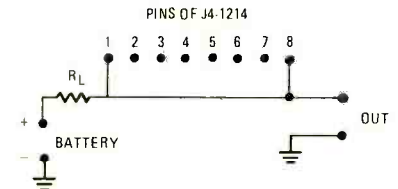


FIG. 3—STRAPPING TECHNIQUE.

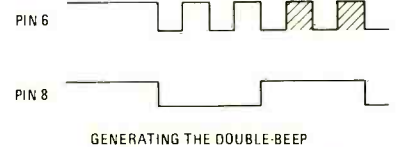


FIG. 4.—GENERATING DOUBLE BEEPS.

the output. The AC output appears across load R_L and the output terminals. The battery is assumed to have negligible impedance.

Figure 3 is a wired-OR connection. If either pin 1 or pin 8 goes to a logic-0 level, there can be no output at the output terminals. As an example, take R *continued on page 98*

PARTS LIST

All resistors 1/4 watt, all capacitors 50 volts.

- R1—47,000 ohms
- R2—20,000 ohms
- R3, R4—10,000 ohms
- C1, C2, C3—.05 μ F
- C4—0.1 μ F
- J1—earpiece jack
- J2—phono jack
- S1, S3—SPST toggle switches
- S2—SPDT toggle switch with "off" center position.
- IC—J4-1214 (Calectro)
- 16-pin DIP socket for above
- IC socket adapter (Radio Shack 276-024)X
- Chassis box, metal, 3/4 \times 2 \times 1 1/2 inches
- Miscellaneous hardware

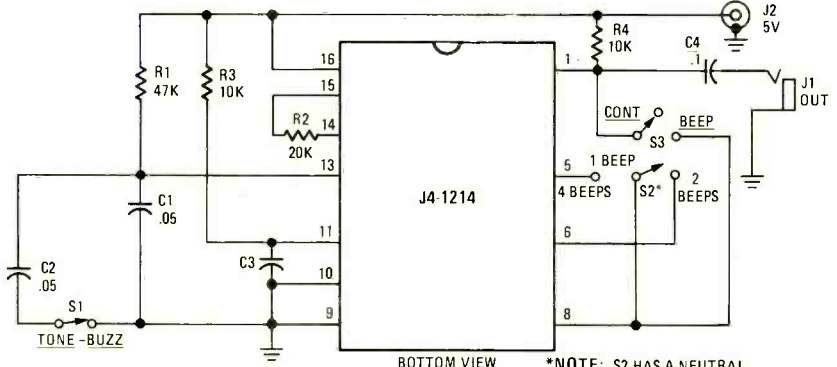


FIG. 5.—BEEPER, the complete schematic.

*NOTE: S2 HAS A NEUTRAL "OFF" POSITION

Long Duration IC Timers That Really Work!

Programmed delays and intervals of many hours can be precisely timed by combining the short-term accuracy of the one-shot multivibrator with the precision of the binary counter. Here is how the setup operates.

JOSEPH J. CARR

ELECTRONIC TIMER CIRCUITS USUALLY CONSIST OF MONOSTABLE multivibrators (i.e., one-shots), and their periods are controlled by an R-C time constant. The 555 IC timer, for example, has an output duration of 1.1 RC. Most literature dealing with 555 applications claim that 10 megohms and 100 μ F are the maximum values for R and C, respectively. However, at these maximum limits, *precision* components are difficult to obtain and, when available, tend to be expensive. Precision resistors in the 10-megohm range are relatively available, but capacitors with better than 5% tolerance in the above 1- μ F range become increasingly difficult to obtain as the capacitance value goes up. In addition, many capacitors, even tantalum types, tend to be leaky; this deteriorates the precision of the R-C circuit. Also, most electrolytic capacitors are rated to be within -20% to +100% in the over 10- μ F region.

Using the maximum values specified by 555 manufacturers produces an output period of $(1.1) (10^7) (10^{-4}) = 1100$ seconds, or about 18 minutes. If you cascade several 555 devices so that the first device triggers the second and so forth, you can obtain long time delays between the occurrence of the trigger and the trailing edge of the last output pulse; these signals could be gated together to produce a long-duration timer using lower-value components of higher precision. However, even if six 555 IC's were connected in that manner, they would only produce a duration of not quite two hours.

A better alternative is to use a *countdown* technique in which a precision timebase oscillator drives a binary or BCD counter. If we build this circuit using TTL or CMOS IC's, the timer will accomplish our purpose, but still have a relatively high IC count.

The Exar Integrated Systems XR-2240, XR-2250 and XR-2260 timers (with counterparts being Intersil's 8240, 8250 and 8260, respectively) contain a timebase that is similar to the 555 timer, as well as a built-in binary or BCD counter.

Figure 1 is the block diagram of these IC's, and Fig. 2 shows the simplified internal circuitry for the XR-2240. The internal circuitry for the XR-2250 and XR-2260 is similar, except that the binary counter is BCD-coded, and the regulator output (pin 15) becomes an overflow output for the counter.

There are three basic sections in the timer IC: the timebase, the binary counter and the control logic (see Fig. 1). The timebase is an R-C multivibrator that produces output pulses with a duration equal to $R \times C$. The counter in the XR-2240 is a straight 8-bit binary counter with outputs weighted in the standard $2^0, 2^1, 2^2, 2^3, 2^4, 2^5, 2^6$ and 2^7 format to produce time

weights of 1, 2, 4, 8, 16, 32, 64 and 128 times the timebase period.

The XR-2250 uses a BCD-weighting in which pins 1 to 4 form the least-significant digit (weighted 1, 2, 4 and 8), and pins 5 to 8 form the most-significant digit (weighted 10, 20, 40 and 80). The XR-2250 produces output times that are weighted 1 to

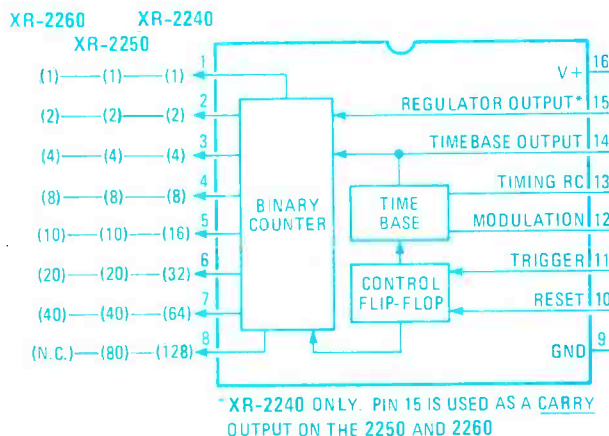


FIG. 1—TIMER IC PINOUT diagram for the XR-2240, XR-2250 and XR-2260. Numbers in parentheses are the binary weighted time durations.

99 times the timebase period.

The XR-2260 timer is similar to the XR-2250, except that the most-significant digit is limited to a maximum count of 5 (with a weighting of 50); output pin 8 on the XR-2260 is not used. This IC is designed for use in hours, seconds and minutes timers.

In all three IC timers, the timebase oscillator produces the pulses that are fed to the counter input, and the control flip-flop triggers the timebase and sets all counter outputs initially to 0.

The XR-2260 timebase, like the popular 555, consists of two voltage comparators, a flip-flop and two NPN transistors (see Fig. 2). This design allows an R-C timer to achieve better stability because there is less dependence upon temperature fluctuations and, especially, on changes in the power-supply voltage.

Transistor Q1 in the timebase is the discharge switch, which keeps capacitor C discharged whenever Q1 is turned on. Transistor Q2 is an open-collector output switch that drives the input

of the binary counter. In normal operation, where the counter is driven by the internal timebase, the collector of Q2 (pin 14) is connected to the regulator output through a 20,000-ohm resistor.

The two voltage comparators are biased by voltage-divider network R1-R3. The noninverting (+) input of comparator 2 is biased to approximately 0.27V (pin 16 is V+).

When a trigger pulse is applied to the control-logic section, the timebase flip-flop is placed in the SET condition, thereby turning off transistors Q1 and Q2. In this condition the timebase output terminal is forced high, and capacitor C starts charging through resistor R from the V+ voltage source. When the capacitor voltage reaches 0.27V, comparator 2 toggles, reset-

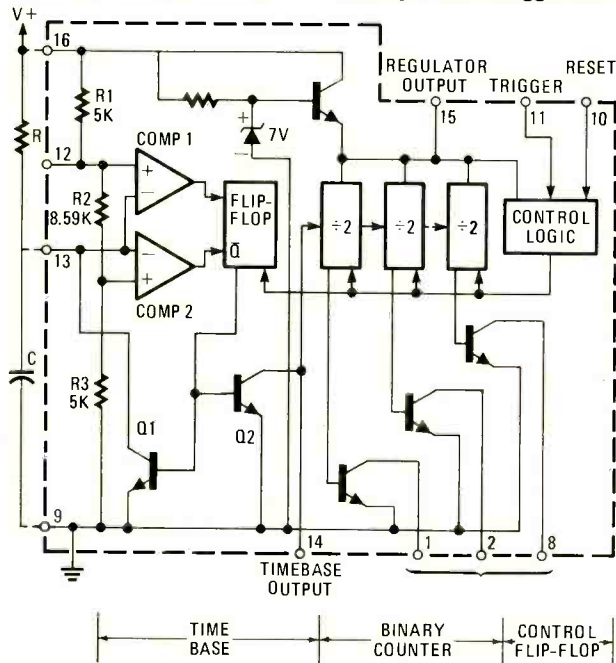


FIG. 2—XR-2240 IC contains a timebase circuit and a binary counter.

ting the flip-flop and turning on Q1 and Q2. When Q1 is turned on, the capacitor is discharged rapidly. When Q2 is turned on, the timebase output terminal drops low. The length of time that the timebase output terminal remains high is the R-C time constant, or $R \times C$ (R is in ohms, C in farads for time in seconds).

Each counter output is an open-collector NPN transistor capable of sinking a maximum of 5 mA, and must be connected to the V+ voltage through a pull-up resistor. Since the IC will operate with V+ supplies between 4.5 volts and 18 volts DC, the minimum value of the pull-up resistor is 900 ohms at the minimum supply voltage and 3600 ohms at the maximum supply voltage. A resistance of 10,000 ohms is typically specified.

The basic timing circuit is shown in Fig. 3-a. When a trigger pulse is received, the timebase begins producing pulses and the counter starts incrementing. It continues to increment until a positive-going reset pulse is applied to pin 10. Figure 4 shows the XR-2240 timing diagram for the first eight timebase pulses (only pins 1 to 4 are involved, pins 5 to 8 remain low until after the eighth timebase pulse).

If two or more outputs are wired together through a single pull-up resistor (a wired-OR configuration), then the output remains low for as long as any single output is low. This feature allows you to program the output period by connecting together the appropriate pins and selecting the timebase frequency. If only pin 1 is used, then the output duration is $1 \times RC$, and if all eight pins are wire-OR'ed together, then the output duration is $255 \times RC$.

If, for example, you wanted a 67-second timer, you could set the R-C time constant to 1 second (i.e., $R = 1$ megohm, $C = 1 \mu F$, or $R = 10$ megohms and $C = 0.1 \mu F$), and then wire-OR

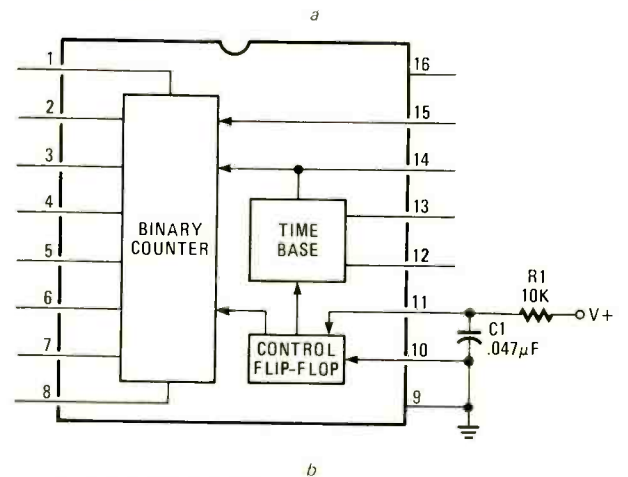
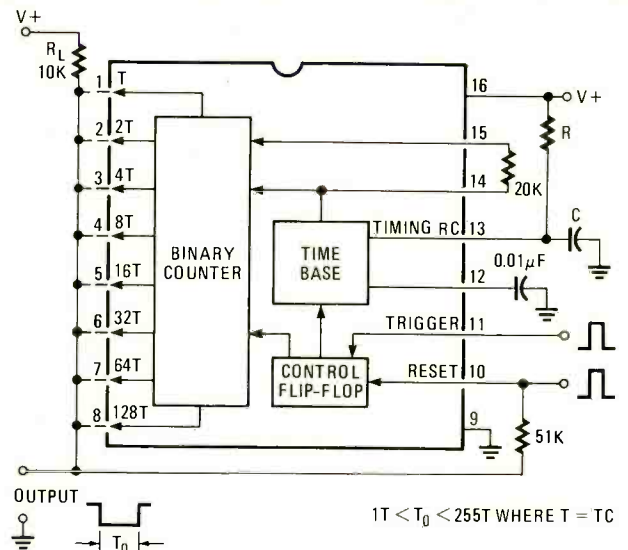


FIG. 3—EXTERNAL CIRCUITRY required for timer IC is shown in a. Timer starts counting after receiving trigger pulse. Modification shown in b is for free-running mode.

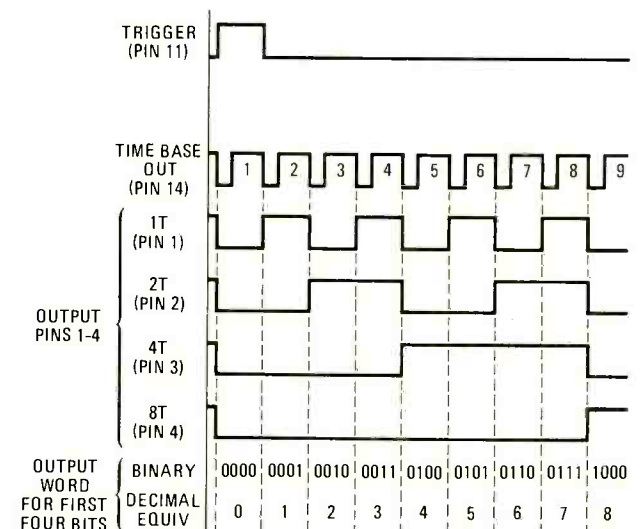


FIG. 4—TIMING DIAGRAM FOR XR-2240 for the first eight clock pulses after triggering.

pins 1, 2 and 7 together to obtain weightings of 1, 2 and 64, respectively. The output duration will then be $(1 + 2 + 64)RC$, or $67RC$. Since the R-C time constant is 1 second, the total output duration is 67 seconds. Any time duration between $1RC$ and $255RC$ can be programmed using the XR-2240, or between $1RC$ and $99RC$ using the XR-2250, or between $1RC$ and $59RC$ using the XR-2260.

The circuit shown in Fig. 3-a is self-resetting because reset

terminal pin 10 is connected to the output through a 51,000-ohm resistor. At the end of the output period, the output terminal snaps high, forming the positive-going reset pulse that shuts off the timebase oscillator.

The modification of the basic circuit shown in Fig. 3-b allows for continuous operation of the timebase. The trigger is tied permanently high, and the reset is grounded and therefore inhibited. You could also use this circuit in other applications besides simple timing procedures.

Precision in an R-C oscillator is often difficult to achieve because of changes in the power-supply voltage and thermal changes in the component values. The comparator timebase used in the Exar IC's and the 555 reduces the effects of power-supply changes because they operate on a fixed percentage of the V+ voltage, rather than being dependent upon the absolute voltage. The timebase can be partially freed of thermal effects by using precision (or at least metal-film) resistors and either silver mica, polycarbonate or polyethylene capacitors in the R-C network.

Figure 5 shows two alternate approaches to achieving greater precision. In Fig. 5-a, the internal timebase is disabled and an external clock is used instead. The external clock can be derived from the 60-Hz power line for moderate precision, or from a crystal oscillator/divider chain similar to the timebase used in a frequency counter for high-precision applications.

Figure 5-b shows an external reference being used to synchro-

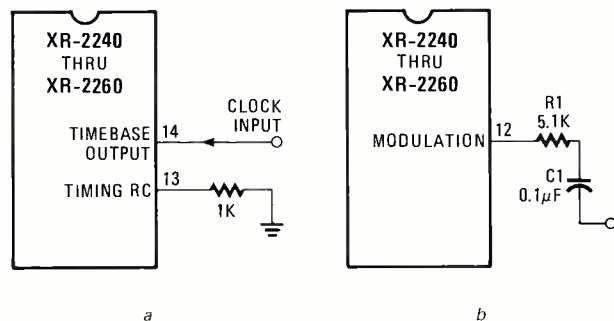


FIG. 5—EXTERNAL TIMEBASE can be connected to timer IC as shown in a. Internal timebase can also be synchronized to an external reference as shown in b.

nize the internal R-C timebase oscillator. A series network consisting of a 5100-ohm resistor and a 0.1- μ F capacitor is connected between the reference frequency source and the timer IC's modulation terminal, pin 12. The frequency source must have a pulse amplitude of at least 3 volts P-P and a duration between 0.3RC and 0.8RC. The repetition rate of the sync pulses can have a 12:1 ratio with the output pulses.

Cascading timers

The output-pulse duration can be extended by cascading two or more timers. Because it is possible to wire-OR the outputs, this procedure is easier than in 555 circuits and results in longer durations-per-IC because of the countdown construction of each timer IC. Figure 6 shows two cascaded XR-2240 timers. In this scheme timer IC1 is the input section, therefore, its timebase oscillator is operating. The time period is set by $R1 \times C1$. The 128 ($R1 \times C1$) terminal (pin 8) forms the external timebase for IC2.

Both timer IC's have their respective trigger and reset terminals connected in-parallel. The IC2 outputs are OR'ed together through resistor R4, forming the timer output terminal. Feedback resistor R5 resets both IC1 and IC2 at the conclusion of the time period.

The total period generated by Fig. 6 is $(2^6) (R1 \times C1)$, or 65,536 ($R1 \times C1$)! If the $R1C1$ time constant is set to 1 second, then the total period is 65,536 seconds, or more than 18 hours! Cascading allows you to use very short timebase periods to produce extremely long output durations. Of course, using short timebase periods means that the values used for R1 and C1 are small, which results in the use of more readily available preci-

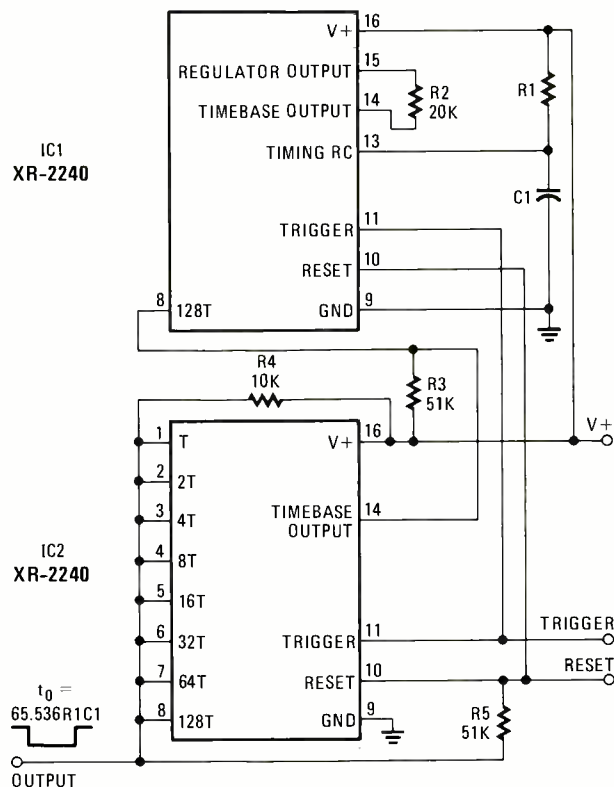


FIG. 6—EXTENDED DELAYS can be obtained by cascading two (or more) timer IC's together.

sion stable components.

For example, let's design a simple 1-hour (3600-second) timer using the circuit shown in Fig. 6. The timebase period (i.e., $R1 \times C1$) must be 3600/65,536, or 0.055 seconds. We must select a combination of R1 and C1 that will produce a 0.055-second time constant. The proper combination can be found by some pencil-and-paper trial and error calculation (thinking with a pencil saves hours of breadboarding and doesn't risk burning out IC's).

First, select a convenient value for C1. You could just as easily select a standard resistance, but it is easier to use a potentiometer to trim a fixed resistor to an oddball value than it is to use a variable capacitor to trim a standard-value capacitor to an oddball value. So, let's select 0.1 μ F for C1.

Next, calculate the value of the resistance required from the equation $R1C1 = 0.055$. Since C1 is 0.1 μ F (10^{-7} Farads), the formula becomes $R1 = 0.055/10^{-7}$, or 5.5×10^5 ohms (550,000 ohms). If R1 is made from a standard 510K resistor and a 100K trimmer (a ten-turn type is preferred), then the timebase period can be adjusted to a precise 0.055 second. Make sure not to use carbon composition resistors (metal film is better) for R1 or unstable capacitors for C1.

Programmable timers

Figure 7 shows a timer that can be programmed using an external digital word. The basic circuit is the one shown in Fig. 3-a, except that the output terminals are connected to a pair of 7485 four-bit magnitude comparators.

The 7485 is composed of a set of TTL exclusive-OR gates connected to compare two 4-bit words and issue outputs indicating whether word A is greater than word B, word A equals word B, or word A is less than word B. Each IC has both magnitude inputs and outputs that allow several 7485 IC's to be cascaded. In this case, you must compare the 8-bit output of the counter with an 8-bit command from an outside source such as a micro-computer output port or a set of thumbwheel switches.

The timing diagram for Fig. 7 is shown in Fig. 8. Since word A programs the timer, the output durations are given in terms of the value of word A and the RC timebase period.

When the trigger pulse is received, word A has previously

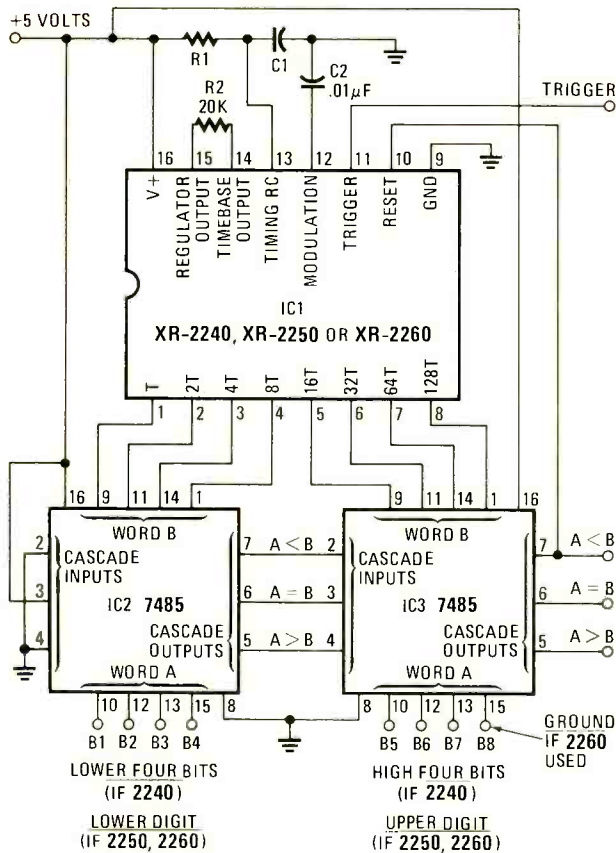


FIG. 7—PROGRAMMABLE DELAY TIMER can be obtained with the addition of two magnitude comparators.

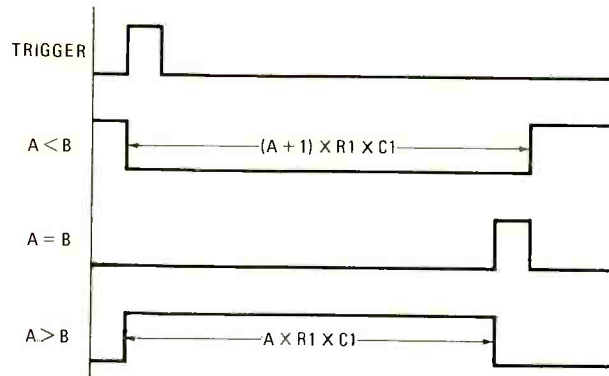


FIG. 8—TIMING DIAGRAM for the programmable delay timer.

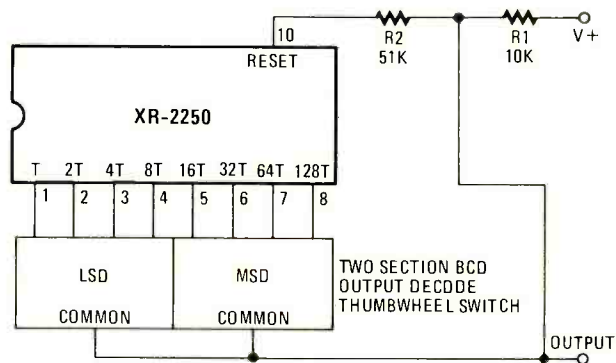
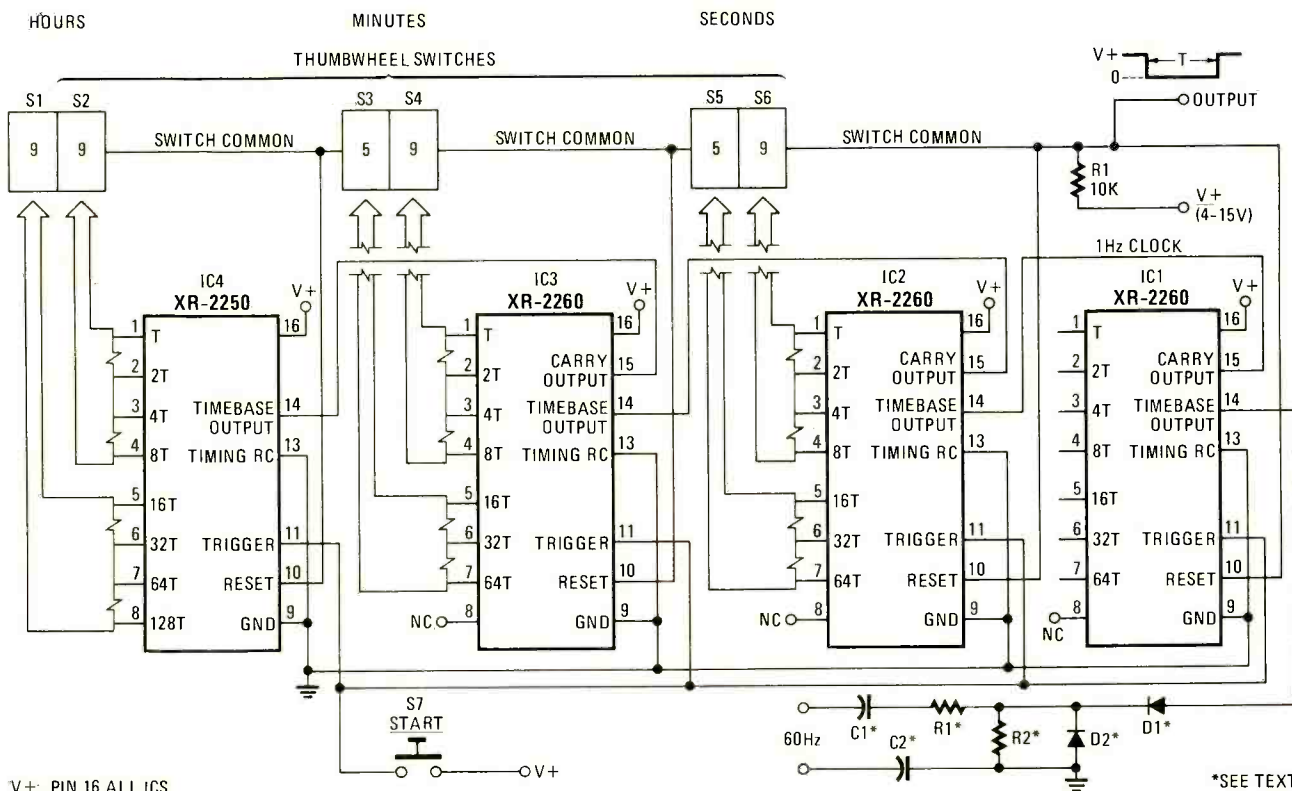


FIG. 9—PROGRAMMABLE DELAY TIMER using BCD thumbwheel switch.



V+: PIN 16 ALL ICs

FIG. 10—100-HOUR PROGRAMMABLE TIMER uses BCD thumbwheel switches.

been set to some value between 0000000_2 (0_{10}) and 1111111_2 (255_{10}). All the timer outputs drop low, so word B is initially 0000000_2 . At this time, word A is either greater than word B, or equal to it if word A is also 0000000_2 . In the former case, the $A > B$ output from IC3 is high and the other two outputs are low. When the counter has incremented so that word A and word B are equal, then the $A = B$ output goes high for one clock period, after which word B is greater than word A, so the $A < B$ output goes high. The timing durations are $(A + 1)R1C1$ for A less than B, and $AR1C1$ for A greater than B. The $A = B$ output produces a single pulse at time $AR1C1$.

Assume that the Fig. 7 circuit is programmed so that word A is 178_{10} (10110010_2), and that the R-C time constant is 5 seconds. What is the duration of the $A > B$ pulse? The duration is $A \times R1 \times C1$, or (178) (5 seconds), or 890 seconds (about 14.8 minutes).

Thumbwheel switches with binary, BCD, or octal-output codes can be used to program these IC timers. Figure 9 shows a BCD thumbwheel switch that programs an XR-2250 timer. The BCD switches are connected to the output lines on the timer IC, and the switches are in turn connected to pull-up resistor R1. The switches provide a simple way to build circuits (such as that of Fig. 3-a) using convenient front-panel switches to change the output duration when desired.

Figure 10 shows the circuit of a precision 100-hour timer using the Exar XR-2250 and XR-2260 timer IC's. Four timers are in cascade and are programmed by thumbwheel switches.

Timer IC1 is used as a timebase to generate a 1-Hz clock from the 60-Hz AC power line. The values of C1, C2, R1 and R2 depend on the amplitude of the 60-Hz line used for the timebase.

The XR-2260 counts to 59, and on the 60th count generates a

carry pulse on pin 15. This output pulse represents a frequency division of 60 for the XR-2260 and a division of 100 for the XR-2250.

As shown in Fig. 10, the *seconds* and *minutes* sections of the circuit are thumbwheel-programmed, cascaded XR-2260 IC's, while the *hours* section is a thumbwheel-programmed XR-2250 IC allowing a maximum of 99 hours. Both XR-2260 IC's use the carry-output pulse to drive the timebase input of the following stage.

The thumbwheel switches program the total duration of the output pulse, but some applications might require a continuous monitoring of the time that has *elapsed* since the trigger pulse started the counters. To do this, you can use an events counter or a clock IC that uses a 1-Hz input. Another alternative is to use an hours-minutes-seconds counter circuit with its reset terminal connected to the timer output. The timer output is active-low and so will allow counting during its time duration, but it will reset the counter when it goes high at the end of the time duration.

Precision is a word not ordinarily used in conjunction with the AC power-line frequency. Digital frequency counters and short-duration timers that use 60-Hz AC power lines as the timebase are often described as being only of moderate precision. The accuracy specification is usually given in terms of 1% or a little less as a result of small variations in the line frequency. The actual frequency varies above and below 60 Hz as the generator speed changes, but the average frequency is quite close to 60 Hz over a period of time. This timer, then, possesses good accuracy for long durations but only moderate accuracy for short time spans. Using a precision crystal-controlled 1-Hz or 60-Hz clock will insure that the accuracy remains the same over the *entire* range.

SWITCHING TRANSISTOR SHORTS

This Quasar TS-938 came in with no picture or sound. Found Q8 (power supply switching transistor) shorted. I replaced it and two days later it was back with Q8 shorted again. I need some help before I blow another one of those expensive transistors! With the replacement in, it worked fine. I could use some help.—H. Y., Baltimore, MD.

You sure could! This can be highly nonhabit forming. I ran across the same thing not very long ago. A friend used the same replacement transistor you did. We checked and found that this was a good transistor but had only a 700-volt peak-surge rating. The original transistor apparently had an 800-volt peak rating. We used a Sylvania ECG-165, which has a 1400-volt peak rating, and it is still working several months later. Always check the peak-voltage rating of this kind of transistor, and use a substitute with the highest rating you can find, of which there are several.

LOSS OF RASTER

This Zenith 19EC45 chassis loses the raster intermittently. I've checked everything I can think of, with no luck. Doesn't seem to damage anything! Where do I go from here?—D. L., Ridgefield, CT.

There is one problem we've run into in this and other Zeniths using the 9-90 horizontal module. Check resistor R808, 330 ohms. If this resistor is bad, it upsets the horizontal sweep. Replace it with a 1-watt resistor.

(Feedback: "That did it! Resistor R808 was almost open. The set is working fine now.")

TURN-ON SQUEAL

This Admiral M10 chassis "popped" at turn-on and finally quit. Resistor R818, in the horizontal driver collector, was open. Replacing this resistor brought the set back on. However, the horizontal oscillator/output would squeal at turn-on, now and then, and also at turn-off. All waveforms became normal after the unit stabilized, but fuse F1000 would blow after 5 to 8 hours

of operation. After quite a bit of checking, I found electrolytic capacitor C811 ($10 \mu\text{F}$, 25 V) was intermittently opening up and allowing some kind of screwball feedback at turn-on and turn-off, and somehow making it blow that fuse.

Thanks to Leon Caldwell, Caldwell's TV, Mena, AR.

NO TINT CONTROL

Several things such as the IC, etc., were replaced in this Zenith 23DC14 chassis after a lightning hit. Now, I do have color and good tints, but no tint control at all. I don't get it.—L. C., Mena, AR.

This chassis uses a DC-voltage tint control, with presets in the color circuitry. Your tint-control DC voltage-supply circuit goes through the chromatic switch. Check for normal DC voltages and follow them through the switch.

(Feedback: "Right, the DC voltage on the tint control wouldn't vary at all. I found a burned-up 470-ohm resistor in the DC voltage-supply circuit. I also finally traced the DC supply voltage through the chromatic switch and found that *both* conductors going to the tint controls were completely blown. Two short pieces of wire fixed that!")

SCOPE PROBLEM

I've got all kinds of odd distortion in a Heath OM-2 scope. I've tried new tubes, with no luck. Do you have any bright ideas?—J. H., Chicago, IL.

Maybe not any *really* bright ideas, but, from my experience in working over old scopes, I recommend checking *all* the coupling capacitors in the vertical amplifier stages. While you're at it, check all the plate, grid and cathode resistors, too, and replace any that are off-value. Doing this fixes most of them.

I had a strange horizontal nonlinearity in an old scope. Replacing the coupling capacitors in the *vertical* amplifier stages cured it. I haven't explained that one yet, but it worked!

(Feedback: "That did it. Changed two 0.1- μF coupling capacitors and now the scope works fine!")

R-E

hobby corner

An easy-to-build crystal oscillator that is a must for your workbench.

EARL "DOC" SAVAGE, K4SDS, HOBBY EDITOR

THIS MONTH WE'LL DISCUSS HOW TO BUILD a device that should be on every workbench. It is a wide-range, multipurpose crystal oscillator that you can use to perform dozens of tasks.

Crystal oscillator

In order for a crystal oscillator to be most useful, it should:

1. Operate over a wide frequency range.
2. Operate with different types of crystals.
3. Have an output containing the fundamental and adequate harmonic frequencies.
4. Require no tuning.
5. Be easy to construct.

The crystal oscillator described here meets these requirements and more. It functions with fundamental crystals from 100 kHz to 17 MHz or more—a very wide range. Overtone crystals will oscillate on their fundamental frequencies—in fact, every crystal I put in causes the unit to take right off. (Incidentally, overtone crystals do not oscillate at an *exact* submultiple of the marked frequency. This is the nature of such crystals—the oscillation will be close but not exact.)

This little oscillator has quite a rich harmonic output for a one-transistor circuit. The harmonics of a 100-kHz crystal can be heard well above 20 MHz.

How you can use it

Before we describe the circuit in detail, let's take a look at some of the ways you can use this oscillator. First, of course, you can use it as a crystal calibrator for a receiver. Most receiver dials are not very accurate, and a 100-kHz crystal in the oscillator will show "marker signals" every 100 kHz across the dial so that you know where you are. Other crystals can also be used for this purpose. For example, a 500-kHz crystal produces a marker every 500 kHz; a 1-MHz crystal, one every 1 MHz; and so forth.

The oscillator can be used for aligning receivers. It is worthwhile even if you have a signal generator because most hobbyist generators are not very accurate themselves. A crystal at the intermediate frequency (IF)—usually 455 kHz—permits you to tune up (align) the IF strip of

the receiver. The fundamental and harmonic frequencies of one or two other crystals will generate the signals you need to align the radio frequency (RF) circuits.

Do you use a transistorized portable radio with shortwave bands or an SWL receiver without a beat-frequency oscillator? If you have ever attempted to tune in ham or commercial transmissions that are CW (code) or SSB (single sideband), you know they sound like gibberish. This little oscillator can clear up the confusion nicely. Simply use a crystal at or very near the intermediate frequency and run its antenna near the IF section of the receiver. Then careful tuning produces readable CW and SSB signals.

Another interesting oscillator application is for frequency spotting. Suppose you want to hear a favorite broadcast or shortwave station that is sometimes hard to locate. Just pop a crystal of the station frequency into the oscillator, and you know just where to look. For example, hams and those studying to be hams sometimes have difficulty locating station W1AW for their code practice sessions on 3580 kHz. An inexpensive 3579.54-kHz crystal of the type used in digital clocks or a 3.58-MHz "rock" as used in TV color oscillators will put you almost right on it.

Often it is desirable to change the frequency coverage on a receiver; for instance, you want to receive a shortwave signal on a broadcast set or the WWV time signals (5, 10, 15 MHz, etc.) on a receiver that does not tune to one of those frequencies. The usual solution is to build a small converter, the main part of which is (you guessed it) a crystal oscillator.

Circuit description

Now that you know some of the ways you can use the crystal oscillator, let's look at the circuit.

Figure 1 shows that only a few standard components are used. In fact, two, C1 and S1, can be omitted entirely and the oscillator will still perk right along. None of the components are highly critical so you can use those with 10% or 20% tolerance, and, in most cases, you can use the next closest standard value.

If you use a socket for the transistor,

the oscillator can be used for a transistor tester provided you put in DPDT switch S1. Then, with the switch in the proper position, pop either an NPN or PNP transistor into the socket for testing. If oscillation occurs, the transistor is good. This test does not tell you *how* good, just

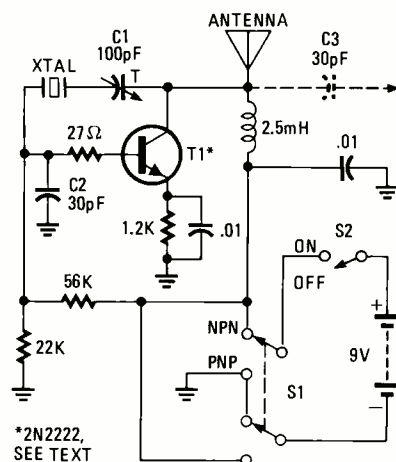


FIG. 1

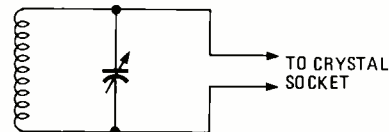


FIG. 2

that it *is* good. However, a transistor that oscillates also amplifies, switches, and does most other tasks it may be called upon to perform.

You can omit switch S1 and the socket if you don't plan to test transistors. In this case, just wire the circuit for an NPN transistor and solder in either a Radio Shack No. RS-2009, a 2N2222, a 2N4124, or any comparable NPN transistor from your junk box.

The current drain on the 9-volt battery is less than 1.5 mA, so it should last for a long time under normal intermittent use.

Variable capacitor C1, if it is used, should be a compression-type padder, and the maximum value can be anything from 40 pF to 200 pF. Space permitting, you can use a physically larger capacitor type. The purpose of C1 is to permit fine adjustments to be made in the oscillation frequency so that it exactly matches a standard; for example, WWV. If this feature is not needed, you can omit C1

and connect the crystal directly to the transistor collector.

Capacitor C3 (shown as two dotted lines in Fig. 1) is needed only if the output is to be connected directly to another circuit. Normally, the short length of wire functioning as the antenna can be placed near a receiver antenna.

Switch S2 is the on-off switch. Depending upon the major use of the oscillator, you may prefer a toggle or a momentary pushbutton switch.

One final note: You may find it advantageous to replace the 30-pF fixed capacitor C2 with a 100-pF compression padder. If you run into a sluggish crystal that is hard to start, adjusting variable capacitor C2 will usually cure it.

Construction

After collecting the components, you should put the circuit together on a breadboard before hard-wiring it. Breadboarding is *always* a good idea because you can find bad components and other problems while they are still easy to correct.

When the circuit is functioning properly, wire it up using your favorite method—wiring pencil, point-to-point, wire-wrap, or circuit board. It should fit easily into a box measuring 1½ × 4 × 2¼-inches unless you have some outsized components.

At least two sizes of crystal sockets should be used. Wire them in parallel. If one will accommodate the HC-6U holder and the other, the FT-243 holder, you can then handle the common crystal sizes.

Capacitors C1 and C2 (if you use a variable capacitor there) should be mounted so that they can be reached for adjustment through holes in either the main panel or a side panel. Don't forget to use an externally mounted socket for the transistor if you also plan to use the oscillator as a tester.

Oscillates without crystal

This circuit will oscillate *without* a crystal. Wire up a coil and a capacitor (as shown in Fig. 2) and you have a variable oscillator; that is, one in which you can change the frequency. For example, a loopstick coil and a 365-pF variable capacitor taken from an old broadcast-band radio enables you to swing the signal frequency across the broadcast band.

You can experiment with other coil and capacitor values to obtain higher and lower frequency ranges. Just plug the coil and capacitor into the crystal socket. You may find it necessary to short-out C1 for some combinations to oscillate, but do not expect every combination of values to create oscillation.

The frequency of the crystal-less or L-C (for inductance-capacitance) oscillator is not very stable. It wobbles and shifts with mechanical vibrations and body capacitance (as when you place your hand nearby). However, for less exacting

requirements, it may serve just as well.

This L-C oscillator has one additional use that may be helpful. Many hobbyists have no means of determining the inductance value of a coil. (A grid-dip meter is the best way to do this at reasonable cost.) However, if you can find a combination of a known capacitance that will oscillate with the unknown inductance, you are in business.

Simply find the oscillating frequency on your receiver and plug the values in the formula below. Make sure that you have the fundamental frequency and not one of the higher harmonics. Also, you must use the correct units of measure-

ment as shown in parentheses below:

$$L_{(\text{microhenrys})} = \frac{25,330}{f_{(\text{MHz})}^2 \times C_{(\text{picofarads})}}$$

The same procedure can also be used to determine the value of an unknown capacitance if it will oscillate with a known inductance. Just use this formula with the same units as before:

$$C = \frac{25,330}{f^2 \times L}$$

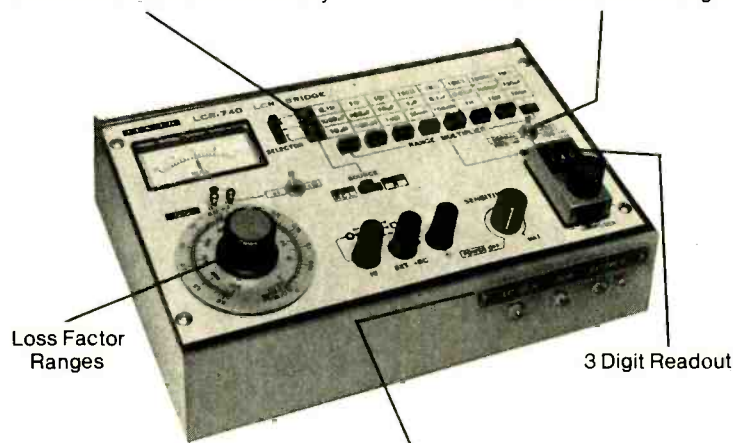
It would be hard to find a more versatile device than this crystal and L-C oscillator. You will surely discover many more uses for it around your workbench, and it is so useful that you may find yourself building more than one. **R-E**

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state of solid state

A look at several interesting applications for National's IC pressure transducers, including a solid-state barometer and blood-pressure meter. **KARL SAVON, SEMICONDUCTOR EDITOR**

THE HYBRID PIEZORESISTIVE IC PRESSURE transducer is the modern replacement for the mechanical sensor. It is a compact device that functions reliably in harsh industrial, consumer, medical, military and automotive environments. The same IC that contains the pressure transducer itself also contains the power-supply regulator, temperature compensation circuits and a low-output-impedance operational amplifier. The transducers are manufactured with standard integrated technology and are laser-trimmed for accuracy. Their frequency response enables them to perform through the audio frequencies. They are insensitive to vibration and operate over wide pressure ranges, from 0 to 5000 psi (pounds-per-square-inch). National Semiconductor manufactures IC pressure transducers for

absolute, gauge and dual-port differential pressure applications. The transducers are available in a hybrid IC package for PC mounting or a zinc alloy housing for mechanical isolation.

Solid-state barometer

The electronic barometer circuit in Fig. 1 uses the LX1701A absolute-pressure transducer. Absolute pressure is referenced to zero pressure (a vacuum). The LX1701A is designed for pressure ranges of 10 to 20 psia (pounds-per-square-inch-absolute), including the 15-psi normal atmospheric pressure. This barometer has a temperature coefficient of ± 0.0054 psi- $^{\circ}\text{C}$ and a stability of ± 0.05 psi. The integrated transducer operates from a nominal supply of 15 volts and can source 20 mA and sink 1 output current of 10

mA. Its output impedance is less than 50 ohms. The output of the LX1701A varies from 2.5 volts to 12.5 volts over its rated input-pressure range.

Figure 1 shows two operational amplifiers connected to the pressure transducer that allow independent adjustment of gain and DC offset. Adjusting the gain and offset calibrates the system to any particular pressure scale, whether it be psi, millibars (1 millibar = 1000 dynes-per-cm²), or millimeters of mercury.

For negative feedback amplifiers, the gain is the ratio of the feedback resistor to the input resistor. Output amplifier A2 in Fig. 1 has a gain of R_4/R_3 . Grounding the noninverting input (the + terminal) results in a zero output voltage for a corresponding zero voltage applied to the input through the input resistor.

Using two inverting amplifiers provides an output-voltage change in the same positive direction as the input voltage. The output-level contribution from the pressure transducer is its own output, V_p , multiplied by the gain through amplifier 1 (R_2/R_1) multiplied by amplifier 2's gain (R_4/R_3). The DC offset at the output is the 6.9 volts from the LM129 reference multiplied by R_2/R_5 and R_4/R_3 . Once R_2 , R_3 and R_4 are selected, the output can be calibrated by choosing R_1 for the scale factor and R_5 for the DC offset.

Blood-pressure meter

Before proceeding with the medically related circuit discussed in this section, it must be emphasized that National Semiconductor does not recommend that its circuits be used in life-support equipment.

Figure 2 is the block diagram of a sphygmomanometer or simply a blood-pressure meter. Here, an LX1601G (-5 to +5 psig) is used for the transducer. The "G" suffix stands for gauge pressure, which is a differential pressure measured between ambient pressure and another source—in this case, the arm cuff. The blood-pressure meter displays diastolic and systolic blood pressure as well as the heart rate. Substituting the electrical output-pressure transducer for the usual mechanical sensor permits its control by a microprocessor, so the blood-pressure machine can be made completely automatic. Blood pressure is taken with respect to air pressure. In other words, when no blood pressure is being taken, the meter output should read zero.

First, the microprocessor must develop

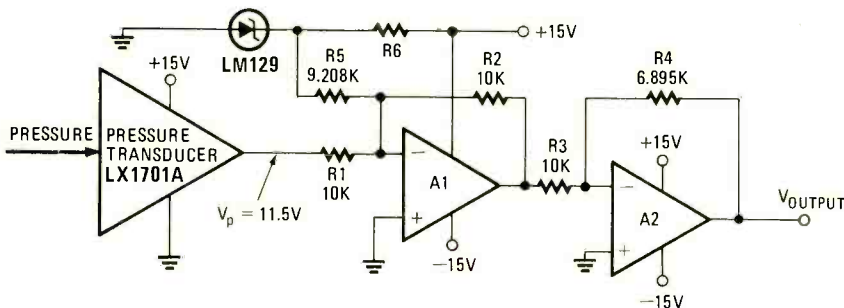


FIG. 1—SOLID-STATE BAROMETER uses two operational amplifiers and pressure transducer. Circuit can be calibrated for any particular pressure scale.

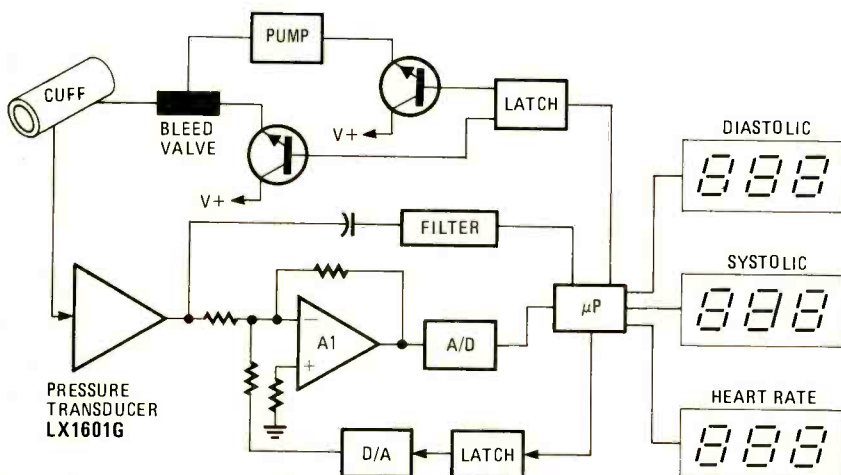


FIG. 2—BLOOD-PRESSURE METER uses an air pump for automatic readings. Circuit operation is controlled by a microprocessor.

an autocorrection signal that gives a zero output reading before the cuff is inflated. A latch circuit stores a multidigit binary number from the microprocessor. The number in the latch is changed into its equivalent DC voltage by the D/A converter.

The circuit is designed so that when the summing amplifier (A1) output is 7.5 volts, the pressure display reads zero. This autocorrection scheme corrects for temperature and offset drifts anywhere in the system. Autocorrection performs the function of the manual zero-adjust screw on an analog meter. Summing amplifier A1 adds the latch correction voltage to the output of the pressure transducer. The amplifier output is connected to the microprocessor through the A/D converter. The output of the summing amplifier and the A/D converter has an AC component due to the pulsing blood pressure.

To take a reading, the cuff is gradually inflated on command of the microprocessor, and a digital display shows the point at which the AC component is at a minimum. This corresponds to the systolic pressure. The cuff is then deflated at a constant rate, and again the minimum AC component corresponding to the diastolic pressure is recorded. There is a direct AC signal path from the transducer to the microprocessor through the filter. This signal is used to measure the frequency of the blood pulses or the heart rate. No DC correction component is necessary since the frequency is not changed by temperature or time drifts in the sphygmomanometer system.

Precision weight scale

Figure 3 shows how you can build a precision weight scale. An open-loop system could be used, in which a hybrid

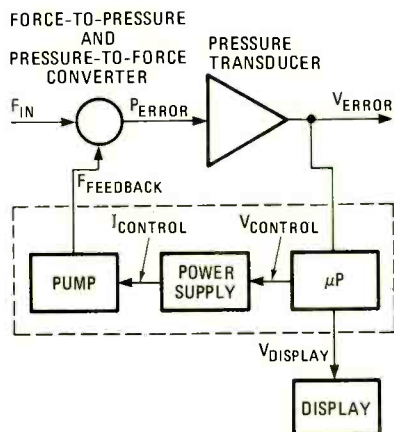



FIG. 3—PRECISION WEIGHT SCALE uses a pump to reduce the transducer's input signal to within range. Circuit is controlled by a microprocessor.

pressure transducer drives a voltmeter through the necessary calibration amplifier. But if you want a true precision scale, the linearity and accuracy of the transducer may not be good enough. The feedback system in the diagram balances the

force produced by a pump against the force produced by the unknown weight. The pump in the feedback circuit, under microprocessor control, produces a force that is equal to the weight, so the output of the differential pressure transducer approaches zero.

The resolution of the system is excellent because the transducer can be much more sensitive than the total weight range of the system. In operation, the weight only has to move the tiny amount necessary to generate the small output from the transducer that is amplified and drives the meter. Initially, there will be a larger movement until the feedback loop

has time to build up the pump pressure. Mechanical limiting mechanisms protect the transducer from exceeding its maximum pressure rating until the pump restores the initial location of the scale platform. Once again, a microprocessor is used, this time as an integrator of the pump control current that keeps track of the accumulated weight. **R-E**



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Data acquisition for microcomputers. PETER R. RONY, JONATHAN

A. TITUS, DAVID G. LARSEN, CHRIS A. TITUS*

THE SOFTWARE IN A PREVIOUS COLUMN provided an example of a program used to acquire a single analog point in digital form. (See October 1977 issue.) We are generally interested in applications in which a series of points are to be acquired, stored, displayed and perhaps manipulated. This month we will explore the use of microcomputers for data acquisition.

It will be assumed that the analog-to-digital converter (A/D) is interfaced as shown in the previous column. The software routine to run the A/D converter is shown in Table 1. The digital value of the analog voltage is returned in the B and C register (register pair B).

In most data acquisition programs, a fixed number of points must be acquired over a fixed period of time. In our example, 100 points will be taken, one every second. The 100 data points will be stored in read/write memory so that they can be used later. In writing data acquisition software we are now faced with three tasks that must be performed in addition to the actual A/D task:

1. Provide a software counter to count 100 points.
2. Provide a one-second timer.
3. Provide software to store the data values.

Software counter

The software necessary to count the 100 acquired points will actually count 100 passes through the data acquisition software. A general-purpose register within the 8080 IC is well suited for this; conditional jump instructions can be used to detect when the count is decremented to zero. The counter can be either incremented or decremented, but decrementing is probably easiest if you are just starting to program microcomputers.

Storing the data in memory is not difficult. Once the converter value is stored in a register pair, the H and L registers (register pair H) can be used as memory pointers to point to a read/write memory location. Note that a complete 16-bit address must be specified for the MOV M,r instructions. Since the data is ac-

*This article is reprinted courtesy American Laboratories. Dr. Rony, Department of Chemical Engineering, and Mr. Larsen, Department of Chemistry, are with the Virginia Polytechnic Institute & State University. Mr. Titus is president of Tychon, Inc.

TABLE 1 TYPICAL ADC INPUT ROUTINE FOR A 10 BIT ANALOG-TO-DIGITAL CONVERTER

```

*100 000
100 000 365 ADC, PUSHPSW /SAVE REGISTER A & FLAGS
100 001 323 OUT /STROBE THE ADC TO START A CONVERSION
100 002 037 037
100 003 333 TEST, IN /INPUT STATUS BIT AND 2 MSB'S
100 004 066 066
100 005 306 ADI /ADD 1 TO THE FLAG BIT
100 006 200 200 /TO CAUSE A CARRY IF IT IS SET
100 007 322 JNC /NO OVERFLOW, CHECK IT AGAIN
100 010 003 TEST
100 011 100 0
100 012 107 MOVBA /OVERFLOW, FLAG=1, SO SAVE MSB'S
100 013 333 IN /INPUT THE 8 LSB'S
100 014 065 065
100 015 117 MOVCA /STORE THEM IN REGISTER C
100 016 361 POPPSW /RESTORE REGISTER A & FLAGS
100 017 311 RET RETURN TO MAIN PROGRAM

```

TABLE 2 100 POINT DATA ACQUISITION ROUTINE FOR ONE POINT PER SECOND

```

DW ADC 100 000

*070 000
070 000 061 START, LXISP /LOAD THE STACK POINTER
070 001 377 377
070 002 070 070
070 003 041 LXIH /LOAD THE DATA STORAGE STARTING
070 004 000 000 /ADDRESS IN REGISTERS H & L
070 005 072 072
070 006 315 CONVRT, CALL /CALL THE ADC SOFTWARE
070 007 000 ADC /SHOWN IN TABLE 1
070 010 100 0
070 011 161 MOVMC /STORE THE 8 LSB'S TO MEMORY
070 012 043 INXH /INCREMENT THE MEMORY POINTER
070 013 160 MOVMB /STORE THE 2 MSB'S TO MEMORY
070 014 043 INXH /INCREMENT THE POINTER AGAIN
070 015 175 MOVAL /GET THE LOW ADDRESS VALUE
070 016 376 CPI /COMPARE IT TO THE 201ST ADDRESS
070 017 310 310 /310 = 200 DECIMAL
070 020 312 JZ /DONE YET?
070 021 047 DONE /YES, JUMP TO "DONE"
070 022 070 0
070 023 315 CALL /NO, DO THE 1 SECOND DELAY
070 024 031 DELAY
070 025 070 0
070 026 303 JMP /AFTER THE DELAY, GET THE NEXT
070 027 006 CONVRT /ADC DATA POINT
070 030 070 0

```

THIS IS THE ONE SECOND TIME DELAY SUBROUTINE

```

070 031 365 DELAY, PUSHPSW /SAVE REG A & FLAGS
070 032 325 PUSHD /SAVE REGISTERS D & E
070 033 021 LXID /LOAD COUNTER REGISTERS
070 034 000 000
070 035 110 110
070 036 033 DEC, DCXD /DECREMENT THE REG PAIR
070 037 172 MOVAD
070 040 263 ORAE
070 041 302 JNZ /IF NOT ZERO, DO IT AGAIN
070 042 036 DEC
070 043 070 0
070 044 321 POPD
070 045 361 POPPSW
070 046 311 RET

```

THE PROGRAM WILL CAUSE THE COMPUTER TO JUMP HERE WHEN IT HAS ACQUIRED ALL THE DATA POINTS. A DISPLAY OR OTHER ROUTINE MIGHT BE PLACED HERE INSTEAD OF THE HALT

```
070 047 166 DONE, HLT
```

quired from a 10-bit A/D converter, two successive memory locations must be used to store each *point*. The INXH instruction (increment register pair H) provides an easy means of pointing to the next successive memory location. We will store the data by placing the eight least significant bits in location *n*, and the two most significant bits in location *n + 1*.

Real-time clock

The one-second timer may present some problems, depending upon the type of system used. It is relatively easy to write a one-second software delay program using a series of register decrementing loops, nested one within the other. However, to accurately time a one-second period, the computer must be doing nothing else. In a system dedicated to data acquisition for the 100-second period, such a procedure is valid.

If interrupts occur or if the computer cannot be allowed to "do nothing" most of the time, an alternate solution is needed. One possibility is to use an external clock, often called a *real-time clock*.

Real-time clocks are unaffected by computer execution times, interrupts, slow I/O devices, etc. Once started, they will continue to run at an accurate rate until they have timed the particular period of interest and sent an interrupt to the microcomputer.

Some clocks are free running, always keeping time; others are programmable or preset for a particular period. The free-running clock interrupts the computer at repetitive intervals while the programmable clock interrupts the computer only once, at the end of its pre-programmed period. Integrated circuits such as the Intel 8253 and Texas Instruments TMS 5501 contain time-keeping circuitry that is easily interfaced to most 8080's.

Data storage

For simplicity, we will use the software clock in our example rather than an interrupt-based real-time clock. The software for the 100-point data acquisition program is shown in Table 2. After completing the program, the computer might be programmed to jump to the type of data display software discussed in the previous column.

If you look at the program carefully, you will not find a separate register to count the 100 passes through the data acquisition software. Since the memory address stored in registers H and L is already a counter, we have chosen to detect the 200th address rather than the 100th loop. This saves an internal register. Instead of decrementing a counter and detecting the zero condition, the contents of register L are compared with the final address and the equality is used to signal the end of the loop.

Analog-to-digital converters are not "instantaneous" devices that take only a few microseconds to perform a conver-

sion. In many real situations, the analog input to the converter will vary while the A/D converter is trying to perform a conversion. This presents the converter with a problem. How does it know what the real value of the voltage is? In most systems the A/D converter module has a *sample-and-hold* device on the analog input. The sample-and-hold circuitry samples the analog voltage when pulsed to provide a steady analog output to the A/D converter for conversion; the A/D converter is then pulsed to start the conversion. The Intersil IH 5110 is a typical sample-and-hold device. R-E



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How to troubleshoot faulty high-voltage shutdown circuits in solid-state color sets.

JACK DARR, SERVICE EDITOR

THERE'S ONE COMMON TROUBLE AREA IN solid-state TV sets (mainly in color sets but in a few black-and-white models) that is all-too-often overlooked. This is the high-voltage regulation circuitry, which includes the high-voltage shutdown circuitry. Note: these are two *separate* circuits, but they interact.

The high voltage in solid-state sets is seldom regulated directly, as in tube sets with 6BK4 tubes, etc. In practically all the solid-state sets, the high voltage is directly proportional to the DC voltage fed to the horizontal-output stage, or B+ supply. So, they do it the easy way and regulate the B+ supply. There are several circuits that can hold the B+ voltage very steady. Here is the simple test that is often overlooked. When there is any problem in or associated with high-voltage regulation (it reads either too high or too low) *always check the B+ voltage first!!*

The B+ value shown in Fig. 1 is critical. If +127 volts is indicated, +127 volts is exactly what is meant and *not* +124 volts or +130 volts. This value must be right on the nose and stay there over a fairly wide range of input-line voltages. If this B+ value is high or low, check out the regulator circuitry. A fault in the B+ line that makes it go too high will make the high voltage rise with it, trip the over-voltage protection circuitry, and shut the whole circuit down. It usually does this by killing the drive to the horizontal-output stage.

The Quasar *model TS-961* and *model TS-963* sets use some typical over-voltage protection circuitry (see Fig. 1). Using the same simple test, read the DC output voltage with an accurate DC voltmeter (digital meters are handy). If the voltage does not read exactly right, go through the voltage regulator and check for a leaky pass transistor or error amplifier transistor, a bad Zener diode, etc. A leaky pass transistor allows the B+ voltage to become too high. If it shorts, you'll see the same B+ output voltage as the DC input voltage.

Here's a helpful hint: The B+ adjustment potentiometer and over-voltage adjustment potentiometers on several Quasar sets are factory-sealed. Several readers have asked, "How do you get these unsealed for adjustment?" You don't,

you take the originals out and put in new ones. A new control, sealant and full instructions can be had in a kit from a Quasar distributor. You cannot break the seal without tearing up the control. (If you take it out, don't throw it away yet! You might have some other problems and, if so, you can put the control back.)

The Quasar part number for the kit used in the *model TS-961* is No. 18M25000A09; for the *model TS-963*, it's No. 18-25000A11. The potentiometer number is R811 for the *model TS-961*, and R812 for the *model TS-963*.

The test for over-voltage protection circuitry in both these models is shown in Fig. 1. If the B+ voltage is normal, then the shutdown circuit should be checked

causing a shutdown.

If this happens, you can readjust the control. Use the same method described above: take out the old control (remember to save it!) and put in the new one from the kit (the same part numbers apply as before for each chassis). If the shutdown circuit cannot be set up to trip correctly, see if the adjustment will help. If not, check out the components used in the circuit. In the *model TS-961*, a silicon-controlled rectifier (SCR) is used; a pair of transistors are used in the *model TS-963*; however, the circuit action is the same. Several Zener diodes are also used, so check them for leakage, shorts, opens, etc. A flyback pulse is fed into the shutdown circuitry through a diode, so you

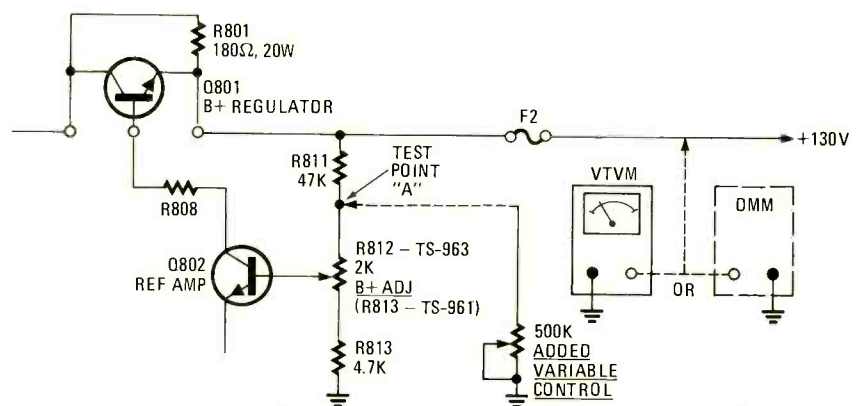


FIG. 1

to see if it's operating. Connect a variable potentiometer across the B+ voltage error-amplifier base circuit, as shown. The value of this control is not critical—from about 100K to 500K. Monitor the B+ supply with an accurate DC voltmeter. Adjust the brightness so that you can barely see a raster because doing this reduces the beam current to a minimum.

For the test, slowly reduce the resistance of the shunt potentiometer. Watch the B+ voltmeter. It will rise slowly and when it reaches the point where the B+ voltage reads from +144 volts to +146 volts, the shutdown circuit should trip and the raster go out. This will happen at about 29.5 kV in the *model TS-961*, and in the *model TS-963*. If the shutdown circuit is not working, the high voltage can be raised above the limit without

should check also for this pulse.

You may run into premature shutdown problems. If the B+ voltage is normal and the high voltage is shut down (for a quick check, look for the drive pulse to the horizontal-output transistor), turn the set off and wait for one minute to be sure the SCR has unlatched and the capacitors have discharged. Then try again. If the circuit shuts down even though the high voltage is within limits, then make the suggested tests described above.

As I said, and continue to stress, check the B+ voltage *first*. Since this voltage can cause so many different problems if it's out of the ballpark, it's advisable to investigate it before you go tearing around through the rest of the circuitry! Just making this one test can save you lots of time.

R-E

service questions

DOUBLE TROUBLE

The symptoms on this Zenith model 12B14C52 are: a very pale picture, plenty of brightness, no snow, color present, and good sound. The contrast and brightness controls react properly. The AGC level control has no effect at all. I check the 6KT8 (AGC) voltages. The cathode voltage varies, the AGC doesn't. After some confusion due to reading on the wrong socket, I finally replace the 6KT8 tube. Now the AGC reacts properly. But the symptoms are still there!

There's a very low signal on the output of the sound-sync detector. This should be 9 volts P-P. I pulled the IF strip and checked it—nothing. There are no normal signals anywhere through the AGC/sync stages. There's a video signal on the 6BA11 pin 6 plate, which should be composite sync.

Finally, I get to the IF strip. The first and third IF transistors show zero emitter volts. This is bad! (The IF strip is now out of the chassis and on the bench, hooked up with jumper leads.) I pull the third IF transistor: the curve tracer says no. I replace it. Is it the first IF transistor? Whoa! I separate the IF strip test shown in manual as being OK for some versions. But if the AGC lead is in the wrong place, the first IF base circuit is open. Therefore, no emitter voltage, no output, just a small amount of signal leaking through the third IF transistor. I add one new jumper. (I use push-on, insulated connectors for these to avoid shorts to the shield.)

The problem here is aggravated by a bad tube in the AGC. This and the IF problem caused similar symptoms for double-trouble.

POWER-LINE FREQUENCY ACCURACY?

When I made a clock, I took the power-line frequency, divided it by 60 for the seconds, then by 60 again for the minutes, then again for the hours. Then I fed this to a binary-number display using lamps. After about a month or six weeks, the clock is always off by two to four minutes. My regular electric clock keeps better time. Isn't it also dependent on line frequency? I don't understand.—F. A., Gulf Breeze, FL.

That's a good question. I have heard of a few cases in which line frequencies were not really accurate. (In the NY area for one.) Now, let's look into that crystal ball.

Your electric clock does depend on the line frequency for its accuracy since it uses a synchronous motor. However, it is possible for a momentary (i.e., a glitch of only a few milliseconds' duration) inter-

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ruption to have no effect on the standard clock because its flywheel effect wouldn't let it slow down too much.

However, digital counter circuits are quite fast. A glitch of a few milliseconds would make them lose a couple of counts, and the cumulative effect over a period of time would result in considerable inaccuracy.

This is pure crystal-ball conjecture, but it could be right. I've heard other cases of complaint about digital clocks that do use a 60-Hz line frequency as half-wave rectified pulses.

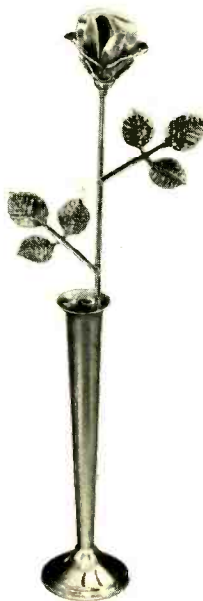
VOLTAGE REGULATOR PROBLEM

I had an oddball problem in a GE transistor TV some time ago. The B+ voltage was quite low, the AC current drain was high and the power transformer heated up—all the classic symptoms of too much current drain! However, the B+ current, at 1.2 amp, wasn't so bad.

Checking the circuits showed they didn't seem to be taking too much current. The set worked fine on an external car battery. So I went back to the power supply and checked the voltage regulator; everything was OK. Then I unhooked the load from the regulator and used external load resistors. When I raised the load current to about 1.2 amp (what it drew in normal operation) the ripple on the output went way up. A closer examination showed the classic *half-wave ripples* in-



**TOP
DEAL**



stead of full-wave ripples.

Additional checks showed no AC on one of the full-wave rectifier diodes. The AC was read at the secondary of the transformer. I discovered a foil break between the transformer and one diode! When this was fixed, there were no more problems. (Thanks to Rodney Schrock, Somerset, PA.)

BAD CAPACITOR

A customer brought in a Zenith model 19FC45 with a bad capacitor in the horizontal output. He had tried to change it himself. The old capacitor is gone and I am not sure just how it should be connected. I don't find this on the Sams service data.—G. H., Panama City, FL.

On the 22-7504-01 replacement for the four-legged capacitor, terminal A goes directly to the collector of the horizontal-output transistor, terminal B goes to the lead from the flyback/damper anode/etc., terminal C goes to the emitter of the output transistor (which is ground), and terminal D goes directly to the ground lug nearby. There should be continuity from terminal A to terminal B, and from terminal C to terminal D, but *never* from one pair to the other!

Check the output transistor, the triple, the damper, etc., for possible shorts. Ordinarily, if this capacitor shorts, it trips the breaker; if it opens the high voltage can do lots of damage!

R-E



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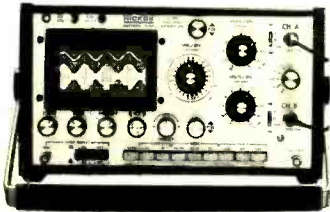


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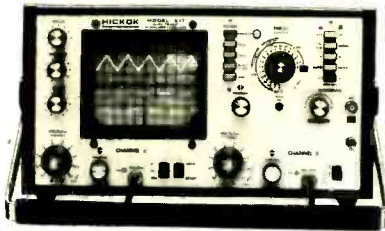
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RCA Receiving Tubes

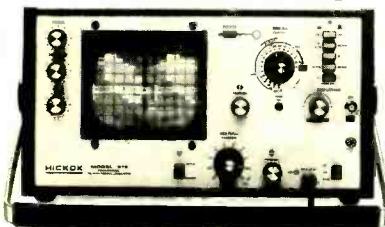
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CIRCLE 67 ON FREE INFORMATION CARD

new products

More information on new products is available from manufacturers of items identified by a Free Information number. Free Information Card is inside back cover.

600-MHz FREQUENCY COUNTER, model *OPTO-8000.1* (prewired) and model *OPTO-8000.1K* (kit), provides 8-digit LED display and guaranteed ± 0.1 PPM TCXO timebase stability, plus selectable input attenuator, selectable timebases (1 sec. and 0.1 sec.) and 9-15 VDC, 115 VAC or 220 VAC, 50-Hz-60-Hz operation. Available options include 220-volt AC operation and a NiCad battery pack. Unit comes in heavy-duty aluminum case with tilt-up bail, and measures 3 H X 7 1/4 W



X 6 1/2 inches D. Prices: Model *OPTO-8000.1* (prewired) \$299.95; model *OPTO-8000.1K* (kit) \$249.95; model 220 VAC option, \$10; model *NCAA 8000* NiCad battery pack with charging circuit, \$19.95; model *P-100 DC Probe*, \$13.95; model *P-101 Lo-Pass Probe*, \$16.95; model *P-102 Hi-Z Probe*, \$16.95. Add 5% shipping, handling, insurance to a maximum of \$10. Outside U.S. and Canada, add 10%.—**Optoelectronics, Inc.**, 5821 N. E. 14th Ave., Ft. Lauderdale, FL 33334.

CIRCLE 105 ON FREE INFORMATION CARD

SOLID-STATE OSCILLOSCOPE, model *PDC-200*, is available as a kit or assembled. The unit measures 4 1/4 X 7 1/4 X 3 inches and weighs 2 lbs, making it ideal for field servicing. The instrument features LED-type display; sweep ranges from 0.1 second to 0.5 μ s-per-division with voltage range



from 100-mV to 100V-per-division; and internal/external triggering. The circuitry is contained on a single PC board and powered by two 9-volt batteries. Probes are also available. Prices: the model *PDC-200* kit, \$99.95; assembled, \$155; probes, \$9.95. Prices do not include shipping.—**Pyramid Data Corp.**, Box 532, Barrington, IL 60010.

CIRCLE 106 ON FREE INFORMATION CARD

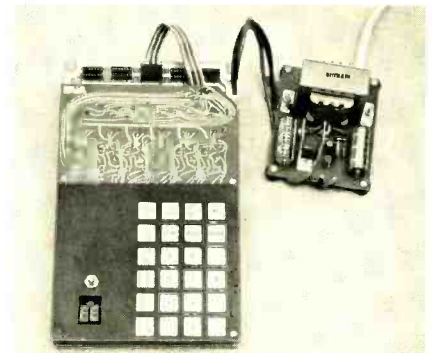
OPEN REEL TAPE DECK, model *A-6600*, provides two speeds (3 3/4 ips and 7 1/2 ips) and four heads (erase, record, playback and reverse playback). The system also incorporates full IC logic control, two-capstan servo tension system, 20-dB mike attenuator, cue control, automatic space



and individual mike and line controls. The front panel contains bias and equalization switches, two VU meters, input- and output-level controls and left- and right-channel selector switches. Also available are optional remote control, mikes, demagnetizer and recorder maintenance kit. Suggested retail prices: model *A-6600*, \$1299; model *RC-80* remote control, \$65; model *MM-100* mike, \$100; model *MM-120* mike, \$120; model *E-1* demagnetizer, \$25; and model *RMK* recorder maintenance kit, \$9.—**TEAC Corp. of America**, 7733 Telegraph Rd., Montebello, CA 90640.

CIRCLE 107 ON FREE INFORMATION CARD

OEM MICROPROCESSOR DEVELOPMENT SYSTEM, 8700 Computer/Controller, comprising CPU board and optional model *PS-87* power



supply, is based on 650x processor family. Socketed, plated-through board has space for 1K bytes of RAM and 1K bytes of PROM, both in 256-byte increments, five 8-bit parallel input

ports, one 8-bit parallel output port, plus connectors for system expansion. Monitor program (*Piebug*, interactive editor debugger) controls code entry and debugging. Other features include relative address, back-space key, pointer high and low keys, and audio beeper (used with optional *model CS-87* cassette interface). A variety of video display options is also available, including upper and lower case ASCII, color and black-and-white graphics. System comes as a kit or assembled, with prices starting around \$90. Other prices: *model PS-87* power supply, \$24.95; *model CS-87* cassette interface, \$22.50.—**PAIA Electronics, Inc.**, OEM Sales, 1020 Wilshire Blvd., Oklahoma City, OK 73116.

CIRCLE 108 ON FREE INFORMATION CARD

AMATEUR TV CONVERTER, model ATVC-10, provides reception of fast-scan ATV in 450-MHz band with any TV set. Device connects to VHF



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CIRCLE 13 ON FREE INFORMATION CARD

Technicians— RCA Flameproof Film Resistors simplify your replacement problems.



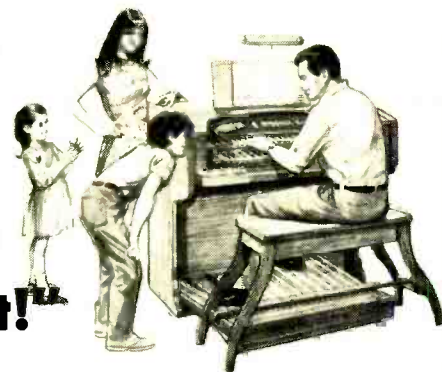
More and more technicians are switching to RCA Flameproof Film Resistors. Here are 5 important reasons why:

- Many of today's circuits require flameproof resistors.
- Simplified inventory: RCA's 2% tolerance flameproof resistors replace many conventional types having 5% and 10% ratings.
- High quality service demands high quality replacement parts.
- They're available in the popular wattage ratings — 1/4, 1/2, 1 and 2 watts.
- Values run from 1.0 Ohm to 1.5 Megohms. Altogether there are 523 film resistors to choose from.

And, RCA's convenient, preloaded and pre-labeled Flameproof Resistor Kit makes it easy to store and locate your resistors. See your RCA Distributor for details or write to RCA Distributor and Special Products Division, 2000 Clements Bridge Road, Deptford, NJ 08096.

RCA Flameproof Film Resistors

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helped assemble
this wonderful
Schober Organ...
now we all play it!"**



Talk about real *family* fun! We all worked together, for a few hours almost every day. Almost too soon, our Schober Organ was finished. Our keen-eyed daughter sorted resistors. Mom soldered transistor sockets, although she'd never soldered anything before. And it did our hearts good to see the care with which our son—he's only 12—installed the transistors. Me? I was the quality control inspector—they let me do the final wiring. And when it came time to finish the beautiful walnut cabinet the easy Schober way, we all worked at it!

Now, we gather around our Schober Organ every evening to play and sing together. Some of us play better than the others, but we're all learning—with the help of the easy Schober Organ playing courses. I might add that I'm especially pleased with all the money we saved. Our completed Schober Organ compares favorably with a "ready-made" one costing twice as much! (The five models range from \$650 to \$2850.) And we didn't even need to pay the whole amount all at once, because we were able

to buy Schober Kits a component at a time, to spread costs out. Or we could have had two-year time payments!

Families like ours have been building Schober Organs for 20 years. How about your family? You can have all the details, without cost or obligation. Just send the coupon for the fascinating Schober color catalog (or enclose \$1 for a 12-inch LP record that lets you hear as well as see Schober quality). Clip the coupon right now—and mail it TODAY!

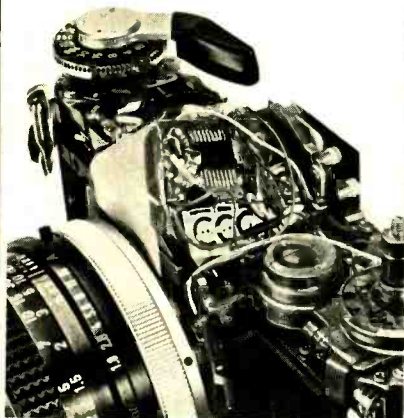
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CIRCLE 15 ON FREE INFORMATION CARD

antenna terminals on TV set. Features high-gain RF amplifier, mixer and oscillator (all varactor-tuned), plus adjustable RF gain. The converter tunes from 420 to 450 MHz and the output can be tuned to VHF Channels 2 through 6. Unit has built-in regulated AC power supply. The converter is housed in a wood-tone aluminum cabinet measuring 1 1/4 x 4 1/4 x 4 1/4 inches, and is available in both prewired and semi-kit form (critical circuits are prewired and aligned). Prices: *model ATVC-10*, \$49.95; kit, \$39.95.—**Science Workshop**, Box 393, Bethpage, NY 11714.

CIRCLE 109 ON FREE INFORMATION CARD

NI-CAD BATTERIES, *Dynacharge* line, come with a five-year warranty for 1000 discharges. Line includes AA penlight cell, C-cell, D-cell, 9-volt transistor cell, plus the *Dynacharger* to be used for all size cells. The *Dynacharger* lets you recharge penlight cells to 100% capacity in five



hours; unit can recharge four cells simultaneously, two different pairs at a time. Point-of-sale battery display contains blister-packed cards of cells plus boxed *Dynachargers*, and can be either pegboard-mounted or used as counter display.—**Dynamic Instrument Corp.**, 933 Motor Parkway, Hauppauge, NY 11787.

CIRCLE 110 ON FREE INFORMATION CARD

EPROM PROGRAMMER, *model POP-1*, low-cost unit interfaces to manufacturer's P-38-1 and P-38-FF EPROM boards which are SS-50 bus compatible. The device is designed to program 2708 EPROM's. Comes with software on audio cas-



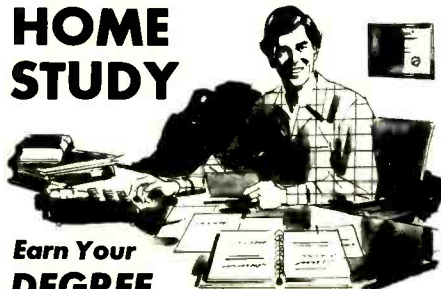
sette; special programming technique allows most 2708's to be programmed in 15 seconds. Power is supplied from separate self-contained source. Price: \$149.—**Smoke Signal Broadcasting**, Box 2017, Hollywood, CA 90028.

CIRCLE 111 ON FREE INFORMATION CARD

POWER BREADBOARDS, *Powerace model 101* (center), *model 102* (right), and *model 103* (left), accept all DIP IC packages, plus TO-5's and discrete components with leads up to .032-inch in diameter. All three models offer 256-5 tiepoint terminals and 16-25 tiepoint buses, fused power supply and ground plane.

The *model 101* has a variable 5-15 VDC, 600 mA, power supply with line- and load-regulation, a 0-15 VDC meter (5% full-scale accuracy). The *model 102* has a fixed 5 VDC 1-amp power supply, 4 slide switches with logic outputs, 2 momentary slide switches with debounce circuit-

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Sensitivity: 50 millivolts RMS at rest jack or 0.5 watts when connected in-line

Input Impedance: 1 Megohm 20 pF. rest jack input

Insertion Loss: Negligible. through in-line connectors

Accuracy: ± 1 digit ± time-base stability

Display: Six 0.30" LED digits


Time Base Frequency: 4 MHz. quartz crystal (factory trimmed to within ± 2 ppm)

Time Base Stability: Less than ± 5 ppm between + 10C and + 40C. after 15-minute warmup

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CIRCLE 80 ON FREE INFORMATION CARD

ry, 4 LED's, a debounce pushbutton and a clock generator (1, 10, 100, 1K, 10K, 100 kHz). The *model 103* has fixed 5 VDC 750-mA, fixed +15 VDC 250-mA and -15 VDC 250-mA power supplies, with tracking. It also includes a 15-0-15



VDC meter, 2 LED's, 2 slide switches with logic outputs, and 2 momentary slide switches with debounce circuitry. Prices: *model 101*, \$84.95; *model 102*, \$114.95; *model 103*, \$124.95.—**A P Products**, Box 110, 72 Corwin Dr., Painesville, OH 44077.

CIRCLE 112 ON FREE INFORMATION CARD

CASSETTE INTERFACE SYSTEM, *model CIS-30+*, which comes in kit form or preassembled, interfaces two cassette recorders and a data terminal to 6800 based computers. Long popular with SWTP 6800 users, newly released technical memos describe interfacing to the MITS 680b. The *model CIS-30+* features include selectable data rates of 30, 60 or 120 bytes/sec., and terminal rates of 300, 600 and 1200 baud; self-clocking cassette data recording; resident firmware; front-



panel "off-line" switch for sending recorded data to terminal; and an instruction manual. Optional program control of recorders is available. Other options include an IC socket kit, remote control kit, and a test cassette with software. Recording format is KC Standard/Bi-Phase-M. Prices: *model CIS-30+* kit, \$79.95; assembled, \$99.95; socket kit, \$4.95; remote control kit, \$14.95; test cassette, \$4.95. Prices include shipping charges; Texas residents add state and local taxes as applicable.—**PERCOM Data Co.**, 318 Barnes, Garland, TX 75042.

CIRCLE 113 ON FREE INFORMATION CARD

SOLDER EXTRACTING SYSTEM, *model SX-300*, *Universal Power Desoldering System*, has



three operational modes—pressure, vacuum and hot-air jet—for many kinds of solder-removal applications. System has built-in vacuum and pressure-generating capabilities, and includes a tip-temperature control, flow controls, pressure

and vacuum regulation, cleaning unit, soldering iron and remote-control pedal switch. Optional power switch prevents transients or spikes from reaching workpiece, making system usable with CMOS and MOS/FET devices. Prices: *model SX-300*, \$460; optional power switch, \$150.—**PACE, Inc.**, 9329 Fraser St., Silver Spring, MD 20910.

CIRCLE 114 ON FREE INFORMATION CARD

TAPE DECK ACCESSORIES, *Head Demagnetizer* (shown) and *Head Cleaner* kit. The *Head Demagnetizer* comes in cassette form with built-



in circuitry, powered by 1.5-volt dry cell battery. Just slip the unit into your tape deck, press PLAY, and red LED display shows when tape heads are completely demagnetized.

The *Head Cleaner* kit includes mirror, brushes, pads and liquid tape head cleaner, also in cassette box. Prices: *Head Demagnetizer*, \$19.95; *Head Cleaner* kit, \$5.95.—**TDK Electronics Corp.**, 755 Eastgate Blvd., Garden City, NY 11530.

CIRCLE 115 ON FREE INFORMATION CARD

BREADBOARDS, *model 45P80-1*, *model 106P106-1*, *model 8801*, permit convenient assembly of custom circuits or S-100 bus-compatible boards. All units feature an isolated array of square solder pads around 0.1-inch-spaced holes, and accept 8- to 64-lead DIP packages plus special modules with leads spaced on irregular multiples of 0.1 inch.

The *model 45P80-1* (measuring 4.5 × 8.08







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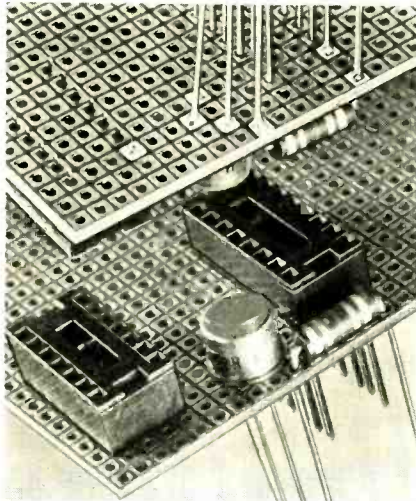
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CIRCLE 43 ON FREE INFORMATION CARD

inches) and the model 106P106-1 (10.6 × 10.6 inches) comes without card connectors and can be cut to any desired shape. The model 8801 is S-100 bus compatible and can accommodate DIP devices, modules and necessary discrete components. This unit has card-edge connectors on 0.125-inch centers, with a copper-free area for attachment of flat cable connectors. One corner

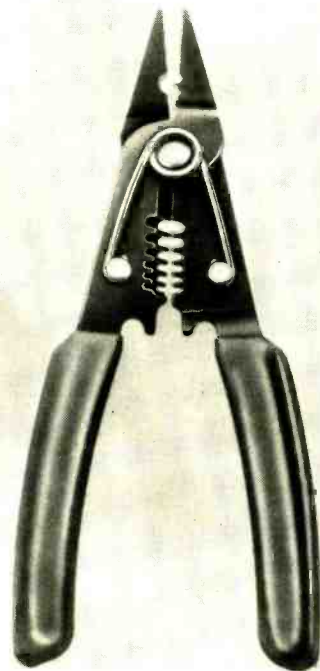


of the model 8801 has an area for two TO-220-packaged regulators mounted in a low-profile heat sink. The leads of one regulator are prewired to power, ground and primary power buses. The leads of the other regulator are uncommitted.

The boards are made of copper-clad blue epoxy glass with solder-tinned pads and buses and gold/nickel-plate edge connectors. The model 8801 sells for \$19.95; the model 45P80-1, \$9.96; and the model 106P106-1, \$18.99.—Vector Electronic Co., Inc., 12460 Gladstone Ave., Sylmar, CA 91342.

CIRCLE 116 ON FREE INFORMATION CARD

FOUR-IN-ONE ELECTRICAL TOOL, model 25502, *The Pike*, functions as a pliers with serrated tips for bending, gripping and pulling; a crimper for solderless terminals; a stripper with 6-wire stripping holes for AWG Nos. 12-22 solid-



wire and Nos. 14-24 stranded wire; and a 3/4-inch cutter. Unit is made of high carbon steel, with heavy plastic handles. Suggested retail price: \$4.74 per unit.—Hunter Tools, 9674 Telstar Ave., El Monte, CA 91731. R-E

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CIRCLE 19 ON FREE INFORMATION CARD

new lit

More information on new lit is available from the manufacturers of items identified by a Free Information number. Use the Free Information Card inside the back cover of this issue.

COMPONENT ASSORTMENT CATALOG, M-946A, is a 12-page tabloid brochure that features the most frequently used electronic components. Assortments come in plastic 6- and 9-drawer cabinets with drawer fronts pre-labeled as to contents, and include aluminum electrolytic, dipped mica, ceramic and tantalum capacitors; carbon-film resistors and Zener diodes; small-signal and power transistors; and switches.—**Sprague Products Co.**, Technical Literature Service, 81 Marshall St., North Adams, MA 01247.

CIRCLE 121 ON FREE INFORMATION CARD

SHORT FORM TEST INSTRUMENT CATALOG is a 12-page brochure describing manufacturer's new line of instruments, including single- and dual-trace oscilloscopes; DMM's, both pushbutton and conventional models; system analyzers, including a 7-instruments-in-one unit; color bar generator; counters, bridges and testers; CB, audio and RF signal generators; audio test instruments; and communications testers, including antenna couplers and impedance/dip meters. Numerous service aids and accessories are also listed.—**Leader Instruments Corp.**, 151 Dupont St., Plainview, NY 11803.

CIRCLE 118 ON FREE INFORMATION CARD

COMPOSITE TEST EQUIPMENT CATALOG, 13 full-color booklets and brochures in one handy looseleaf binder. Contains latest specs and information on products by B&K-Precision, Continental Specialties, Data Precision, ECD Corp., Fluke, Hickok, Leader, Nonlinear Systems, Philips, Sawa, Simpson, Triplett and Tif Instruments. Available for \$5.00; separate mail order form.—**Advance Electronics**, 54 West 45th St., New York, NY 10036.

ELECTRONIC SWITCH CATALOG, No. C-77, contains 47 pages of information on a wide spectrum of switches, ranging from keyboard types to thumbwheel and lever wheel models. Featured is the *Rotocode* switch that combines thumbwheel and rotary characteristics. Complete thumbwheel product listing and truth tables are in the back. Comes with a separate price list.—**Cherry Electrical Products Corp.**, 3600 Sunset Ave., Waukegan, IL 60085.

CIRCLE 120 ON FREE INFORMATION CARD

COMPONENT CATALOG, No. C-651, provides 28 pages of information on manufacturers *Q-Line* components most used in TV/radio servicing, hobbyist and experimental electronics, lab breadboarding and home audio and marine installation and repair. Catalog lists all types of capacitors (including trimmers), carbon-film and vitreous-enamel resistors, silicon and germanium transistors, rectifiers, diodes, IC's, quartz crystals, LED's, switches, wiring components, pulse transformers and CB noise filters.—**Sprague Products Co.**, Technical Information Service, 81 Marshall St., North Adams, MA 01247. R-E

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CIRCLE 69 ON FREE INFORMATION CARD

DIGITAL DARKROOM TIMER

continued from page 36

minates internal calculator modes by activating the Equals function. The 9 output from IC10 recalls the memory's contents and the 0 output clears the memory and resets latch IC14, which, in turn, halts IC10 with its 0 output high.

The 1 and 8 logic outputs of IC10 are not used. By skipping them, more time is allowed for the calculator IC to prepare for its following operation. This gives even a slow calculator IC enough time to complete its preceding operation before the next one is initiated.

Operation in the minutes/seconds range is similar to the operation in the seconds range. The outputs from IC6 and IC10 are combined by IC13 so the algorithm in Fig. 1 is executed.

AND gates IC9 and 12, combined with the 10-Hz signal from IC1-5, are required to satisfy the calculator IC's requirement that to recognize successive inputs, there must be a period (about 6 ms) after each input during which no inputs are allowed. The 10-Hz signal disconnects IC9 and IC12's inputs from IC8 and IC11 for 50 ms each cycle.

The electronic keyboard is formed by IC8 and IC11. Each contains four single-pole single-throw switches, which are

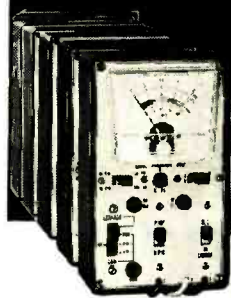
closed when "pressed" by a high at the gate inputs (pins 5, 6, 12 and 13).

The circuit which detects a minus (this means that only segment "g" is on) in position D8 works as follows. If you make a table that shows which segments are on for each of the numerals from 0 to 9, you will see that segment "b" is on for all numbers except 5 and 6. If we connect segment "b" to one input of a two-input OR gate and connect the gate's second input to any of the segments which are on (high) for both numerals 5 and 6 (segment "f" is used here), the output will be high if any numeral is being displayed. This also means that the output is low only if no numerals are being displayed. Thus by looking at only two of the seven segment signals, we can determine if the display is blank. (Eight other pairs of segments could also have been used—ab, ac, bc, bd, cd, ce, cg or bg.) By combining IC16-4 with D8 and segment "g," IC16-3 will be low only when a minus is in the D8 position.

Transistors Q2-Q4 and their base and collector resistors amplify the small 1.7-volt signal change across the on segments to the higher levels required by CMOS gates for reliable switching.

Next month we conclude the circuit description and will go into details of assembling, testing and operating this versatile timer. **R-E**

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next month

AUGUST 1978

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Want to use your TV set as a computer terminal, or as a monitor for a TV camera, or as the playing field for a TV game? Then you need a video modulator. Here's a rundown on every video modulator we could locate.

■ All About RF Generators

A special Forest Belt "What You Need To Know" section.

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Darkroom Timer Construction Details

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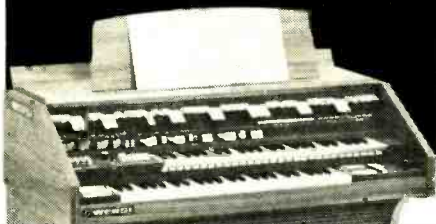
JULY 1978

87

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CIRCLE 31 ON FREE INFORMATION CARD

PICTURE TUBE BRIGHTENERS

continued from page 49

taminated, limiting emission. This contamination is caused by out gassing of tube elements or from slow air leaks. Examining the cathodes of such tubes reveals heavy crystalization in the center of the cathode. When a contaminated tube is turned on, the electron cloud under the grid aperture opening is rapidly depleted and the tube becomes emission-limited. In some cases, after the tube is on for some time, heat builds up around the cathode and raises the temperature of the cathode sufficiently to increase emission.

The effect of low emission is to limit the beam current and diminish picture brightness. If the brightness level is too low for normal viewing, the tube has apparently reached end-life.

Using a brightener on a contaminated tube causes a rapid decontamination of the cathode surface and raises the emission level. A satisfactory beam current is obtained almost instantly.

Restorers

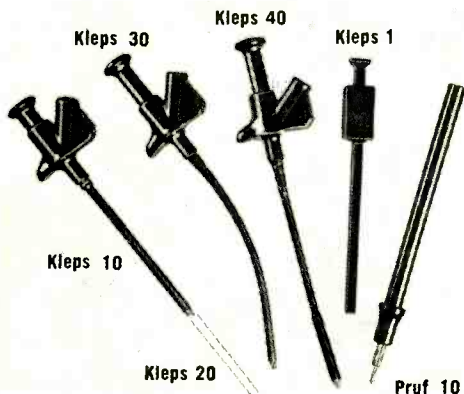
Devices known as restorers or rejuvenators are also designed to improve picture tube emission by removing contamination and restoring the emitter surface under the grid aperture. This technique involves elevating the heater voltage and simulta-

neously applying a high positive voltage between the grid and cathode, thereby causing an abnormally high cathode current. Unless this technique is carefully controlled, the cathode can be easily damaged. Several early restorers were hazardous and caused a fair number of picture tubes to be destroyed. In the capacitive discharge method, an arc is actually produced that provides intense localized cathode heating. However, the field, can be strong enough to actually remove material from the cathode.

Improved rejuvenators have automatic controls and can apply the required high voltage levels in increasing steps to limit the energy level. However, they generally do so at some sacrifice in the duration of the improvement.

The most recently developed rejuvenators use more effective control circuits so that the duration of the high cathode current is long enough to assure good restoration, but short enough to reduce damage to a minimum. Automatic controls are used to prevent user error.

One of the problems of the rejuvenator is that the cathode current flows from the entire cathode surface during rejuvenation. However, to be successful, the emission restoration must occur in the cathode area directly beneath the grid aperture. Fortunately, this is usually the area of greatest heat because the heater is in the center of the cathode cylinder.



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To know if rejuvenation has been successful it is necessary to measure the quality of the beam current after rejuvenation. Unfortunately, the grid current cannot indicate emission quality from the cathode area directly under the grid aperture; only a true beam-current-emission test can do this. This is why you may obtain disappointing results, even with tests indicating success.

In color sets, restoring one gun may mean one or both of the other guns must be rejuvenated. Rejuvenating the low-emission gun of a color picture tube can increase the emission to the original beam-current level, so that it may not track with the other aged guns whose beam currents may have declined from the original level, even though their brightness level is satisfactory.

Unsatisfactory picture tube performance can often be diagnosed visually by a repair technician, and modern test equipment and performance data help substantiate the visual judgment. A tube brightener can always be tried without risk (unlike many restorers), and it can be used in all applications in which a restorer is effective and in many in which it might not be. A brightener will work longer than a restorer; therefore, it can be implemented in all low-brightness-level situations.

R-E

**1
out of
2 who
have it
don't
know
it...**

1 out of 2

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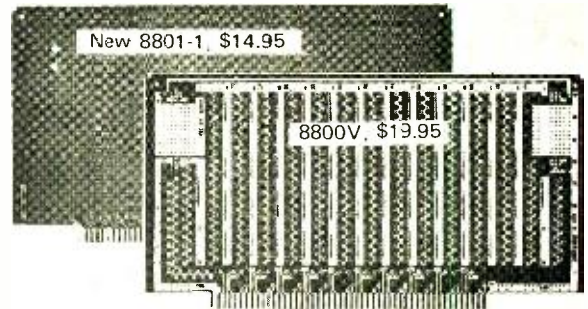
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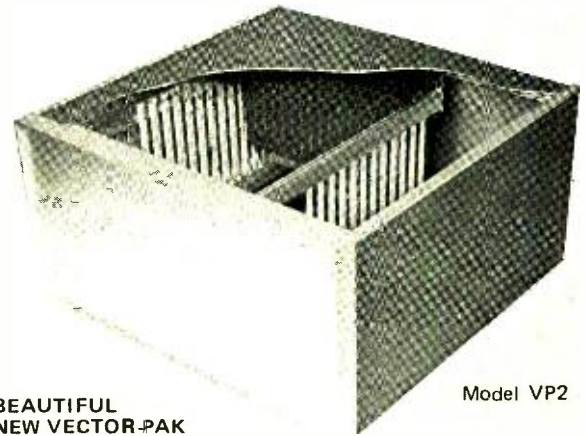
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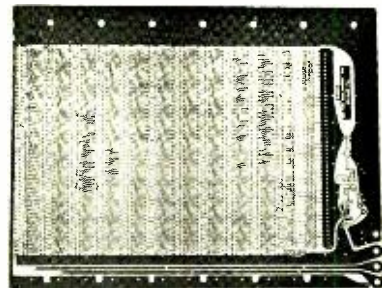
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2SA 440 9.0	2SA 815 1.00	2SC 184 .59	2SC 735 5.9	2SC 1211D 5.9	2SC 1738 1.90	2SD 370 3.00	LA 1201 2.50	UPC 41C 3.70			
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2SA 483 3.00	2SA 818 1.30	2SC 283 .59	2SC 737 .45	2SC 1213 1.70	2SC 1745 7.0	2SD 381 7.60	LA 3300 3.00	UPC 555H 2.25			
2SA 484 2.50	2SA 837 2.80	2SC 284 1.20	2SC 772 .70	2SC 1213 1.40	2SC 1885 7.0	2SD 381 1.40	LA 3301 3.00	UPC 563 3.70			
2SA 485 2.00	2SA 839 2.15	2SC 317 1.20	2SC 774 2.00	2SC 1215 7.0	2SC 1909 3.90	2SD 382 1.40	LA 4030P 3.40	UPC 566H 1.60			
2SA 489 1.60	2SA 841 .59	2SC 352A 2.50	2SC 775 1.95	2SC 1222 4.5	2SC 1913 1.20	2SD 388 3.40	LA 4031P 3.00	UPC 576 3.70			
2SA 490 9.0	2SA 847 .59	2SC 353A 2.50	2SC 776 2.65	2SC 1226A 9.0	2SC 1945 5.60	2SD 388A 3.40	LA 4032P 3.40	UPC 577H 2.25			
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2SA 562 .59	2SB 77 .45	2SC 388A 1.00	2SC 828 4.5	2SC 1327 .59	2SC 2098 4.90	FET		STK 011 9.50			
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2SA 565 1.10	2SB 175 4.5	2SC 403 .59	2SC 830H 5.90	2SC 1335 7.0	2SD 16 5.90	2SK 19BL .90	STK 016 8.50	SN 7400 .19			
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2SA 627 3.60	2SB 202 1.60	2SC 461 .59	2SC 870 .59	2SC 1347 .70	2SD 81 3.00	2SK 40 1.10	STK 502 8.80	IS 332 4.5			
2SA 628 5.9	2SB 220 .70	2SC 478 1.10	2SC 871 .59	2SC 1358 5.90	2SD 90 1.60	2SK 41 1.10	STK 503 9.60	IS 953 .25			
2SA 634 .70	2SB 303 5.9	2SC 481 1.60	2SC 895 4.90	2SC 1359 .59	2SD 91 1.60	2SK 49 1.30	STK 507 3.00	IS 1209 4.5			
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2SA 721 .59	2SB 523 1.00	2SC 645 .70	2SC 1061 1.40	2SC 1586 7.60	2SD 255 1.60	BA 511 2.90	STK 507 3.00	IS 10D10 .60			
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2SA 738 .70	2SB 529 .90	2SC 680 2.80	2SC 1096 1.00	2SC 1647 1.00	2SD 291 2.80	HA 1151 4.40	STK 507 3.00	IS 10D10 .60			
2SA 740 2.65	2SB 530 4.40	2SC 681A 5.80	2SC 1098 1.10	2SC 1667 3.40	2SD 300 5.60	HA 1151 4.40	STK 507 3.00	IS 10D10 .60			
2SA 743A 1.60	2SB 531 3.40	2SC 684 1.10	2SC 1111 3.40	2SC 1669 1.60	2SD 300 5.60	HA 1151 4.40	STK 507 3.00	IS 10D10 .60			
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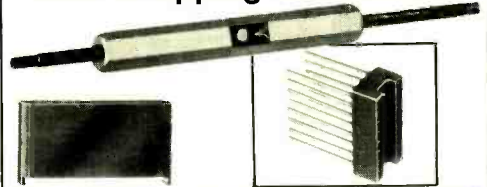


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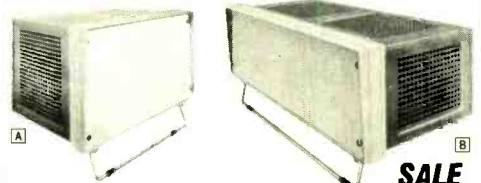
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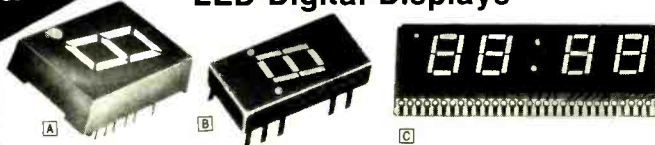
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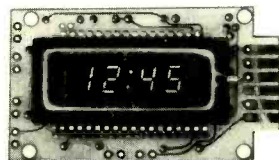
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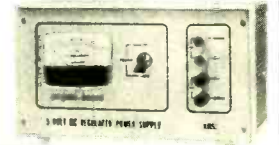


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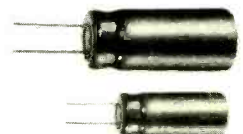
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Gate time: 1 second, 1/10 second, with automatic decimal point positioning on both direct and prescale option
Display: 8 digit red LED, 4" height
Accuracy: 2 ppm, internal TCXO standard
Input: BNC, 1 megohm direct, 50 Ohm with prescale option
Power: 110 V ac 5 Watts or 12 V dc @ 300.400 ma
Size: Approx. 6" x 4" x 2", high quality aluminum case

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Phone orders accepted 6 days, until 7 p.m.



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Resolution—1 Hz to 60 MHz; 10 Hz to 600 MHz
Decimal Point—Automatic
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Display—8 digit LED
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Approximate Size—3" h x 7 1/2" w x 6 1/2" d
Approximate Weight—2 1/2 pounds
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ELF II features an RCA COSMAC COS/MDS 8-bit microprocessor addressable to 64k bytes with DMA, interrupt, 16 registers, ALU, 256 byte RAM, full hex keyboard, two digit hex output display, 5 slot plug-in expansion bus (less connectors), stable crystal clock for timing purposes and a double-sided, plated-through PCB board plus RCA 1801 video IC to display any segment of memory on a video monitor or TV screen.

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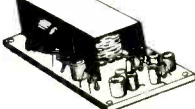
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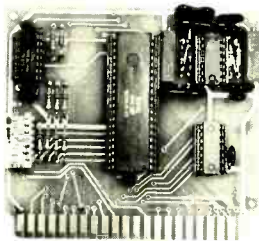
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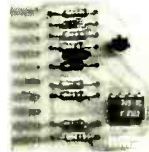
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Part no. 111

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- No coils

Part no. 107

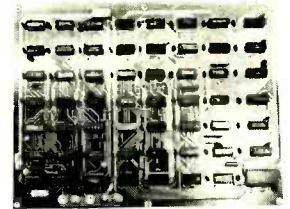
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LITTLE PRO BEEP
continued from page 60

= 47,000 ohms and $C = .05 \mu F$ as the timing components (Fig. 1). Then $T = 2.35 \text{ ms}$. A complete cycle at pin 1 would last 4.7 ms, which converts to a frequency of about 210 Hz. After 7 stages of divide-by-2, the frequency would be 210/128 or nearly 2 Hz, at pin 8. We mentioned previously that when any of the counter outputs goes to a logic-0 level, there can be no signal output regardless of what happens at any other counter output. Therefore, during the long "0" state of pin 8, we cannot hear the audio from pin

1. The 210 Hz becomes available *only* while pin 8 is at a logic-1 level. We hear beeps, intermittent sound.

Now suppose we strap pins 1, 6 and 8 to the output terminal. (See Fig. 4.) We know that the logic-1 state of pin 8 will block all output. During the logic-1 state, there are four alternations at pin 6: two of them logic-1 and two logic-0. As we know, there can be no output during the logic-0 intervals. However, the tone or buzz from pin 1 *will* be audible during the 2 logic-1 intervals, shown shaded. This provides the *double* beep.

To obtain a 4-beep sequence, we can strap pins, 1, 5, 8 to the output, and so on. If you strap pins 1, 4, 6, 8, you will obtain a sequence of 2 double-beeps.

Figure 5 is the schematic of the beeper. Use a 16-pin DIP socket for the IC, and a socket adapter for the capacitors and resistors. Most of the wiring can be done at or near the socket. The switches, earpiece jack (for output) and power plug can be mounted on a small metal box, 3 1/4 x 2 x 1 1/2". A source of 5 volts at 6 mA will operate the beeper.

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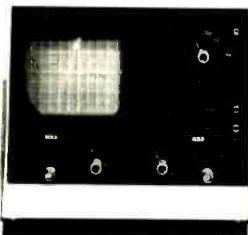
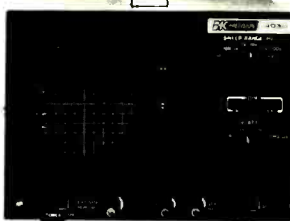
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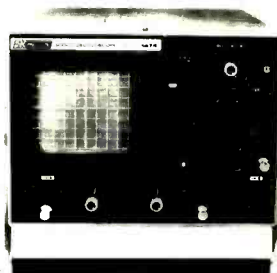
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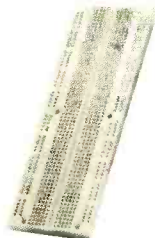
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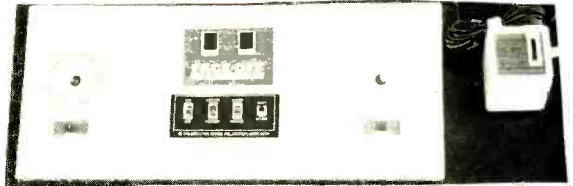
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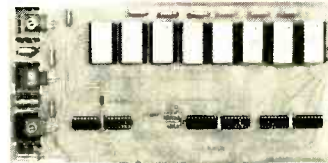
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- ★ ★ S-100 bus compatible
- ★ ★ Optional wait states: none, one or two (selectable with plug-in jumpers)
- ★ ★ Epoxy board with plated thru holes (double sided)
- ★ ★ Address any one of 8 different 8K blocks or any one of 16 different 4K blocks
- ★ ★ All sockets provided
- ★ ★ Gold plated connectors
- ★ ★ On board regulation (heat sinked)
- ★ ★ Convertible for use with 2716 EPROMS
- ★ ★ Complete instructions



2708 EPROM BOARD

AVAILABLE SOON: (S-100 BUS)

★★ Phase encoded cassette interface with programmable serial port: Dual recorder, tarbell or K-C method supported ★★ 2708 Programmer

4th OF JULY SPECIAL (July & August)

ASSEMBLED & TESTED (without EPROMS) \$55.50

COMPLETE KIT (without EPROMS) \$42.50

BARE BOARD ONLY \$25.50

2708 SPECIAL 8/\$68. (with purchase of board)

TVT-III (with 2K memory capability)
This system provides 16 lines of 32 characters each. It is composed of all TTL except for the 2513 and 2102's. Basic kit includes: Main board, 2K Memory board, six 2102's (1K of memory), plus other components. (Keyboard, case and power supply not included.)

All boards are double sided and plated thru. Small boards plug into main board at right angles.

	KITS	ASSEMBLED	BOARD ONLY
Basic TVT-III	\$99.95	\$135.45	\$32.50
Screen Read	12.65	17.30	8.10
Manual Cursor	9.95	13.95	6.50
U.A.R.T.	29.95	39.95	15.50

Construction packet sold separately: \$2.95

AVAILABLE NOW! 64 CHARACTER CONVERSION AND/OR SCROLLING OPTION FOR THE TVT-III.
Complete documentation. Kit: \$6.95. Data only: \$1.00

MEMORY BARGAIN:
16K STATIC RAM ON ONE BOARD: (32 ea. μPD410); S-100 BUS; Low Pwr., 200NS 4K Chips. Kit includes all IC's, discrete parts & sockets: \$295.00

PROGRAMMING SERVICE

1702A: \$5. original program, copies \$2. (same order)	2708: \$10. original program, copies \$4. (same order)	2716: \$20. original program, copies \$8. (same order)
--	--	--

WE REQUIRE A WRITTEN PROGRAM IN EITHER HEX OR SPLIT OCTAL or a CHIP ALREADY PROGRAMMED.

CPU's	EPROMS		
Z-80A	\$19.50	2716, 16K	\$29.50
8080A	9.50	2708, 8K	9.50
6800	19.95	1702A, 2K	3.95

RAMS

21L02 (250ns)	1.50 ea.; 8/\$11.50
21L02 (450ns)	1.35 ea.; 8/\$10.00
2107B, 4K DYNAMIC	\$3.50
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2513 upper case	5.50
MM5330 4 1/2 DIGIT DVM CHIP	7.95
CT7001 clock/cal. chip	5.75

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LM1889 TV video mod.	10.95
MC1488, 1489—RS232 & TTL interface	1.95

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6 PIN ROCKER	1.80
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DUAL GOLD PLATE EDGE CONN. (.156")

6 POS. 12 CONTACT	\$.40
10 POS. 20 CONTACT	.50
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2.097152 MHz	MM5378	\$8.50
	MM5379	
	Auto Clock	
2.4576 MHz	34702 Bit Rate Generator	\$8.50
2.667 MHz	Pace	\$8.50
3.00 MHz	MCS 48	\$8.50
3.20 MHz	RCA 1803	\$8.50
3.2768 MHz	Intersil ICM 7205	\$8.50
3.579545 MHz	NSC MM5369 (TV Color Burst)	\$1.25
4.0 MHz	Intel 4040/4004/4201	\$4.95
4.0 MHz	Pace	\$4.95
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8.0 MHz	Signetics	\$4.95
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All capacitors are Brand New U.S. made in standard size

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11000MFD	35V	\$3.20 EA.
20000MFD	55V	\$3.50 EA.
23000MFD	20V	\$3.00 EA.
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(Size of a Match Box)

This unit converts the video signal to RF signal. Ideal for computer terminal or TV games.

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Preset at 61.25 MHz (channel 3)
Frq. adjustable ± 3 MHz
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(On-Off Contacts)

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5 positions	\$1.60
6 positions	\$1.70
7 positions	\$1.70
8 positions	\$1.80
10 positions	\$2.00

8K STATIC RAM MEMORY BOARD (BARE) \$24.50 PER BOARD

Memory Chips: 21L02 X 64 pcs.
Fully buffered; all support chips are LS type; total of 26 bypass capacitors on the board; four 340T-S voltage regulators give more than enough power to the board.

16K STATIC RAM MEMORY BOARD (BARE) \$24.50 PER BOARD

MEM-2 uses 2114 any speed any power
Start address and stop address selected on board so board can be used portionally populated; capable of block address, can be used in SOL_{TM} and systems without front panel.

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This Quiet Mother is standard 2-100 BUS with 13 slots; Kluge area; extensive ground plane; passive terminal for all lines; double solder mask; silk screen parts layout.

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With 8 level vector interrupt. CPU chip 8080A; clock chip: 8224; X'tal freq.: 18 MHz; vector interrupt chip: 8214.

RTC-1 REAL TIME CLOCK BOARD for S-100 BUS \$24.50 BARE

On Board Oscillator: 1 MHz Crystal
Number of Interrupt: 2, one is decade; it steps from 100 μ Sec to 100 Sec.
Second is 16 bit counter, it is selectable from 10 μ Sec to 655, 360 μ Sec in 10 μ Sec step.

All above boards include all documents and detailed parts list and instructions.



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Includes all parts, PC Board and Transformer

Take advantage of this new state-of-the-art counter featuring the many benefits of custom LSI circuitry. This new technology approach to instrumentation yields enhanced performance, smaller physical size, drastically reduced power consumption (portable battery operation is now practical), dependability, easy assembly and revolutionary lower pricing!

- 0.5" red LED 6 digits display
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\$6.95 EACH KIT

This new model FM wireless MIC kit uses 3 high freq. transistors, works in the FM range (88-108 MHz). It transmits the sound wave fidelity clearly over long distances (up to 2500 ft.). Kit comes with all electronic parts, P.C. Board and mini microphone!

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All parts completed on a PC Board SCR will turn on relay, buzzer or trigger other circuit for 2-10 sec. (adjustable). Ideal for use as door alarm, sound controlled toys and many other projects. Supply voltage 4.5V-9V D.C. \$1.75 ea./2 for \$3.00

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12V DC POWERED
Lights up 8-15 Watt Fluorescent Light Tubes
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Kit includes high voltage coil, power transistor, heat sink, all other electronic parts and PC Board, light tube not included!

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6 Functions with % and memory
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*Built-in X'tal controlled stop watch count to 1/10 of a second.
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BATTERIES NOT INCLUDED



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All parts are pre assembled on a mini PC Board
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Kit includes: 2 pcs. Fisher PA 301 Hybrid IC all electronic parts with PC Board. Power supply ± 16V DC (not included). Power band with (KF=1% ± 3dB) Voltage gain: 33dB, 20Hz-20KHz.

Super Buy Only **\$22.50 each kit**

5W AUDIO AMP KIT

2 LM 380 with Volume Control
Power Supply 6-18V DC
only \$5.00 ea.



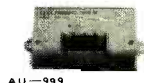
TIMER KIT

Time Controlled from 1-100sec.
Ideal to be used as time delay unit for burglar alarm, photo service, and other purposes.
Max. loading 110V, 2 AMP
Supply voltage 12-18V D.C.
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COMPLETE UNIT
Ideal for use as an Alarm Unit, or hook up to your car back up to make a reverse indicator Light Output up to 130 dB.
Voltage Supply 6-12V
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Designed for Calculator. Can be used for Computer Data Entry Pad or Digital Lock. All key tops blank with super good touch feeling. **\$0.95 ea.**



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400R RELAY AND KEY PAD NOT INCLUDED



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0-30V D.C. REGULATED
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MA1003, 12V DC CLOCK MODULE



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Built in X'TAL controlled time base. Protected against automotive volt transients. Automatic brightness control with 0.3" green color display. Display turnoff with ignition "OFF".

12V DC MINI RELAY

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2AMP Contact SPDT \$1.30 EA
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12V D.C. AUTO DIGITAL CLOCK

Complete Unit Not a Kit!

0.4" blue color 4 digits display. Turn off readouts when car is not running. X'tal controlled time base for time accuracy. Special designed case for easy mounting on top of your dashboard. Ideal for car, boat and campers.

ONLY \$28.50 ea.



ELECTRONIC SWITCH KIT

CONDENSER TYPE
Touch on Touch Off
use 7473 IC and 12V relay
\$5.50 each

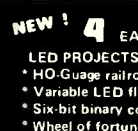


NEW! 4 EASY-TO-BUILD KIT

LED PROJECTS ON ONE BOARD

- HO-Gauge railroad crossing blinker
- Variable LED flasher
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Kit includes all electronic components and PC Board. Especially designed for experimenters. Guaranteed - they all work!! (Battery not included)
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LED ALARM CLOCK

COMPLETE UNIT NOT A KIT

- 0.5" RED LED READ OUT
- 24 HRS. ALARM SET
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- AM/PM ALARM INDICATORS
- SECOND DISPLAY SWITCH
- AUTOMATIC BRIGHTNESS CONTROL
- COMPACT AND HANDSOME PACKING
- 110V AC 60HZ INPUT

\$17.50 EACH



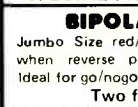
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5 Blade Type
110V 50/60HZ
Case made of Die-Cast metal used but almost brand new
SPECIAL PRICE \$9.60 ea.



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Jumbo Size red/green change color when reverse polarity of voltage. Ideal for go/nogo indicator.
Two for **\$1.50**



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Same as the E-Z clips With 20" Long Leads In Black and Red Colors
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5 Sub C pack	SIZE	\$4.90 EA

*12V Mini Size Motorcycle rechargeable w/rt battery 6AMP/Hour Brand new, but ltd quantity **\$12.50 EA.**



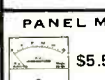
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Size 60MM X 68MM
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0.1MA 0.50V
0.5MA 0.100V
0.100MA



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24V CT	0.8AMP	\$1.80 EA.
12V CT	0.8AMP	\$1.80 EA.



AC POWER SUPPLY

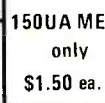
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12V AC	Output	200MA	\$2.75 EA.
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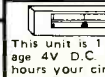
150UA METER

only **\$1.50 ea.**



HOUR INDICATOR

This unit is 1 5/8" long operating voltage 4V D.C. will tell you how many hours your circuit or machine has been in service up to 100 hours.
Limited Quantity **ONLY \$1.75 ea.**



MINI-MINI TOGGLE SWITCH

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NEW! 16K x 8 ECONORAM IV™ KIT (\$329)

Guaranteed current consumption 2000 mA. Manual write protect switches for 4K blocks; use with or without phantom line. Fully static. Add \$35 for assembled/tested.

NEW! 24K x 8 ECONORAM VII™ KIT (\$490)

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NOTE: The above 3 boards are guaranteed compatible with S-100 systems running at 2 MHz. However, due to our conservative design, users report excellent results in 4 MHz systems also.

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2N718	.25	2N4093	.85
2N930	.25	2N4124	.16
2N956	.30	2N4126	.16
2N1302	1.25	2N4141	.20
2N1305	.75	2N4142	.20
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2SB324	.60	2SC1018	1.00	2SC1975	3.50	3SK22Y	2.20
2SB370	.65	2SC1061	1.25	2SC2028	.80	3SK40	2.00
2SB405	.60	2SC1079	3.95	2SC2029	3.40	3SK45	2.20
2SB415	.65	2SC1096	.80	2SC2091	2.50	3SK49	2.20
2SB461	1.60	2SC1098	1.00	2SC2092	3.25		
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16 pin	39	38	36	32	27
24 pin	42	42	39	35	30
18 pin	53	54	47	42	35
20 pin	80	75	70	65	53
24 pin	90	85	80	70	57
24 pin	90	84	78	68	58
28 pin	110	100	90	84	71
36 pin	150	140	120	104	89

SOCKETS PURCHASED IN QUANTITIES OF 50 per type may be combined for better prices

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8803

MOTHER BOARD FOR S100 BUS MICRO-COMPUTERS

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- Working side viewer. Component side bare epoxy glass with white markings for component locations
- Grid epoxy glass board with 2 ounce copper solder plated area. 138 diameter holes for leads
- Solder mask with solder windows on critical circuits to avoid accidental shorts
- Mounts 11 receptacles with 100 contacts (2 rows) on 125 centers with 250 row spacing
- Vertical part number 8801-2 or mounts 10 receptacles plus interconnections to smaller mother board for expansion
- Includes etched circuits and instructions for option of active pull up or floating terminations
- Large buses - 50 mil (DIP AMP'S) + 12V or 16V (17 AMP'S). Current ratings as per MIL-STD-275 with 100% test
- Fits in Vector case enclosures
- Fits in MS-1 8800 microcomputer as expansion board

Price: \$29.50

Vector

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8800V Universal Microcomputer/processor plugboard use with S-100 bus. Complete with heat sink & hardware. 5.3" x 10" x 1/16"

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8801-1 Same as 8800V except plain, less power buses & heat sink

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P pattern plugboards for IC's Epoxy Glass 1/16" 44 pin con. spaced 156

3690-12 CARD EXTENDER

Card Extender has 100 contacts-50 per side on 125 centers-Attached connector-is compatible with S-100 Bus Systems \$25.00
3690 6.5" 22/44 pin .158 ctrs. Extenders \$12.00

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Sockets are End & Side stackable. closed entry

1/16" Vector BOARD

.042 dia holes on 0.1 spacing for IC's

Phenolic

PART NO.	SIZE	PRICE
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169P44XXXP	4.5x17"	\$3.51 3.16

Epoxy Glass

PART NO.	SIZE	PRICE
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84P44	4.5x8.5"	\$2.10 1.89
169P44	4.5x17"	\$4.30 3.87
169P84	8.5x17"	\$7.65 6.89

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P180 with two 100' spools of 28 ga wire \$24.50

P180-AT includes charger wire \$75.00

SLIT-N-WRAP WIRE NO. 28 GAGE INSULATED WIRE, 100' SPOOLS

W28 2 Pkg 3 Green	W28 2 Pkg 3 Blue
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"Day clean" runs separate and sets to use. Manual or power operation

2708 8K 450 ns

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25 + Call For Price

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- For Auto, Home, Office
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- Clocks mount anywhere with either 3M double-sided tape or VELCRO, included
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Perfectly balanced, fluorescent lighting with precision magnifier lens. For profit, technician, hobbyist. Has die cast protective shade, instant start 3 collector lens 42" reach

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MICRO-KLIP for .042 dia holes (all boards on this page)

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P-149 hand installing tool \$ 2.03

12:28

8" LED ALARM CLOCK

12 hr. LED Alarm Clock uses 3 1/2" digit 8" LED Display with AM/PM indicators and colors. Direct drive, PIN to PIN interface with AT959A IC. Just add switches, AC Supply Alarm Display and IC only

\$7.95 or 2/\$15.00

Price Breakthrough! \$17.50

MA1003 CAR CLOCK

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SPECIAL

14CS2 100 for \$14.00
16CS2 100 for \$16.00
14 pin CS2 10 for \$2.25
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Vector

WRAP POST for .042 dia. holes (all boards on this page)

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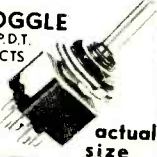
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24 PIN DIP PLUGS WITH COVERS

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EXTRA SMALL D.P.D.T.
3A 125VAC CONTACTS
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8 - 10.
100 - \$105.



\$1.35 10-\$12
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Single pole double throw
center off (3 position)
toggle switch.....

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POWER TRANSISTOR MOUNTED ON A
FACTORY ASSEMBLED HEAT SINK -
3.8 WATTS @ AMBIENT OF 25C
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1 AMP 20 WATT DC CONTACTS
LONG LIFE (1 X 10⁹) OPER.
HIGH SPEED (500 HZ)
HERMETICALLY SEALED 3/8X3/8X3/4"
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ultra sensitive relay
Single pole normally
open dry reed relay.
Unbelievable sensi-
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with 3 volt pull-in
(that's less than 3
ma.) A TTL gate
will drive many of
these. A CMOS buffer
will directly actuate
this nifty relay....
NPN, unknown, but I
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these. A CMOS buffer
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this nifty relay....
600 available
\$2.95 ea.



ELEC-TROL 12 vdc
REED RELAY 440 Ω
coil
dpst n.o.
1.1"
PICTURE IS APPROXIMATION. RELAY HAS
FOUR PINS ON ONE END AND TWO ON OTHER.
PINS ARE SPACED .1" X 1". 1/2A CONTACTS.
79¢ ea. 10 - \$6

POTTER BRUMFIELD
Type KHP Relay
4 PDT 3A Contacts
24VDC COIL
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\$1.60 ea.

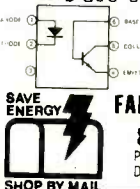


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multiplexed 7 seg. 1.25
MM5369 time base osc. chip 1.75
T.V. crystal for above.... 1.75
1101A 256 x 1 MOS RAM59
93410 256 x 1 45NS RAM85
75451 peripheral driver40
7402 TTL nor gate16
7430 TTL NAND gate14
CERAMIC FILTER 10.7 MHz.1.20
MM5058 1024 bit static SR.1.50
UA715 60MHz. op-amp1.20
74S206 256 x 1 45NS RAM90
78L2.6 2.6V TO92 .1A +regul.29
78L05 5.0V " " " "29
78L6.2 6.2V " " " "29
78L12 12V " " " "29
78L15 15V " " " "29
78M05 5V tab .5A + regul.59
78MG 4 term. adj. .5A + reg. 1.29
79MG " " " " - reg. 1.39
7805T tab 1A 5v regul.89
7812 to-3 1A +12V regul.59

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7404	.19	74126	.65	745112	.65	8859	1.00
7406	.40	74132	1.25	745114	1.58	8865	1.50
7407	.40	74141	1.15	745134	.85	8866	1.50
7410	.25	74145	1.10	745153	1.89	8867	1.85
7411	.18	74154	1.25	745157	1.55	8869	1.75
7413	.78	74157	1.00	8000 TTL		8879	2.25
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7417	.38	74163	1.30	8094	.60	8884	2.45
7420	.18	74164	1.45	8095	.80	8973	2.95
7421	.35	74165	1.35	8096	.90	8974	2.95
7427	.35	74173	1.70	8098	.90	8976	3.25
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7432	.30	74177	.95	8136	3.25	75450	1.00
7437	.44	74182	.95	8220	3.25	75451	.80
7440	.18	74192	1.45	8231	2.25	75452	.80
7442	1.00	74193	1.35	8242	1.75	75453	.80
7445	.70	74195	1.00	8250	1.75	75491	1.25
7446	.70	74196	1.10	8250	2.25	75492	1.40
7447	.70	74197	1.10	8281	1.00	75494	1.50
7450	.25	74199	2.25	8284		8284	
7451	.25	74367	.90	8285		8334	4.00
7453	.25			8334	4.00	8553	6.50
7454	.35			8553	6.50	8556	3.25
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7472	.40	74H04	.40	8599	3.25	Interface	
7473	.40	74H08	.40	0025	3.50	8251	11.50
7474	.40	74H10	.40	0026	1.75	8255	10.50
7475	.55	74H11	.40	8640	1.25	2513	9.50
7476	.55	74H30	.65	8641	2.75	2516	9.50
7483	1.05	74H40	.65	8641	2.75	1013	6.50
7485	1.10	74H53	.65	8806	3.00		
7486	.43			8819	1.25	561	5.00
7489	2.00	74S00	.40	8820	5.00	566	1.70
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\$16.95

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Assembled & Tested **\$29.95**

Automatically Starts & Stops Tape Recorders
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7402	13	7444	59	7492	43	74160	55
7403	13	7447	68	7493	43	74161	65
7404	15	7448	71	7494	67	74163	65
7406A	29	7450	13	7495	67	74164	85
7407A	44	7451	13	7496	67	74165	89
7408	16	7453	13	74100	30	74174	85
7408	19	7460	19	74104	49	74175	85
7410	13	7470	27	74107	28	74180	67
7411	18	7472	21	74109	31	74181	93
7412	26	7473	29	74121	29	74182	68
7416	15	7474	29	74123	48	74191	98
7420	13	7477	47	74132	99	74192	79
7421	25	7476	21	74136	95	74193	81
7425	29	7480	31	74538	95	74194	81
7433	26	7481	55	74141	75	74195	69
7437	23	7482	57	74151	61	9316	85
7438	23	7483	67	74153	61	9601	35/1
7440	13	7485	89	74154	98	9104	35
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7408	10:1	00	that you 74141	3:1	00
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74LS03	21	74LS54	26	74LS139	71	74LS266	26
74LS04	28	74LS55	26	74LS145	100	74LS279	55
74LS05	28	74LS73	35	74LS151	70	74LS293	61
74LS08	21	74LS76	35	74LS153	70	74LS293	61
74LS09	28	74LS74	49	74LS155	69	74LS295	95
74LS10	21	74LS83	73	74LS156	70	74LS298	95
74LS11	21	74LS85	135	74LS157	75	74LS365	55
74LS13	45	74LS86	86	74LS158	71	74LS366	55
74LS14	99	74LS90	55	74LS160	85	74LS367	55
74LS15	26	74LS92	55	74LS161	85	74LS368	55
74LS20	24	74LS93	55	74LS162	85	74LS390	1.75
74LS21	28	74LS99	38	74LS163	85	74LS393	1.75
74LS22	28	74LS112	38	74LS164	149	74LS670	2.40
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74LS27	32	74LS114	38	74LS169	85	74LS193	95
74LS30	26	74LS122	99	74LS170	119	74LS194	95
74LS32	32	74LS124	99	74LS173	110	74LS195	85
74LS37	32	74LS125	47	74LS174	100	74LS196	85
74LS38	32	74LS126	47	74LS175	81	74LS197	85
74LS40	26	74LS127	79	74LS190	85	74LS251	85
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CD4011	19	CD4024	75	CD4049	35	CD4518	1.10
CD4012	78	CD4025	19	CD4050	39	CD4520	69
CD4013	32	CD4027	39	CD4051	1.19	CD4528	85
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WITH COLONS & AM/PM INDICATOR
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\$3.95 Ea.

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9006/7408	6/1 00
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ITT 508	Hex Digit Dr.	49
ITT 509	8 Seg. Dr.	49
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ITT 514	8 Digit Dr.	59

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IN 1958	8.2V 400M	4/\$1
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HOUSE # ZENER

4.7V	500 MW	10/\$1
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10V	1 Watt	8/\$1
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22V	5 Watt	6/\$1

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22PF 10	10/\$1		
CERAMIC DISC			
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100PF 100V	002FMD 200V		
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330PF 50V	33PF 500V		
001MFD 500V	100PF 500V		
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Part No.	No. of Contacts	Length	Price
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924005-18R	40	18"	8.27 ea.
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924006-18R	50	18"	10.31 ea.
924006-06R	50	6"	9.15 ea.

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923863-R	26	straight	\$1.28 ea.
923873-R	26	right angle	1.52 ea.
923865-R	40	straight	1.94 ea.
923875-R	40	right angle	2.30 ea.
923866-R	50	straight	2.36 ea.
923876-R	50	right angle	2.82 ea.

DIP JUMPERS

Mates with standard IC sockets
24" length • Fully Assembled & Tested

CONTS DESCRIPTION	PART NO.	PRICE
14 sgl end	924-102-24	\$ 1.92
14 dbi end	924-106-24	3.02
16 sgl end	924-112-24	2.13
16 dbi end	924-116-24	3.34
24 sgl end	924-122-24	3.30
24 dbi end	924-126-24	5.20
40 sgl end	924-132-24	5.53

Also Available in 12" and 36" lengths

CRYSTALS

THESE FREQUENCIES ONLY

PART NO.	FREQUENCY	CASE	PRICE
CY1A	1.000MHz	HC33	5.95
CY1.84	1.8432MHz	HC33	5.95
CY2A	2.000MHz	HC33	5.95
CY2.01	2.010MHz	HC33	1.95
CY2.50	2.500MHz	HC33	4.95
CY3.27	3.2768MHz	HC33	4.95
CY3.57	3.579545MHz	HC33	4.95
CY3A	4.000MHz	HC18	4.95
CY4	4.916MHz	HC18	4.95
CY4.91	5.000MHz	HC18	4.95
CY7A	5.000MHz	HC18	4.95
CY5.18	5.185MHz	HC18	4.95
CY6.14	6.144MHz	HC18	4.95
CY6.40	6.400MHz	HC18	4.95
CY6.55	6.5536MHz	HC18	4.95
CY6.55	10.000MHz	HC18	4.95
CY12A	14.31813MHz	HC18	4.95
CY19A	18.432MHz	HC18	4.95
CY18.43	20.000MHz	HC18	4.95
CY22A	32.000MHz	HC18	4.95
CY30A	32.000MHz	HC18	4.95

TRIMMERS

10MM size trimmers - .394" Dia.
Part No. 1-9 10-24 25-49 100+
TR-11 (value) .35 .30 25 20

TRIMPOTS

Single-Turn - 1/2 Watt
Square - Top Adjust - 3/8" Size
Part No. 1-9 10-24 25-49 50-99
63P (value) .99 .89 80 70

15-Turn - 3/4 Watt

Rectangular Side Adjust 3/4" x 1/4" Size
Part No. 1-9 10-24 25-49 50-99
43P (value) 1.35 1.25 1.20 1.15

1/16 VECTOR BOARD

PHENOLIC	Part No.	Hole Spacing	P. Pattern	L	W	1-9	10 up
	64P4 062XXXP	4.50	6.50	1.72	1.54		
	169P4 062XXXP	4.50	17.00	3.69	3.32		
EPoxy	64P4 062WE	4.50	6.50	2.07	1.86		
GLASS	64P4 062WE	4.50	8.50	2.56	2.31		
	169P4 062WE	4.50	17.00	5.04	4.53		
	169P4 062WE	8.50	17.00	9.23	8.26		
EPoxy GLASS	169P4 062WEC1	4.50	17.00	6.00	6.12		

ELITE WRAP

Model P180 includes 2-100' spools #28 AWG wire wrap wire

Supplies insulated wire from spool to wrap-posts without prestrapping and precutting using "daisy chain" method.

Model P180

REPLACEMENT WRAP BIT for P180 Slit N Wrap No. P180A \$12.95 each

Replacement wire-wrap wire for P180 #28 AWG (pkg of 3) \$2.75 each
W28-2-A green W28-2-B red
W28-2-C clear W28-2-D blue

INSTRUMENT/CLOCK CASE

Injection molded unit Complete with red bezel 4 1/2" x 4" x 1-9/16"

\$3.49

MICROPROCESSOR COMPONENTS

Part No.	Description	Price	Part No.	Description	Price
P8085	CPU	\$29.95	CDP 1802	CPU	\$19.95
8080A	CPU	10.95	Z80	CPU	24.95
8212	8-Bit Input/Output	4.95	2650	MPU	26.50
8214	Priority Interrupt Control	7.95	MC6800	MPU	19.95
8216	Bi-Directional Bus Driver	4.95	MC6810A	128 x 8 Static Ram	5.95
8224	Clock Generator/Driver	5.95	MC6820	Periph. Interface Adapter	11.50
8228	System Controller/Bus Driver	5.95	MC6821	Periph. Interface Adapter	11.50
8251	Prog. Comm. Interface	9.95	MC6830L8	1024 x 8 Bit ROM	14.95
8255	Prog. Periph. Interface	10.95	MC6850	Asynchronous Comm. Adapter	14.95

SPECIAL REQUESTED ITEMS

TELEPHONE	ICM CHIPS	MEMORIES	MISCELLANEOUS
KEYBOARD CHIPS	ICM7045 \$24.95	11C90 \$19.95	MK40240 \$17.50
AY-5-9100 \$14.95	ICM7205 19.95	MC166571 \$13.50	DS0226C4 3.75
AY-5-9500 14.95	ICM7207 7.50	MC14987L 4.95	TL1708 10.20
AY-5-9500 4.95	ICM7208 19.95	MC166575 13.50	MC1493L8 5.75
AY-5-2376 14.95	ICM7209 6.95		93H160 11.95
HD0185 7.95	TV GAME CHIP SET		L0110/111 \$75.00/set
74C922 9.95	AY-5-8500-1 Chip and 2,010 MHz Crystal \$7.95		MC401S(74416) 7.50

PARATRONICS Logic Analyzer Kit Model 100A \$229.00/kit

- Analyzes any type of digital system
- Checks data rates in excess of 8 million words per second
- Trouble shoot TTL, CMOS, DTL, RTL, Schottky and MOS families
- Displays 16 logic states up to 8 digits wide
- See ones and zeros displayed on your CRT, octal or hexadecimal format
- Tests circuits under actual operating conditions
- Easy to assemble — comes with step-by-step construction manual which includes 80 pages on logic analyzer operation. (Model 100A Manual - \$4.95)

PARATRONICS TRIGGER EXPANDER - Model 10

Adds 16 additional bits. Provides digital delay and qualification of input clock and 24-bit trigger word — Connects direct to Model 100A for integrated unit.

Model 10 Kit - \$229.00
Caseplate - \$9.95
Model 10 Manual - \$4.95

3 1/2-Digit Portable DMM

- Overload Protected
- High LED Display
- Battery or AC operation
- Auto Zeroing
- 1mv 1/2A, 0.1 ohm resolution
- Overrange reading
- 10 meg impedance
- DC Accuracy 1% typical
- Ranges: DC Voltage - 0-1000V AC Voltage 0-1000V
- Freq. Response 50-400 Hz
- DC/AC Current 0-100mA
- Resistance 0-10 meg ohm
- Size: 6 1/4" x 4 1/4" x 2"

100 MHz 8-Digit Counter

- 20 Hz-100 MHz Range
- 6 LED Display
- Crystal-controlled timebase
- Fully Automatic
- Portable — completely self-contained
- Size — 1 7/8" x 7 3/8" x 5 5/8"

MAX-100 \$134.95

ACCESSORIES FOR MAX 100:

Mobile Charger/Eliminator use power from car battery Model 100 — CLA \$3.95
Charger/Eliminator use 110 V AC Model 100 — CAI \$9.95

KEYBOARDS

63-Key Unencoded Hexadecimal Encoder

This is a 63-key, terminal keyboard newly manufactured by a large computer manufacturer. It is unencoded with SPST keys, unattached to any kind of PC board. A very solid molded plastic 13 x 4 1/4" base suits most application. IN STOCK

\$29.95/each

19-key pad includes 1-10 keys, ABCDEF and 2 optional keys and a shift key. \$10.95/each

Jameco ELECTRONICS

MAIL ORDER ELECTRONICS — WORLDWIDE
1021 HOWARD AVENUE, SAN CARLOS, CA 94070
Advertised Prices Good Thru July

The Incredible "Pennywhistle 103"

\$129.95 Kit Only

The Pennywhistle 103 is capable of recording data to and from audio tape without critical speed requirements for the recorder and it is able to communicate directly with another modem and terminal for telephone "barring" and communications for the deaf. In addition, it is free of critical adjustments and is built with non-precision readily available parts.

Maximum Data Rate: 300 Baud
Data Format: Asynchronous Serial (return to mark level required between each character)

Receive Channel Frequencies: 2025 Hz for space 2225 Hz for mark
Transmit Channel Frequencies: Switch selectable low (normal): 1070 space 1270 mark High: 025 space 2225 mark

Receive Sensitivity: 46 dbm acoustically coupled
Transmit Level: 15 dbm nominal Adjustable from 6 dbm to 20 dbm

Receive Frequency Tolerance: Frequency reference automatically adjusts to allow for operation between 1800 Hz and 2400 Hz

Digital Data Interface: EIA RS-232C or 20 mA current loop (receiver is unipolar and non-polar)

Power Requirements: 120 VAC, single phase, 10 Watts

Physical Characteristics: All components mount on a single 5" by 9" printed circuit board. All components included.

Requires a VOM, Audio Oscillator, Frequency Counter and/or Oscilloscope to align

The Original the 3rd Hand \$9.95 each

Leaves two hands free for working

- Clamps on edge of bench, table or work bench
- Position board on angle or flat position for soldering or clipping
- Sturdy, aluminum construction for hobbyist, manufacturer or school rooms

DIGITAL STOPWATCH

- 8-Digit LED Display
- Times to 59 minutes 59.99 seconds
- Crystal Controlled Time Base
- Three Stopwatches in One
- Times Single Event — Split & Taylor
- Size 4 1/2" x 2 1/8" x 90 (1 1/2 ounces)
- Uses 3 Penrite Cells

Kit — \$39.95
Assembled — \$49.95
Heavy Duty Carry Case \$5.95

Stop Watch Chip Only (7205) \$19.95

IMK 3 1/2 DIGIT DPM KIT

- New Bipolar Unit
- Auto Zeroing
- 5" LED
- Auto Polarity
- Low Power
- Single IC Unit

Model KB500 DPM Kit \$49.00
Model KB503 5V Power Kit \$17.50

JE700 CLOCK

The JE700 is a low-cost digital clock but is a very high quality unit. The unit features a simulated walnut case with dimensions of 11" x 7 1/2" x 2 1/2". It plays a MAN72 high brightness red-out, and the MM5314 clock chip

12 or 24 Hour
KIT ONLY \$16.95

JE803 PROBE

The Logic Probe is a unit which is for the most part indispensable in trouble shooting logic families. TTL, DTL, RTL, CMOS. It drives the power inputs to operate directly off of the circuit under test drawing a scant 10 mA max. It uses a MAN3 readout to indicate any of the following states by these symbols: (H) 1 (LOW) 0 (PULSE) P (The Probe can detect high frequency pulses to 15 MHz. It can be used at MOS levels or circuit damage will result)

printed circuit board

\$9.95 Per Kit

TPL 5V 1A Supply

This is a standard TTL power supply using the well known LM309K regulator. It provides a solid 1 AMP of current at 5 volts. We try to make things easy for you by providing everything you need in one package including the hardware for only

JE225 \$9.95 Per Kit

PROTO BOARDS

PROTO BOARD 6 \$15.95 (6" long X 4" wide)

PB100 - 4.5" x 6" \$ 19.95
PB101 - 5.8" x 4.5" 29.95
PB102 - 7" x 4.5" 39.95
PB103 - 9" x 6" 59.95
PB104 - 9.5" x 8" 79.95
PB203 - 9.75" x 6 1/2" x 2 1/4" 80.00
PB203A - 9.75" x 6 1/2" x 2 1/4" 129.95 (includes power supply)

PROTO CLIPS

14 PIN \$4.50
16 PIN 4.75
24 PIN 8.50
40 PIN 13.75

7400 TTL

SN7400N	16	SN7427N	28	SN74160N	89
SN7401N	18	SN7428N	35	SN74161N	89
SN7402N	18	SN7429N	35	SN74162N	1.95
SN7403N	18	SN7430N	35	SN74163N	89
SN7404N	18	SN7431N	35	SN74164N	89
SN7405N	20	SN7432N	35	SN74165N	89
SN7406N	20	SN7433N	5.00	SN74166N	2.95
SN7407N	20	SN7434N	5.00	SN74167N	1.95
SN7408N	20	SN7435N	5.00	SN74168N	1.95
SN7409N	20	SN7436N	5.00	SN74169N	1.95
SN7410N	18	SN7437N	5.00	SN74170N	1.95
SN7411N	25	SN7438N	5.00	SN74171N	1.95
SN7412N	25	SN7439N	1.75	SN74172N	1.95
SN7413N	40	SN7440N	4.5	SN74173N	1.95
SN7414N	70	SN7441N	59	SN74174N	89
SN7415N	25	SN7442N	43	SN74175N	1.95
SN7416N	25	SN7443N	65	SN74176N	1.95
SN7417N	25	SN7444N	65	SN74177N	1.95
SN7418N	20	SN7445N	43	SN74178N	1.95
SN7419N	20	SN7446N	65	SN74179N	1.95
SN7420N	20	SN7447N	65	SN74180N	1.95
SN7421N	29	SN7448N	65	SN74181N	1.95
SN7422N	39	SN7449N	65	SN74182N	1.95
SN7423N	25	SN7450N	65	SN74183N	1.95
SN7424N	25	SN7451N	65	SN74184N	1.95
SN7425N	29	SN7452N	65	SN74185N	1.95
SN7426N	29	SN7453N	65	SN74186N	9.95
SN7427N	25	SN7454N	65	SN74187N	9.95
SN7428N	25	SN7455N	65	SN74188N	9.95
SN7429N	29	SN7456N	65	SN74189N	9.95
SN7430N	25	SN7457N	65	SN74190N	9.95
SN7431N	25	SN7458N	65	SN74191N	25
SN7432N	25	SN7459N	65	SN74192N	79
SN7433N	25	SN7460N	65	SN74193N	79
SN7434N	25	SN7461N	65	SN74194N	79
SN7435N	25	SN7462N	65	SN74195N	69
SN7436N	25	SN7463N	65	SN74196N	89
SN7437N	25	SN7464N	65	SN74197N	89
SN7438N	25	SN7465N	65	SN74198N	1.95
SN7439N	25	SN7466N	65	SN74199N	1.95
SN7440N	20	SN7467N	65	SN74200N	5.79
SN7441N	75	SN7468N	65	SN74201N	1.59
SN7442N	49	SN7469N	65	SN74202N	1.59
SN7443N	75	SN7470N	65	SN74203N	1.59
SN7444N	75	SN7471N	65	SN74204N	1.59
SN7445N	75	SN7472N	65	SN74205N	1.59
SN7446N	69	SN7473N	65	SN74206N	1.59
SN7447N	59	SN7474N	65	SN74207N	1.59
SN7448N	59	SN7475N	65	SN74208N	1.59
SN7449N	59	SN7476N	65	SN74209N	1.59
SN7450N	20	SN7477N	65	SN74210N	1.59
SN7451N	20	SN7478N	65	SN74211N	1.59
SN7452N	20	SN7479N	65	SN74212N	1.59
SN7453N	20	SN7480N	65	SN74213N	1.59
SN7454N	20	SN7481N	65	SN74214N	1.59
SN7455N	25	SN7482N	65	SN74215N	1.59
SN7456N	20	SN7483N	65	SN74216N	1.59
SN7457N	29	SN7484N	65	SN74217N	1.59
SN7458N	29	SN7485N	65	SN74218N	1.59
SN7459N	29	SN7486N	65	SN74219N	1.59
SN7460N	20	SN7487N	65	SN74220N	1.59
SN7461N	20	SN7488N	65	SN74221N	1.59
SN7462N	20	SN7489N	65	SN74222N	1.59
SN7463N	20	SN7490N	65	SN74223N	1.59
SN7464N	20	SN7491N	65	SN74224N	1.59
SN7465N	20	SN7492N	65	SN74225N	1.59
SN7466N	20	SN7493N	65	SN74226N	1.59
SN7467N	20	SN7494N	65	SN74227N	1.59
SN7468N	20	SN7495N	65	SN74228N	1.59
SN7469N	20	SN7496N	65	SN74229N	1.59
SN7470N	20	SN7497N	65	SN74230N	1.59
SN7471N	20	SN7498N	65	SN74231N	1.59
SN7472N	20	SN7499N	65	SN74232N	1.59
SN7473N	20	SN7500N	65	SN74233N	1.59
SN7474N	20	SN7501N	65	SN74234N	1.59
SN7475N	20	SN7502N	65	SN74235N	1.59
SN7476N	20	SN7503N	65	SN74236N	1.59
SN7477N	20	SN7504N	65	SN74237N	1.59
SN7478N	20	SN7505N	65	SN74238N	1.59
SN7479N	20	SN7506N	65	SN74239N	1.59
SN7480N	20	SN7507N	65	SN74240N	1.59
SN7481N	20	SN7508N	65	SN74241N	1.59
SN7482N	20	SN7509N	65	SN74242N	1.59
SN7483N	20	SN7510N	65	SN74243N	1.59
SN7484N	20	SN7511N	65	SN74244N	1.59
SN7485N	20	SN7512N	65	SN74245N	1.59
SN7486N	20	SN7513N	65	SN74246N	1.59
SN7487N	20	SN7514N	65	SN74247N	1.59
SN7488N	20	SN7515N	65	SN74248N	1.59
SN7489N	20	SN7516N	65	SN74249N	1.59
SN7490N	20	SN7517N	65	SN74250N	1.59

C/MOS

CD4000	23	CD4070	55
CD4001	23	CD4071	23
CD4002	23	CD4072	23
CD4006	1.19	CD4073	49
CD4007	25	CD4074	23
CD4009	49	CD4075	23
CD4010	49	CD4076	23
CD4011	23	CD4077	23
CD4012	25	CD4078	23
CD4013	39	CD4079	23
CD4014	1.39	CD4080	23
CD4015	1.19	CD4081	23
CD4016	49	CD4082	23
CD4017	1.19	CD4083	23
CD4018	99	CD4084	23
CD4019	49	CD4085	23
CD4020	1.19	CD4086	23
CD4021	1.19	CD4087	23
CD4022	1.19	CD4088	23
CD4023	23	CD4089	23
CD4024	23	CD4090	23
CD4025	23	CD4091	23
CD4026	2.25	CD4092	23
CD4027	69	CD4093	23

7400

7400	59	7403	3.25
7401	65	7404	3.25
7402	65	7405	6.49
7403	65	7406	3.00
7404	65	7407	2.75
7405	65	7408	2.75
7406	3.00	7409	2.75
7407	65	7410	2.75
7408	65	7411	2.75
7409	65	7412	2.75
7410	2.15	7413	3.00
7411	4.75	7414	3.00
7412	5.90	7415	3.00
7413	1.15	7416	1.50

LINEAR

LM3000	1.75	LM733N	1.00
LM3001	1.75	LM734N	1.00
LM3002	1.75	LM735N	1.00
LM3003	1.75	LM736N	1.00
LM3004	1.00	LM737N	1.00
LM3005	1.00	LM738N	1.00
LM3006	1.00	LM739N	1.00
LM3007	1.00	LM740N	1.00
LM3008	1.00	LM741N	1.00
LM3009	1.00	LM742N	1.00
LM3010	1.00	LM743N	1.00
LM3011	1.00	LM744N	1.00
LM3012	1.00	LM745N	1.00
LM3013	1.00	LM746N	1.00
LM3014	1.00	LM747N	1.00
LM3015	1.00	LM748N	1.00
LM3016	1.00	LM749N	1.00
LM3017	1.00	LM750N	1.00
LM3018	1.00	LM751N	1.00
LM3019	1.00	LM752N	1.00
LM3020	1.00	LM753N	1.00
LM3021	1.00	LM754N	1.00
LM3022	1.00	LM755N	1.00
LM3023	1.00	LM756N	1.00
LM3024	1.00	LM757N	1.00
LM3025	1.00	LM758N	1.00
LM3026	1.00	LM759N	1.00
LM3027	1.00	LM760N	1.00
LM3028	1.00	LM761N	1.00
LM3029	1.00	LM762N	1.00
LM3030	1.00	LM763N	1.00
LM3031	1.00	LM764N	1.00
LM3032	1.00	LM765N	1.00
LM3033	1.00	LM766N	1.00
LM3034	1.00	LM767N	1.00
LM3035	1.00	LM768N	1.00
LM3036	1.00	LM769N	1.00
LM3037	1.00	LM770N	1.00
LM3038	1.00	LM771N	1.00
LM3039	1.00	LM772N	1.00
LM3040	1.00	LM773N	1.00
LM3041	1.00	LM774N	1.00
LM3042	1.00	LM775N	1.00
LM3043	1.00	LM776N	1.00
LM3044	1.00	LM777N	1.00
LM3045	1.00	LM778N	1.00
LM3046	1.00	LM779N	1.00
LM3047	1.00	LM780N	1.00
LM3048	1.00	LM781N	1.00
LM3049	1.00	LM782N	1.00
LM3050	1.00	LM783N	1.00
LM3051	1.00	LM784N	1.00
LM3052	1.00	LM785N	1.00
LM3053	1.00	LM786N	1.00
LM3054	1.00	LM787N	1.00
LM3055	1.00	LM788N	1.00
LM3056	1.00	LM789N	1.00
LM3057	1.00	LM790N	1.00
LM3058	1.00	LM791N	1.00
LM3059	1.00	LM792N	1.00
LM3060	1.00	LM793N	1.00
LM3061	1.00	LM794N	1.00
LM3062	1.00	LM795N	1.00
LM3063	1.00	LM796N	1.00
LM3064	1.00	LM797N	1.00
LM3065	1.00	LM798N	1.00
LM3066	1.00	LM799N	1.00
LM3067	1.00	LM800N	1.00
LM3068	1.00	LM801N	1.00
LM3069	1.00	LM802N	1.00
LM3070	1.00	LM803N	1.00
LM3071	1.00	LM804N	1.00
LM3072	1.00	LM805N	1.00
LM3073	1.00	LM806N	1.00
LM3074	1.00	LM807N	1.00
LM3075	1.00	LM808N	1.00
LM3076	1.00	LM809N	1.00
LM3077	1.00	LM810N	1.00
LM3078	1.00	LM811N	1.00
LM3079	1.00	LM812N	1.00
LM3080	1.00	LM813N	1.00
LM3081	1.00	LM814N	1.00
LM3082	1.00	LM815N	1.00
LM3083	1.00	LM816N	1.00
LM3084	1.00	LM817N	1.00
LM3085	1.00	LM818N	1.00
LM3086	1.00	LM819N	1.00
LM3087	1.00	LM820N	1.00
LM3088	1.00	LM821N	1.00
LM3089	1.00	LM822N	1.00
LM3090	1.00	LM823N	1.00
LM3091	1.00	LM824N	1.00
LM3092	1.00	LM825N	1.00
LM3093	1.00	LM826N	1.00
LM3094	1.00	LM827N	1.00
LM3095	1.00	LM828N	1.00
LM3096	1.00	LM829N	1.00
LM3097	1.00	LM830N	1.00
LM3098	1.00	LM831N	1.00
LM3099	1.00	LM832N	1.00
LM3100	1.00	LM833N	1.00
LM3101	1.00	LM834N	1.00
LM3102	1.00	LM835N	1.00
LM3103	1.00	LM836N	1.00
LM3104	1.00	LM837N	1.00
LM3105	1.00	LM838N	1.00
LM3106	1.00	LM839N	1.00
LM3107	1.00	LM840N	1.00
LM3108	1.00	LM841N	1.00
LM3109	1.00	LM842N	1.00
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Slide Switch
Smooth action, 4-pole, triple throw, J-5100 **80¢**



Toggle Switch
Heavy duty, SPST toggle, rated @ 15A/125 volts. J-5004 **95¢**



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Miniature glass type high speed switching diode. Case AA. J-1002 **12 for \$1.00**



Connector
J-1000 **\$1.00**



Transformer
8V @ 500 ma or 12V @ 190ma. W/ specs. J-5005 **\$1.50**



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22 ga. magnet wire w/ solder-through insulation. Appx. 500'. J-7003 **\$2.50**



Meteorological balloons
100 grms **\$2.99** J-50



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Standard, solder-type, low profile 28 pin IC socket. J-7101 **40¢**



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6 x 6 (J-8018) **5 for \$3.60**
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Trimmer Cap
1-3pf miniature PC trimmer cap for crystal circuits, etc. J-3000 **75¢**



2N3055
Popular audio type power transistor. J-1001 **50¢**



Dual Pot
100k dual, 1 shaft **75¢ each** J-600



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Sensitive 12 VDC reed relay, DPST, 1 1/2" long, with P-C leads. J-5047 **\$1.50**




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Standard size hex nuts for use on pots, switches, etc. 3/8" I.D. **25 for \$1.00** J-7006



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.1uf at 6KV
New Sprague, **70¢** J-503



Potentiometer
Miniature 1/8" shaft, 1/4" mount potentiometer. 100 ohm value. J-2100 **70¢**



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Horizontal type, 1K-ohm PC trimmer. J-2101 **25¢**




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4007-16 4019-	37 4030-	33 4071-	24
4009-37 4020-	84 4025-	97 4076-	97
4010-37 4021-	77 4042-	58 4518-	89
4011-16 4022-	85 4046-	145 74C01-	22
4012-16 4023-	16 4047-	150 74C193	1.50
4013-29 4024-	66 4049-	34	
4015-76 4025-	16 4050-	34	

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MM5271 AXI DYN	\$ 3.45
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8264	\$ 1.95
8265	\$ 1.95
8266	\$ 1.95
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8268	\$ 1.95
8269	\$ 1.95
8270	\$ 1.95
8271	\$ 1.95
8272	\$ 1.95
8273	\$ 1.95
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8280	\$ 1.95
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8283	\$ 1.95
8284	\$ 1.95
8285	\$ 1.95
8286	\$ 1.95
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8288	\$ 1.95
8289	\$ 1.95
8290	\$ 1.95
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4 1/2" x 8 1/2" SINGLE SIDED EPOXY BOARD 1/16" thick, unetched **\$52.60** \$ 60 ea.

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100	06	14	30	80	370	5.00
200	07	20	35	115	425	6.50
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LM 376	\$.60	340T 5.6, 6.8, 12	
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or 24V	\$.95		

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2N404 PNP GE TO 5	5/\$ 1.00
2N3919 NPN Si TO 3 RF	\$ 1.50
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7401-13	7446-68	74153-61
7402-13	7447-58	74154-94
7403-13	7448-68	74155-58
7404-15	7450-15	74157-55
7406-13	7472-25	74161-55
7408-13	7473-28	74163-55
7407-20	7474-28	74164-85
7408-18	7475-45	74165-95
7409-18	7476-30	74170-168
7410-13	7480-31	74173-120
7411-18	7483-65	74174-95
7412-13	7485-87	74175-85
7413-36	7486-28	74176-75
7416-25	7490-42	74177-75
7416-22	7491-58	74180-65
7417-25	7492-43	74181-90
7420-13	7493-43	74190-100
7425-25	7494-67	74191-100
7426-22	7495-65	74192-79
7427-19	7496-65	74193-79
7437-13	7497-28	74194-80
7432-22	74121-29	75375-150
7437-21	74122-38	74196-86
7438-21	74123-45	74279-55
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PRV	2A	6A	25A	DIP SOCKETS
100			1.30	8 PIN .22 24 PIN .40
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600				



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National NSL452 photoconductive cells. 3/8" diameter. Changes resistance with light intensity. 4 megohms in total darkness - 45,000 ohms in bright light. Plastic encapsulated. (2 oz.) L413DR.

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L400MS (1 oz.) 3.5mm (mini)
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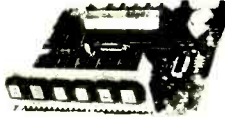
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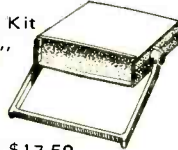


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\$54.95
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KIT INCLUDES: Detailed Instructions (22 pages). All parts including transformer.

Now Low Cost! Instrument Case for Kit

H. 2 7/16" W. 8 1/2" D. 9 1/4" and in 6 varying heights also compatible as other instrument housing



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5V 10A Power Supply Kit for your TTL Circuits!

Kit Includes: Extra Large Heat Sink, Power Tr., IC Regulator, P.C. Board, with OVP Circuitry KIT # T-500



With Optional Rectifiers and Filtering Capacitor. \$17.50 (X'former not available)

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- Features:
- A. Fairchild 0.8" FSC8000 Display Array
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 - C. P.C. Board, Transformer, Speaker & all parts included (less case)
 - D. Detailed Instructions
- THIS IS A BIG ONE!
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- Features:
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 - B. Display Board may be remote
 - C. X'tal time base
 - D. P.C. Boards, speaker, IC's and all parts
 - E. Detailed Instructions
- KIT # T-1302

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MINIATURE SLIDE SWITCH

DPDT .20 each
10 for \$1.75
100 for \$15.00

PUSH BUTTON SWITCH

White, green and yellow
30¢ ea.
4/\$1.00

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TI TIP 2955 PNP
TIP 3055 NPN
10 AMP 60 VOLT
90 WATTS
\$2.25 PER PAIR

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On board 2708 EPROM addressable to any 4K boundary above 32K. Power-on-Jump to any 4K boundary above 32K, or the on-board 2708. A wait state may be added to any:

- 1) M1 cycle
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- 3) On-board ROM cycle (for use at 4 MHz)
- 4) Input cycle
- 5) Out-put cycle

On-board run-stop flip-flop and optional generation of MEMORY WRITE allow front panel-less operation. Selectable 10 addressing mode: 1) 8080 Mode where peripheral address byte is duplicated on high and low order address bytes. 2) Z-80 mode where the peripheral byte appears on the low order address, and the contents of the accumulator appears on the high order byte, allowing simultaneous I/O. DMA Grant tri-states all signals from the processor board. 8224 clock generator provides 8080 look-alike $\phi 1$ and $\phi 2$ for the S-100 bus. Status signals SINP, SOUT, SMEMR, & SWO are latched per S-100 bus specs.

\$135.00 EA. (2MHZ) **\$149.95 EA. (4MHZ)**

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ASSEMBLED & TESTED: ADD \$50 to each KIT PRICE
KITS INCLUDE ALL PARTS AND 2708 UNPROGRAMMED

PUT YOUR HAM GEAR OR CB IN YOUR HOUSE WITH THIS SPECIALLY DESIGNED POWER SUPPLY KIT!

KIT INCLUDES: Transformer, PC Board, Large heat sink, Large filtering capacitor and all parts with detailed instruction.

- Features:
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 - Output Current = 3.5A Min.
 - Load Regulation = 0.01%
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 - Internal Current Limiting
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T-8K/ Static Ram Memory Kits

Memory Chips
2102 low power chips. Boards operate with the fastest 2102 on the market (250ns).
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Input and output data lines fully buffered. All possible support chips are low power schottky. Support chips at bottom of board for strong drive signals to bus.

Bypass
26 bypass capacitors on the board.
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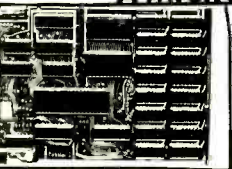
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7417	74117	74217	74317	74417	74517	74617	74717	74817	74917
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7424	74124	74224	74324	74424	74524	74624	74724	74824	74924
7425	74125	74225	74325	74425	74525	74625	74725	74825	74925
7426	74126	74226	74326	74426	74526	74626	74726	74826	74926
7427	74127	74227	74327	74427	74527	74627	74727	74827	74927
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7432	74132	74232	74332	74432	74532	74632	74732	74832	74932
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7435	74135	74235	74335	74435	74535	74635	74735	74835	74935
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MP3921	MP3921	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3922	MP3922	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3923	MP3923	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3924	MP3924	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3925	MP3925	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3926	MP3926	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3927	MP3927	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3928	MP3928	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3929	MP3929	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3930	MP3930	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3931	MP3931	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3932	MP3932	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3933	MP3933	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3934	MP3934	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3935	MP3935	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3936	MP3936	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3937	MP3937	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3938	MP3938	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3939	MP3939	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3940	MP3940	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3941	MP3941	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3942	MP3942	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3943	MP3943	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3944	MP3944	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3945	MP3945	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3946	MP3946	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3947	MP3947	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3948	MP3948	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
MP3949	MP3949	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
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SN7407	.39	.40	SN74151	1.29	1.30
SN7409	.39	.40	SN74153	1.29	1.30
SN7410	.25	.26	SN74154	1.75	1.76
SN7411	.59	.60	SN74157	.79	.80
SN7413	.99	1.00	SN74158	.39	.40
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LM309K	1.49	1.50	LM710M	.79	.80
LM311H	.79	.80	LM713N	.29	.30
LM318H	1.49	1.50	LM714H	.29	.30
LM320H-5	.99	1.00	LM747H	.59	.60
LM320H-12	.99	1.00	LM1304	1.49	1.50
LM320H-15	.99	1.00	LM1340	1.79	1.80
LM320T-5	1.49	1.50	LM1312	2.49	2.50
LM320T-6	1.49	1.50	LM1414V	.19	.20
LM322N	1.19	1.20	LM1458V	.39	.40
LM324H	1.79	1.80	LM1820N	.79	.80
LM339N	1.79	1.80	LM3028H	.69	.70
LM340K-5	1.49	1.50	LM3900N	.45	.46
LM340K-6	1.49	1.50	LM3909V	1.75	1.76
LM340K-8	1.49	1.50	LM555N	.76	.77
LM340K-12	1.49	1.50	LM75451	.69	.70
LM340K-15	1.49	1.50	LM75453	.69	.70
LM340K-18	1.49	1.50	LM75459	.80	.81
LM340K-24	1.49	1.50	LM75492	.80	.81
			LM75494	.60	.61
			PA263	1.50	1.51
			DM8864N	1.29	1.30

RIBBON CABLE:

Order by Cat. 7R 3939 and Conductors

Cond.	20	26	34	40
7-R	1.98	1.98	1.98	1.98
14-R	1.99	1.99	1.99	1.99
6-R	1.99	1.99	1.99	1.99

25 AMP BRIDGE

Order RECTIFIERS by Cat. No. 7R 2273 & voltage

V	Each	1c SALE
50	1.00	1.00
100	1.49	1.50
200	1.69	1.70
300	2.25	2.26
400	2.25	2.26
600	2.50	2.51

25 AMP POWER

Order SHUT SCRS, TRIACS by Cat. No. 7R-14A147 (SCRs) Cat. No. 7R-60B86A (Triacs)

V	Each	1c SALE
50	\$.95	2 for \$.96
100	1.25	2 for 1.26
200	1.95	2 for 1.96
400	2.25	2 for 2.26
600	2.95	2 for 2.96

1N4000 Epoxy Rectifiers

Order by Cat. No. and Type No.

Cat. No.	Type No.	Price
7R 2377	1N4001	50 for \$.65 20 for .76
7R 2378	1N4002	100 for .75 20 for .86
7R 2379	1N4003	100 for .99 20 for 1.00
7R 2380	1N4004	100 for 1.29 20 for 1.40
7R 2381	1N4005	100 for 1.39 20 for 1.50
7R 2382	1N4006	100 for 1.49 20 for 1.60
7R 2383	1N4007	1000 for 1.39 20 for 1.40

BULLET RECTIFIERS!

Order by Cat. No. 7R60B84 and voltage

V	10 for	50 for	20 for
50V	\$.59	.70	.80
100V	.69	.79	.89
200V	.79	.89	.99
400V	1.09	1.19	1.29
600V	1.19	1.29	1.39
800V	1.19	1.29	1.39

10 AMP POWER TABS!

Order by Cat. No. and Type No.

Type	Price
7R1448 TRIACS	50 for \$.55 2 for .66
7R1730 SCR	100 for .69 2 for .80
7R1590 QUADRACS	400 for 1.19 2 for 1.20
	600 for 1.59 2 for 1.60

1.5 AMP BULLET RECTIFIERS!

Order by Cat. No. 7R60B84 and voltage

V	10 for	50 for	20 for
50V	\$.59	.70	.80
100V	.69	.79	.89
200V	.79	.89	.99
400V	1.09	1.19	1.29
600V	1.19	1.29	1.39
800V	1.19	1.29	1.39

MICRO MINI TOGGLE SWITCHES!

3 Amps. 125 VAC contacts with chrome handles, complete with mounting hardware.

Cat. No.	Type	Price
7R3936	SPST	\$1.19 1.20
7R5085	SPDT	1.39 1.40
7R4037	DPDT	1.45 1.46

BUY 1 AT OUR SALE PRICE GET 2 FOR 1c MORE

LYTIC KLINIC

Order by Cat. No. 7R 3269 and value

UF	V	Style	1c SALE
10	15	P.C.	10 for \$1.00
20	15	P.C.	8 for 1.00
50	25	P.C.	8 for 1.00
100	25	P.C.	6 for 1.00
200	25	P.C.	6 for 1.00
500	25	P.C.	10 for 1.00
1000	25	P.C.	10 for 1.00

6 AMP CARTWHEEL RECTIFIERS

Order by Cat. No. 7R 3584

V	50	100	200
50	\$.36	2 for \$.37	
100	.45	2 for .46	
200	.75	2 for .76	
400	.85	2 for .86	
600	.90	2 for .91	
800	1.19	2 for 1.20	
1000	1.39	2 for 1.40	

ZENERS!

Order by Cat. No. 7R5210 & voltage & wattage!

Watts	5 for	1c SALE
1 Watt	5 for	1.10 for 1.01
8.2V	5 for	1.10 for 1.01
9.1V	5 for	1.10 for 1.01
12.5V	5 for	1.10 for 1.01
15.0V	5 for	1.10 for 1.01



BUY ONE AT SALE PRICE. GET 2ND FOR ONLY 1c MORE.

Easy to build! Both units are 12 hour, uses MM516 clock chip and a minimum of external components. Includes options for radio, snooze, and more. Includes IC sockets, transistors, readouts and all other necessary components and line cord, transformer, and line cord (not included).
4 DIGIT
 Wt. 1lb.
 Your Choice
\$9.95
 2 for \$9.96

FLUORESCENT - LED CLOCK BASICS

- Cat. No. #7R3791 Fluorescent Basic Clock Kit \$9.95 2 for \$9.96
- Cat. No. #7R3792 LED Basic Clock Kit \$9.95 2 for \$9.96
- Cat. No. #7R3412 12V 300 ma Transformer 1.49 2 for 1.50
- Cat. No. #7R3792 Components for Optional Alarm 2.95 2 for

Order by Cat. No. in parenthesis

Quan.	Description	Price	1c SALE
1	PANCAKE PHOTOCELLS, 600 to 15K ohms (7R2939)	1.00	10 for 1.01
1	100KHZ MARKER CRYSTALS, approx for marker gen. (7R3896)	1.95	2 for 1.96
1	MOTHEBOARD EDGE CONNECTOR, 12 pin, 15' (7R3987)	3.50	2 for 3.51
1	48-PIN EDGE CONNECTOR, .156" spacing (7R3963)	1.95	2 for 1.96
1	WOODGRAIN CABINET, 3 1/2 x 10 x 1/2 deep, sokrs, alarms. (#7R5201)	\$1.49	2 for \$1.50
1	DUAL GATE MOSFET, 4 pin, with knob (#7R3064)	1.00	4 for 1.01
1	YOSTICK, LED basic includes 57-TUBES, LED basic includes 45' RED LED READOUTS	4.00	2 for 4.01
1	ECCO THUMBWHEEL SWITCH, BCD, 0-7 (#7R2870A)	1.49	2 for 1.50
1	B-TRACK TAPE TRANSPORT, with pump (#7R3010)	9.95	2 for 9.96
1	PLESSEY TV SIDEBAND FILTER, for chan. 3 or 4 (#7R3975)	1.85	2 for 1.86
1	30-M. WIRE WRAP WIRE, 30 gauge, 100' (7R3803)	1.00	2 for 1.01
1	METER, 50uA, 1 1/2" square, 0-20db (#7R3705)	1.19	2 for 1.20
1	SPST RELAY, norm. open 12-24 VDC, 1250 ohms, dip style (#7R5175)	1.00	2 for 1.01
1	VEEDER ROOT COUNTER, 000-999, resettable, panel mt. (#7R5081)	1.49	2 for 1.50
1	DUAL GATE MOSFET, 4 pin, with knob (#7R3064)	1.00	4 for 1.01
1	B-TRACK TAPE HEAD, with plug 'n' cord (#7R3468)	1.00	4 for 1.01
1	CALCULATOR KEYBOARDS, 20 keys and more (7R3524)	\$2.00	20 for \$2.01
1	SLIDE VOLUME CONTROLS, ass'd values (7R3057)	2.00	20 for 2.01
1	RESISTOR SPECIAL, max int. C.R.S., max int. C.R.S. (#7R3250)	2.00	20 for 2.01
1	MOLEX IC SOCKETS, on a strip, cut to length (#7R3144)	2.00	20 for 2.01
1	TERMINAL STRIPS, from 2 lugs up (7R3136)	2.00	20 for 2.01
1	WEZNEON LAMPS, all 100% good (7R2613)	2.00	80 for 2.01
1	SHIELDED CABLE, 1 cond. milk, phone (#7R3577)	2.00	80 for 2.01
1	TRANSISTOR ELECTROS, ass'd values, types (7R2747)	2.00	80 for 2.01
1	SOUND TRIGGERS, sound triggers scr w/amp (7R3625)	2.00	6 for 2.01
1	CB CRYSTALS, for phase lock loop, HAM, HC/18 (#7R5050)	2.00	20 for 2.01
1	DISC CAPACITORS, low ESR, marked, ass'd (#7R2598)	2.00	200 for 2.01
1	CAPACITOR SPECIAL, discs, marked, ass'd (#7R2598)	2.00	200 for 2.01
1	VOLTAGE REGULATORS, hobby LM320, 340, TO-3 (#7R3330)	2.00	20 for 2.01
1	PANEL SWITCHES, slides, rotaries, mod. 3P (#7R2688)	2.00	60 for 2.01
1	RESISTOR SPECIAL, 1/4 to 1W, carbon, metal (7R3054)	2.00	400 for 2.01
1	HALF WATERS, 1/4 to 1W, carbon, metal (7R3046)	2.00	400 for 2.01
1	NATIONAL IC BONANZA, linears, 7400s ROMS (7R2860)	2.00	200 for 2.01
1	NOBBY LEDS, ass'd types, mostly usable (7R2859)	2.00	80 for 2.01
1	LM340T VOLTAGE REGULATORS, 5 to 24V, TO-220 (#7R2635)	2.00	30 for 2.01
1	TWO WATERS, resistors, carbon, metal (7R2735)	2.00	200 for 2.01
1	POLYSTYRENE CAPS, ass'd values, voltage, hi-Q (#7R2729)	2.00	200 for 2.01
1	THERMISTORS, resistors that change with temp (7R4089)	2.00	100 for 2.01
1	BRIDGES, untested, 2, 4, 6, 10, amp. full wave (7R4022)	2.00	40 for 2.01
1	LAMP SOCKETS, 14, 18 pin, solder tail (#7R3921)	2.00	50 for 2.01
1	MIXED READOUTS, hobby, untested, 127, .3, .5, etc. (7R3619)	2.00	30 for 2.01
1	QUARTER WATERS, resistors, metal film, marked (7R3413)	2.00	300 for 2.01
1	PLASTIC TRANSISTORS, untested, TO-92 (#7R2604)	2.00	200 for 2.01
1	PREFORMED RESISTORS, 1/4, 1/2, 1W, marked, ass'd (7R2608)	2.00	200 for 2.01
1	PRECISION RESISTORS, 1/4, 1/2, 1W, 1%, 2% marked (7R2428)	2.00	400 for 2.01
1	DIPPED MYLARS, shiny finish, ass'd values (7R2597)	2.00	120 for 2.01
1	IC SOCKETS, 14, 18 pin, solder tail (#7R3921)	2.00	300 for 2.01
1	PREFORMED DISC CAPACITORS, marked values, ass'd (#7R2605)	2.00	60 for 2.01
1	TRIMMER CAPACITORS, mica compression, piston, ass'd (#7R3714)	2.00	120 for 2.01
1	YELLOW JACKET MYLAR CAPACITORS, marked, pop values (#7R3476)	2.00	150 for 2.01
1	PLASTIC TRANSISTORS, untested, better than 50% yield (#7R2740)	2.00	300 for 2.01
1	CARBON FILM RESISTORS, 1/4 to 1W, marked, ass'd (#7R3432)	2.00	400 for 2.01
1	UNMARKED CAPACITORS, polystyrene, molded, pop values (#7R3805)	2.00	500 for 2.01
1	SILICON SIGNAL DIODES, micro, glass, zeners too, 50% yield (#7R2628)	2.00	600 for 2.01
1	IC SOCKETS, 14, 18 pin, solder tail (#7R3921)	2.00	16 for 2.01
1	1N4148/914 SWITCHING DIODES, 50% yield, untested! (#7R2418)	2.00	200 for 2.01
1	PC TRIM POTENTIOMETERS, thumbwheel, screwdriver ass't (#7R3345)	2.00	60 for 2.01
1	SLIDE VOLUME CONTROLS, ass'd values, duals, singles (#7R3057)	2.00	30 for 2.01
1	IC SOCKETS, 14, 18 pin, solder tail (#7R3921)	2.00	150 for 2.01

INTERNATIONAL ELECTRONICS UNLIMITED

SUMMER SALE

SPECIAL SALE
LOW POWER SCHOTTKY 74LS00
 First Quality - Full Spec
 Buy \$15 - any mix- deduct 10%
 Buy \$25 - any mix- deduct 15%

LINEAR CIRCUITS
 Buy \$10.00 - any mix- deduct 10%
 Buy \$25.00 - any mix- deduct 15%

74LS00	19	75LS04	20	74LS157	59	74LS251	79
74LS01	19	74LS158	21	74LS158	69	74LS252	79
74LS02	19	74LS159	24	74LS160	79	74LS256	1.15
74LS03	19	74LS161	75	74LS161	79	74LS257	69
74LS04	21	74LS162	79	74LS162	79	74LS258	69
74LS05	21	74LS163	79	74LS163	79	74LS259	1.39
74LS06	21	74LS164	79	74LS164	79	74LS260	29
74LS09	20	74LS165	90	74LS165	1.39	74LS266	29
74LS10	20	74LS166	90	74LS166	89	74LS268	29
74LS11	20	74LS167	90	74LS167	89	74LS275	50
74LS12	20	74LS168	90	74LS168	89	74LS276	72
74LS13	39	74LS169	90	74LS169	89	74LS283	72
74LS14	39	74LS170	90	74LS170	1.49	74LS290	59
74LS15	39	74LS171	97	74LS171	97	74LS291	64
74LS16	39	74LS172	97	74LS172	97	74LS292	64
74LS17	39	74LS173	97	74LS173	97	74LS295	89
74LS18	39	74LS174	97	74LS174	97	74LS296	89
74LS19	39	74LS175	97	74LS175	97	74LS298	89
74LS20	39	74LS176	97	74LS176	97	74LS299	89
74LS21	39	74LS177	97	74LS177	97	74LS300	89
74LS22	39	74LS178	97	74LS178	97	74LS301	89
74LS23	39	74LS179	97	74LS179	97	74LS302	89
74LS24	39	74LS180	97	74LS180	97	74LS303	89
74LS25	39	74LS181	97	74LS181	97	74LS304	89
74LS26	39	74LS182	97	74LS182	97	74LS305	89
74LS27	39	74LS183	97	74LS183	97	74LS306	89
74LS28	39	74LS184	97	74LS184	97	74LS307	89
74LS29	39	74LS185	97	74LS185	97	74LS308	89
74LS30	39	74LS186	97	74LS186	97	74LS309	89
74LS31	39	74LS187	97	74LS187	97	74LS310	89
74LS32	39	74LS188	97	74LS188	97	74LS311	89
74LS33	39	74LS189	97	74LS189	97	74LS312	89
74LS34	39	74LS190	97	74LS190	97	74LS313	89
74LS35	39	74LS191	97	74LS191	97	74LS314	89
74LS36	39	74LS192	97	74LS192	97	74LS315	89
74LS37	39	74LS193	97	74LS193	97	74LS316	89
74LS38	39	74LS194	97	74LS194	97	74LS317	89
74LS39	39	74LS195	97	74LS195	97	74LS318	89
74LS40	39	74LS196	97	74LS196	97	74LS319	89
74LS41	39	74LS197	97	74LS197	97	74LS320	89
74LS42	39	74LS198	97	74LS198	97	74LS321	89
74LS43	39	74LS199	97	74LS199	97	74LS322	89
74LS44	39	74LS200	97	74LS200	97	74LS323	89
74LS45	39	74LS201	97	74LS201	97	74LS324	89
74LS46	39	74LS202	97	74LS202	97	74LS325	89
74LS47	39	74LS203	97	74LS203	97	74LS326	89
74LS48	39	74LS204	97	74LS204	97	74LS327	89
74LS49	39	74LS205	97	74LS205	97	74LS328	89
74LS50	39	74LS206	97	74LS206	97	74LS329	89
74LS51	39	74LS207	97	74LS207	97	74LS330	89

TTL - 7400
 Buy \$10.00 - any mix- deduct 10%
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7400 Series TTL
 tested and passed logic test-
 not full parametric specs
 minimum 10 of item

7400	11	7441	85	7403	48	74100	1.19
7401	11	7442	58	7404	75	74101	1.90
7402	14	7443	85	7405	75	74102	1.10
7403	14	7444	73	7406	75	74103	1.40
7404	16	7445	85	7407	75	74104	1.10
7405	19	7446	81	7408	37	74105	95
7406	29	7447	59	7409	37	74106	84
7407	28	7448	79	7410	37	74107	84
7408	22	7449	17	7411	45	74108	85
7409	17	7450	17	7412	54	74109	2.30
7410	18	7451	20	7413	54	74110	85
7411	25	7452	20	7414	54	74111	85
7412	43	7453	20	7415	54	74112	85
7413	43	7454	20	7416	54	74113	85
7414	85	7455	20	7417	54	74114	85
7415	25	7456	35	7418	54	74115	85
7416	35	7457	35	7419	54	74116	85
7417	35	7458	35	7420	54	74117	85
7418	35	7459	35	7421	54	74118	85
7419	35	7460	35	7422	54	74119	85
7420	18	7461	35	7423	54	74120	85
7421	37	7462	35	7424	54	74121	85
7422	37	7463	35	7425	54	74122	85
7423	37	7464	35	7426	54	74123	85
7424	37	7465	35	7427	54	74124	85
7425	35	7466	35	7428	54	74125	85
7426	22	7467	35	7429	54	74126	85
7427	35	7468	35	7430	54	74127	85
7428	35	7469	35	7431	54	74128	85
7429	35	7470	35	7432	54	74129	85
7430	35	7471	35	7433	54	74130	85
7431	35	7472	35	7434	54	74131	85
7432	35	7473	35	7435	54	74132	85
7433	35	7474	35	7436	54	74133	85
7434	35	7475	35	7437	54	74134	85
7435	35	7476	35	7438	54	74135	85
7436	35	7477	35	7439	54	74136	85
7437	35	7478	35	7440	54	74137	85
7438	35	7479	35	7441	54	74138	85
7439	35	7480	35	7442	54	74139	85
7440	35	7481	35	7443	54	74140	85
7441	35	7482	35	7444	54	74141	85
7442	35	7483	35	7445	54	74142	85
7443	35	7484	35	7446	54	74143	85
7444	35	7485	35	7447	54	74144	85
7445	35	7486	35	7448	54	74145	85
7446	35	7487	35	7449	54	74146	85
7447	35	7488	35	7450	54	74147	85
7448	35	7489	35	7451	54	74148	85
7449	35	7490	35	7452	54	74149	85
7450	35	7491	35	7453	54	74150	85
7451	35	7492	35	7454	54	74151	85
7452	35	7493	35	7455	54	74152	85
7453	35	7494	35	7456	54	74153	85
7454	35	7495	35	7457	54	74154	85
7455	35	7496	35	7458	54	74155	85
7456	35	7497	35	7459	54	74156	85
7457	35	7498	35	7460	54	74157	85
7458	35	7499	35	7461	54	74158	85
7459	35	7500	35	7462	54	74159	85
7460	35	7501	35	7463	54	74160	85

7460	0.05	7460	.05
7401	.07	7472	.12
7402	.05	7473	.12
7403	.05	7474	.10
7404	.06	7476	.12
7405	.06	7478	.18
7406	.09	7485	.25
7407	.09	7486	.12
7409	.07	7491	.20
7410	.05	7493	.14
7411	.12	7495	.20
7412	.12	7496	.20
7420	.06	7497	.12
7421	.12	7498	.12
7422	.12	7499	.12
7423	.12	7499	.12
7424	.12	7499	.12
7425	.12	7499	.12
7426	.09	7499	.12
7427	.09	7499	.12
7428	.07	7499	.12
7429	.08	7499	.12
7430	.07	7499	.12
7431	.07	7499	.12
7432	.07	7499	.12
7433	.07	7499	.12
7434	.07	7499	.12
7435	.07	7499	.12
7436	.07	7499	.12
7437	.07	7499	.12
7438	.07	7499	.12
7439	.07	7499	.12
7440	.07	7499	.12
7441	.07	7499	.12
7442	.07	7499	.12
7443	.07	7499	.12
7444	.07	7499	.12
7445	.07	7499	.12
7446	.07	7499	.12
7447	.07	7499	.12
7448	.07	7499	.12
7449	.07	7499	.12
7450	.07	7499	.12
7451	.07	7499	.12
7452	.07	7499	.12
7453	.07	7499	.12
7454	.07	7499	.12
7455	.07	7499	.12
7456	.07	7499	.12
7457	.07	7499	.12
7458	.07	7499	.12
7459	.07	7499	.12
7460	.07	7499	.12

LM300H	21	LM372N	2.93	LM733N	79
LM301CN	29	LM376CN	59	LM739N	99
LM302H	85	LM380N	1.29	LM741CN(4)	32
LM304H	80	LM380CN	95	LM741CN(14)	32
LM307CN	28	LM381N	1.75	LM741CH	35
LM307H	30	LM382N	1.75	LM747H	79
LM308H	89	NE555K	45	LM747N	71
LM309H	1.05	NE531A	2.90	LM748CN	35
LM309N	99	NE540	2.90	LM4310N	2.80
LM310CN	1.07	NE568A	1.09	LM1414N	1.59
LM311CN	89	NE555A	75	LM1458CN	1.29
LM311H	89	LM555CN	30	LM1458CN	59
LM311N	89	NE556A	85	LM1498N	89
LM313N	1.13	LM580N	2.95	LM1800N	2.48
LM320N	1.52	LM581N	2.95	LM200N	89
5.5 2 12 15	1.19	LM585N	1.19	LM2209	89
5.5 2 15	1.19	LM585H	1.39	CA3048	1.19
LM322N	1.59	LM586H	1.39	CA3081	1.49
LM324N	1.59	LM586CN	1.19	LM3909N	49
LM324N	1.59	LM587CN	1.29	LM752N	79
LM324N	1.59	LM587H	1.49	LM752N	79
LM340N	1.52	LM700N	2.95	LM752N	79
LM340N	1.52	LM700CN	39	LM752N	79
5.6 8 12	1.29	LM709N	29	LM752N	79
5.6 8 12	1.29	LM709H	29	LM752N	79
5.6 8 12	1.29	LM710N	29	LM752N	79
5.6 8 12	1.29	LM711N	29	LM752N	79
5.6 8 12	1.29	LM723N	55	LM752N	79
5.6 8 12	1.29	LM723H	55	LM7	

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LM370	\$3.40	LM1820	\$1.25
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LM382	\$1.40	LM1889	\$5.35
LM384	\$1.60	LM2111	\$1.40
LM386	\$0.90	LM2113	\$1.50
LM389	\$1.08	LM2907	\$1.95
LM390	\$1.70	LM2917	\$1.95
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LM1304	\$1.75	LM3065	\$1.00
LM1305	\$1.60	LM3070	\$1.95
LM1307N	\$1.25	LM3071	\$1.95
LM1307E	\$2.00	LM3075	\$1.80
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S-100 EXPANDABLE MOTHER BOARD

• 8-SLOT EXPANDABLE BACKPLANE—in line male and female connectors enable backplanes to be plugged together, or the female may be used in place of an extender board.

QUIET—ground plane decouples all signal lines.

RELIABLE—SAE 8100 phenolic body, gold contact connectors.

COMPLETE KIT \$66.00 ASSEMBLED \$89.00

UNIVERSAL "U DESIGN" WIRE WRAP BOARDS

ALL BOARDS ARE G-10 GLASS EPOXY. HAVE VCC AND GROUND PLANES. PLATED THROUGH HOLES. & GOLD PLATED EDGE CONNECTORS

No. 1 BEST ON THE MARKET! MICRO CPU CARD WITH S-100 BUS. 3 ON BOARD REGULATORS FITS ALL STANDARD I.C. SOCKET CONFIGURATIONS. 1700+ HOLES. SIZE 5"x10" \$23.95

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with connector

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UNENCODED—MOUNTED ON G-10 GLASS EPOXY BOARDS—A BLACK METAL FRAME KEEPS KEY SWITCHES SECURELY IN PLACE.

No. 1 .53 key main keyboard .10 auxiliary & cursor control keys .11 key numeric pad .Bank of 5 auxiliary power and control, rocker arm switches— one of them lights up. \$39.95

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OUR NEW ASCII KEYBOARD

HAS ON BOARD UV PROM, A MAIN KEYBOARD SECTION OF 58 KEYS, A HEX PAD OF 15 KEYS AND 16 MORE PERIPHERAL KEYS. 89 KEYS TOTAL & ASCII ENCODED for only \$99.95

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DIODES

IN4001	50V at 1A	6c
IN4003	200V at 1A	8c
IN4007	1000V at 1A	12c
IN250	60V at 20A	95c
IN3909	50V at 30A	\$1.25

BRIDGES

FAST RECOVERY AVALANCHE BRIDGE IN4436/T	200V at 10A	\$4.25
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VOLTAGE REGULATORS

UA723 VARIABLE 2V to 3.7V	PLASTIC	METAL
78L05 5V at 100mA	TO 92	3/98c
340T-6 6V at 1 AMP	TO 220	2/98c

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MJE3055	10A PLASTIC	89c
2N3055	10A TO-3	95c
2N5301	30A TO-3	\$1.95

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BUILD YOUR OWN PAPER TAPE READER

1/10" CENTER STACKABLE

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8 PIN	10/\$1.59	10/\$1.35
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2 LEVEL 10/98c 100/\$7.40 1000/\$64.00 3 LEVEL 10/\$1.25 100/\$8.60 1000/\$72.00

HI-REL GOLD WIRE WRAP SOCKET PIN

10/\$1.20 100/\$9.00 1000/\$79.00 LIST PRICE 29c

OP AMPS

SINGLE 709	10c
741	12c
DUAL MC1458	39c
LM3900	49c

DIODES

IN91	15c
IN270	12c
IN914	15c
IN3600	15c
IN4148	10c

TRANSISTORS

2N2222	12c
2N3904	12c
2N3906	12c
2N3053	49c

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SINETICS NE5008 \$9.95

SUB-MINIATURE CRYSTAL FILTER

455 KHZ WITH DATA \$2.95

LEDS

YELLOW, GREEN, OR AMBER (SPECIFY COLOR) 3/88c

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No. 1 0.4" CHARACTER WITH CONNECTOR \$6.95

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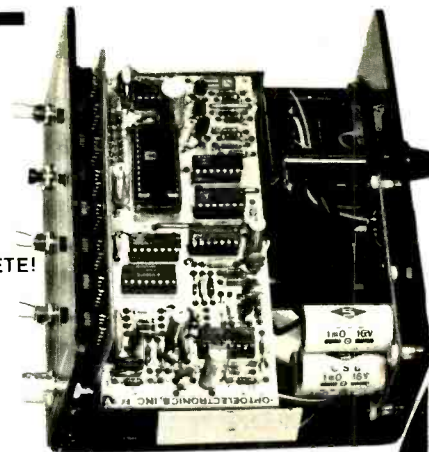
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NEW LSI TECHNOLOGY
FREQUENCY COUNTER

TAKE ADVANTAGE OF THIS NEW STATE-OF-THE-ART COUNTER FEATURING THE MANY BENEFITS OF CUSTOM LSI CIRCUITRY. THIS NEW TECHNOLOGY APPROACH TO INSTRUMENTATION YIELDS ENHANCED PERFORMANCE, SMALLER PHYSICAL SIZE, DRASTICALLY REDUCED POWER CONSUMPTION [PORTABLE BATTERY OPERATION IS NOW PRACTICAL], DEPENDABILITY, EASY ASSEMBLY AND REVOLUTIONARY LOWER PRICING!

- KIT #FC-50C 60 MHZ COUNTER WITH CABINET & P.S. **\$119⁹⁵ COMPLETE!**
- KIT #PSL-650 650 MHZ PRESCALER (NOT SHOWN) 29.95
- MODEL #FC-50WT 60 MHZ COUNTER WIRED, TESTED & CAL. 165.95
- MODEL #FC-50/600WT 600 MHZ COUNTER WIRED, TESTED & CAL. 199.95



SIZE:
3" High
6" Wide
5 1/2" Deep

FEATURES AND SPECIFICATIONS:
 DISPLAY: 8 RED LED DIGITS .4" CHARACTER HEIGHT
 GATE TIMES: 1 SECOND AND 1/10 SECOND
 PRESCALER WILL FIT INSIDE COUNTER CABINET
 RESOLUTION: 1 HZ AT 1 SECOND, 10 HZ AT 1/10 SECOND.
 FREQUENCY RANGE: 10 HZ TO 60 MHZ. [65 MHZ TYPICAL].
 SENSITIVITY: 10 MV RMS TO 50 MHZ, 20 MV RMS TO 60 MHZ TYP.
 INPUT IMPEDANCE: 1 MEGOHM AND 20 PF.
 [DIODE PROTECTED INPUT FOR OVER VOLTAGE PROTECTION.]
 ACCURACY: ± 1 PPM [± .0001%], AFTER CALIBRATION TYPICAL.
 STABILITY: WITHIN 1 PPM PER HOUR AFTER WARM UP [.001% XTAL]
 IC PACKAGE COUNT: 8 [ALL SOCKETED]
 INTERNAL POWER SUPPLY: 5 V DC REGULATED.
 INPUT POWER REQUIRED: 8-12 VDC OR 115 VAC AT 50/60 HZ.
 POWER CONSUMPTION: 4 WATTS



KIT #FC-50C IS COMPLETE WITH PREDRILLED CHASSIS ALL HARDWARE AND STEP-BY-STEP INSTRUCTIONS. WIRED & TESTED UNITS ARE CALIBRATED AND GUARANTEED.

PLEXIGLAS CABINETS

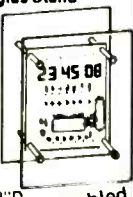
Great for Clocks or any LED Digital project. Clear-Red Digital serves as Bezel to increase contrast of digital displays.

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3"H, 6 1/4"W, 5 1/2"D Black, White or Clear Cover
- CABINET II**
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RED OR GREY PLEXIGLAS FOR DIGITAL BEZELS
3"x6"x1/8" 95¢ ea 4/93

SEE THE WORKS Clock Kit
Clear Plexiglas Stand

- 6 Big .4" digits
- 12 or 24 hr. time
- 3 set switches
- Plug transformer
- all parts included



Plexiglas Pre-cut & drilled
Kit #850-4CP
Size: 6"H, 4 1/4"W, 3"D

Assembled \$23⁵⁰ ea. 2/*45. \$29⁹⁵

60 HZ.

XTAL TIME BASE:
Will enable Digital Clock Kits or Clock-Calendar Kits to operate from 12V DC.
1"x2" PC Board
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Easy 3 wire hookup
Accuracy: ± 2PPM
#TB-1 (Adjustable)
Complete Kit \$4⁹⁵
Wir & Cal \$9.95

SPECIAL PRICING!

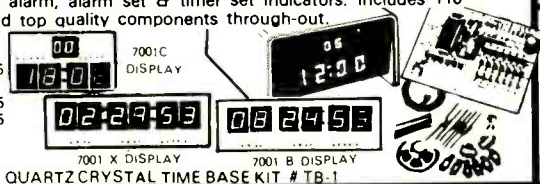
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LOW POWER - FACTORY FRESH
1-24 \$1.75 ea. 100-199 \$1.45 ea.
25-99 1.60 ea. 200-999 1.39 ea.
1000 AND OVER **\$1.29** ea.

6-DIGIT LED CLOCK CALENDAR KIT
DATE-TIME-SNOOZE ALARM & MORE... KIT 7001

FOR THE BUILDER THAT WANTS THE BEST. FEATURING 12 OR 24 HOUR TIME — 29-30-31 DAY CALENDAR. ALARM, SNOOZE AND AUX. TIMER CIRCUITS

Will alternate time (8 seconds) and date (2 seconds) and may be wired for time or date display only, with other functions on demand. Has built-in oscillator for battery back-up. A loud 24 hour alarm with a repeatable 10 minute snooze alarm, alarm set & timer set indicators. Includes 110 VAC/60Hz power pack with cord and top quality components through-out.

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- KIT - 7001C WITH 4 .6" DIGITS & 2 .3" DIGITS FOR SECONDS \$42.95
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KITS ARE COMPLETE (LESS CABINET)
ALL 7001 KITS FIT CABINET I AND ACCEPT QUARTZ CRYSTAL TIME BASE KIT #TB-1

PRINTED CIRCUIT BOARDS for CT-7001 Kits sold separately with assembly info. PC Boards are drilled Fiberglass, solder plated and screened with component layout.

Specify for 7001
B, C or X - \$7.95

AUTO BURGLAR ALARM KIT

AN EASY TO ASSEMBLE AND EASY TO INSTALL ALARM PROVIDING MANY FEATURES NOT NORMALLY FOUND. KEYLESS ALARM HAS PROVISION FOR POS & GROUNDING SWITCHES OR SENSORS. WILL PULSE HORN RELAY AT 1HZ RATE OR DRIVE SIREN. KIT PROVIDES PROGRAMMABLE TIME DELAYS FOR EXIT ENTRY & ALARM PERIOD. UNIT MOUNTS UNDER DASH. REMOTE SWITCH CAN BE MOUNTED WHERE DESIRED. CMOS RELIABILITY RESISTS FALSE ALARMS & PROVIDES FOR ULTRA DEPENDABLE ALARM. DO NOT BE FOOLED BY LOW PRICES! THIS IS A TOP QUALITY COMPLETE KIT WITH ALL PARTS INCLUDING DETAILED DRAWINGS AND INSTRUCTIONS OR AVAILABLE WIRED AND TESTED



KIT #ALR-1 \$9.95
#ALR-1WT WIRED & TESTED \$19.95

VARIABLE REGULATED 1 AMP POWER SUPPLY KIT

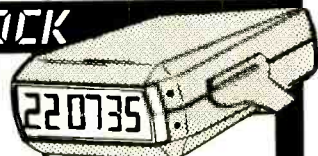
- VARIABLE FROM 4 TO 14V
- SHORT CIRCUIT PROOF
- 723 IC REGULATOR
- 2N3055 PASS TRANSISTOR
- CURRENT LIMITING AT 1 Amp

KIT IS COMPLETE INCLUDING DRILLED & SOLDER PLATED FIBERGLASS PC BOARD AND ALL PARTS (Less TRANS. FORMER) KIT #PS-01 \$8.95
TRANSFORMER 24V CT will provide 300MA at 12V and 1 Amp at 5V. \$3.50

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12/24 HR .4" DIGITS!

MODEL 12 VOLT AC OR #2001 DC POWERED



- 6 JUMBO .4" RED LED'S BEHIND RED FILTER LENS WITH CHROME RIM
- SET TIME FROM FRONT VIA HIDDEN SWITCHES • 12/24-Hr. TIME FORMAT
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- MOUNTING BRACKET INCLUDED

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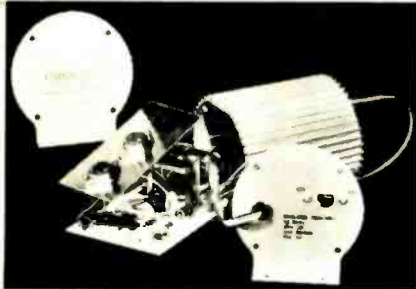
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ZOOM LENS

9 TO 30 mm f:1.9 Lens



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- USE INDOORS OR OUTDOORS
- ALL SOLID STATE

15 FOOT CABLE AND PISTOL GRIP
\$195.00
Complete

NOT A KIT
Fully Guaranteed

DIGITAL AUTO INSTRUMENTS
SEVEN MODELS!

- #1 TACHOMETER
- #2 WATER TEMP.
- #3 FUEL LEVEL
- #4 SPEEDOMETER*
- #5 OIL PRESSURE
- #6 OIL TEMP.
- #7 BATTERY MONITOR



- KIT INCLUDES:
- CASE & ALL HARDWARE
 - PRESSURE & TEMP. SENDERS
 - ASSEMBLED MAIN PC BOARD
- FEATURES:
- 4 ORANGE LED'S
 - 4 1/2" x 4 1/2" x 2" ABS CASE

*ADD \$10 FOR REQUIRED SPEED SENDER \$15 FOR SPEED SENDER ALONE
KIT: \$49.95 ASSEMBLED: \$59.95

SPECIFY
12 OR 24
HOURS

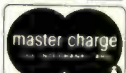
3 1/2 DIGITAL CLOCK

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