

Lowdown on electronic burglar alarms

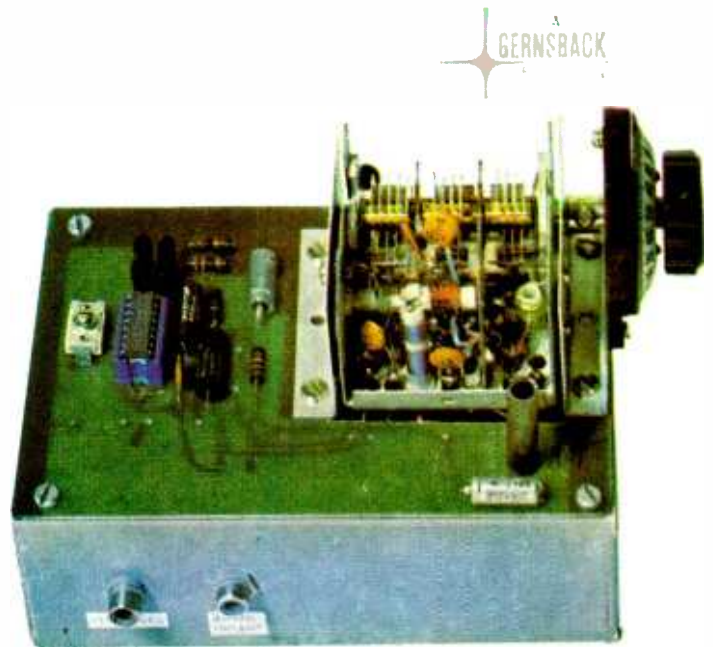
60c ■ NOV. 1970

Radio-Electronics^{IND}

FOR MEN WITH IDEAS IN ELECTRONICS

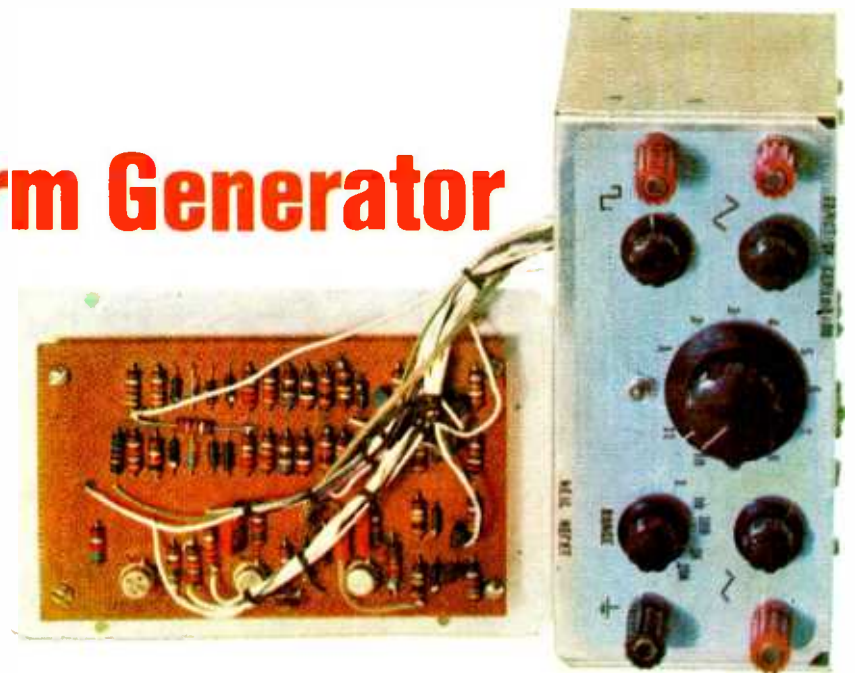
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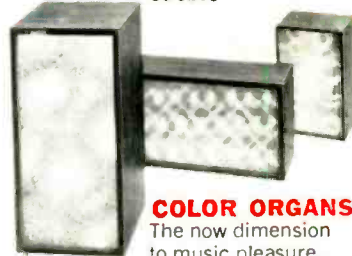
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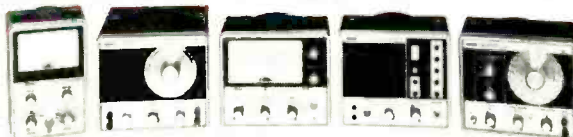
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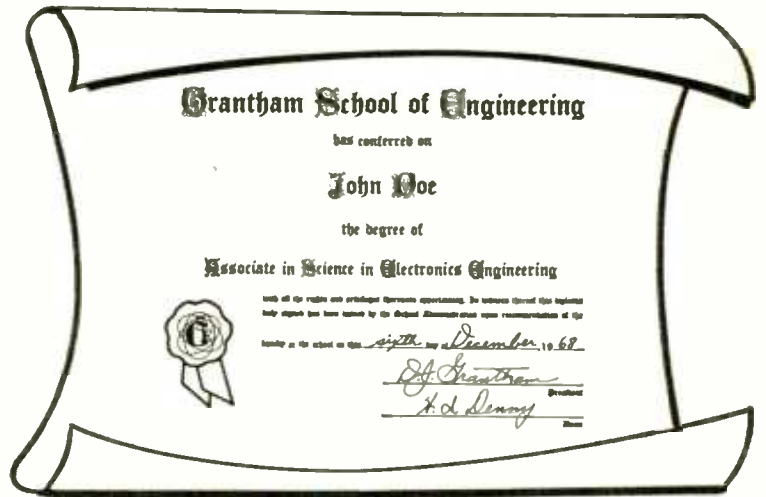
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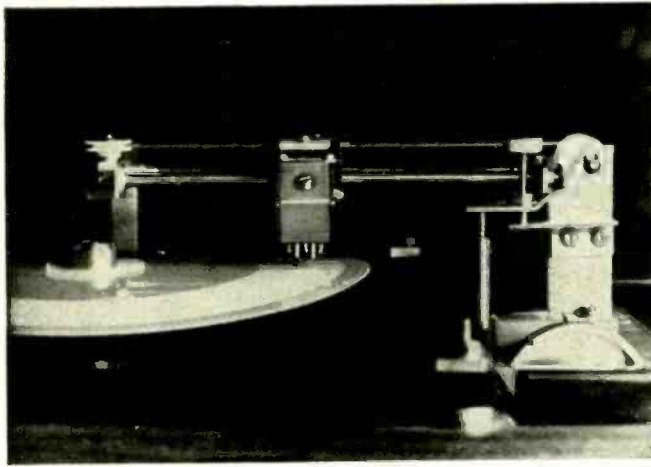
Volume 41 Number 11

RADIO-ELECTRONICS . . . FOR MEN WITH IDEAS IN ELECTRONICS

November 1970

TV PICTURES ON A DISC

Thin plastic disc rotates at 1500 rpm, floats on air and produces TV pictures. Made by AEG-Telefunken, the unit looks a lot like a conventional record player. The unit shown handles only black-and-white, but color is in the works. For more details see Looking Ahead on page 4. ★



BATTERY-POWERED CONTINUOUS LASERS

MURRAY HILL, N.J.—A new laser, smaller than a grain of sand, that can be powered by ordinary dry-cell batteries, has been built by Bell Laboratory Scientists. It emits a beam of near-visible infrared light.



Designed by Izuo Hayashi and Morton Panish, the new laser is a semiconductor device that operates continuously at normal room temperature (75-80°F). This is about the average operating temperature of typical transmission equipment used in communications systems.

In the past, the heat generated in semiconductor lasers was too great to permit operation at room temperature for more than a fraction of a second at a time without destroying the semiconductor itself. Even when these lasers were mounted on heat sinks, room-temperature operation was limited to about 1/10,000 second.

By building a new semiconductor laser structure the designers eliminated the need for additional cooling.

The new semiconductor laser consists of four layers of semiconductor materials. Two of the layers, each about 60 millionths of an inch thick, confine laser light to a thin central layer of the structure. This layer, called the active region is about 20 millionths of an inch thick.

Laser activity, effectively confined to such a small region, can be produced with much less current, resulting in less heating. This makes the continuous laser operation possible. ★

Supermagnets To Speed Tomorrow's Transit

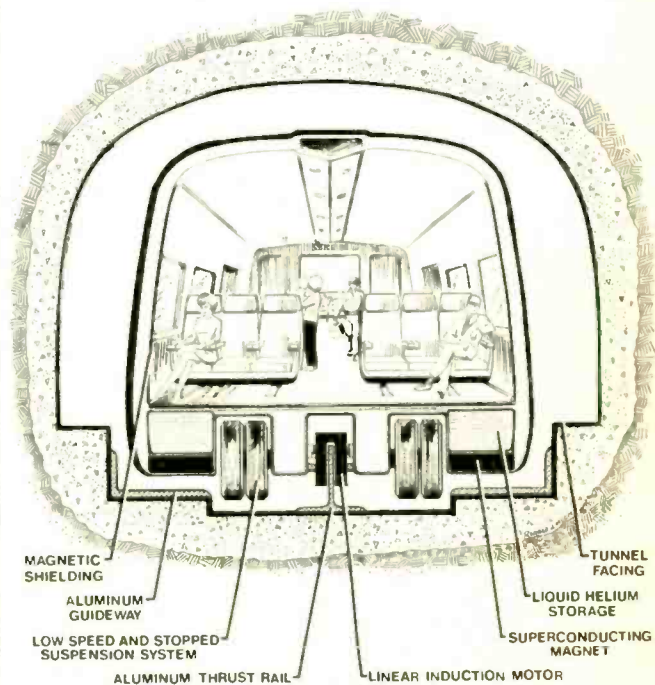
MENLO PARK, CALIF.—A new rapid transit vehicle which can race along an invisible magnetic "track" at over 300 miles-an-hour is under study at Stanford Research Institute. The new transit mode would use the tremendous power of superconducting magnets to lift the weight of the vehicle off its track and propel it to extremely high speeds.

Key to the system is magnetic levitation, a phrase which refers to the use of magnetic repulsion to lift the vehicle free of any physical

inventing the time-consuming auto trip between airport and downtown destinations this new transit system could compete with airliners operating in high-density travel corridors.

SRI researchers reasoned that a transit vehicle carrying powerful superconducting magnets could be levitated as much as a foot above a metal-lined roadway. The same magnet forces could be used to keep the car centered in the roadway.

The suspended vehicle could be propelled by linear-



support, allowing it to achieve very high speeds easily and economically. Pushed by linear induction motors, which also use magnetic forces, the vehicle could travel at high speed in virtual silence.

Operating between downtown urban centers such a vehicle could whisk passengers between New York and Washington in about an hour. The trip between downtown San Francisco and downtown Los Angeles could be made in about 1-1/2 hours. By elim-

ination motors. They use magnetic power, are silent, and produce no contaminating exhaust.

In theory the only limit to the speed of such a vehicle would be the resistance of the surrounding air. This could be solved by having the vehicle run in a tunnel from which enough air had been removed to create a partial vacuum. Passengers in the vehicle would be as comfortable as those riding in the pressurized cabin of an airliner. (continued on page 12)

Radio-Electronics

FOR MEN WITH IDEAS IN ELECTRONICS

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IC unit delivers sinewaves, square waves, rectangular waves

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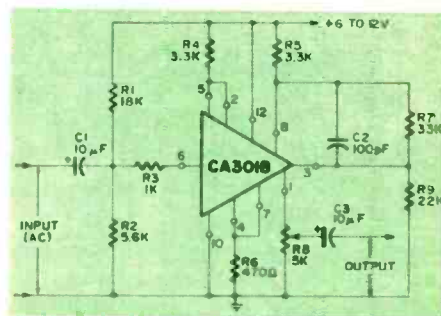
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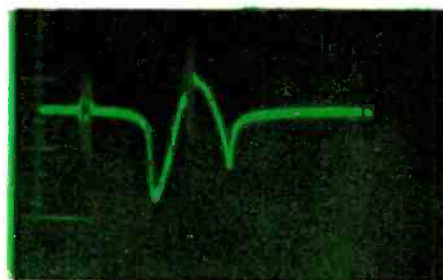
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The CA3018 has four transistors inside its case. Here are more ways to put it to work for you. . . . see page 37



Computer technicians are in demand. Discover the opportunities in this exciting growing industry. . . . see page 52



Kwik-Fix waveforms like this one will help you make quadrature repairs a cinch. For complete details see page 41

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LOOKING AHEAD

Volume 41 Number 11

RADIO-ELECTRONICS . . . FOR MEN WITH IDEAS IN ELECTRONICS

November 1970

by **DAVID LACHENBRUCH**
CONTRIBUTING EDITOR

Video disc system

A new entry in the videoplayer sweepstakes has some engineers very excited. It's the long-rumored video disc system developed by German manufacturer AEG-Telefunken with British Decca and demonstrated recently in Berlin. The playback unit looks, at first sight, like a conventional phonograph. The grooved discs are extremely thin flexible plastic records measuring eight and 12 inches in diameter, and playing for five and 12 minutes respectively in black-and-white. (See photo in *New & Timely*, page 2.)

Each disc contains from 120 to 140 grooves per millimeter, and spins at 1,500 revolutions per minute, producing one frame per revolution. On the turntable, the discs ride on a cushion of air, and a new type of pickup responds to the variations in groove elevation. Telefunken is developing longer-playing discs, a disc changer and a system which plays back in color.

American engineers who witnessed the Berlin demonstration say picture quality was very good. Among the claimed advantages of the disc system over other videoplayer techniques: (1) Economy—discs may cost five dollars each, or less; players would be priced somewhere between \$140 and \$280. (2) Fast duplication—discs may be pressed in a conventional manner, about 1,000 in 12 minutes. (3) Playback ease—the phonograph is still the most familiar type of playback equipment, and the consumer would not have the problem of learning new procedures for playing visual material.

For color consistency

A new series of signals is going out over the air along with color TV broadcasts. It's an experimental system being tested by the television broadcasters as part of an all-out effort to achieve uniformity in color from program-to-program, scene-to-scene and channel-to-channel.

On selected programs of all three networks, a test is being made of the use of part of the "vertical interval" between frames to transmit reference signals for contrast, tint and color intensity. The vertical interval reference signals are inserted in line 20 of field 1. If the tests work out, stations will be supplied with a method for absolute calibration of equipment to produce uniformity in colors. This and other procedures and standards being developed by a top-level engineering committee representing TV broadcasters, set manufacturers, film and tape companies and the telephone company, eventually could lead to color uniformity so good that the tint control could be eliminated from color TV receivers. At least, that's the hope of the engineers.

Standard amplifier measurements

The rule faced by the purchaser of high-fidelity equipment—particularly the unknowledgable one—has always been "let the buyer beware." Claims for amplifier power, in particular, have been difficult or impossible to compare. There have been several so-called "standard" ways of measuring—the Institute of High Fidelity (IHF) standard, the Electronic Industries Association (EIA) "music power" standard, and something that some manufacturers dreamed up and called "peak music power." Generally, the IHF standard was designed to apply to audio component ampli-

fiers, the EIA and "peak" standards to packaged consoles—and there was no real way for the layman to compare various equipment.

Finally, after some 15 years of discussion and debate, a standard system for describing audio amplifier power appears to be in sight. The EIA has modified its system to the point where it would be compatible with the IHF system. Under the proposal, which had not yet been officially approved by EIA members at press time, all manufacturers would use the same measurements and terminology.

The original EIA "music power output" (MPO) standard, developed in 1960, specified five percent total harmonic distortion. The new proposed standard permits the manufacturer to use any total harmonic distortion level he wishes, but not to exceed five percent. In addition, the EIA has changed test procedures, renamed MPO "momentary power output" and added a continuous power output test and rating procedure.

EIA's proposed new rule also includes definitions and procedures for establishing rated intermodulation distortion, intermodulation distortion-low level, frequency response, channel separation, signal-to-noise ratio, input overload margin, input impedance, output impedance, damping factor and tracking error.

The EIA-IHF efforts to arrive at a common standard were accelerated by a new move by the Federal Trade Commission to step in and police amplifier output claims through a new Trade Regulation Rule. The proposed rule is still in exploratory stages but would require that manufacturers use standard measurements whenever they made claims for amplifier power. EIA and IHF hope that the FTC will accept their proposed joint standard, giving it—in effect—the force of law, and ending at least some of the confusion in the amplifier "horsepower race."

Home videoplayer sweepstakes

Keep your eye on the hot home video-player field. It could end up as biggest boom—or biggest bust—in the history of consumer electronics. It appears that almost everybody has a system, and hardly any one is compatible with any other. In addition to the Telefunken system described above, it's believed that some 20 or more different systems will have been exposed before the year is over. Among non-compatible approaches are RCA's holographic-tape Selecta-Vision, CBS's electronically-recorded EVR film, Sony's Video-cassette magnetic tape system, and other non-compatible tape approaches by Cartridge Television (Cartrivision), Ampex, Arvin, Philips and others. In addition there is the ABTO color-encoded film system, plus electronically-scanned Super-8 systems introduced by several manufacturers.

As each new system is introduced, the question of standards becomes more urgent. The promising new field of pre-recorded audio-visual records for the home is in danger of strangling on consumer confusion over the multiplicity of systems. How many years would audio recording have been set back if there had been introduced simultaneously cylinder recordings in eight different speeds, discs which played clockwise and discs which played counter-clockwise in five different speeds, magnetic tapes in four widths, grooved tapes recorded by four different non-compatible methods and optical film recordings in several configurations? It's enough to bring back vaudeville, band concerts in the park and other innocent non-electronic diversions.

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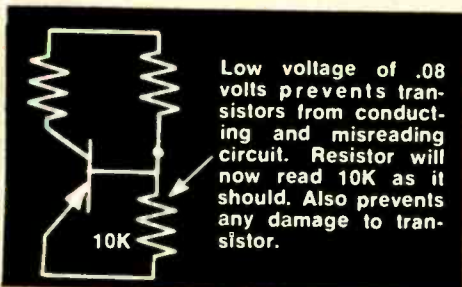
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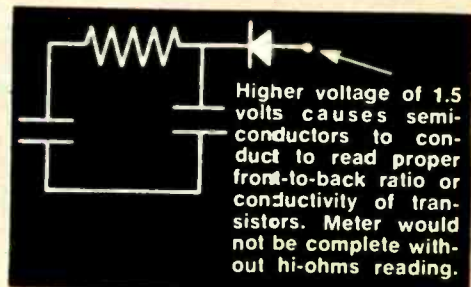
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Editorial — TO THE CONSTRUCTOR

FROM DAY TO DAY, OUR PERSONAL EXPERIENCES AND CALLS and letters from you indicate that it is becoming increasingly difficult to obtain all the parts needed to complete any, except the simplest of construction projects. With present-day construction techniques tending toward increased use of semiconductors, printed circuits and miniaturization, it is often impossible to substitute more-readily available components.

Take, for example, a magazine article describing a photoelectric control unit with a PC board laid out to accept a particular relay and power transformer. Its cabinet is just large enough to house the electronic circuitry. You, like most readers, want to build the unit exactly as the author described it. The relay, transformer and cabinet—the critical parts in this project—are all standard replacement or consumer-type items listed in manufacturers' catalogs but you can expect some difficulty in obtaining at least one of these parts. This is especially true when you try to buy them locally.

Most of you go to local electronics parts houses for components for your projects. There are three basic types of parts suppliers. One is set up to serve the local TV and radio service industry or possibly the needs of nearby electronics or industrial plants. This type of dealer stocks only those parts experience has shown can be sold fast and with the least difficulty. The variety of components often needed by experimenters and other nonprofessionals are not carried in stock. Experimenters, CB'ers and do-it-yourselfers are not welcome because they often ask for assistance or advice which the dealer is not prepared (or able) to give.

The second type of parts house specializes in hi-fi or CB gear and stocks only those IC's, transistors, tubes, resistors, capacitors and other components that might be required to maintain the equipment it sells. You may find "blister packs" of general-purpose semi-conductors and other components for beginning students and experimenters but you are not likely to find all the parts needed for an advanced construction project.

Mail-order houses—the third type—are generally the best source of parts for advanced construction projects. Most of these serve all phases of the electronics industry, schools and experimenters and are more likely to have the

parts you need in stock. If a component is out of stock, they can supply it without undue delay. Note well that not all mail-order dealers carry all brands of components and some will carry the full line of a particular manufacturer while others do not.

More and more parts manufacturers are gearing their product and sales toward the industrial market and sell only through wholesale (OEM-type) distributors. With the exception of some mail-order houses many distributors are basically wholesalers who have a minimum-order sales policy. That is, they don't accept orders totaling less than a specific amount—usually \$8 to 10. RADIO-ELECTRONICS has turned down many articles on equipment requiring a special but inexpensive part obtainable only through sources which have a minimum-order figure many times the value of that part. If a special part for a construction project cannot be readily obtained at reasonable cost, there is no place for that story in RADIO-ELECTRONICS.

If you want to complete your construction projects without the frustration of not being able to locate a critical component, you will have to rely more and more on mail-order houses. Naturally, you will have to await the arrival of the parts from the mail-order house but this is not nearly as frustrating as waiting for one part ordered through a local dealer.

I have nothing against local dealers. It is just that most are specialists who do not stock the type of parts required for many of your projects.

If you enjoy constructing your own electronic gear from magazine articles, here are a few tips we hope will enhance your enjoyment of future projects.

(1) Before beginning a construction project from a magazine, read the article and parts list again. Then, go over it once more. Often, in your anxiety to get started, you overlook authors' or editors' notes on obtaining a particular part.

(2) Obtain catalogs from at least six mail-order houses. Study them until you are familiar with the types and brands of components listed in each.

(3) When ordering, follow instructions to the letter. List page and stock numbers and a description of each

(continued on page 14)

Radio-Electronics

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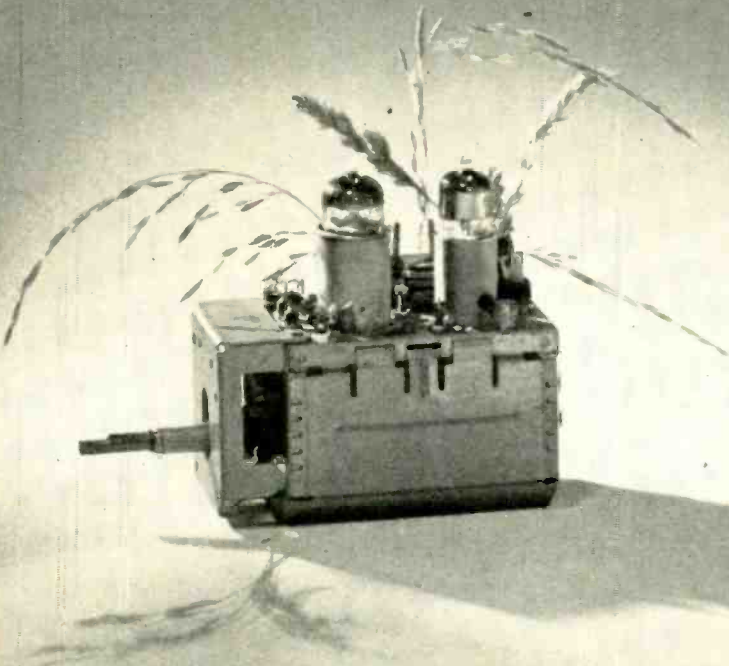
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The rest of our new miracle service chemical line

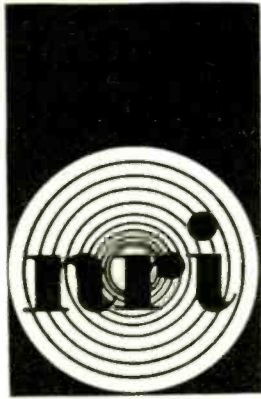
50 BELOW Circuit Cooling Spray...cools to at least 20 degrees below any other spray on the circuit. TAPE HEAD CLEANER-penetrates deep to clean and protect tape heads, reduce noise and improve fidelity. TUNER CLEANER-dispenses under 80 pounds pressure to do the job of an ultrasonic bath.

Special Intro Offer: Your first 5 can order of Channel Master service chemicals comes packed in a free metal tote box with fold out storage tray...Grab onto one!



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Whatever your reason for wanting knowledge of Electronics, you'll find the NRI "3 Dimensional" method makes learning exciting, fast. You build, test, experiment, explore. Investigate NRI training plans, find out about the NRI Achievement Kit. Fill in and mail the postage-free card. No salesman will call. NATIONAL RADIO INSTITUTE, Electronics Division, Washington, D. C. 20016



ELECTRONICS COMES ALIVE AS YOU LEARN BY DOING WITH CUSTOM TRAINING EQUIPMENT

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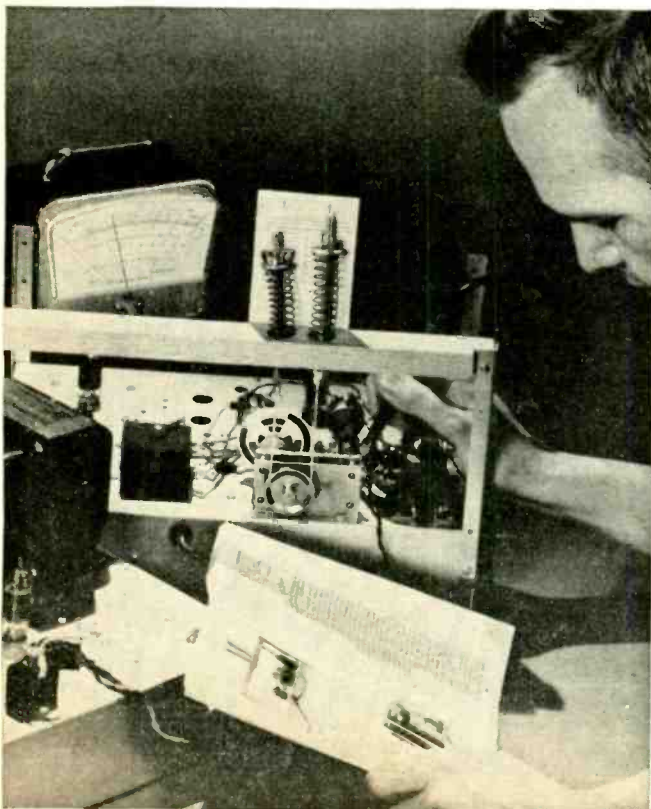
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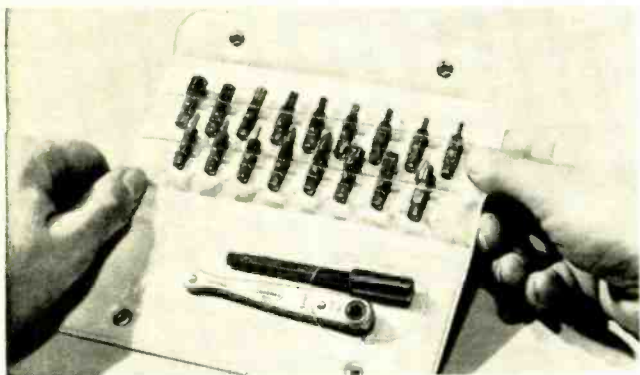
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custom training kits “bite-size” texts



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Stainless steel Midget Ratchet handle, 17 Drivers, 3 $\frac{3}{4}$ " extension, two slotted and two Phillips bits, 12 hexagonal bits from 0.050 to 5/16", one square adapter for $\frac{1}{4}$ " wrench sockets—all in pocket-sized vinyl case.

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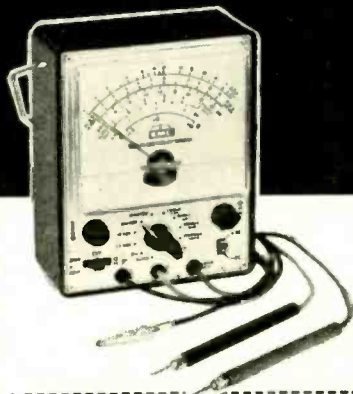
NEW BATTERY-OPERATED FET SOLID-STATE VOLT-OHMMETER #116

Easy-to-build KIT

\$29⁹⁵ #116K

Factory-Wired & Tested

\$39⁹⁵ #116W



Now you can get all the benefits of a VTVM (laboratory accuracy, stability and wide range) but with its drawbacks gone: no plugging into an AC outlet, no waiting for warm-up, no bulkiness. New Field Effect Transistor (FET) design makes possible low loading, instant-on battery-operation and small size. Excellent for both bench and field work.

Compare these valuable features:

- High impedance low loading: 11 meg-ohms input on DC, 1 megohm on AC
- 500-times more sensitive than a standard 20,000 ohms-per-volt VOM
- Wide-range versatility: 4 P-P AC voltage ranges: 0-3.3, 33, 330, 1200V; 4 RMS AC voltage ranges: 0-1.2, 12, 120, 1200V; 4 DC voltage ranges: 0-1.2, 12, 120, 1200V; 4 Resistance ranges: 0-1K, 0-100K, 0-10 meg., 0-1000 meg.; 4DB ranges: -24 to +56DB.

Sensitive easy-to-read 4 $\frac{1}{2}$ " 200 micro-amp meter. Zero center position available. Comprises FET transistor, 4 silicon transistors, 2 diodes. Meter and transistors protected against burnout. Etched panel for durability. High-impact bakelite case with handle useable as instrument stand. Kit has simplified step-by-step assembly instructions. Both kit and factory-wired versions shipped complete with batteries and test leads. 5 $\frac{1}{4}$ "H x 6 $\frac{3}{4}$ "W x 2 $\frac{7}{8}$ "D. 3 lbs.

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New & Timely

(continued from page 2)

SRI scientists call their concept MAGLEV (for Magnetic Levitation). They first worked it out two years ago in a theoretical design for an experimental rocket sled ca-

pable of speeds above 6,000 miles an hour. The magnetically-suspended sled, freed of friction and weight, was designed to be accelerated along a special track to tremendous speeds by a fast-burning rocket motor. ★

ZENITH SUBSCRIPTION TV APPROVED BY FCC

CHICAGO, ILL.—Zenith Radio Corporation's Phonevision system for over-the-air subscription TV (STV) is the first system granted technical approval by the Federal Communications Commission. Joseph S. Wright, Zenith's board chairman, said, "With FCC technical approval, Teco, Inc., Zenith's licensee for commercial development of Phone-

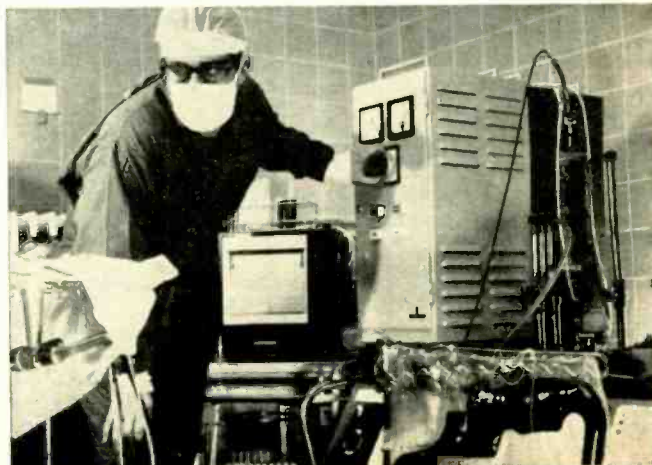
vision in North America, is in a position to implement plans to introduce this new box office TV service."

Under the FCC rules for over-the-air STV, it was noted, no station application for regular STV broadcasting can be granted unless the station has an agreement to use an FCC-approved system. ★



ELECTRONIC HOME TV BOX OFFICE—Zenith's subscription TV (STV) decoder is part of the first STV system to receive FCC technical approval. This compact solid state unit can enable subscribers to see first run movies, blacked-out sports events, Broadway shows and others features not available on advertising-sponsored TV, for a modest charge. To select an STV feature, the subscriber uses a subscriber "ticket" and dials a simple code.

Electrons Control Body Temperature

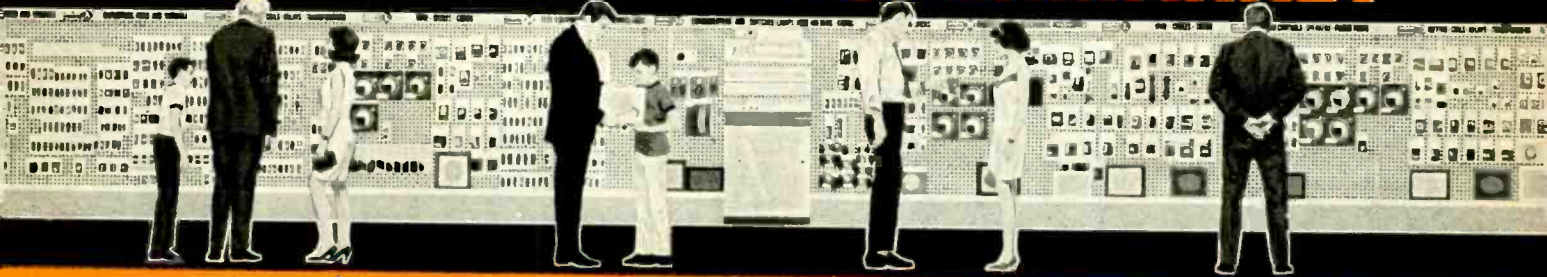


ERLANGEN, GERMANY—Operations on the heart and aorta are generally performed with

the aid of a heart-lung machine. This ensures a proper supply of oxygen and keeps

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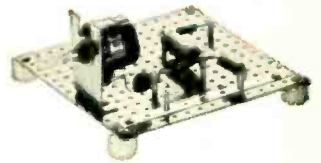
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Diagonals, long nose pliers, soldering iron and solder, solder aid tool, heat-sink, and screwdriver. A deal gift item. Cat. No. H3-378 \$7.95 Net.



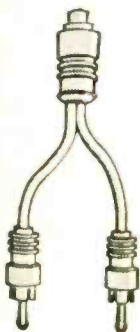
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"Y" Audio Adaptor has a phono pin jack on one end and dual phono pin plugs on the other. Cat. No. Q4-288 \$.99 Net



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Easy to Assemble Printed Circuit Board Kit — \$49.95
Factory Assembled — \$65.00

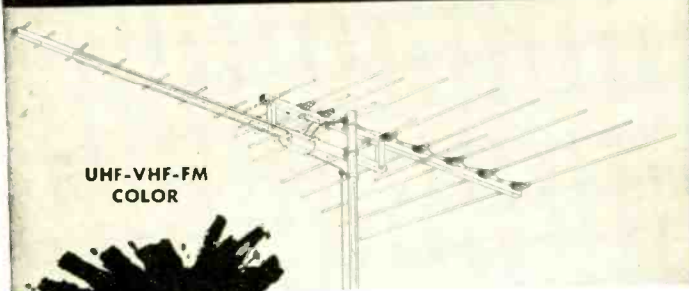
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New & Timely

(continued from page 12)

the patient's blood circulation intact. In special cases it is advantageous to lower the temperature of the patient's body during such complicated operations to reduce metabolism as much as possible and thus any post-operative consequences.

Scientists of the Siemens Research Centre in Erlangen (Federal Republic of Germany) have developed a continuous-flow heat exchanger of stainless steel (capacity 190 ml) which is fitted with Sirigor cooling blocks (Peltier elements).

A Peltier element consists of two semi-conductor legs of which the one is of the n-type and the other of the p-type. The two legs are connected at the face end by a copper strap. If dc flows in a given direction through the

Peltier element, the copper strap absorbs thermal energy from the immediate surroundings, and cools off. The thermal energy absorbed by the cold side of the element—boosted by the current input—is given off on the other side of the element. The Peltier element thus "pumps" thermal energy from its cold to its warm side. The direction of the heat transfer can be reversed by reversing the direction of the current. A Sirigor cooling block is made up of several Peltier elements connected meander-fashion in series. Although the elements are connected in series electrically, they are in parallel thermally so that all the cold copper straps are on the one side and all the warm ones on the other side of the cooling block.

R-E

Editorial (continued from page 6)

item when asked for on the order blank. Failure to do this will delay your order. Parts are frequently stocked in bins by catalog number, not by name, brand or part number. If you omit catalog numbers, someone will have to find time to look up the numbers and perhaps rewrite your order in a way that it can be handled by workers in the stockroom. In many cases, the person filling your order may not be able to tell a resistor from an i.f. transformer.

(4) Read new-literature columns in this and other magazines and write for catalogs and data sheets from components manufacturers. Many contain substitution guides, dealer and distributor listings and a wealth of technical information that will prove useful when selecting parts for future projects. You should have catalogs from at least two makers of each type of common electronic component.

(5) Pay particular attention to the parts list when planning a construction project from an old issue. Some parts which were readily available at time of publication may have been discontinued and there may not be an acceptable substitute. If there is any indication of special components in the device, check with the manufacturer or listed source of supply before purchasing any of the more readily available components.

(6) Don't rush headlong into a construction project whose operation depends on a special surplus component such as a meter, relay, transformer or filter network. Make sure that the part is available before you begin. The source may have "dried up" and there is no commercial equivalent available.

Construction projects can be fun when you know the ropes. Follow the tips above and then dig in and tackle some of the equipment on the following pages.

Robert F. Scott

SENIOR TECHNICAL EDITOR

RADIO-ELECTRONICS



**B & K Precision
Model E-200D RF Generator**



**B & K Precision
Model E-310B Sine/Square
Wave Generator**

NEW GENERATION FROM B&K

Go ahead. Compare them with any others. These new all solid state Precision generators from B&K do more things and do them better. Because B&K makes products that really work.

Take the Model E-200D RF generator, for instance. You get wide-range coverage from 100 KHz to 216 MHz in 5 bands. A completely shielded RF output provides continuously variable monitored attenuation . . . calibrated to less than a microvolt on fundamentals to 54 MHz. And built-in crystal calibration assures you of frequency accuracy.

You get a giant 5" vernier dial with anti-backlash 12:1 gear ratio. So it's easy to obtain and repeat the results you want.

Or consider the Model E-310B. This new sine/square wave generator gives you step attenuation of both outputs.

And a variable control lets you adjust continuously from maximum down to less than 0.25 millivolt.

It also features extended frequency ranges: 20 Hz to 2 MHz for sine wave, 20 Hz to 200 KHz for square wave. Plus constant-voltage output over each band and band-to-band.

Both new generators are based on earlier models that made us famous. Now you can have all the additional benefits only transistors make possible.

See the new solid state generation from B&K at your local distributors. Or write us for complete details.

B & K Precision MODEL E-200D RF GENERATOR \$159.95

Other features:

- Four outputs: unmodulated RF, modulated RF (400Hz), externally modulated RF, 400 Hz audio signal.
- Covers full VHF TV range; can be used as crystal-calibrated marker (100 KHz to 54 MHz on fundamentals, 54 to 216 MHz on harmonics).
- FM and TV channels identified on dial.
- Three-wire AC line cord.
- RF cable.

B & K Precision MODEL E-310B SINE/SQUARE WAVE GENERATOR \$119.95

Other features:

- All solid state—100% silicon transistors.
- Continuously variable output control plus accurate 6-step attenuator for precise voltage division of both sine and square wave outputs.
- Fast square wave rise time (less than 100 nanoseconds).
- Wide range permits low frequency RF applications.
- Large 5" dial with vernier tuning for maximum accuracy
- Three-wire AC line cord.



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have to solder,
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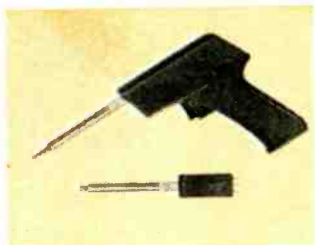
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equipped with soldering tools that perform as well as those in your production line? If not they should be . . . it's costing you money!

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By JACK DARR
SERVICE EDITOR

Up-Down-Positive-Negative

For a long time, old-timers like me have said that a voltage "going up" means it goes more positive, and "going down" means it goes more negative. This worked fine with tubes because plate voltages were always positive and that's what we did most of the work with. Now, we've got to change our "thinking!"

We've got to learn to think "goes more positive" or "goes more negative." This is true even in tube circuits used nowadays, let alone solid-state stuff. And, the way we *think* about the circuits determines the speed with which we can analyze their operation and find faults in them! You do not do this with voltmeters or any kind of test equipment. You do this with that mass of blue mud between your ears! The proper thinking-methods will speed up the process of getting the information, forming a conclusion, and making the diagnosis.

There are circuits we work with every day that demand this type of thinking. The bias on color picture tubes is a good example. How many times have you been fooled by a "no raster" symptom, only to find out (sometimes very painfully) that there *was* plenty of high voltage after all? A biased-off picture tube has a dark screen, but lots of high voltage on the ultor to bite unwary fingertips.

All of these things can be traced back to circuit reactions as simple as current flowing through a resistor—Ohm's Law. In the video output stage used in millions of color TV sets, the voltage on the video output tube plate and CRT cathode are determined by the amount of current flowing through the plate load resistor. (Incidentally, we can still say "up" and "down" speaking of current, no matter which way it's flowing!)

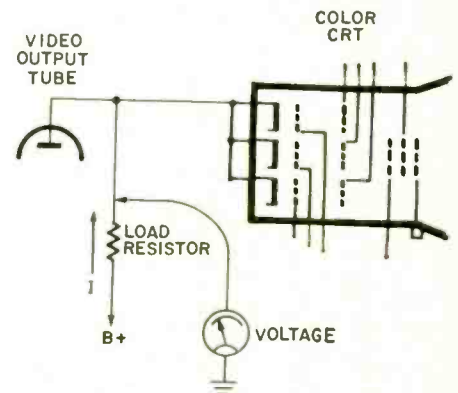
If the video output tube plate current increases, the voltage at the plate goes more negative. So the cathode of the picture tube goes more negative since it is directly coupled to the plate.

A negative-going voltage on the cathode of all vacuum tubes is the same as a positive voltage on the grid. (We're assuming now that the grid voltage is fixed.) The important relationship here is the voltage between cathode and grid (*not to ground*). This determines the conduction of the tube. So if the cathode goes toward a greater negative voltage, the grid-cathode voltage has gone more positive. Result; the raster gets brighter (more beam current).

A positive-going cathode voltage has the opposite effect—the grid has been

made more negative with respect to the cathode and beam-current is reduced. A dimmer raster results.

There's a dandy example of the value of thinking of the circuit reactions in terms of "positive-going" or "negative-going" instead of "up" or "down". The terms will remind you of the *effect* that the voltage-changes have on the circuit operation. You can get the same effects in the CRT grid circuits, of course. A more positive plate voltage on a color-difference amplifier tube (directly coupled to one CRT grid) makes that grid more positive. What does a more



positive grid voltage do in all tubes? Makes them conduct more current. In this case the raster will be brighter, because the CRT is being biased to draw more beam current.

This kind of thinking can be most helpful in solid-state circuits. Why? Because we can find transistors with "negative plate voltages"! The collector voltage of a pnp transistor in a common-emitter circuit *is* negative. If this transistor conducts more heavily, draws more current through its load resistor, the voltage will go more positive. (Oh yes it will too. Assume that the normal collector voltage was -15 volts, from a -20 volt supply through a load resistor. If this changes to -5 volts, a -5 volt reading is 10 volts more positive than -15 volts, isn't it?)

In cases like this, you might have a -20 volt supply, with only 5 volts dropped across the load resistor in normal operation. This leaves us -15 volts on the collector. A higher collector current might cause a 15-volt drop across the load resistor, leaving us only -5 volts on the collector. This is a *symptom*, not a *cause*. The cause is whatever is making this transistor draw three (continued on page 86)

C

orrespondence



PROGRESS WITH SCR'S

It is noted that the Noteworthy Circuits item, "Portable SCR Tester," in the March 1970 RADIO-ELECTRONICS (page 94) reproduces, with editorial changes, GE Application Note 201.3.

Our policy is to request that our application notes not be copied without permission. New product introductions constantly require updating of circuits, and when requests are received to publish circuits disclosed in our publications, it affords us an opportunity to inform the author of the most recent advances in the art.

Our suggestion for replacing the C5U SCR described is to employ a C106Q (any type 1-4). For the electronic hobbyist, it may be convenient to pick up a GEMR-5.

CAROL O. THOMAS

Patent Attorney

General Electric Co.

We appreciate the suggested update for the SCR tester, it will aid our readers. Our normal procedure is to credit sources of information that appear in RADIO-ELECTRONICS. The oversight in this instance was due to the absence of any credit or reference to General Electric in our file copy.

FORGOTTEN AMMETER

The autotransformer story by James Squires (Sept. 1970) was of interest because I don't believe a technician's bench should be without one.

However, after reading the item, there was one item that should have been included in the circuit—an ac ammeter. The ammeter is necessary to know the power drawn by the unit under repair, and quickly lets you know if there's a short in the work.

Use an ac ammeter and voltmeter with a Variac variable line transformer built into my bench, connected to an output plug. By starting the transformer at zero, I've saved many receivers from burning with this rig. I advise anyone building one of these units to wire in the ammeter. I wired a shunted switch across mine so it reads to 2 and 20 amps.

PETER LEGON

Malden, Mass.

(continued on page 24)

RADIO-ELECTRONICS



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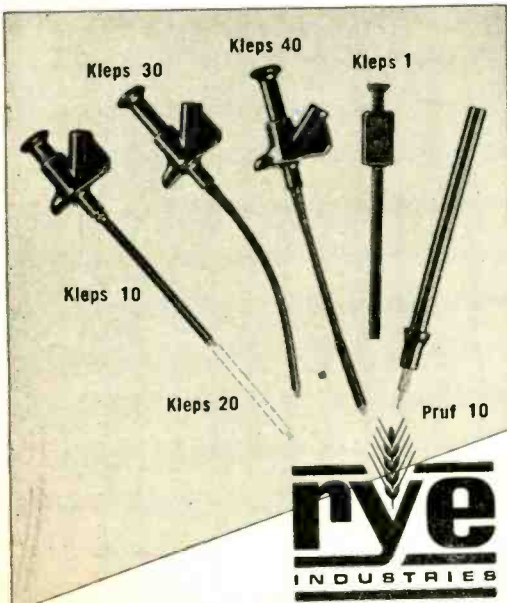
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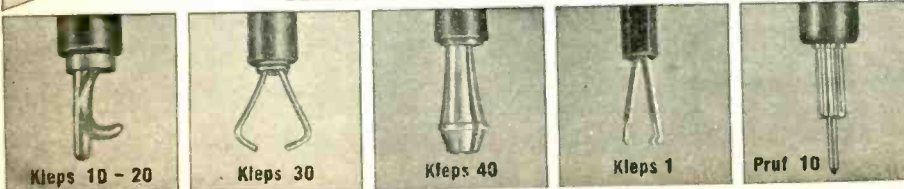
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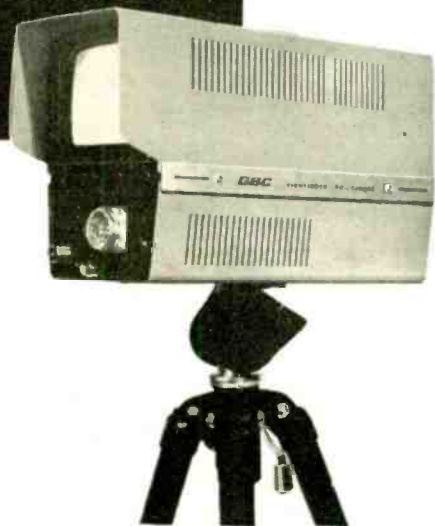
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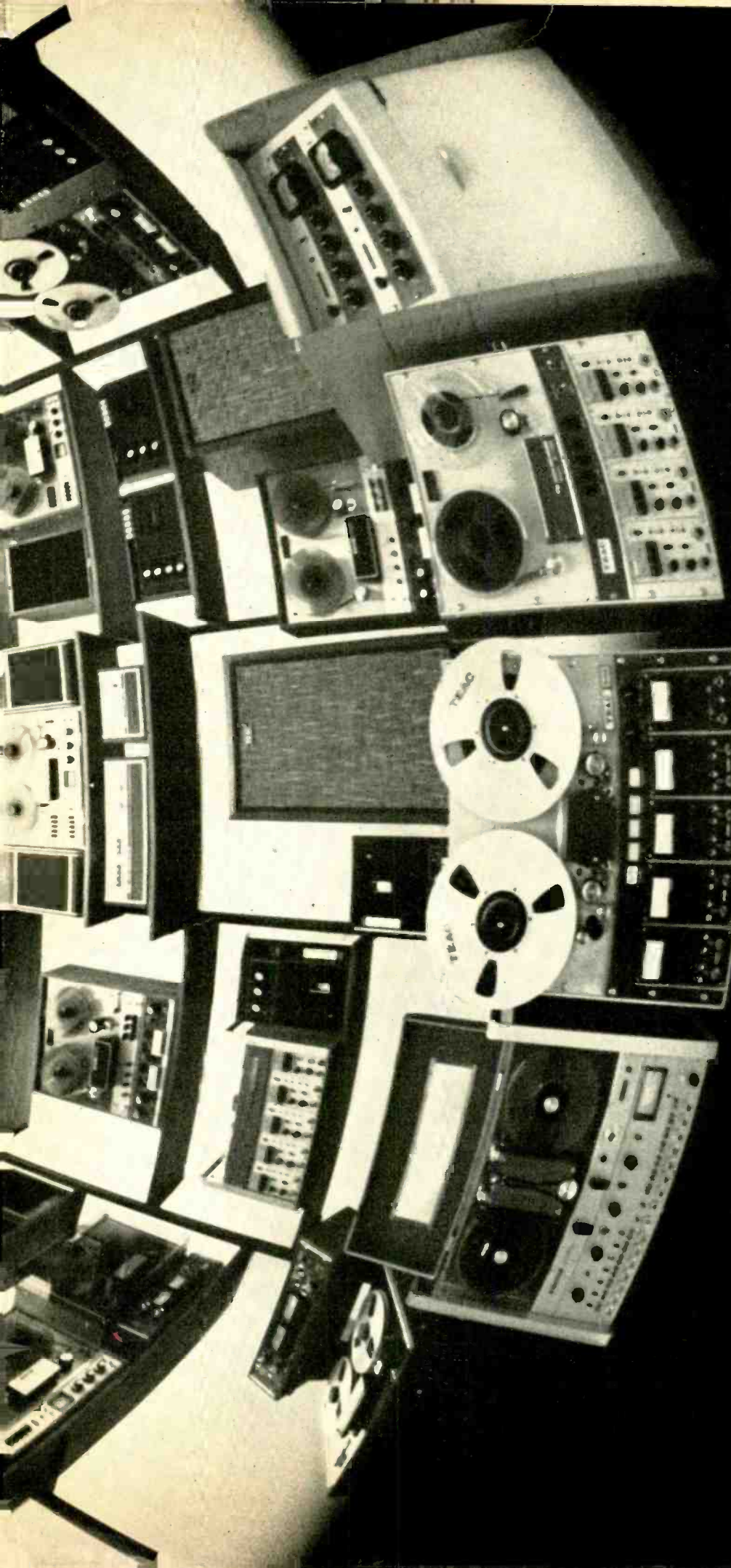
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Circle 15 on reader service card

Don't shoot till you see the tenths of a volt.

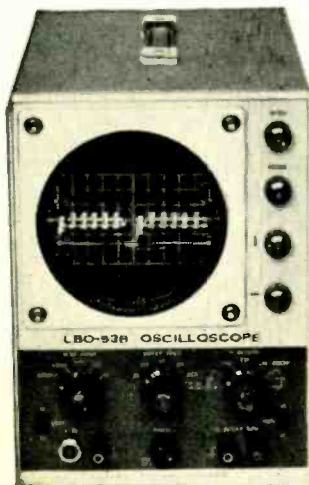
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Circle 16 on reader service card

CORRESPONDENCE

(continued from page 18)

CAR TUNE-UPS

As the article stated, I tried to use a vtvm instead of a vom ("Tune Up With A Vom," April 1970 RADIO-ELECTRONICS). However, my vtvm read in the minus 0 range and, reversing the polarity, in the plus infinity range. What creates this problem?

ROBERT KILL

Worthington, Ohio

The article on tune-up with a vom was interesting, but I feel that aligning the time marks by pulling on the fan is unsafe. If the blades are bent and then restraightened, metal fatigue could cause the blade to break when the engine is operated.

ROBERT DAVIS

Ottawa, Kan.

My guess is that the vtvm user has a bad diode. The blocking diode, preferably of several hundred PIV, should be checked for a large resistance difference between forward and reverse values. The vtvm volt function will show the polarity of the coil terminal that connects to the points, and the vtvm and diode can be connected in series properly.

When you use the vtvm you will need to modify your calculations for the dwell setting. Vtvm's read down-scale from infinity rather than upscale as vom's do. Make the calculation for the voltage corresponding to your dwell setting exactly as described on page 24 in the article.

I got a reading of 80 on a 120 volt scale for a 30° dwell. For a vtvm subtract the voltage you get from the full-scale value and use that as the mark to set by. For example, $120 - 80 = 40$. With the values given in the article, you'd use the 40V position on the 120V scale with a vtvm, not 80V.

The fan-blade problem is correct in some cases, but most blades I've encountered are heavy enough to pull the engine into place for timing. A blade would be dangerous if restraightened, but common sense should guide you here.

LOUIS E. FRENZEL, JR.

Silver Spring, Md.

HELP!

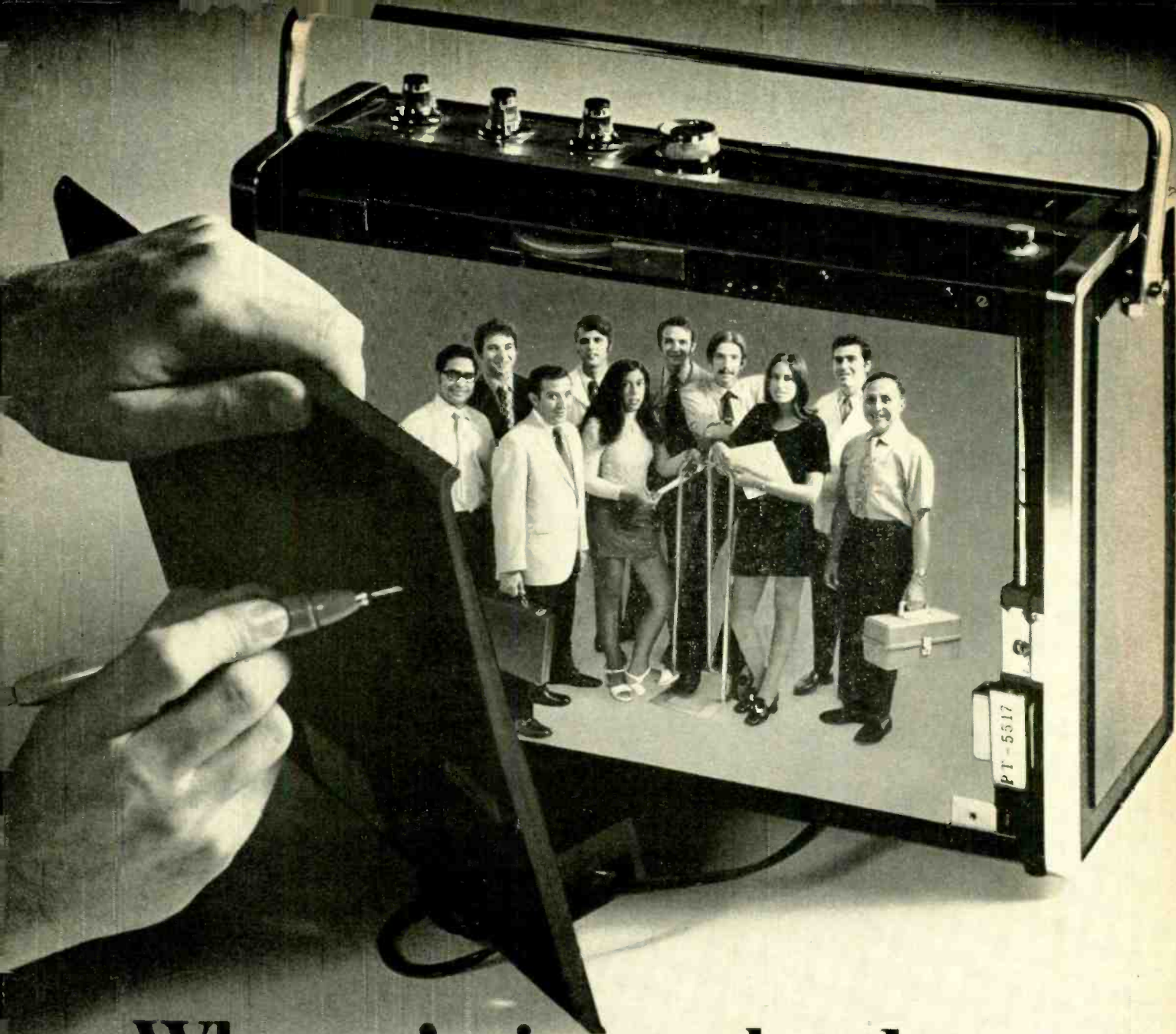
I have endeavored, thus far without success, to locate a copy of the instruction manual and schematic drawing for a Smith-Lawrence, Inc. Model 851, Potentiometric DC Voltmeter.

I would appreciate any assistance your readers can provide in making this manual and schematic available.

RAYMOND HADLEY
207 Royal Palm Drive
Ft. Lauderdale, Fla.

R-E

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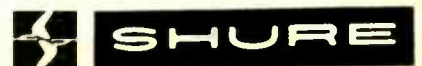
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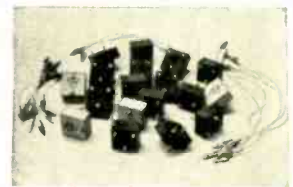
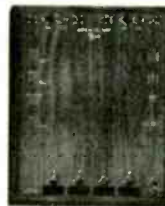
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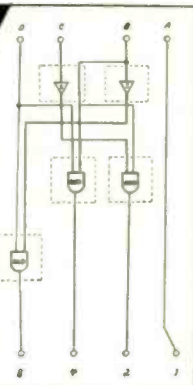
"SINK THE BATTLESHIP"

2	1	7
3	8	4
6	9	5

This is an entertaining problem, easy to set up. You'll see it programmed on the board in the photograph, and a schematic diagram at right is shown at right. With the logic blocks in place and wired, you tell another person to select any number in the squares at left, but not to divulge this number to you. You then ask him the following questions.

- Is the number odd?
- Is the number in the first or third vertical row?
- Divide the number by four; is the remainder less than two?
- Is the square number greater than 3?

As he answers each question, you feed the answer into the appropriate inputs ("A", "B", "C", "D"). Each answer is instantly retaped and/or stored in the transistorized logic blocks. When all the answers have been fed in, you then read the computer output and it tells you exactly which of the numbers the other person had in mind. It never fails; it is infallible and precise, operating as it does on principles of pure logic. In the easy-to-read and easy-to-understand instruction manual which accompanies your Compulogical Tutor, you'll find hundreds of similar problems, and others more interesting and complex.



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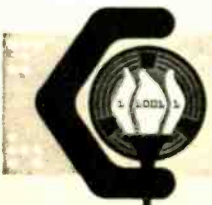
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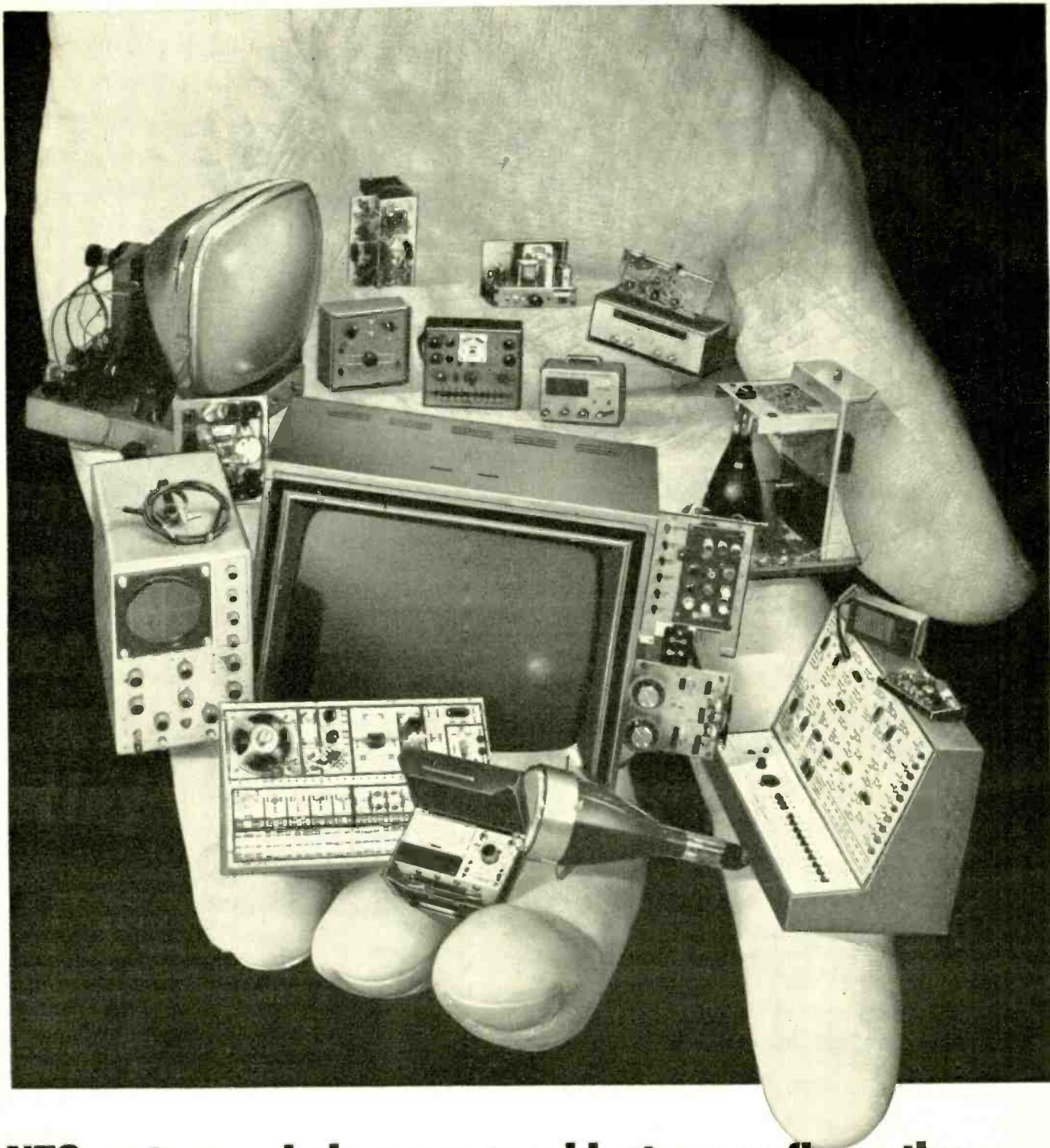
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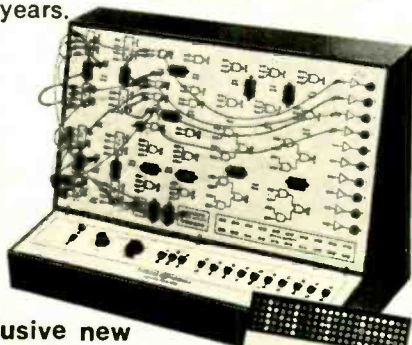


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A broadcast/recording monitor console in miniature.

The TA-2000 resembles professional sound consoles in more than its VU meters. In addition to the 20 jacks and seven input level controls provided on its rear panel for permanent connections to the rest of your hi-fi system, the TA-2000 boasts a professional patch board in miniature on its front.

Thus, you can feed the inputs from microphones, electric guitars, portable recorders or other signal sources into your system without moving the preamplifier or disturbing your normal system connections in the least. And a front-panel Line Out jack feeds signals for dubbing or other purposes into an external amp or tape recorder, with full control of tone and level from the front-panel controls and VU meters.

The tone correction and filtering facilities are also reminiscent of professional practice, allowing a total of 488 *precisely repeatable* response settings, including one in which all tone controls and filters are removed completely from the circuit.

The amplifier — no mere "black box"

A power amplifier can be considered simply as a "black box" with input and output connections, a power cord, and an on/off switch; and such an amplifier can perform as well (or poorly) as the next one. But in designing the TA-3200F Sony took pains to match the amplifier's facilities to the preamplifier's.

Thus to complement the TA-2000's two pairs of stereo outputs, the TA-3200F has two stereo pairs of inputs, selected by a switch on the front panel. Other front panel controls include independent input level controls for both channels, a speaker

selector switch, and a power limiter (in case your present speaker should lack the power handling capacity of the next one you intend to buy).

Circuitry unusual, performance more so

The single-ended, push-pull output circuitry of the TA-3200F amplifier is supplied with both positive and negative voltages (not just positive and "ground") from dual balanced power supplies. This system allows the amplifier to be coupled directly to the speakers with no intervening coupling capacitors to cause phase shift or low-end roll-off (A switch on the rear panel does let you limit the bass response below 30Hz if you should want to, otherwise, it extends all the way down to 10Hz.)

The individual stages within the amplifier are also directly coupled with a transformerless complementary-symmetry driver stage, and Darlington type capacitorless coupling between the voltage amplifier stages.

As a result, in part, of this unique approach, the TA-3200F produces 200 watts of continuous (RMS) power at 8 ohms, across the entire frequency range from 20 to 20,000 Hz; IHF Dynamic Power is rated at 320 watts into 8 ohms (and fully 500 watts into a 4-ohm load).

But more important by far is the quality of the sound; intermodulation and harmonic distortion levels are held to a mere 0.1% at full rated output, and 0.03% at the more likely listening level of one-half watt. The signal-to-noise ratio is an incredible 110dB. And the full damping factor of 170 is maintained down to the lowest, most critical frequencies (another advantage of the capacitorless output circuit).

The companion TA-2000 preamplifier also boasts vanishingly low distortion and a wide signal-to-noise ratio, but this is less unusual in a preamplifier of the TA-2000's quality (and price). What is unusual is the performance of the phono and tape head preamplifier circuits; for though they have sufficient sensitivity (0.06mV) for the lowest-output cartridges (even without accessory transformers), these preamplifier circuits are virtually immune to overload — even with input signals 80 times greater than normal.

Their sole vice: they are hardly inexpensive

Of course, at a price of \$329.50 (suggested list) for the TA-2000 preamplifier, and \$349.50 (suggested list) for the TA-3200F power amp, this system cannot be considered other than a luxury. But then, it was intended to be. For there are those to whom fulfillment of prosaic tasks is

unfilling. And among them are not only many of our customers, but also many of our engineers. Sony Corporation of America, 47-47 Van Dam Street, Long Island City, New York 11101.

SONY®





special report

ALARMS STOP BURGLARS

Photo courtesy Malory Distrbuto Products

by **BILL GRACIE**

Police statistics show that your chances of becoming a burglary victim are growing rapidly. Last year in Los Angeles more than 65,000 burglaries were attempted as compared to only 36,000 in 1960. Across the nation, burglary was the most frequently committed crime. A properly installed burglar alarm will help shift the odds in your favor. This article shows how these systems work and what they can do for you. Turn the page and see . . .

WHILE CRIME DOESN'T PAY, FEAR OF CRIME DOES. AND while no American likes to admit that he isn't safe in his own home, a great majority of home owners realize they aren't and are installing burglar alarm systems at a rapid rate.

Last year in Los Angeles, the city with the highest crime rate in the nation, more than 65,000 burglaries were attempted as opposed to only 36,000 in 1960. Of the 65,000, over 36,000 were committed against residences and apartments. Across the nation, burglary was the most frequently committed crime.

Statistics such as these are responsible for the boom in burglar or intruder systems. Among the vast array of available devices you're bound to find the one that fits both your pocketbook and needs. Though not all of the available systems will do an adequate job of protecting you, most of them will when properly installed.

What do you really want to protect?

Since the majority of home owners and apartment dwellers insure the contents of their homes against theft, it's really our lives that we wish to protect. This should be kept in mind when considering a system. Protecting life is the single most important function of an alarm system. Unfortunately, not all of the available devices are designed with this in mind. However, a combination of techniques will do the job well.

How sophisticated a system you need depends upon the size and design of your home and how much you are willing to spend. The variety of products on the market range from battery operated single-station units that attach to a door or window, to space-age ultrasonic and microwave detectors similar to the ones used in Vietnam to detect troop movements. They'll do everything from ringing a bell and lighting a light, to dialing the local police station. Costs vary just as widely, ranging from \$5.95 to over a thousand dollars.

Unless you want to protect valuable art treasures, jewels, or furs, a self-installed system that will adequately protect your family should run from \$100 to \$500 for a one-family house and from \$100 to \$200 for the average four-room apartment.

What you'll get for this amount of money is protection from the amateur or petty thief. Police officials claim that this type of thief scares off quickly when an alarm goes off or a light suddenly flashes on. And if this type of thief knows your home is protected, he may not even attempt to enter it. Why should he take a chance when your neighbor's home is unprotected.

Vandals will also be kept out of your home. And while vandals are not generally after monetary items, they can give you a bigger headache than thieves. A recent vandalism case reported in Long Island, New York resulted in over \$10,000 worth of damage to a house, performed over a period of one week while the owners were away on vacation.

Petty thievery and vandalism are often committed by neighborhood teenagers. So upon successful installation of your system, demonstrate it for the neighbors. Word will quickly spread that your house or apartment is protected.

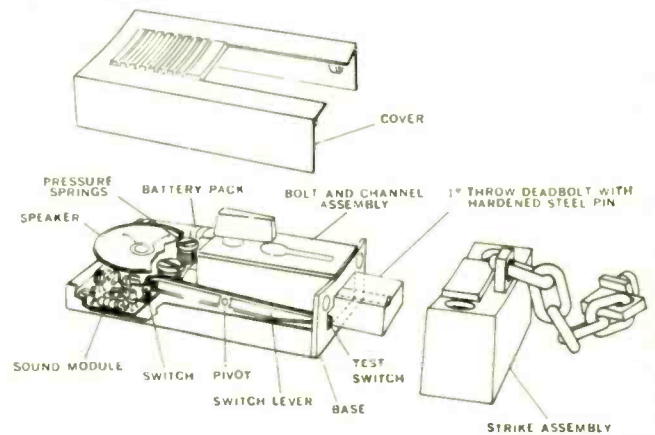
What's Available to do the Job?

Basically there are three different types of systems

that are suited to the homeowner. One or a combination of two techniques will afford you adequate protection no matter what the design of your home or apartment. The three systems include electromechanical or perimeter, photo-electric, and ultrasonic and rf systems.

Electromechanical or perimeter systems are the least expensive, but the most difficult to install. Photo-electric devices are seemingly easy to install and are comparable in price to perimeter systems, however, generally they will only cover a small area. Total protection generally necessitates several units, costing more money.

As might be expected, ultrasonic and rf units are the costliest, yet they can virtually be taken out of their box and plugged in for instant protection. So their high



INTERESTING PERIMETER ALARM is this special entry lock made by 3-M. Pressure on the door, forcing the chain or even attempting to pick the cylinder sets off the battery operated, electronic alarm.

cost is off-set by low installation time and costs. The system you select will largely depend upon how much you want to spend and the design of your home.

Fire protection is an added feature of most systems. Low cost sensors (\$5 to \$10) will convert your intruder system into an effective fire alarm system as well.

Perimeter systems

Though the most difficult to install, electromechanical or perimeter systems offer the greatest degree of protection and flexibility for the money. The reason they are sometimes referred to as perimeter systems is that a properly installed system will surround your home with security. In a complete system all doors and windows are fitted with magnetic or contact switches that trigger the alarm when one is broken. Unless someone saws a hole through your roof or cuts a hole in a sliding glass door, the average thief is detected by such a system.

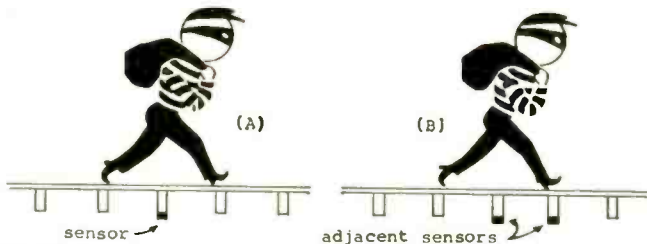
Since nearly every room is wired, adding fire protection to a perimeter system is easy. Heat or smoke sensors are simply wired into the same circuit the switches are on.

A basic perimeter system consists of a central control box or console (sometimes containing a key switch, batteries and horn) external bell or siren and window and door switches. An advantage of a perimeter system is that it may easily be expanded as desired or required. Some perimeter systems on the market today are single-station devices, that is, they

only protect a single door or window. However, in most cases, they can easily be expanded to afford complete protection.

Low cost perimeter systems are generally battery operated and contain an integral siren or horn, located in the console or control box. Before you purchase any system, ask to hear the alarm. Some units aren't loud enough to wake a sleeping baby.

More expensive and sophisticated perimeter alarms are both ac and dc operated. If there is a power failure or an intentionally cut power line, such units switch over to dc automatically and the system is still



PRESSURE SENSORS DETECT MOVEMENT on a floor. They are not very practical for home use and tend to find applications in bank vaults.

Diagram courtesy Systron Donner

on. Such a system is valuable if you live in an apartment building where your fuse box is located outside your apartment. Ac operation also allows you to hook up a hefty alarm and operate internal or outside lights.

An effective system of any kind should include both integral and external alarms or bells. An external signal will alert neighbors and police while you are away. The internal alarm tells the intruder he's been detected and that it's time to get out—fast! Incidentally, if you plan to go away, instruct a dependable neighbor on the operation of your system, in case it accidentally goes off. And if you add fire protection to your system, use a different-sound alarm so you can easily distinguish between a fire and an intruder. If your system does not allow you to hook up ac accessories, a simple relay set-up will do the trick.

The actuating key switch is generally located in the control box inside your home or apartment. A delay of from 10 to 30 seconds is built into many systems, allowing you ample time to enter your house and deactivate the system (a feature of some systems is an instant or delay switch for use when you are inside the house). This delay works in reverse when leaving. Activation upon closing of a door is another variation in some systems.

If for some reason you prefer to have the actuating keyswitch outside the house, optional externally mounted key panels are available with some systems. If you intend to eventually install a complete and comprehensive system, look for a manufacturer that offers a complete line of accessories. It will make add-ons much easier.

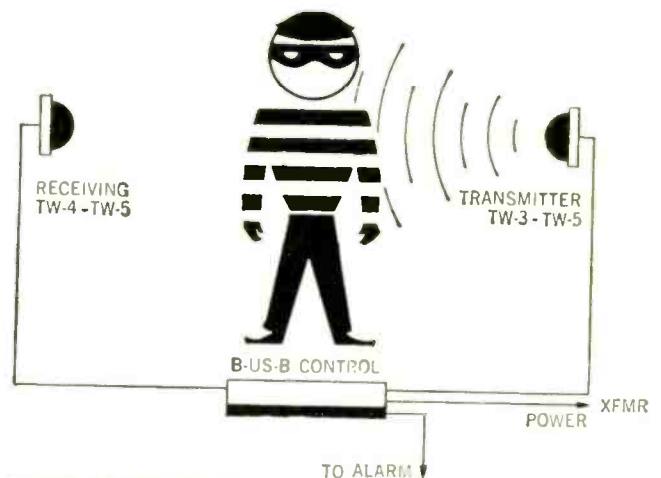
An important component in any system and especially easy to include in a perimeter system is a series of panic buttons, which upon activation, will sound the signalling device. For the greatest degree of protection, the panic button should sound an alarm even when the system is deactivated. Thus, should a housewife not have the system activated during the day when doors and windows are open, she can sound the alarm should

an unwelcome visitor show up. Place panic buttons throughout the house and especially in rooms where a direct view of the entrance foyer is not available.

Windows and doors can be protected by either of three basic means. Magnetic or reed actuated switches have become very popular because of their reliability and ease of installation. They are especially convenient for doors, and cost around \$3.50 each. For windows the lever or contact switch is used for its simplicity and low cost (about \$1.50). Break-away wire is the most economical "switch" but it can only be used where a particular window or door is virtually unused. Break away wire is very thin and can not be seen at night. The device you choose will depend upon each situation. More exotic and costly switches which are activated by vibration or pressure are available but they are prone to false alarms. Use them only where such protection is essential.

Naturally, the more you spend the more features you'll get. For instance, some of the deluxe systems have power indicators, battery life testers, etc. The one thing you should add to your system is fire protection. This is done simply by wiring into the same circuit, small thermostats or temperature sensors which activate when a room reaches 130 to 135°F. For warmer climates or rooms, 190° units are also available. Trouble spots like boiler rooms, kitchens and laundry rooms are likely places to install these units.

If you are planning to install a perimeter system in a new, yet to be constructed house, ask the builder if



WHEN INTRUDER MOVES INTO protected area he upsets the existing energy field setting off the alarm.

Sketch courtesy Systron-Donner

you can wire the skeleton before the plasterboard is installed. This will greatly reduce the amount of work you'll have to do in cramped attics and crawl spaces.

Photoelectric systems

Unlike perimeter systems, photo-electric protection devices are designed to cover a single room or area. To provide the breadth of protection afforded by perimeter systems a photo-electric set-up would require several sets of photo-electric pairs. In addition, the optical alignment of such systems is tricky, and must be doublechecked. There are instances though, where a single photo-electric system strategically placed, will cover the critical portions of a home and at a very nominal cost.

Photo-electric systems usually consist of a light source and photocell sensor. While older systems used easy-to-fool incandescent light sources, modern versions utilize infrared, ultra-violet and modulated light sources which the average thief will find very difficult to "jimmy". A few systems even use relatively new IR (infrared) light emitting diodes as light sources. These gallium-arsenide diodes emit IR in the range of 9100 angstroms and provide the reliability and low power drain of semiconductors.

All photo-electric systems operate on the same basic principle—when a beam is interrupted, the resistance of the photo-cell increases and triggers an alarm. IR, UV (ultraviolet) and modulated beams are used to make the devices tamper proof.

Burglars vs space age electronics

It's doubtful that any thief will find a way to get by the space age protection afforded by ultrasonic, radar and microwave intruder systems. Representing the high technology portion of the burglar business, these devices are difficult to fool, easy to install but relatively costly to buy. Prices range from \$75 to about \$500.

Each of the three systems operate on the same principle—the Doppler effect. Simply, the Doppler effect dictates that movement within an energy field of an rf or ultrasonic signal will cause a change in the phase, frequency and amplitude of the energy pattern.

Operation of a typical system is illustrated by the diagram shown at right. In this system, an oscillator generates a 20 to 40 KHz signal (depending upon the particular model) which is fed to a piezoelectric transmitter and subsequently radiated throughout a room. Reflected ultrasonic waves from moving and stationary objects are received by a second piezoelectric crystal. Since a moving object causes a change in the characteristics of the original signal, a simple detector/reference circuit detects the difference in the signal and triggers an alarm. Filters help reduce the possibility of false alarms caused by extraneous signals.

Solid state electronics has made it possible to package these systems in a single box, looking something like a small hi-fi speaker. Other systems require the use of detectors (which must be connected to the transmitter) located in various parts of a room. Except for these multi-part set-ups, all the wiring required for a single-part system are bell, light and panic button circuits similar to those used in perimeter or photo-electric systems. Some of these systems are small enough, and portable enough, to take along on a trip to guard your motel or hotel room.

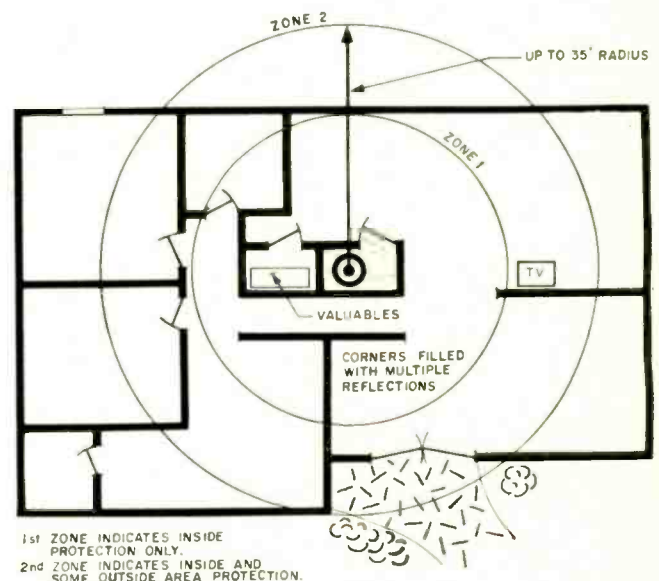
The basic difference between ultrasonic, radar and microwave systems lies in the frequency range within which they operate. Ultrasonic units operate at about 20 to 40 kHz; radar in the 450 to 500 MHz region, and microwave goes up in frequency where radar leaves off. These different frequency ranges bring about some inherent advantages and disadvantages for each.

Ultrasonic systems seem to be best suited for the average home. Their greatest advantage is that the radiated signal will not penetrate walls and is thus confined to a single room and is not disturbed by outside movement. However, so effective is the protection within a room that air turbulence caused by hot air or air condi-

tioning vents can cause a false alarm under some circumstances. Reducing the sensitivity setting will control this problem. However, to help overcome these unwanted problems, air turbulence compensators and sensitivity adjustments are provided. You'll have to find the right setting for your room.

Radar or higher rf intruder systems have their problems too. Their high frequencies and shorter wave lengths permit the signals to penetrate walls and glass panes, thus posing the problem of outside movements interfering with the system's operation. Also, commercial communications systems such as citizen's band and police radios might possibly trigger an alarm.

Microwave devices operating near and in the Gigahertz region are not at all disturbed by air turbulence or commercial communications systems. However, these units are generally more expensive than ultrasonic units and their radiating pattern is spherical in shape and therefore does not "fill" a room as ultrasonic units do.



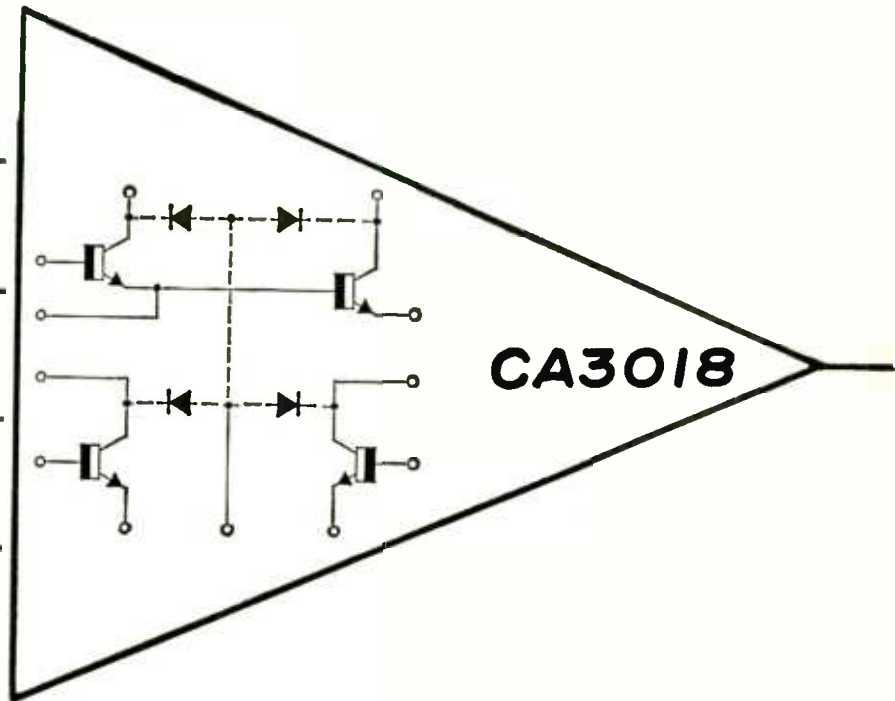
RF ALARM PENETRATES WALLS, so field of coverage must be carefully restricted or movement outside the house might trigger a false alarm. Sketch courtesy Systron-Donner.

Coverage accorded by all three techniques is about the same ranging from 15'X15' areas for the lower cost units, to 35' diameter areas for the higher power ones. Temperature and humidity changes will cause some variation in their range. Long corridors are relatively easy to protect with some units able to secure a 150-foot long space.

Good protection is afforded by a combination of perimeter and ultrasonic units. There are several ultrasonic units available that have provisions for adding perimeter system for supplementary protection. But no system is better than the installation job. Carefully consider possible entry points and make sure they are covered. Remember, an alarm positioned where it can't detect an intruder is worthless.

For a complete listing of burglar alarm manufacturers turn to pages 83, 94, 95, and 97. If you come up with a manufacturer we haven't listed, let us know. Write Burglar Alarms, c/o RADIO-ELECTRONICS, 200 Park Ave South, New York, N.Y. 10003. **R-E**

30 new IC circuits you can use



by R. M. MARSTON

Part II: Versatile four-transistor IC operates from dc to vhf. Build ultra-sensitive switches, ice alarms, smoke alarms, simple robot brain, automatic car light controls, heater and air conditioner controls, and more.

Ultrasensitive switches

The CA3018 is ideally suited for ultrasensitive switching applications. It can be used as a temperature-operated switch triggered by thermal changes of less than 0.5°C. It can become a light-operated switch that senses changes in light level too small to be detected by the human eye, such as those caused by a puff of cigarette smoke at a range of several feet.

In this section, we present seven ultrasensitive switching applications. All of them use the same basic operating principle.

In the basic circuit of an emitter-coupled dc differential amplifier (Fig. 16-a) Q1 and Q2 share common-emitter resistor R5. Transistor Q1's base is set at 6 volts by R1-R2; Q2's base is variable about 6 volts via R3-R4.

Due to the basic emitter-follower actions of Q1 and Q2, the voltage at the top of R5 is about 0.65 volt less than the larger of the two base voltages. So, if the two bases are at identical potentials, Q1 and Q2 are equally forward-biased and conduct by equal amounts. Under this condition the circuit is balanced.

If, on the other hand, Q2's base is made more positive than that of Q1, Q2 becomes more heavily forward-biased, and turns on harder than Q1. Simultaneously, R5's voltage rises in proportion to Q2's base voltage, reducing the forward bias of Q1.

ward bias of Q1.

Transistor Q1 becomes less forward-biased under this condition. Thus, the emitter coupling of the two transistors causes Q2 to conduct more and Q1 to conduct less when Q2 base voltage is increased. The reverse occurs when Q2's base voltage is lower than Q1's.

Note that the degree of conduction of the two transistors is dictated mainly by the difference between the two base

voltages, rather than by the absolute base potentials. Both Q1 and Q2 give a current and voltage gain between collector and base. If each transistor has a voltage gain of $\times 100$, a differential change of 10 mV on Q2's base causes a 1-volt change on Q2's collector.

Resistors R1 through R4 are in the form of a bridge, so the balance point of the circuit (at which Q1 and Q2's base potentials are equal) is determined solely by the relative values of these resistors and is independent of variations in supply voltage.

Since Q1 and Q2 are connected as a differential amplifier, thermal variations in the forward base-emitter voltages of the two transistors automatically cancel, particularly if Q1 and Q2 are part of an integrated circuit. The balance point of the circuit is thus independent of ambient temperature.

Finally, if R4 is adjusted so the circuit is near balance, Q2's collector voltage can be shifted 1 volt by differentially increasing Q2's base voltage by only (say) 10 mV. Since a standing bias of 6 volts is normally applied to Q2's base via voltage divider R3-R4, this shift in voltage is obtained by changing R4's value by less than 0.2%. The circuit balance is thus very sensitive to changes in the bias values of resistors R1 to R4, but is virtually independent of variations in actual supply voltage and ambient temperature. As a result, this circuit can be used as the basis of an exceptionally stable but sensitive resistor-operated electronic switch.

The basic differential circuit can be modified to act as a regenerative switch that drives a relay output (Fig. 16-b). Here, Q2's collector is direct-coupled to the base of pnp transistor Q3, and Q3 drives a relay load in its collector. Part of Q3's collector voltage is also fed back to Q2's base via R7 and the R4-R8 junction.

Suppose R4 is adjusted so that Q2

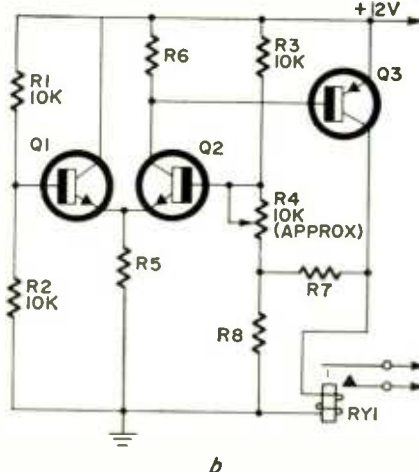
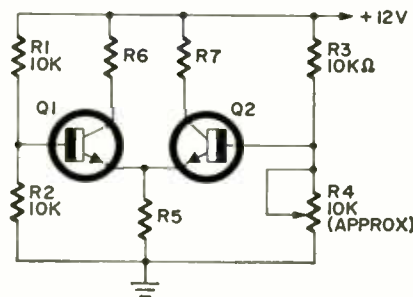
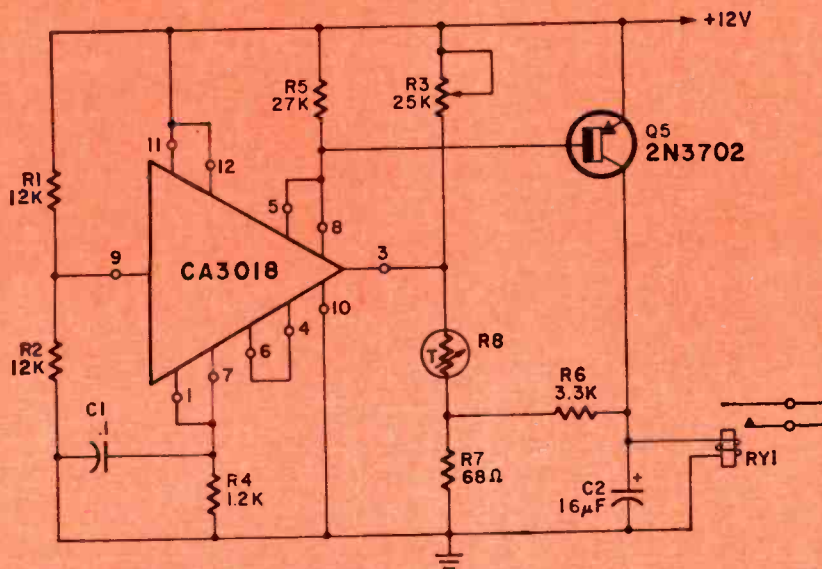
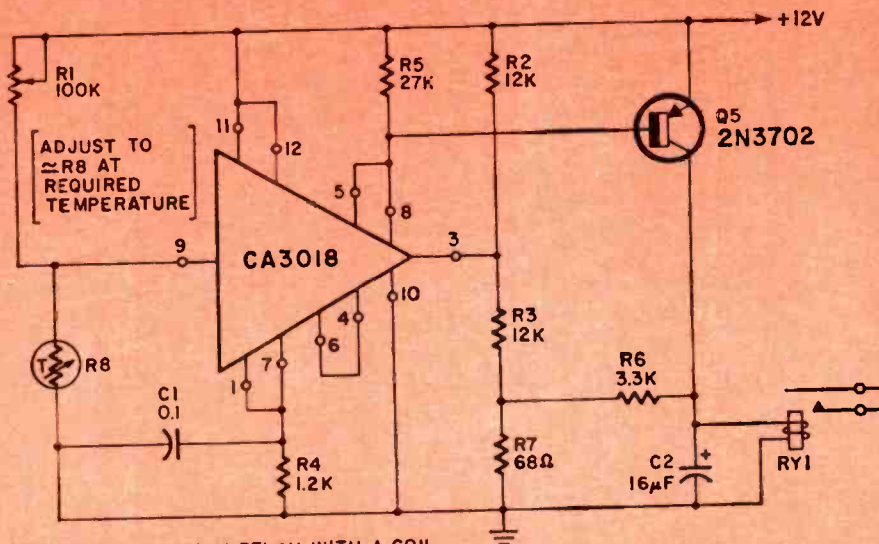


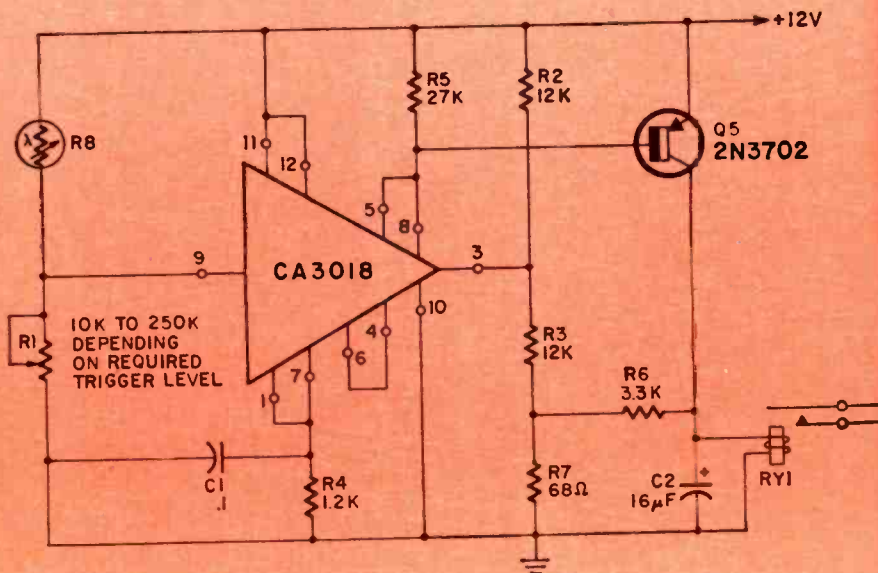
FIG. 16-a—EMITTER-COUPLED dc-differential amplifier. (b) same circuit modified to act as a regenerative switch that drives a relay.



NOTE: RY1 = ANY 12-V RELAY WITH A COIL RESISTANCE GREATER THAN 120Ω.
R8 = CARBON ROD THERMISTOR, MULLARD VA1066S, RCA KD2108, OR SIMILAR.



NOTE: RY1 = ANY 12-V RELAY WITH A COIL RESISTANCE GREATER THAN 120Ω.
R8 = CARBON ROD THERMISTOR, MULLARD VA1056S, MULLARD VA1067S, RCA KD2109, OR SIMILAR.



NOTE: RY1 = ANY 12-V RELAY WITH A COIL RESISTANCE GREATER THAN 120Ω.

is cut off. Now Q2 passes zero drive current into Q3's base and Q3 and the relay are off. Now adjust R4 so Q2 is biased on enough to apply a very small drive current to Q3's base. As Q3 starts to turn on, its collector moves towards the positive supply voltage.

A fraction of this rising voltage is fed back to the R4-R8 junction via R7, causing Q2's base voltage to increase. This increased base voltage turns Q2 and Q3 on even harder. A regenerative action takes place and Q3 is driven rapidly to saturation, and the relay goes on.

By selecting the value of R8, the triggering backlash of the circuit can be varied to suit individual needs—if R8 is given a very small value, the circuit can be adjusted so the relay turns on when R4 is 10,000 ohms, but does not turn off again until R4 is reduced to (say) 9900 ohms, giving a backlash of 100 ohms.

The overall action of the circuit of Fig. 16-b is fairly simple. R1 through R4 are wired in the form of a bridge which is balanced when R4 equals 10,000 ohms. If R4 is (say) 0.5% or more below 10,000 ohms, the bridge goes out of balance in such a way that Q2 is turned off, so Q3 and the relay are off also.

If, on the other hand, R4's value increases to within less than 0.5% of 10,000 ohms or is made greater than 10,000 ohms, the bridge either balances or goes out of balance and turns Q2 on. Either way, a regenerative action is set up between Q2 and Q3, and the relay is driven sharply on.

The circuit drives the relay on whenever R4's value reaches a pre-selected magnitude. This value can be set with an accuracy of a fraction of a percent. By replacing R4 with a thermistor or LDR (light dependent resistor), the circuit can be used as an ultrasensitive temperature- or light-operated electronic switch.

Fig. 17 shows how to use the CA3018 as a high-performance ice-warning indicator. Transistors Q3-Q4 (within the IC) are Super-Alpha-connected and used as one half of the differential stage (with input to pin 9). Transistors Q2-Q1, also Super-Alpha-connected, are used as the other half, with input to pin 3. Capacitor C1 prevents high-frequency instability of the differential stage, and C2 insures that supply transients do not cause erratic triggering of the regenerative section of the design.

In action, the relay goes on as soon as the thermistor temperature falls below its preset value. The triggering temperature can be varied over quite a wide range via R3.

When the unit is used as an ice-warning indicator, R3 is set so the relay goes on as soon as R8 falls below 0°C. The thermal backlash of the circuit is 0.5°C.—if the relay goes on when the temperature falls to 0°C, it goes off again when the temperature rises to 0.5°C.

FIG. 17 (top)—ICE-WARNING INDICATOR. Relay goes on as temperature falls.

FIG. 18 (center)—ULTRASENSITIVE light-operated switch is modification of Fig. 17.

FIG. 19 (bottom)—LIGHT-OPERATED switch turns on when it gets dark.

$\pm 0.5^\circ\text{C}$. Backlash can be increased, if required, by increasing R7's resistance.

This circuit is very valuable as an ice-warning indicator for a car. The thermistor is mounted on the front underside of the vehicle, about 10 inches above the road surface.

Alternatively, the unit can be used at home to turn on a heater when room temperature falls below a preset value.

The circuit can be modified for use as an excess-temperature indicator (Fig. 18). Now the relay goes on as soon as R8's temperature exceeds its preset value.

The circuit can be made to operate at any temperature between about 30°C to 110°C , using the thermistor types shown. Thermal backlash is 1°C at an operating temperature of 100°C . Other operating ranges require different thermistor types. Their resistive values must be between 2000 and 80,000 ohms at the selected trigger temperature.

This circuit can be used to give an indication of engine, gear-box, or brake-drum overheating in a car. It can automatically turn on air conditioning equipment, etc., when temperatures exceed a preset value.

Fig. 19 shows how the circuit can be made to work as an ultrasensitive light-operated switch, which turns the relay on as soon as light level falls below a preset value. The action can be modified so the relay goes on when the light rises above a preset value, by transposing R8 and R1.

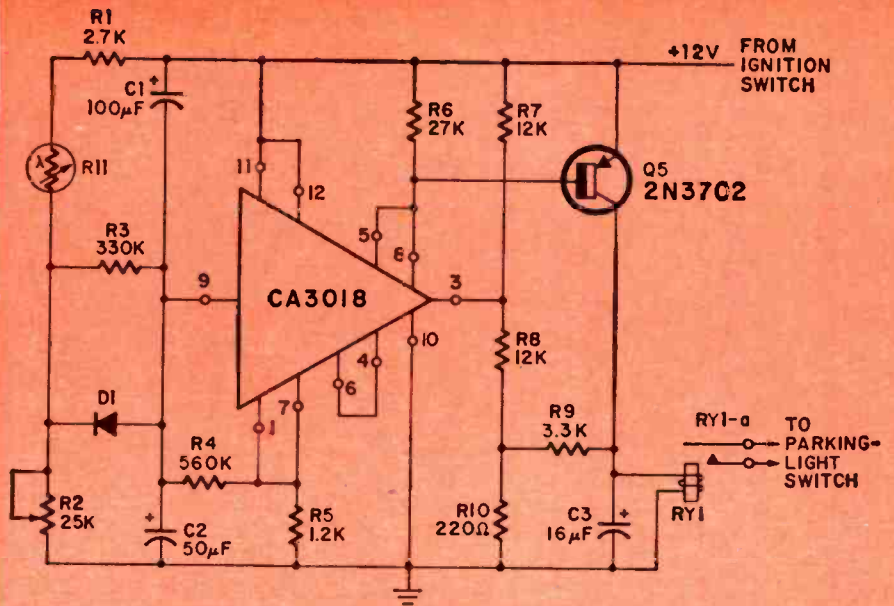
Circuit sensitivity is so high that the relay can be operated by changes in light level caused by a puff of cigarette smoke at a range of several feet (changes too small to be detected by the human eye). As a result the unit makes a good ultrasensitive detector of smoke, fog, mist, etc. When used in applications of this type, the brilliance of the source illumination must be stabilized.

The circuit can be modified for use as a lighting slave (Fig. 20), which operates a car's parking lights automatically. It turns them on when the car is driven at night, and keeps them off when the car is used in daylight. It turns lights off automatically when the car is parked at night.

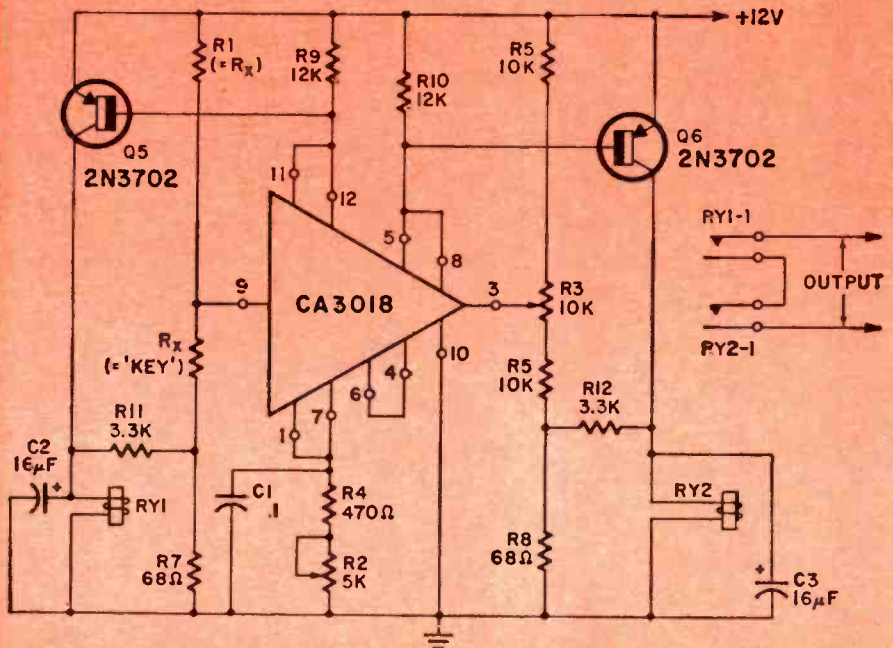
Power is applied to the unit via the car's ignition switch, and the relay contacts are wired in parallel with the vehicle's parking-light switch. R1-R11-R2-R3-D1-C1-C2 and R4 are wired as a dual-time-constant light-level integrating network to control the variable input voltage to the differential stage. The regenerative section of the unit has a moderately large value of backlash (via R10).

Because of the integrating network (when the car is in use) the relay is energized when the light level on R11 falls briefly below a value preset by R2. But the relay releases again only if the light fails to fall below that level again

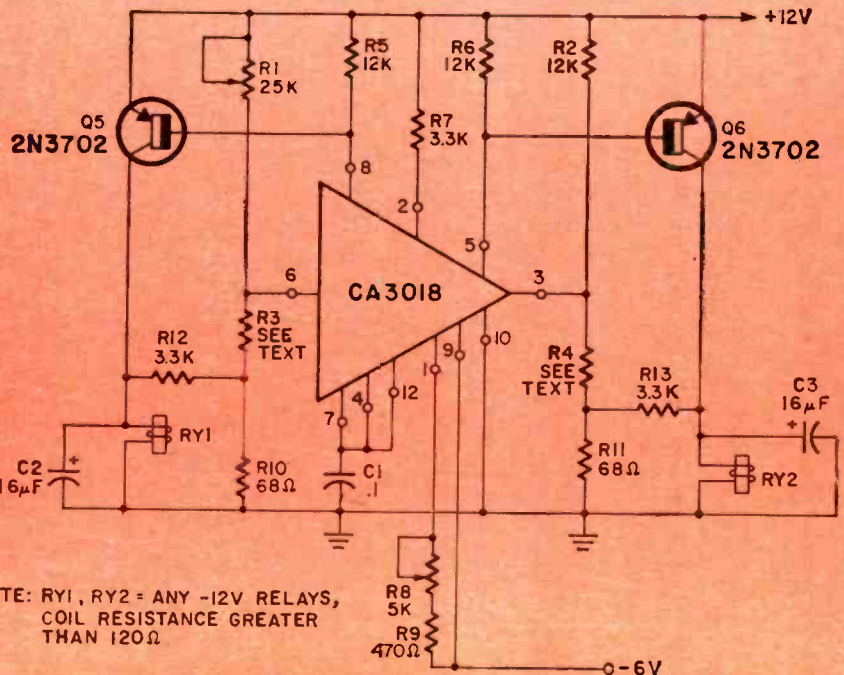
FIG. 20 (top)—TURN CAR LIGHTS ON automatically with this simple IC circuit.
FIG. 21 (center)—ELECTRONIC LOCK uses variable resistor as its adjustable key.
FIG. 22 (bottom)—DIFFERENTIAL BALANCE DETECTOR can be used as a robot brain.



NOTE: D1 = GENERAL PURPOSE SILICON DIODE.
 RY1 = ANY 12-V RELAY WITH A COIL RESISTANCE GREATER THAN 120Ω .



NOTES: R_x ('KEY') = R1 = ANY VALUE IN THE RANGE 100Ω TO 100K .
 ALL RESISTORS ARE 5% OR BETTER HI-STABILITY TYPES.
 RY1 AND RY2 = 12-V RELAYS, COIL RESISTANCE GREATER THAN 120Ω



NOTE: RY1, RY2 = ANY 12V RELAYS, COIL RESISTANCE GREATER THAN 120Ω .

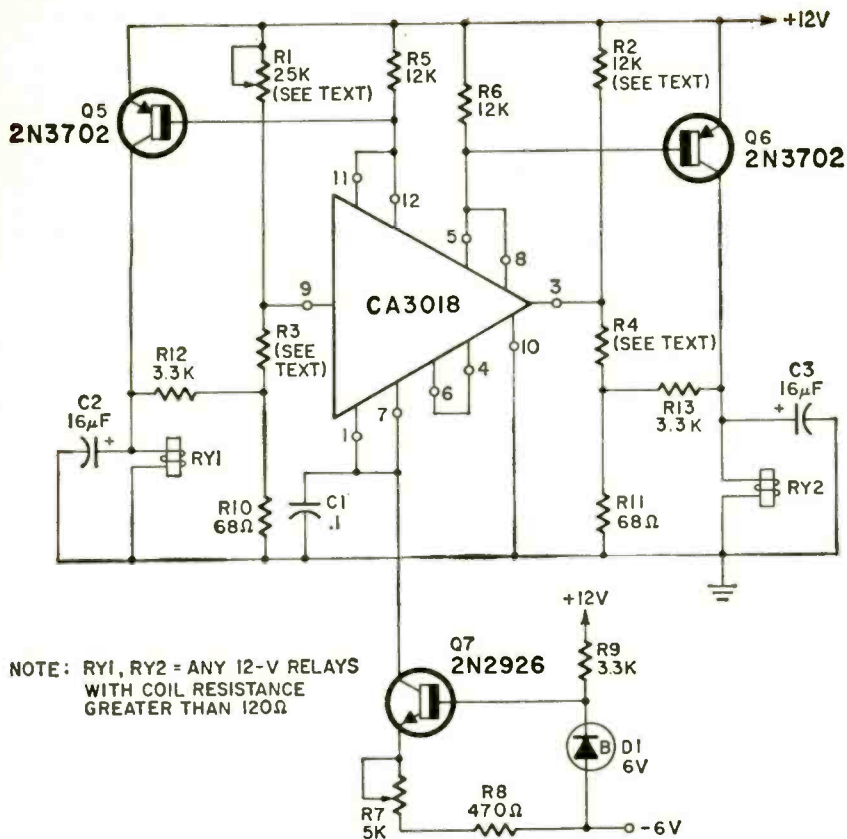


FIG. 23—HIGH-IMPEDANCE version of the differential balance detector of Fig. 22. Q7 and Zener diode D1 are used as a constant current source in this circuit.

within the following 30 sec. or so. The circuit is unaffected by sudden increases in light level, such as lightning flashes, etc. R2 is set so once the relay is on, it only turns off again when the light level reaching R11 rises to normal dusk or dawn values, and remains at those values for at least 30 seconds.

When the car is used at night, the relay and the lights turn on automatically in response to the low illumination on R11, and are unaffected by sudden transient increases in light level caused by street lights and lightning flashes. When the car is parked, power to the unit is cut off by the car ignition switch, so the lights go off automatically.

When the car is used in daylight, the lights are normally held off via R11. They will turn on automatically, however, if you drive under a bridge or through tunnel in which the light is below the preset level. They turn off again about 30 sec. after the vehicle emerges from the dark area.

Finally, when the vehicle is used on the threshold of dusk, the car lights turn on as soon as R11 illumination falls momentarily to a value perceptibly below that of the mean dusk level (when a shadow caused by a tree, a building or a passing vehicle falls on R11's face). The moderately high backlash of the circuit then insures that the lights do not turn off again until the illumination is held above the mean dusk level for at least 30 sec. (until the following morning). Thus, the cars lights operate without the slightest sign of flicker, even on the verge of dusk (or dawn).

When this unit is used in a vehicle, the LDR (light dependent resistor R11) should be mounted on the dash, with its

face looking into the car interior, so it is insensitive to bright external lights. When setting the trigger levels of the circuit via R2, shade the LDR face so the relay goes on, and then carefully trim R2 so that the relay just turns off again under dusk or dawn illumination conditions. Because of the long time constants of the unit's integrator network, at least 2 min. should be allowed to elapse between each trial adjustment of R2.

The CA3018 can be connected as a precision resistor-operated electronic lock too (Fig. 21). The circuit uses two relays and has a regenerative switching stage wired to each side of the differential amplifier. The two sets of relay contacts are wired in series, so there is an output only when both relays are switched on simultaneously. Resistor R_x is the "key" to the lock, and can have any value between 100 and 100,000 ohms. Resistors R1 and R_x must have nearly equal values, R3 balances small differences between R1 and R_x .

The two relays go on simultaneously only if the key (R_x) is the correct value. The tolerance of the circuit can be adjusted between about $\pm 0\%$ and $\pm 5\%$ via R2. If a wrong key is used, only one of the two relays operates—if R_x is too large, RY1 operates; if R_x is too small, RY2 is energized.

If this circuit is to be used as a practical electronic lock, the output should be taken via an additional 5- or 10-sec. delay relay, so the lock cannot be fiddled open with a variable resistor as the key.

The circuit can be modified for use as a differential balance detector (Fig. 22) which drives RY1 and RY2 on si-

multaneously only if R3 and R4 are equal. Balance is independent of the absolute values of R3 and R4 between 470 and 20,000 ohms.

This independence is obtained by using Q3 and Q4 as a constant-current generator in the tail of the Q1-Q2 differential amplifier stage, so voltage gain is independent of pin 3's and pin 6's bias voltages.

This particular circuit is intended for use as a "brain" in a simple robot animal. Simply replace R3 and R4 with a pair of matched thermistors or LDR's. In the LDR application, the LDR faces point toward a white line drawn on the ground, and the relay contacts control track-driving motors that move the animal forward. Relay RY1 operates the left-hand track, and RY2 the right-hand track.

When the animal is squarely over the white line, both LDR's are illuminated equally and both motors operate simultaneously to drive the animal forward. If the animal moves to the left of the line, the right-hand LDR is illuminated more than the left-hand one, RY2 drops out, and the left-hand track automatically moves the animal to the right until it is squarely over the white line again.

Similarly, if the animal moves to the right of the line, RY1 drops out, and the right-hand motor drives the animal back over the line. Thus, the animal automatically follows the path of the white line.

If you use thermistors, the animal can be made to automatically move toward or away from sources of heat.

Circuit sensitivity is very high, but can be controlled to a limited degree by R8. Sensitivity can be greatly reduced, if required, by wiring a fixed resistor (between 10 and 470 ohms) in series with the emitter of each of the differential transistors, Q1 and Q2.

Finally, in Fig. 23, there is a high-impedance version of the differential balance detector; here R3 and R4 can have any value between 1000 and 200,000 ohms. Resistor R1 must be $2 \times R3$, and R2 must be equal to R4.

This circuit is similar to that of Fig. 22, except that the differential stage is Super-Alpha-connected, and Q7 and Zener diode D1 are used as a constant-current source. (continued on page 69)



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Sound I.F. and
Quadrature Detector





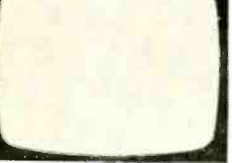

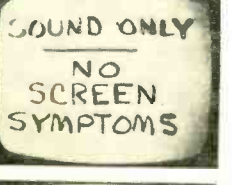

With i.f. feedback coil

Kwik-Fix™ picture and waveform charts

by Forest H. Belt & Associates*

SCREEN SYMPTOMS AS GUIDES

WHERE TO CHECK FIRST

SYMPTOM PIC	DESCRIPTION	VOLTAGE	WAVEFORM	PART
	No sync, accompanied by no sound	—	—	T1, if in output circuit of a sync-and-sound-i.f. amplifier stage
	Sound very weak; buzz and noise with volume control up	—	—	C1 open R6 high C6 shorted T1 open C7 open C8 leaky C9 open R3 open
	No sound	—	—	C2 open R6 open C4 shorted R9 open C10 shorted L1 open C10 leaky C11 open R5 open R5 high
	Sound very weak; may be distorted	—	—	R2 open R9 low C10 leaky
	No video; no sound	—	—	T1, if in output of video detector or video amp stage
	Sound critical at fine tuning; volume about normal	—	—	C3 open C5 open C6 open C8 open
	Sound critical at fine tuning; volume weak	—	—	C4 open
	Sound critical at fine tuning; volume above normal	—	—	C4 leaky

DC VOLTAGES AS GUIDES

change	to zero	very low	low	slightly low	slightly high	high
V1-cathode, pin-1 Normal 0.7 V	C3 shorted C4 shorted R3 open	C4 leaky R3 high	C3 shorted C4 leaky R2 low R3 high			R2 open R2 high R3 low
V1-grid, pin-2 Normal 0V	C4 shorted C4 leaky *R3 open *R3 high					
V1-plate, pin-3 Normal 75V	C4 shorted R3 open	C4 leaky R3 high	C3 shorted C4 leaky R2 low R3 high			R2 open R2 high R3 low
V2-cathode, pin-2 Normal 3.5 V	C7 shorted R5 shorted	R6 open	C8 leaky R6 high L1 open	C8 open		R5 open
V2-plate, pin-5 Normal 120 V	C10 shorted R9 open	C10 leaky L2 open	C6 shorted C7 shorted C8 open C10 leaky R5 low R5 shorted L2 shorted	C5 open R2 open R6 slightly high	C7 open C8 leaky R6 high	C7 open R5 open R6 open R9 low
V2-screen, pin-6 Normal 100 V	C8 shorted R6 open	C8 leaky R6 very high	C7 shorted C8 leaky R5 low R6 high		C8 open R6 low L2 open	R5 open
V2-quad-grid, pin-7 Normal 0V	*C7 shorted *C8 open *R5 low					

NOTES:

*Goes negative

Use this guide and the Waveforms Guide to help you pinpoint the faulty part.

All voltages are taken with sweep-generator signal applied (see Waveforms Guide notes.)

Measure each of the seven key voltages with a vtom.

For each, move across to the column that describes whatever change you find in that voltage.

Notice which parts might cause that change.

Finally, notice which parts are repeated in the combination of changes you find.

Test those parts individually for the fault described.

Signal and Control Effects

Station signal doesn't have much effect on dc voltages in these stages, except as already described for the bias at quad-grid-pin-7. It becomes more negative with stronger station signals.

Tuning the coils affects voltages only slightly.

Quick Troubleshooting

As is so common with stages that handle rf and i.f. signals, tuning is frequently the quickest way to spot a trouble. Before you try anything else, adjust the coils by the station-sound method.

Tune in the weakest TV station you get. Leave the antenna lead-in coupled loosely if necessary, but be sure you have a weak and noisy station signal.

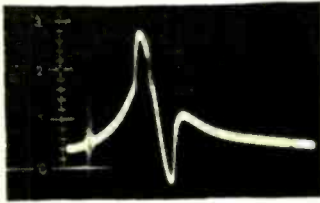
Peak T1 and L1 for the loudest sound with least noise; a compromise between noise and sound strength may be necessary at first. Then tune quadrature coil L2 for loudest sound with least distortion. Finally, go back over all three adjustments. Try to end up with loudest sound, least noise, and least distortion.

If any coil doesn't respond, suspect it or the components near it in the circuit. If the coils seem to tune ok, but still don't produce normal sound, make quick voltage checks at V1-plate-pin-3, V2-plate-pin-5, and V2-screen-pin-6. If those tests don't tell you anything, consult the Voltages Guide above and Waveforms Guide on the next two pages.

Sweep-response waveforms are used here because they are more informative of stage operation than any other kind of waveforms would be.

Turn page for waveform guide

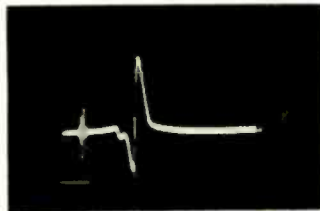
WAVEFORMS AS GUIDES



WF 1 Normal 0.3 V p-p

Taken at V1-grid-pin-2, this is response curve of the input tuned circuit. As indicated by the markers, this may not be centered at 4.5 MHz, because responses of tuned circuits farther along affect the overall response of the sound i.f. section. Amplitude may vary 30 to 50 percent. Curve shape is what's important (see Notes).

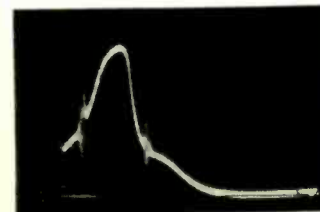
V p-p low	V p-p high	V p-p zero C1 open C2 open	 T1 mistuned up	 T1 mistuned down	 0.5 V p-p R2 open	 0.15 V p-p C4 open	 0.3 V p-p C5 open
			 0.3 V p-p C6 open	 0.25 V p-p R5 low	 0.3 V p-p R6 low	 0.2 V p-p C8 open	



WF 2 Normal 3.0 V p-p

Taken at V1-plate-pin-3, this curve includes reflected response of L1. Also includes effect of feedback from extra turns of L1, through C5 to grid-pin-2. Still is not the shape you might expect, but note that 4.5-MHz (right-hand) marker is near center of irregular curve. The 4.08-MHz marker (left) gives you some reference from which to gauge the width of these response curves.

V p-p low	V p-p high	V p-p zero C1 open C2 open C6 shorted R2 open R3 open L1 shorted	 T1 mistuned up	 T1 mistuned down	 3.0 V p-p L1 mistuned	 2.0 V p-p R2 low	 1.5 V p-p C3 open
			 2.0 V p-p R3 low	 0.5 V p-p R3 high	 0.5 V p-p C4 open	 0.7 V p-p C4 leaky	 0.4 V p-p C5 open
			 0.5 V p-p R5 low	 0.2 V p-p R5 open	 0.2 V p-p C7 open	 0.3 V p-p R6 high	 0.3 V p-p C8 open

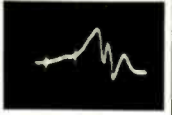

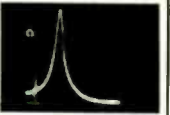
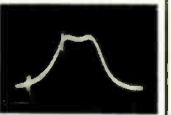
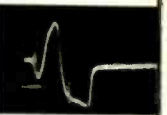


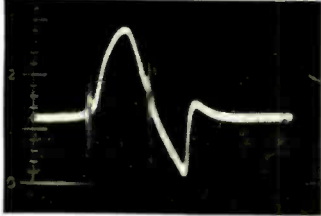
WF 3 Normal 3.5 V p-p

Taken at V2-grid-pin-1, this curve includes the response with all effects of L1. Shape is more like you might expect, combining the responses of the entire sound i.f. stage (V1). Curve is still slightly to left (down-frequency) of 4.5-MHz center, but that's normal. Don't let sweep width (on the generator) be too wide when you look at this waveform, or the curve will take an inaccurate shape.

V p-p low	V p-p high	V p-p zero C1 open C2 open C6 shorted C8 open R2 open R6 open R6 high L1 shorted	 L1 mistuned	 L1 mistuned	 3.5 V p-p R2 low	 0.5 V p-p C3 open	 1.0 V p-p C5 open
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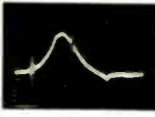


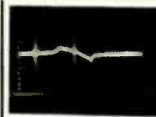


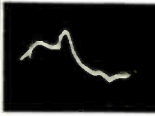
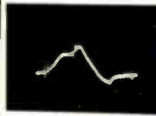
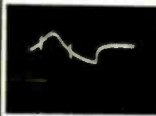

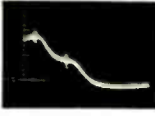

WAVEFORMS AS GUIDES

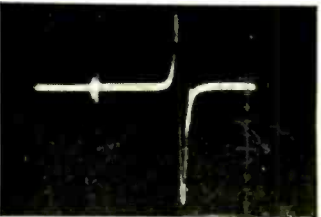
V p-p low	V p-p high	V p-p zero	 0.7 V p-p C6 open	 1.5 V p-p R4 open	 5 V p-p R5 open	 0.25 V p-p C7 open	 3.5 V p-p R6 low
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WF 4 Normal 3.0 V p-p




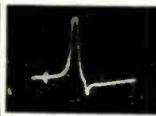





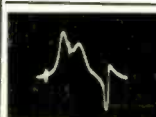



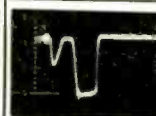


This is taken at quad-grid-pin-7. You can use a direct probe, but the shape shown here is with a demodulator probe. The 4.5-MHz marker is in the center of the slanted straight portion of the curve. The top left is rounded slightly, the bottom right is pointed. Patterns below show how mistuning L2 affects this waveform.

V p-p low	V p-p high	V p-p zero	 L1 mistuned	 L1 mistuned	 L2 mistuned	 L2 mistuned CW	 L2 mistuned CCW
		C1 open C2 open C4 open C6 shorted C7 open C7 shorted R2 open R3 open R5 open R6 open L1 open	 1.0 V p-p R3 high	 1.3 V p-p C5 open	 0.7 V p-p C6 open	 0.7 V p-p R5 high	 2.0 V p-p R5 low
			 0.5 V p-p R6 high	 2.0 V p-p C8 open	 1.5 V p-p C9 open		



WF 5 Normal 15 V p-p

Taken following the dc-blocking capacitor at V2-plate-pin-5, this is the S-curve response of the quadrature detector. As you can see from the waveform patterns below, the shape and centering of the S-curve depends on *all* the tuned circuits in both stages. The slanted straight portion of the curve is very steep; the marker spreads itself all up and down it, making the marker somewhat indistinct.

V p-p low	V p-p high	V p-p zero	 station sound not cut off	 L2 off CCW*	 L2 off CW*	 L1 off CCW*	 L1 off CW*
		C1 open C2 open C4 open C6 shorted C10 shorted C11 open R2 open R3 open R9 open L2 open	 T1 off CCW*	 T1 off CW*	 R3 high	 C5 open	 C6 open
		 R5 high	 R5 low	 3.0 V p-p C7 open	 2.5 V p-p R6 high	 C8 open	 C9 open

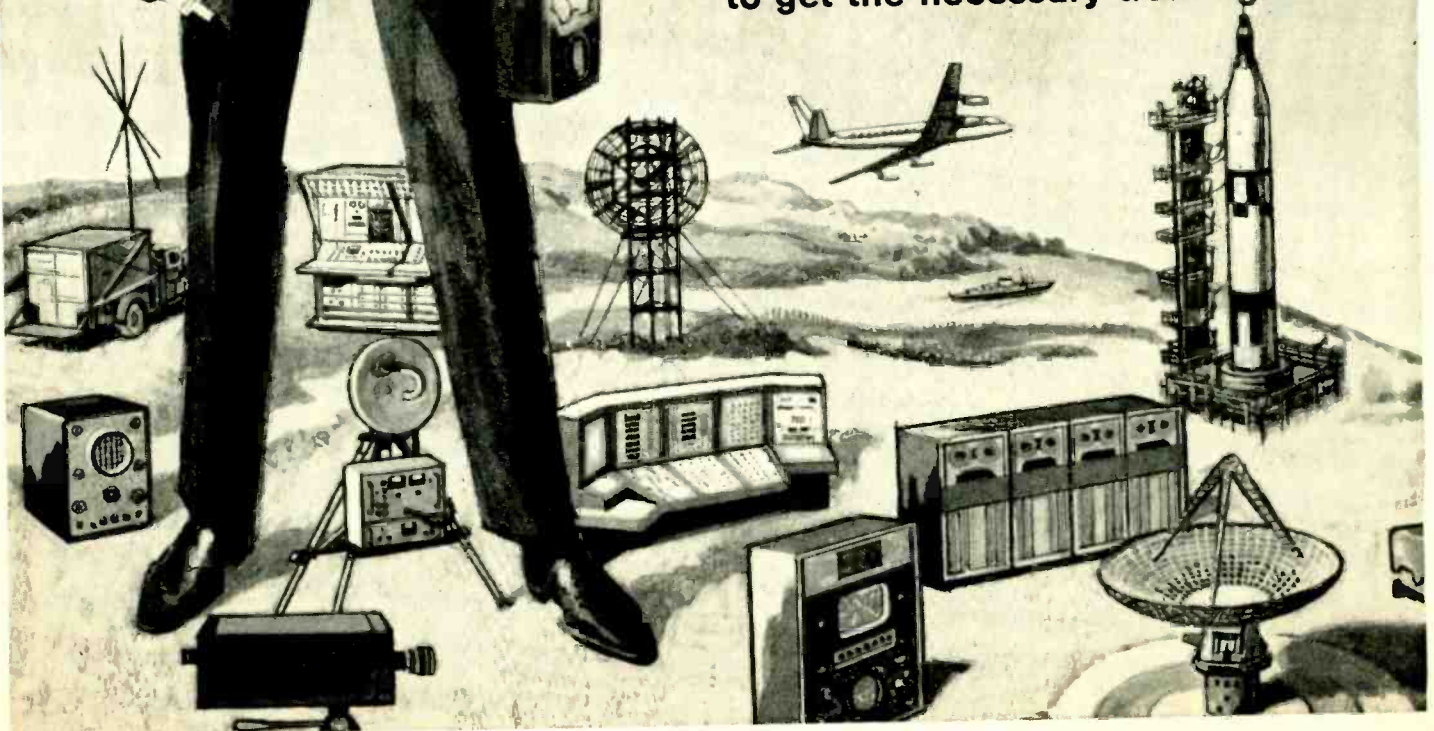
NOTES: *from bottom of coil
Use this guide and the Voltages Guide to help pin down fault possibilities.
Before using this guide, carefully go through preliminary alignment of the sound i.f. section by the station-signal method (see

Quick Troubleshooting in text).
Check all five key waveforms. Readjust the scope input to suit the p-p size of each waveform.
Note amplitude. If it's low or high, check the parts listed under those columns.

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off voltage, the oscillation stops. The oscillations terminate after approximately 1 second.

The D13T can also be used as a Schmitt trigger, as shown in Fig. 17. The circuit can be considered a pnp transistor operated in the common-base mode, driving a common collector npn. With the variable resistor shorted out only linear amplification occurs. As the resistor is increased, regeneration makes the circuit perform as a Schmitt trigger.

The complementary SCR's ability to drive a load in the gate makes anti-coincidence circuit (Fig. 18) easy to implement. This circuit provides ground only to the one load selected first. If other loads are selected, the gates are reverse biased on the associated CSCR's until the first load is released. Note that two loads cannot be driven at one time since any load that is driving deprives all others of triggering voltage. Any number of loads may

be operated with this circuit at supply voltages up to 50 volts and load currents to 50mA. The 1N4156 stabistor (low-voltage reference diode) and 10K resistor provide gate trigger voltage for the CSCR, while the 1K resistor provides holding current.

A temperature-compensated battery voltage monitor using the CSCR is illustrated in Fig. 19. If battery voltage is low, trigger voltage is not applied to the C13Y gate by the voltage divider and the lamp stays off.

High current SCR triggers

Figures 20, 21 and 22 are also thyristor applications for the D13T1. Generally, it is necessary to trigger a high-current SCR from a pulse transformer secondary.

The circuit in Fig. 20 uses the slow-rising waveform on the transformer secondary to charge the 0.1- μ F capacitor through the diode. The capacitor stops charging at the peak

of the pulse. As the voltage across the secondary collapses, the D13T1 gate becomes forward biased and the device turns on. This quickly discharges the capacitor into the gate of the SCR. Measurement with a 0.1- μ F capacitor and a 20-ohm load resistor (instead of SCR gate) showed rise-times of 50 nsecs to 1 amp of pulse current.

The circuit in Fig. 20 requires that the total energy needed to trigger the SCR be delivered by the pulse transformer. Sometimes this is not possible. The circuit in Fig. 21 derives its trigger energy from the power line thereby using the pulse transformer secondary to trigger the D13T1. The voltage across C2 is initially derived through C1 and R1. When the voltage across C2 reaches the Zener breakdown voltage, the rest of the line voltage is absorbed across C1. Dissipation in the Zener diode is minimized by the time constant of R1 and C1. When a trigger pulse is applied to the transformer secondary across the D13T1 gate, the D13T1 triggers thereby discharging C2 rapidly into the SCR.

This same trigger technique can be used to convert a wide, low-amplitude pulse from a dc circuit into a sharp trigger pulse for ordinary SCR's. Figure 22 shows the arrange-

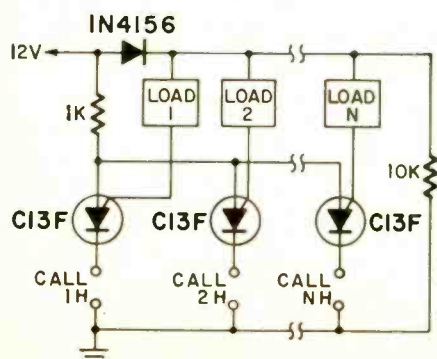
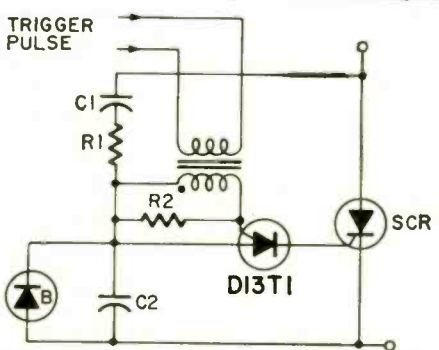
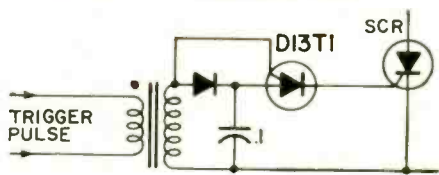
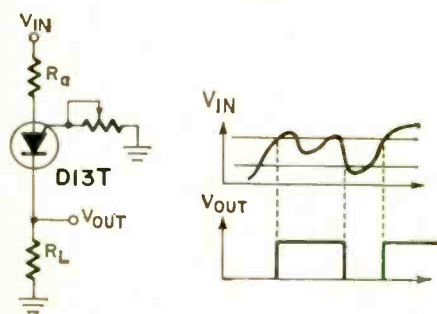
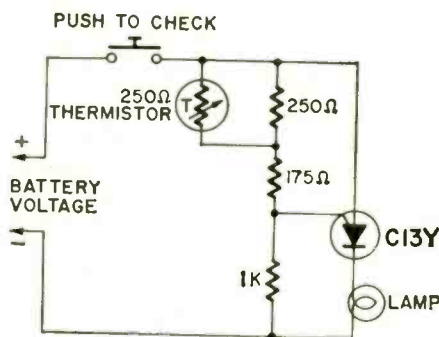
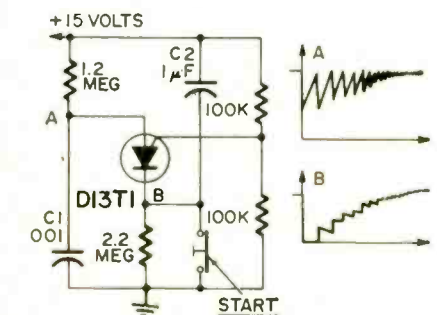


FIG. 16 (top left)—CHIRP OSCILLATOR output may be taken from PUT anode or cathode. FIG. 17 (middle left)—SIMPLE SCHMITT TRIGGER is equivalent to common-base pnp transistor driving a common-collector npn. FIG. 18 (left)—ANTI-COINCIDENCE CIRCUIT drives only the load selected first.

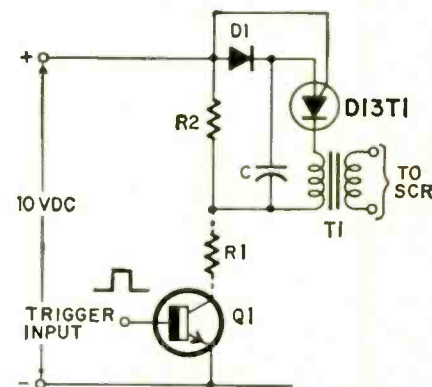


FIG. 19 (top left)—THREE-VOLT BATTERY TESTER. Temperature compensated circuit lights lamp when voltage is above about 2.5. FIG. 20 (middle left)—ISOLATED-GATE TRIGGER circuit for driving a high-current SCR. FIG. 21 (bottom left)—ISOLATED-GATE TRIGGER circuit for use when pulse transformer cannot supply total energy for SCR gate. FIG. 22 (above)—PULSE SHARPENER converts slow dc pulse into sharp a trigger for SCR's.

ment necessary. Here the dc circuit is considered as a 10-volt dc supply switched on via Q1 and R1. When Q1 is turned on, the C1, charges via diode D1. During this period, the D13T1 gate is back biased. When Q1 is switched off, the anode gate becomes forward biased via R2 and discharges C through the pulse transformer primary very rapidly. A circuit of this sort could be the interface between low-power integrated circuits such as the PA424 zero-voltage switch and high power SCR's.

CAREERS in ELECTRONICS

blueprint to your future



by **GEORGE G. GARMUS**

Kenneth E. Strandberg, at the left is manager of Technical Operations for G-E's Field Engineering Department in Phoenix, Ariz. He was interviewed by the author for this article.

Computer maintenance is really big business. Your electronic background may be all you need to get in on the ground floor.

THE EXPLOSIVE GROWTH OF THE computer industry has created numerous career opportunities for those trained in electronics. A new spectrum of technical specialties has been spawned by the computer industry in the past decade. These new specialists range from logic designers, systems engineers, programmers and input-output designers to computer field engineers. Attractive salaries, fringe benefits and job security are a common denominator of these specialties.

One of the most important and interesting jobs in the computer industry is computer systems maintenance. The electronic specialists who maintain and troubleshoot computer systems in the field are usually called field or customer engineers. Some computer manufacturers call their maintenance specialists field service engineers or representatives. However, the important fact is that computer maintenance men are responsible for keeping complex, expensive computer systems "on the air."

RADIO-ELECTRONICS zeroed-in on the facts about a career in computer maintenance by arranging an interview with an expert who maintained one of the very first commercial digital computer systems and now heads up General Electric's commercial computer maintenance training operations, Kenneth Strandberg. Ken is manager of Technical Operations for G-E's Field Engineering Department headquartered in Phoenix, Arizona.

Maintenance is a business

The role of the computer main-

tenance specialist is extremely vital. A billion dollar corporation's business activities can come to a standstill when its computer system is "down." The maintenance engineer must be able to think and act coolly under pressure while he's trying to correct a computer system problem. He must have a solid understanding of electronics, computer theory, programming and a high mechanical aptitude. Since computer systems are extremely complex he must be able to think in a logical, analytical manner.

Computer maintenance is at least a \$100 million business, ranking on par with manufacturing and sales. Most of the giant computer vendors have recognized this by establishing their computer maintenance operations as separate divisions or departments. In some instances more profit is made from maintenance contracts than from actual sales or rentals of computer systems.

The maintenance of computer systems is truly a "Big Picture" job. The typical field engineer's activities involve electronic circuit analysis, electromechanical adjustments, computer programming and analysis, complex diagnostic systems analysis, customer relations, logistics and business administration. He may, during the course of a month's activities, deal with people in design engineering, manufacturing, sales, training, programming, product planning and several levels of management at the customer's installation.

Computer maintenance is an evolving profession. The growing complexity of computer hardware and software requires special skills and insights to

achieve greater operational efficiency and reliability. Maintenance engineers are now considered an integral of the computer systems design team. Engineering courses on systems maintainability are becoming an increasingly important part of an engineering curriculum. Maintainability principles were developed by maintenance engineers. Also, maintenance engineering is now recognized as a professional specialty by engineering and trade associations such as BEMA (Business Equipment Manufacturers Association).

Qualifications for computer maintenance

The typical computer field engineer has at least 2 years of formal electronics training or armed forces training. Some of the best qualified men have had experience in the military as fire control technicians or in the army as missile systems electronic technicians. This type of training involves exposure not only to complex electronic systems but to intricate electromechanical systems as well.

The computer man must be equally skilled in maintaining both electronic and mechanical equipment. High-speed printers, magnetic and paper tape readers, card processors and similar equipment combine electronic, electrical, mechanical and electromechanical elements in their operations.

The candidate must be thoroughly familiar with pulse circuitry theory and be able to use oscilloscopes and voltmeters. He should be able to analyze complex waveforms—quickly and accurately.

When a computer system "goes down" it can be due to a number of reasons—component malfunctions or program errors. Therefore the candidate, in addition to possessing skills in electronics and mechanical aptitudes, must also have a strong talent for logical and analytical thinking. Unless he has this combination of skills he wouldn't be able to isolate the source of a computer trouble. He might spend precious minutes or hours chasing a trouble in an electronic unit—when it was actually a goof in the computer program! Or he might get hung up making adjustments on one of the high-speed printers.

He should have a clear understanding of present electronic theory and be able to read and correctly interpret



schematic diagrams. Equally important is his ability to get along with people effectively. Working under pressure at a customer's facility requires tact, empathy and integrity. Blowing up under pressure could result in having a computer system thrown out!

The computer field engineer should be a "self-starter"—requiring a minimum of supervision. Ideally, he should have many of those talents required to run your own business. That is, he recognizes the importance of time and money in his sphere of activities. He knows how to schedule the use of his time effectively.

Screening and training

At G-E, candidates are screened to determine their electronics knowledge, mechanical aptitudes and customer relations potential. He must present a personal appearance that is above average because of his constant contact with all levels of customer personnel. The candidate must be able to effectively communicate with a wide range of people, in both oral and written communications. He must be in good health and be bondable. The job interview and screening is usually done by a field engineering manager in the city in which the candidate is located. The general policy is hire candidates from cities in which they will work.

The salary ranges from approximately \$125 to \$300 per week for 40 hours. An inexperienced new trainee generally starts at the low end of the salary range. There is additional compensation at a rate of 1.5 or 2 times straight pay for overtime in excess of the straight 40 hours. The company provides life and medical insurance, sick leave and paid vacations to all field engineer reps.

The objectives of the training program are to provide the field engineer with a thorough educational program covering: data processing equipment installation and servicing; customer relations and site management.

The student's training program begins the day he becomes employed and continues through his career.

The new employee's first phase of training is an orientation period. Initially, he is assigned to work with field engineering personnel in the area where

he will be assigned for a period of 3 to 4 weeks. During this time he is familiarized with the computer maintenance business, its operations, maintenance procedures and is acquainted with the men with whom he will work.

G-E has found this is one of the most practical approaches to training computer field engineers—"seeing it the way it is." In the past, many students became confused when they thought what they learned in school didn't seem to jibe with what they saw in the field. Another benefit of this approach is that it instills a higher degree of confidence in the new student in the formal training program.

The second phase of training begins when the student reports to the Phoenix training center. His training at this time is centered around basic mechanics, diagnostic troubleshooting and equipment maintenance procedures. He works on actual computer systems and performs adjustments on electromechanical assemblies.

He is taught basic programming, computer operating methods, and is familiarized with software and typical applications programs. It is extremely important that the field engineer be able to determine why a type of trouble is causing a computing error. Is it a bug in the program? An operator goof? An electronic component malfunction? A mechanical unit out of adjustment? These are the questions that come up when a field engineer is zeroing in on a computer trouble.

The length of the student's training program ranges from 3 to 25 weeks, depending on the specific equipment to which he is initially assigned. There are an average of 12 to 14 students in each class at Phoenix. The school has an international flavor with students attending from Japan, France, Italy, Canada, Australia, Sweden, Holland and Mexico.

Upon completion of his training, the field engineer is returned to his district to assume his duties of equipment maintenance, customer relations and site management.

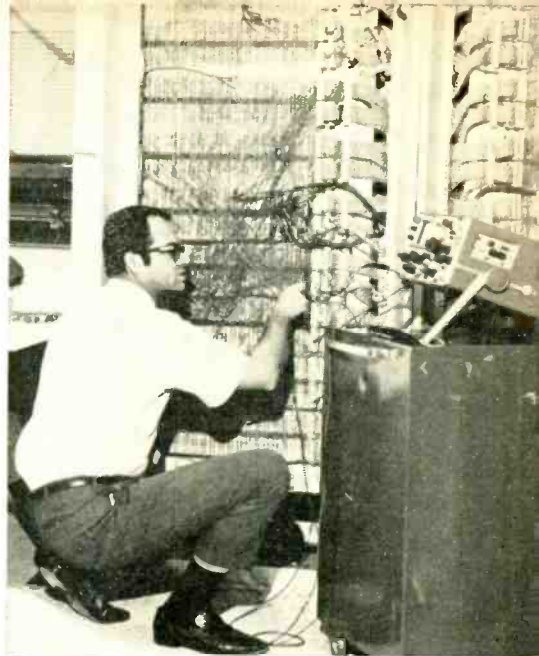
Continuous training and advancement

Computer technology is probably one of the fastest changing fields in engineering. Therefore, the field engineer must constantly keep his professional knowledge up to date. The experienced field engineer receives additional instruction through a comprehensive field training program. Field training is conducted at sites especially designed for the use of audio-visual training techniques. This

(turn page)

AT GE'S PHOENIX CENTER (top left) students study actual equipment to gain mastery of computer maintenance. SCOPE AND VOLTMETER (top right) combine with logical analysis to track down troubles. AUDIO AND VISUAL aids are used (2nd photo from top) to introduce actual equipment and techniques. TEAMWORK IS THE KEY (2nd from bottom) to maintaining the more complex equipment. GOOD MANUAL DEXTERITY (bottom) and a semiconductor background are prerequisites.

Photos by G-E.



portion of his training is presented on audio tape with instruction books and associated materials. He is assisted in this period by a qualified tutor (another field engineer) who has had special training and is experienced on the equipment the student is studying.

Students return to the Phoenix training center for additional training on an average of once every 2.3 years. Special audio-visual training classes are also conducted at Phoenix for experienced field engineers. The teaching system is essentially the same used in the field training programs.

Training programs in the form of formal classes, seminars and workshops are conducted regularly in the field. "In a business such as ours where the technology is constantly changing," says Strandberg, "continued training is a must to maintain an organization of competent professionals."

There is only one criterion for advancement and this is performance. The prime factor in selecting the field engineer for promotion is how he puts his technical knowledge to work. Basically, there are two paths of advancement open to the field engineer. That of the specialist and that of management.

The field engineer progresses to specialist because of demonstrated technical, customer relations and administrative abilities. The specialist may then join the headquarters staff in technical operations where he will assist in the design and development of new equipment from the maintainability standpoint, or he may enter training operations where he will contribute to the development of new training programs.

The field engineer may proceed along the path toward a management goal, seeking a position as a district, regional or departmental manager. The path is of his own choosing and how he travels along this path depends on his ability and initiative.

The plusses for a career in com-

puter maintenance are that it is a job that uses a wide range of an individual's talents and presents an everlasting challenge to the "problem solver" . . . who also likes to work with his hands. It ties in that rare combination of dealing with very complex equipment and customer relations.

The challenge and variety of a field engineer's duties are hard to beat. Dealing with programming, sales, engineering, logistics, administrative and customer relations problems makes a field engineer's life very interesting. The status of the field engineer ranks high in the computer field and is continually improving. Good salaries and job security are another plus for a career in field engineering.

Like any profession, field engineering has its negative aspects. Working on rotating shifts is probably the most negative aspect of computer maintenance work. However, some field personnel find it to their liking. Being able to go golfing or fishing while everyone is busy at the office can be a plus to many outdoor enthusiasts. Another consideration is periodic relocation for those moving up the promotion ladder.

Computer maintenance outlook

The phenomenal growth of field engineering has increased the demand for qualified men. In 1960, there were approximately 10,000 field engineers in commercial computer systems maintenance in the entire industry. The estimated total in 1970 is about 25,000! The industry estimates that there will be a rock-bottom minimum of 50,000 required in 1980. If anything, most past estimates were entirely too light.

With the growing complexity of computer software, the use of micro-miniaturization techniques and solid-state technology, the professional requirements for field engineers will be raised to higher levels of technical competence. In all likelihood, we'll soon wit-



COMPLEX COMPUTER CIRCUITS take trained troubleshooting skills to repair.

ness the emergence of maintenance engineering majors in many college engineering programs.

There is now a trend toward developing field engineering specialists in computer software and programming. The growing complexity of computer software and the increasing use of integrated circuits is making it more difficult for the field engineer to zero in on the specific source of computer errors. Therefore, we will probably see the emergence of a team approach to computer troubleshooting. The software specialist's prime function will be to determine the most likely source of the trouble—hardware or software through rigorous diagnostic analysis. Once the basic area of the trouble is determined, the maintenance crew will then use further diagnostic techniques to isolate and correct the cause of the malfunction. **R-E**

POWER TOOLS

This issue, starting on the facing page, is the newest section of your Radio-Electronics Reference Manual. It continues the section on tools and should be inserted following the end of section one.

Other material, including nomographs, special subject reports and important reference tables and charts, will be published in this same format in future issues. Each month we will present 8 new pages for your manual. To put your book into shape, cut out the next four pages (two sheets) following the dashed lines. Then arrange the pages in the proper order (follow the numbers at the bottom of each manual page). Next fasten them together at the top.

If you wish you can purchase a special hardcover binder to keep your Reference Manual pages together. It has a dark blue fabric cover and is gold stamped Radio-Electronics Reference Manual. The cost is \$1.25, postpaid. Order from N. Estrada, 17 Slate Lane, Central Islip, N. Y. 11722.

POWER TOOLS

by **THOMAS R. HASKETT**

Unless your interest or work in electronics is very limited, you probably use power tools. They enable you to drill holes, cut a chassis or speaker-mounting hole, and sand a speaker enclosure.

Electric drills are versatile

Almost indispensable for cutting holes in metal, the common electric drill consists of a motor in a pistol-grip housing (Fig. 1). The motor

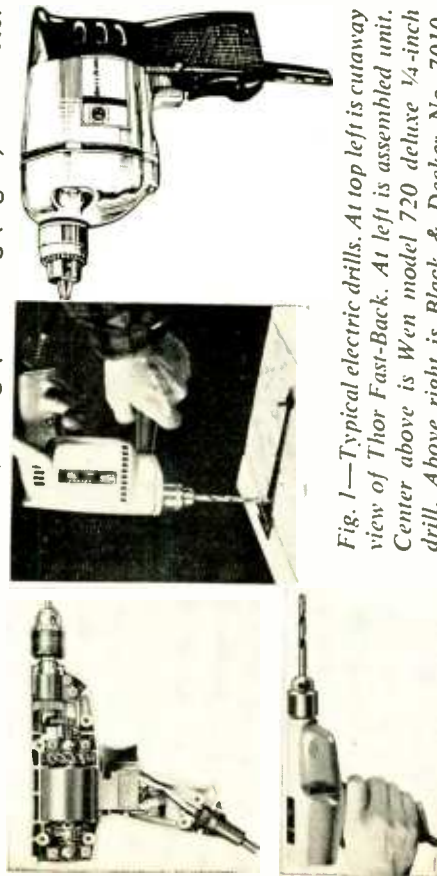


Fig. 1—Typical electric drills. At top left is cutaway view of Thor Fast-Back. At left is assembled unit. Center above is Wen model 720 deluxe 1/4-inch drill. Above right is Black & Decker No. 7010.

shaft is geared down to drive the chuck and the drill bit, which does the cutting. Most drills today have keyed chucks, which you turn with a key (furnished with the drill), causing the chuck to grip the bit. Some inexpensive drills use a knurled chuck which you twist with your hand. Such a chuck is not desirable.

Drills are rated by their chuck capacity and class of service. The three standard chuck sizes are 1/4", 3/8", and 1/2". Larger drills usually turn more slowly, since there is more work to be done. The average speed of a 1/4" drill is 2200 rpm; a 3/8" drill, 1000 rpm; and a 1/2" drill, 600 rpm. Of course, the larger drill has a larger horsepower rating, too (1/7 HP for 1/4" vs 2/5 HP for 1/2").

A standard or light-duty drill will do for general work, but a heavy-duty drill (in any of the above sizes) has a larger-horsepower motor and somewhat slower bit speed. Light-duty drills aren't intended for contin-

kit containing one each of the fractional sizes over a certain range. It's more useful to have several of the same often-used sizes, in case you break one when you can't easily obtain another.

Twist drills are made of one of two types of metal: **Carbon steel**, or **high-speed steel**. Carbon-steel (or **butter**) drills are useful only in soft materials such as wood or plastic, and don't work well in metal. High-speed twist drills, although more expensive, can be used in metal, wood, or plastic. They are marked HS or HSS.

Most drills come in standard lengths, and the length increases as the diameter of the drill does. Extra-long (**jobbers**) sizes are available, but are useful only when you have to cut a deep hole. Short shanks are better for close, precise work, as the drill is working closer in.

Normally, **straight-shank** drills are most useful, but to drill a hole larger than the chuck capacity of your drill, you may have to use a reduced-shank twist drill or bit. You don't get something for nothing, however; a reduced-shank type won't drill through as hard a material as a straight-shank model. Also, the tip tends to wobble and may cut a ragged hole. Also available are chuck adapters.

Many bits come with a quick-cut or split point; such a drill starts its own hole without walking around on the work.

The most convenient way of carrying and keeping bits is the kit mentioned previously. It has spaces for drills from perhaps 1/64" to 1/4" or 1/2". Cases are available for various size ranges.

To drill concrete or masonry, you must use a twist drill tipped with an extra-hard metal such as **carbide** or **Carbology**.

Notes on drilling

When you're drilling holes in a painted surface, cover it with masking tape or paper and lay out the hole centers with a pencil. Use a center punch first to indent the chassis; then drill through the paper and surface. Don't remove the paper until you've drilled all the holes. That way you're not as likely to mar the painted surface.

If you're drilling a hole in existing equipment, remove all glass ahead of time—vacuum tubes, meter movement, glass dial faces, etc. To prevent metal chips from clogging tube sockets and other receptacles, cover them with masking tape.

Drill holes larger than 1/4" in two or more steps. Drill a pilot hole first—say 1/8". Then enlarge it with the larger drill or reamer. For precision, cut larger holes (3/4" diameter and above) with chassis punches.

uous operation over long periods of time, and they can't be used for working thick, heavy metals without running the risk of overheating and damaging the motor. Heavy-duty drills can be used all day, and will do large jobs with ease, but they are a little bulky for small work and are more expensive than those rated for light duty.

For most chassis work, a 1/4" standard drill is adequate. A 1/2" drill is useful in construction and electrical work, and for industrial and broadcast electronics. It has a spade grip and crossbar handle for working with both hands. The 3/8" drill is something of a compromise between the other two, and usually has an auxiliary handle on the side of the case.

Some drills have a built-in **variable-speed control**, which lets you change speed to fit the hardness of the work. Some drills are **reversible**, a valuable feature when you have to back a drill out of a deep cut.

Most drills have universal motors, which will run on 117 Vac or dc. Another feature of interest is the position of the chuck key on the power cord. It's usually attached at the plug end of the cord. It should remind you to always pull the plug before changing bits. If the plug is in, a short at the trigger switch could start the motor and injure your hand.

At least one manufacturer—Black and Decker—offers a **cord-**



Fig. 2—Battery-powered drill is great for outdoors work. Unit shown here is made by Black & Decker. A rechargeable battery is used in this drill.

less or battery-powered 1/4" drill (Fig. 2). The batteries are rechargeable.

Drill bits aren't all alike

Also called **twist drills**, **drill bits** are available to cut holes from .0135" to 1". The most useful sizes for electronics work at No. 33 (to pass 4-40 screws), No. 28 (to pass 6-32 screws), No. 18 (to pass 8-32 screws), No. 11 (to pass 10-24 screws) and 1/4". Many people buy a

When drilling aluminum or brass, you can remove the burrs of these soft metals with a sharp knife. To remove burrs from a steel hole, use a countersink or a cold chisel and hammer.

Only sharp drill bits do a good job. It's not easy to sharpen bits, as the tip angle is critical. Kits are available to hold the bit at the correct angle for a grinder, but many people simply replace small bits when they become dull, and have the larger ones sharpened at a hardware store.

Drill accessories are numerous

You can extend the usefulness of any drill by acquiring special attachments designed for specific jobs. You can also buy a drill and several attachments and bits as a kit. A **wire wheel** or **cup brush** (Fig. 3) is handy for removing rust and scale, or for rough-polishing metal.

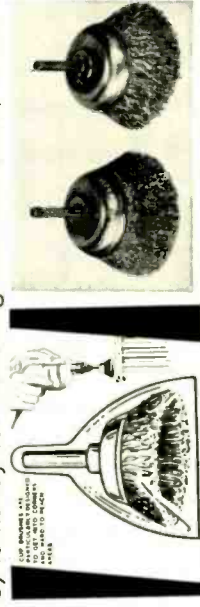


Fig. 3—Wire cup brushes are handy for removing rust and scale. Unit at left is made by Costal Abrasive. At right are brushes by Black & Decker.

Sandpaper discs and a **lambswool buffing wheel** are used with a **rubber disc** for backing. Both are handy for smoothing and polishing. A **grinding wheel** (Fig. 4) is made of abrasive stone and can be used to



Fig. 4—Grinding wheels are great for sharpening tools of all kinds. Costal Abrasive wheel is at left. Black & Decker unit is on the right. When using a grinding wheel in a hand drill, you must use a stand for the drill.

sharpen edged tools. The easiest way to perform this job is to clamp the drill body in an accessory **horizontal drill stand** (Fig. 5) and the stand in a vise. The **countersink** (Fig. 6) was mentioned above, and is used for putting a dimple in the work to take a flathead screw, as well as for cleaning the edges of a rough hole.



Fig. 5—Horizontal drill stands convert hand drill to a bench unit. Two units are shown above. At far left is Costal Abrasive stand. In the center is the Skill 45192 horizontal drill stand. Fig. 6—(above left) Costal countersinks.

To cut a hole larger than the largest drill bit you have, you can use a hole saw (Fig. 7), which comes in various sizes and uses serrated teeth to saw out the circle. It comes with an arbor and a pilot drill to start the hole. The same job can also be done with a fly or circle cutter (Fig. 8), which spins a cutting bit in a circle around the pilot drill. Unfortunately, a fly cutter is prone to wobble and cut a ragged hole.

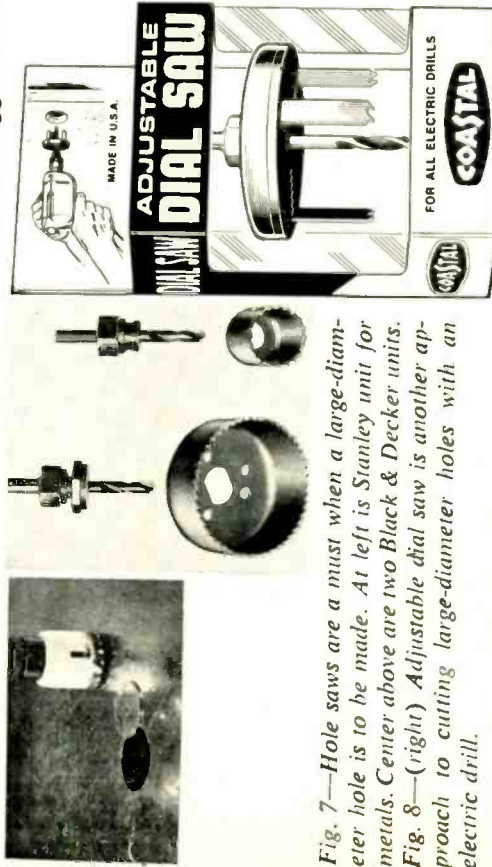


Fig. 7—Hole saws are a must when a large-diameter hole is to be made. At left is Stanley unit for metals. Center above are two Black & Decker units. Fig. 8—(right) Adjustable dial saw is another approach to cutting large-diameter holes with an electric drill.

Screwdriver bits (Fig. 9) are useful for extracting and installing tight screws when used with a special adapter.

One of the most recent additions to drill kits is the SCR speed

Less expensive saws run at a fixed speed, although some let you switch between two speeds. But it's sometimes useful to vary blade speed with work thickness, and the variable-speed saw does this job nicely.

Often a saber saw is used for cutting small curved areas and you have to turn the drill as you curve the cut. A few models are available which allow you to rotate the blade (Fig. 12) for easier curve cutting.



Fig. 12—Tilt while cutting circles makes the job a bit easier. Stanley 90486 speed-control saw is shown here.

Blades for your saber saw

A variety of blade types are made for saber saws. For scroll (curved) cutting of soft wood and plastic, thin, narrow blades are best. For straight cuts through plywood and hardwood, wider blades are used. Tougher blades with different teeth are recommended for hard, green, or wet wood. You have to use a hacksaw blade, of course, for metal. A knife blade is useful for cutting rubber, leather, tile, etc.

In any case, a coarse-toothed blade (less teeth per inch) is used for rough work, while a fine-toothed blade (more teeth per inch) is used for finish work. Most blades are about 3" long, but some are 4" or even 6", for deep work. In Fig. 13 you see the typical blade compared to the flush-cut blade. To use this type, you remove the guard at the front of the saw (it screws on) and insert the flush-cut blade in the saw. Then



Fig. 13—Saber-saw blades come in wide variety of sizes, shapes and types. There are blades for wood and metal, flush cut and standard. So pick the blade that meets your requirements. Units at left are made by Skill. Blades at right are P.A. blades.

control (Fig. 10), which gives you continuously variable speed, letting you match bit rpm to the hardness and thickness of the work. Some drill bodies are available with SCR speed control, while accessory units are made for drills not already equipped.

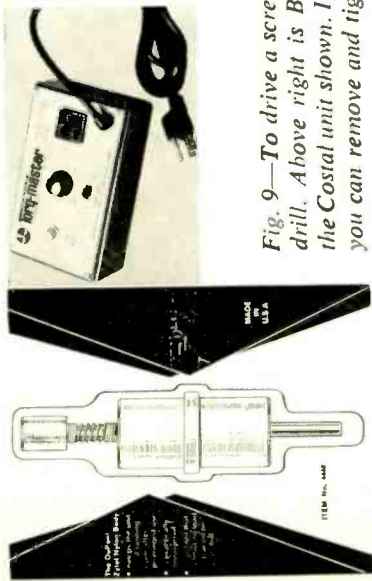


Fig. 10—Add on speed control turns most ac-powered tools into variable-speed units. Just insert control box between tool and wall outlet. Unit shown is made by Chicago Wheel.

Fig. 9—To drive a screwdriver bit with an electric drill. Above right is Black & Decker No. 7010, the Costal unit shown. If you have a reversible drill, you can remove and tighten screws with ease.

Electric saber saw

If you do any cabinet, carpentry, or chassis work, a very useful tool is the **saber saw** (Fig. 11). Also called a **portable jig saw**, it contains an electric motor which drives a reciprocating cam that moves a saw blade back and forth across a limited area. The best saws have universal motors which work on both ac and dc. The blade is in the front of the saw, on the bottom. Some models have a light above the blade so you can see where you're going.

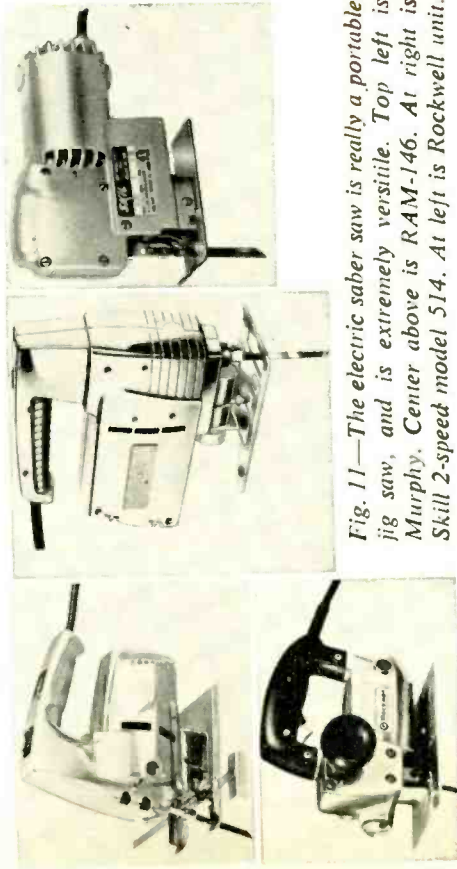


Fig. 11—The electric saber saw is really a portable jig saw, and is extremely versatile. Top left is Murphy. Center above is RAM-146. At right is Skill 2-speed model 514. At left is Rockwell unit.

you can cut up against a wall or corner.

Most saws run at a fixed speed, but **variable-speed** models are on the market. Of course, you can use most saber saws with accessory SCR speed controls.

Other power tools for you

For woodwork, cabinetry and carpentry, a **circular saw** (Fig. 14) is



Fig. 14—The circular saw is a powerful tool for all kinds of cabinetry and carpentry. It is not often used by a technician but should be a part of most power-tool kits. From left to right are the Ram Industrial Power Saw, the Rockwell Speedmatic, and the Thor E-732.

essential. It guides itself through a straight cut, can make cuts at a varying vertical angle, and cuts fast. Several blade types are used—**rip**, **crosscut**, **combination**—depending on whether you're making a rough or finish cut with or against the grain of the wood.

An **electric hand grinder** is useful for chassis and hobby work where you wish to do grinding, buffing, and shaping. The kit usually includes wheels and bits of several shapes and grits. You can deburr, carve, rout, and sand small areas with this tool.

A **bench grinder** is simply a bench-mounted grinding wheel about 6" in diameter for keeping edged tools sharp. Unless you have frequent need for this tool, you can probably get by with a grinding wheel and your hand drill.

It's a good idea to mark your tools for identification to prevent mistakes when lending or borrowing them. Using a sharp awl or pick is tedious; much easier is Dremel's Electric Engraver. This handy little tool is a small vibrator with a hard point. A variable control lets you adjust stroke length for the depth of cut you want and the hardness of the material. The tool cuts metal, ceramic, and glass.

More specialized tools exist—such as routers—but are seldom used in electronics.

R-E

New FM tuner

One IC replaces entire I.F. strip

- From rf to audio
- No tuned circuits
- Narrow bandpass
- High linearity
- Low distortion
- Phase-locked loop

An old circuit in a new IC may spell revolution in many radio circuits

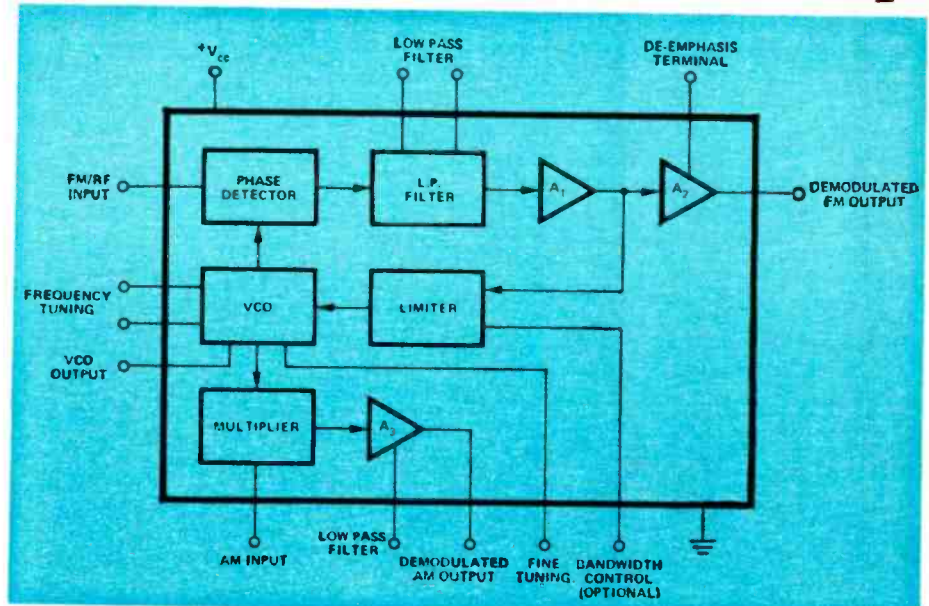
by **LARRY STECKLER**
MANAGING EDITOR

A new kind of IC, a phase-locked-loop (PLL) has been developed by Signetics and Motorola. Other units by other manufacturers are soon to follow. This new unit makes it possible to replace tuned circuits in i.f. strips, demodulators, and other inductor-dependent circuits with an IC arrangement that completely eliminates the need for coils or transformers.

Shown on our cover this month is an FM tuner built around one of the Signetics IC's. It consists of a conventional FM front end of the type used in stereo FM tuners, which is coupled to a PLL IC. This IC takes the 10.7-MHZ output from the front end and converts it to audio. While the quality of this circuit is not equal to the performance of a conventional stereo tuner, it opens the door to a new approach to FM.

The PLL will likely find its way into many types of electronic equipment such as radio receivers, appliances, TV sets, tape recorders. Commercial uses include signal generators, detecting and decoding of tones, etc.

One major advantage of PLL in FM tuners is that mistuning and drift are eliminated. The PLL locks onto a station and its built-in VCO (voltage-controlled oscillator) keeps it on station.



PLL in itself, is not a new idea. It has been used for several years to track satellites, stabilize the frequency of klystrons, and filter information out of noise. Unfortunately, in its discrete-component form the circuit is too costly and complicated to be used in consumer electronics.

As an IC, however, even the cost of this complex circuit is low, and we are likely to find the PLL appearing in many new kinds of equipment.

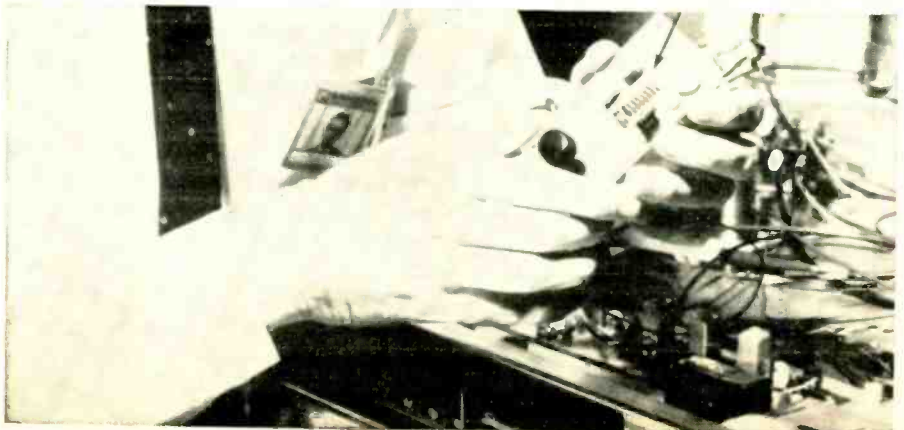
As shown in Fig. 1, the PLL is made up of four basic sections—the phase comparator, the low-pass filter, the amplifier, and a voltage controlled oscillator (VCO).

The phase comparator is a multiplier that mixes the input signal with the VCO signal to produce sum and difference frequencies

$f_s \pm f_o$. When the loop is in lock, the VCO duplicates the input frequency and the difference frequency ($f_s - f_o$) is zero. At this time, the output of the phase comparator contains a dc component.

The output of the comparator is filtered to remove the sum frequency component ($f_s + f_o$), while the remaining dc component is amplified and fed back to the VCO. It is this error voltage that drives the VCO keep f_s equal to f_o when the loop is in lock. The size of this error voltage is a direct function of the phase relationship between the VCO and the input signals.

If you want to see how error voltage is affected by the phase difference between the two signals, take a look at Fig. 2. Here the VCO signal is a square wave. When the loop is out of lock, and the differ-



ence between f_i and f_o is large, both sum and difference signals are filtered by the low-pass filter and the VCO runs at its preset frequency.

But as the difference between f_i and f_o approaches the cutoff frequency of the low-pass filter, a beat note forms which tends to force the VCO frequency toward the input frequency. When this dc component of the beat note grows large enough, the VCO frequency reaches the input frequency and the loop locks. Therefore, lock sensitivity is a function of the lowpass filter. Once in lock, the loop will track signals over its entire lock range.

PLL applications

Some of the expected applications for PLL IC's include FM demodulation in FM tuners, FM telemetry and telephony, frequency-shift keying (FSK), and store-casting music (SCA) detection.

Other potential uses are in signal conditioning where the PLL can be used to filter unwanted sidebands from the input signal. The useful PLL output in this kind of application is from the VCO output signal.

Frequency multiplication is another application. For low orders of multiplication with small FM deviations the loop can be simply locked to a harmonic of the input signal. For very wide FM deviations or higher multiplication, the loop can be broken between the VCO and the phase comparator, and a frequency divider inserted. Now, when the circuit is in lock, the output of

the divider is at the same frequency as the input signal and the VCO runs at a multiple of this frequency, thus forming a precise frequency multiplier which tracks the input signal over a wide range.

Some circuits to try

Now let's take a look at some of the practical ways we can expect to use a PLL right now. In each of the four examples that follow, a Signetics SE/NE565 IC is used.

FM demodulation

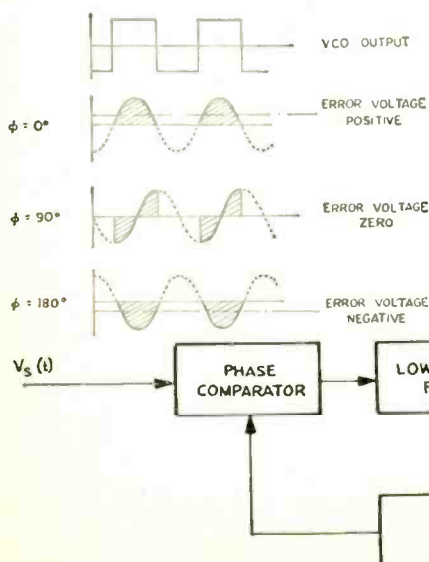
The 565IC phase Locked Loop is a general-purpose circuit designed for high-linearity FM demodulation. During lock, the average dc level of the phase comparator output signal is directly proportional to the frequency of the input signal. As the input frequency shifts, it is this output signal which causes the VCO to shift its frequency to match that of the input. Consequently, the linearity of the phase comparator output with frequency is determined by the voltage-to-frequency function of the VCO.

Because of its unique and highly linear VCO, the PLL can lock to and track an input signal over a very wide range (typically $\pm 60\%$) with very high linearity (typically within 0.5%).

A typical FM demodulator circuit is shown in Fig. 3. The VCO free-running frequency is represented approximately by $f_o = 1/4R1C1$, and should be adjusted to be at the center of the input signal frequency range. Capacitor C1 can be any value, but resistor R1 should be within the range of 2000 to 20,000 ohms with an optimum value in the order of 4000 ohms. The source can be direct coupled if the dc resistances seen from pins 2 and 3 are equal and there is no dc voltage difference between the pins.

FIG. 1—PLL BLOCK DIAGRAM (below) illustrates basic operation of the phase-lock-loop.

FIG. 2—CORRECTION SIGNALS (left) are developed as shown here.



A short between pins 4 and 5 connects the VCO to the phase comparator. Pin 6 provides a dc reference voltage that is close to the dc potential of the demodulated output (pin 7). Thus, if a resistance is connected between pins 6 and 7, the gain of the output stage can be reduced with little change in the dc voltage level at the output. This allows the lock range to be decreased with little change in the free-running frequency.

In this manner the lock range can be decreased from $\pm 20\%$ of f_o (at $\pm 6V$).

A small capacitor (typically 0.0001 μ F) should be connected between pins 7 and 8 to eliminate possible oscillation in the control current source.

A single-pole loop filter is formed by capacitor C2, connected between pin 7 and positive supply, and resistance inside the IC of approximately 3600 ohms.

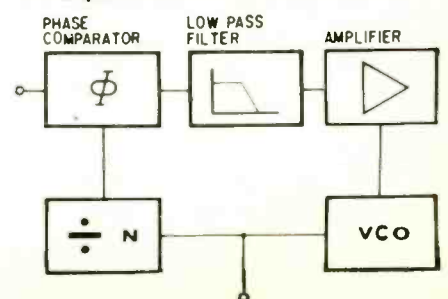
Frequency shift keying

FSK (frequency-shift keying) refers to data transmission by means of a carrier which is shifted between two preset frequencies. This frequency shift is usually handled by driving a VCO with the binary data signal so that the two resulting frequencies correspond to the "O" and "I" states (commonly called space and mark) of the binary data signal.

A simple circuit using the PLL to receive FSK signals of 1070 Hz and 1270 Hz is shown in Fig. 4. As the signal appears at the input, the loop locks to the input frequency and tracks it between the two frequencies with a corresponding dc shift at the output.

Loop filter capacitor C2 is chosen smaller than usual to eliminate overshoot on the output pulse, and a three-stage RC ladder filter is used to remove the carrier component from the output. The band

FIG. 3—FM DEMODULATOR using the PLL IC. Experimenters will enjoy this one.



edge of the ladder filter is set approximately half way between the maximum keying rate (in this case 150 Hz) and twice the input frequency (approximately 2200 Hz). The output signal can now be made logic compatible by connecting a voltage comparator between the output and pin 6 of the loop. The free-running frequency is adjusted with RI to result in a slightly-positive voltage at the output when $f_{in} = 1070\text{Hz}$.

The input connection is typical for cases where a dc voltage is present at the source and therefore a direct connection is not desirable. Both input terminals are returned to ground with identical resistors (in this case, the values are chosen to effect a 600 ohm input impedance).

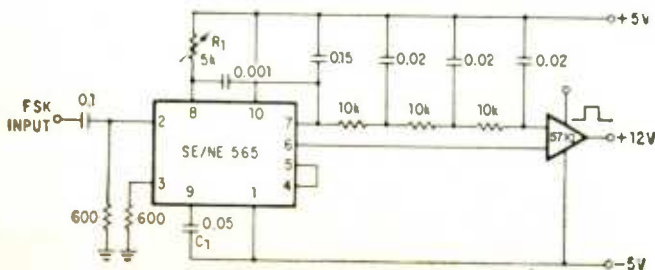
Frequency multiplication

There are two methods of doing this using the 565:

1. Locking to a harmonic of the input signal.
2. Including a digital frequency divider or counter in the loop between the VCO and phase comparator.

The first method is the simplest, and is used by setting the free-running frequency of the VCO to a multiple of the input frequency. A limitation to this scheme is that the lock range decreases as successively higher and weaker harmonics are used for locking. If the input frequency is to be constant with little tracking required, the loop can generally be locked to any one of the first 5 harmonics. For higher orders of multiplication, or for cases where a large lock range is desired, the second scheme is more desirable. An example of this might be where the input signal varies over a wide frequency range and a large multiple of the input frequency is required.

FIG. 4—FREQUENCY SHIFT KEYING is easily handled by a PLL in the circuit below. FIG. 7—SCA DECODER (bottom right) is another circuit that the PLL simplifies.



A block diagram of the second scheme is shown in Fig 5. Here the loop is broken between the VCO and the phase comparator, and a frequency divider is inserted. The fundamental of the divided VCO frequency is locked to the input frequency in this case, so that the VCO is actually running at a multiple of the input frequency. The amount of multiplication is determined by the frequency divider.

A typical circuit is shown in Fig. 6. To set up the circuit, the frequency limits of the input signal must be determined. The free-running frequency of the VCO is then adjusted by means of RI and CI (as discussed under FM demodulation) so the output frequency of the divider is midway between the input frequency limits. Filter capacitor C2 can be neglected, in this case, since it is generally desirable to have the capture and lock ranges as large as possible. The output can now be taken as the VCO square-wave output, and its fundamental will be the desired multiple of the

input frequency as long as the loop is in lock.

SCA decoder

Some FM stations are authorized by the FCC to broadcast uninterrupted background music for commercial use. To do this, a frequency modulated 67-kHz subcarrier is used.

The SCA signal can be filtered out and demodulated with the NE565 PLL without the use of any resonant circuits. A circuit for this application is in Fig. 7. This circuit

also serves as an example of operation from a single power supply.

A resistive voltage divider is used to establish a bias voltage for the input (pins 2 and 3). The demodulated (multiplex) FM signal is fed to the input through a two-stage high-pass filter, both to effect capacitive coupling and to attenuate the strong signal of the regular channel. A total signal amplitude, between 80 mV and 300 mV, is required at the input. Its source should have an impedance of less than 10,000 ohms.

The PLL is tuned to 67 kHz with a 5000 ohm potentiometer. Only approximate tuning is required, since the loop will seek the signal.

The demodulated output (pin 7) passes through a three-stage low-pass filter to provide de-emphasis and attenuate the high-frequency noise which often accompanies SCA transmission. Note that no capacitor is provided directly at pin 7. Thus, the circuit is operating as a first-order loop. The demodulated output signal is about 50 mV and the frequency response extends to 7 kHz.

R-E

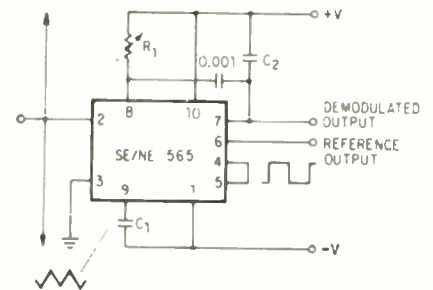
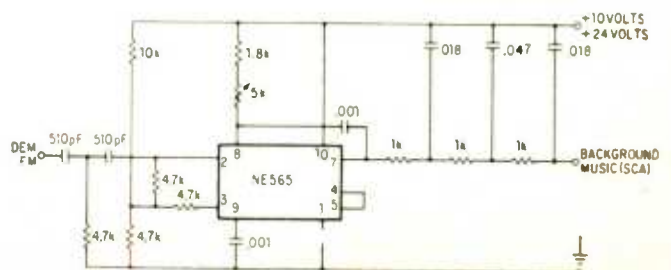
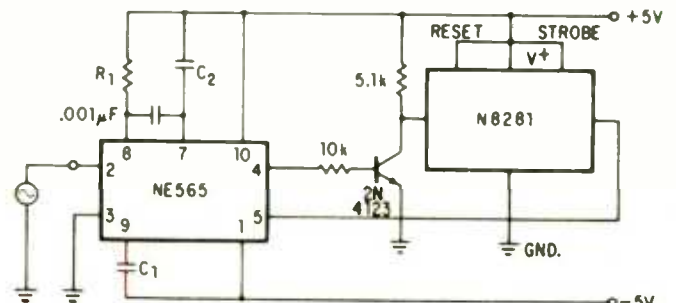
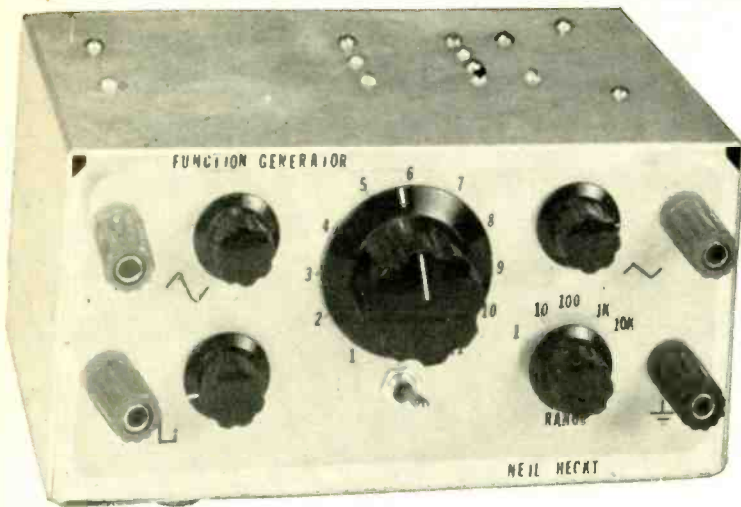


FIG. 5—(above) FREQUENCY MULTIPLIER that can be built around the PLL IC. FIG. 6—(below) ANOTHER FREQUENCY MULTIPLIER circuit variation.





Build with IC's

3-way

■ Sinewaves

by NEIL HECKT

For about \$40.00 this wide-range function generator will prove a most valuable addition to the work bench of the electronic experimenter or audiophile. It provides sine waves, triangle waves, and square waves simultaneously, from 1 to 110,000 Hz. Each output amplitude is constant regardless of frequency, and adjustable from zero to twenty volts P-P or more. Tuning is variable over one decade, with ten percent overlap on each range. A range selector provides five decade multiples of the vernier tuning.

The heart of the generator is three Motorola MC1709CG integrated-circuit operational amplifiers. Each contains the equivalent of 13 transistors, 2 diodes, and 15 resistors in a standard TO-5 package.

Theory of operation

The unit (Fig. 1) can be divided into three major parts. The oscillator sine-wave converter, and output amplifiers. The oscillator is a form of relaxation circuit called a hysteresis oscillator. It is composed of two integrated-circuit operational amplifiers IC and IC2.

Integrated circuit IC2 is connected as a voltage comparator with hysteresis feedback. Because of the high gain of the amplifier, if the non-inverting (+) terminal is slightly more positive than the inverting (-) terminal, the output will be the positive saturation level of +14 volts. Conversely, if the non-inverting input is more negative, the output will be the

negative saturation level, -14 volts.

Assuming the output is +14 volts, positive feedback applied to the non-inverting input via R9, keeps the amplifier latched in the positive saturation state. An increasingly negative voltage is applied to the input resistor R7. The ratio of R7 and R9 is such, that when the input reaches -10 volts, it cancels the +14 volt feedback. As the input goes slightly more negative, the amplifier switches to the negative saturation level. It remains in that state until the input exceeds +10 volts when a similar, but opposite reaction occurs causing the amplifier to latch again at the positive level.

Integrated circuit IC1 is connected as an integrator. If a positive voltage (E_{in}) is applied to the input resistor, R5, the output will try to swing negative. The voltage across the timing capacitor C_1 , (C1-C5) cannot change instantaneously, however, and the negative output thru the capacitor cancels out the positive input in an attempt to maintain the inverting terminal at zero volts. The cancelling current charges the capacitor at a linear rate toward -14 volts at $E_{in}/R5C_1$ volts per second. The opposite effect is true of negative inputs.

If comparator IC2 is latched in the +14-volt state, and some portion of the +14 volts is applied to the integrator input resistor, the integrator output will run down linearly toward -14 volts. Since the integrator output is connected to the comparator input, the comparator will switch to -14 volts when the integrator output equals -10 volts. This applies the opposite polarity to the integrator, and its output runs up linearly toward +14 volts. At +10 volts the comparator switches back to +14 volts com-

pleting one cycle. The cycle then repeats and the circuit oscillates at a constant rate.

The frequency of oscillation is a function of the applied voltage to the integrator (E_{in}), the integrator input resistor (R5), the feedback capacitor (C_1), and the trigger levels of the comparator ($E_o \text{ max}$) according to $F = E_{in}/4 E_o \text{ max} \times R5 \times C_1$. $E_o \text{ max}$ is set at 10 volts, R5 is 25,000 ohms, therefore $F = E_{in}/C_1$ where C_1 is in μf . Resistors R1, R2 and R3 are chosen to provide E_{in} from \pm to ± 11 volts, and C_1 is switched from 1 μF to 100 PF providing frequencies from 1 Hz to 110,000 Hz.

The output from the comparator is a 28-volt p-p square wave while the output of the integrator is a triangle of 20 volts p-p. The triangle wave output of the integrator is converted to a sine wave by the non-linear resistor/diode network R15-R30 and D7-D14.

Assuming the input to R29 and output across R30 are both zero initially, all the diodes are reverse biased by the bias voltage divider R15-R24. If the triangle wave is rising linearly toward +10 volts, its slope is reduced at the output by the voltage dividing action of R29 and R30. When the output reaches the bias level of D10, R25 is switched in parallel with R30, further reducing the slope. As the output increases still more, diodes D9, D8 and D7 conduct in turn (Fig. 2), each reducing the output slope by switching in R26, R27 and R28. The input triangle is therefore "distorted" into a quarter sine wave. As the triangle voltage decreases back toward zero, each diode is cutoff in reverse order to produce the second quarter sine wave. For negative half-cycles,

Waveform Generator

■ Squarewaves

■ Triangular waves

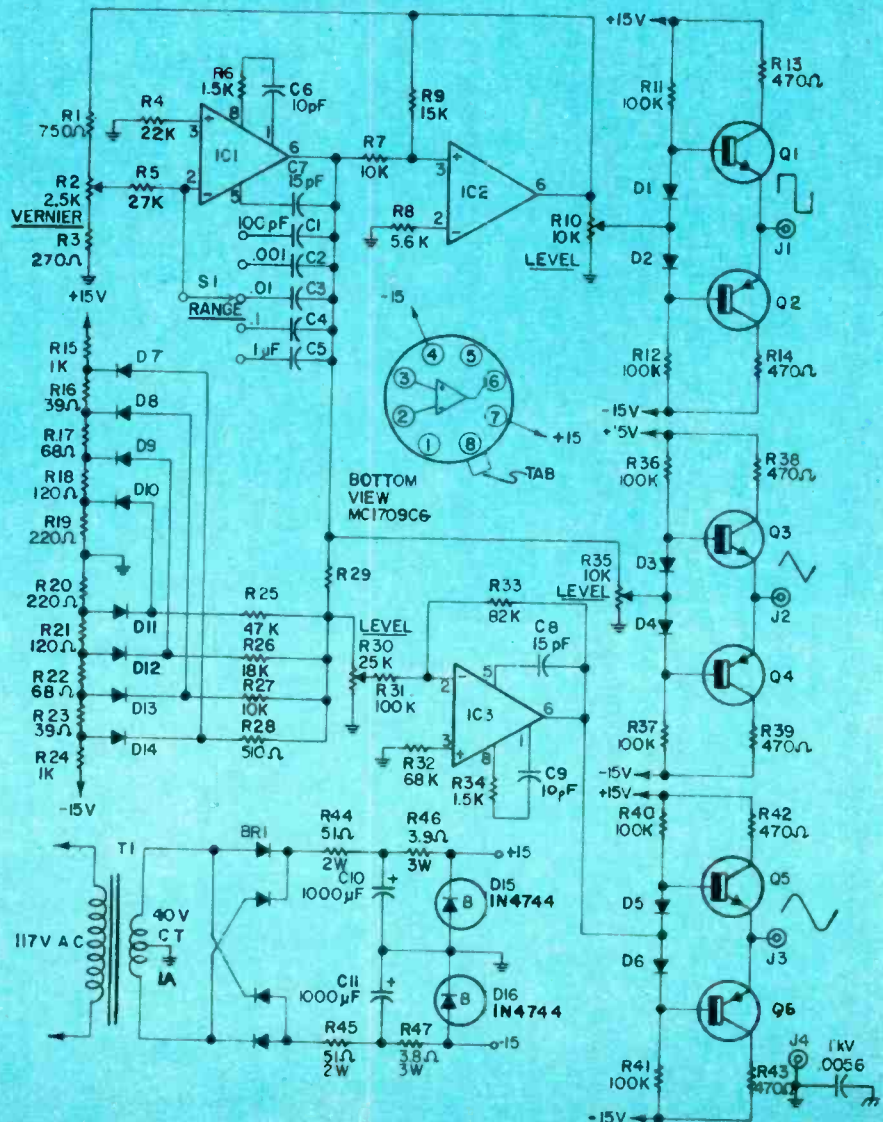
Add this solid-state waveform generator to your electronic workbench. 3 IC's keep it simple

PARTS LISTS

All resistors 1/2 watt, 10% unless noted

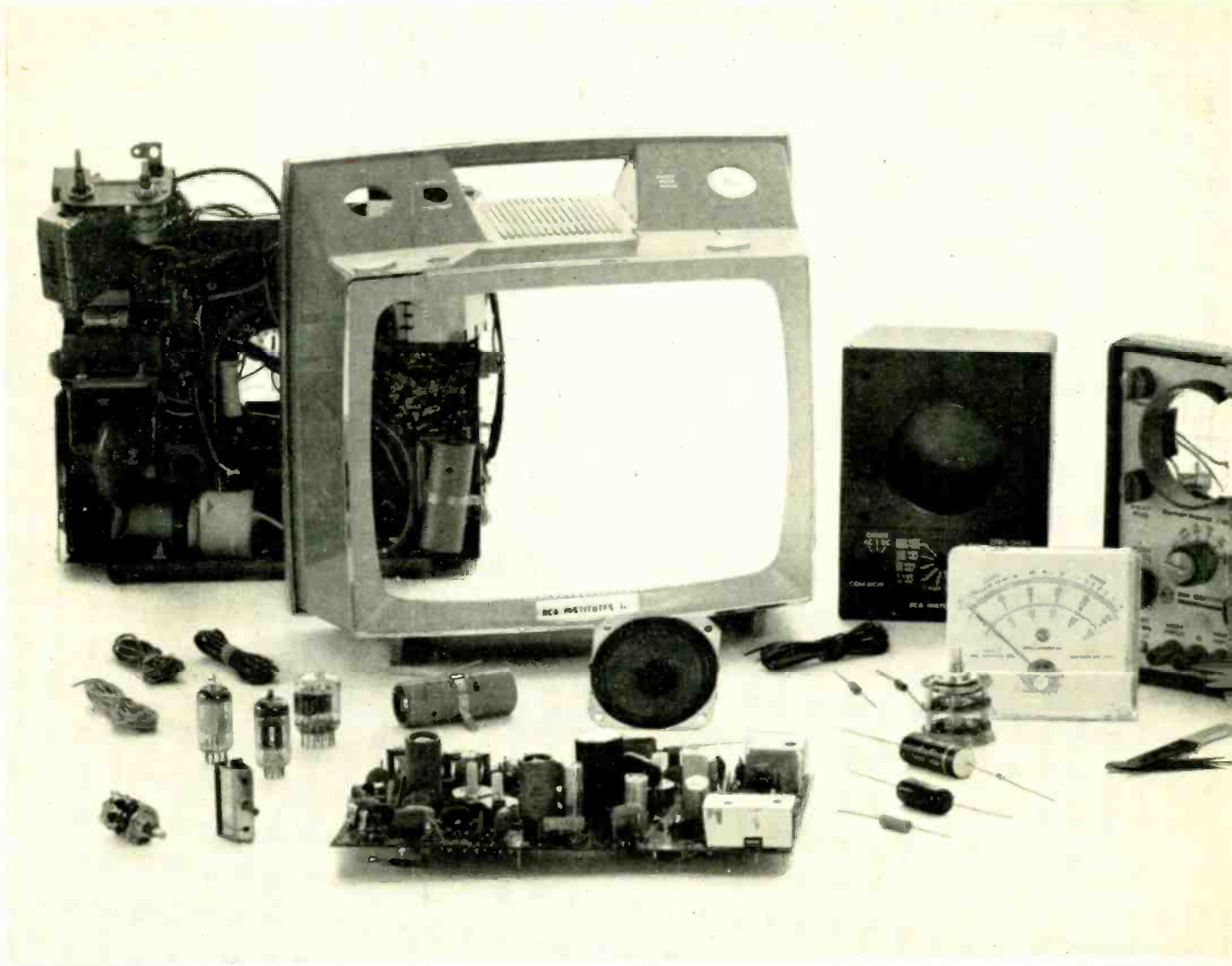
- R1—750 ohms
- R2—potentiometer, 2500 ohms, linear taper
- R3—270 ohms
- R4—22,000 ohms
- R5—27,000 ohms
- R6, R34—1500 ohms
- R7—10,000 ohms
- R8—5600 ohms
- R9—15,000 ohms
- R10, R35—potentiometer, 10,000 ohms, linear taper
- R11, R12, R31, R36, R37, R40, R41—100,000 ohms
- R13, R14, R38, R39, R42, R43—470ohms
- R15, R24—1000 ohms
- R16, R23—39 ohms
- R17, R22—68 ohms
- R18, R21—120 ohms
- R19, R20—220 ohms
- R25—47,000 ohms
- R26—18,000 ohms
- R27—10,000 ohms
- R28—510 ohms
- R29—27,000 ohms
- R30—potentiometer, 25,000 ohms, linear taper
- R32—68,000 ohms
- R33—82,000 ohms
- R44, R45—51 ohms, 2 ohms, 5%
- R46, R47—3.9 ohms, 3 watts
- C1—100 pF
- C6, C9—10 pF
- C2—0.01 μF
- C3—0.01 μF
- C4—0.1 μF
- C5—1.0 μF
- C7, C8—15 pF
- C10, C11—1000 μF, 30 volts, electrolytic
- C12—0.0056 μF
- J1, J2, J3, J4—4-way jacks
- T1—transformer, 118-volt primary, 40-volt center-tapped secondary, 1 amp
- BR1—full-wave bridge rectifier, 100 pIV, 1A or four 1N4004 or similar diodes
- D1 to D14—1N914
- Q1, Q3, Q5—2N4123
- Q2, Q4, Q6—2N4125
- IC1, IC2, IC3—MC1709CG IC op amp
- Miscellaneous: 6" x 4" x 3" utility box, knobs, stand-off terminals, hookup wire.

- D1 TO D14—1N914
- Q1, Q3, Q5—2N4123
- Q2, Q4, Q6—2N4125
- BR1—MDA 920A-3
- IC1, IC2, IC3—MC1709CG



COMPLETE CIRCUIT OF WAVEFORM GENERATOR. Regulated power supply delivers +15 and -15 volts. Three IC amplifiers simplify construction drastically.

(continued on page 68)



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3-WAY GENERATOR
(continued from page 63)

diodes D11-D14 are cut in in a similar manner. The points at which the slope changes are not sharp, but are rounded by the non-linear forward resistance characteristics of the diodes as the current thru them increases.

The output of the converter is about 10 volts p-p. Operational amplifier IC3 is connected to give a gain of 2 so that 20 volts p-p are available.

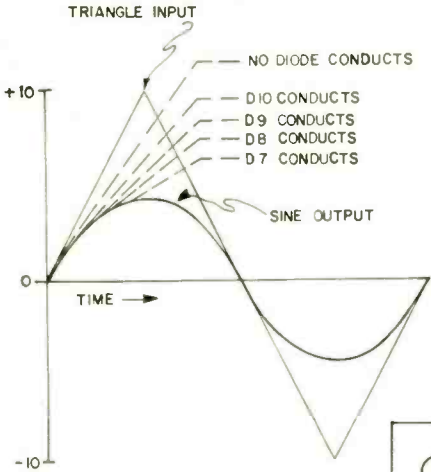


FIG. 2—DIODES MAKE A SINE WAVE out of a triangular wave.

INSIDE THE GENERATOR (top right) you can see how parts are set up.

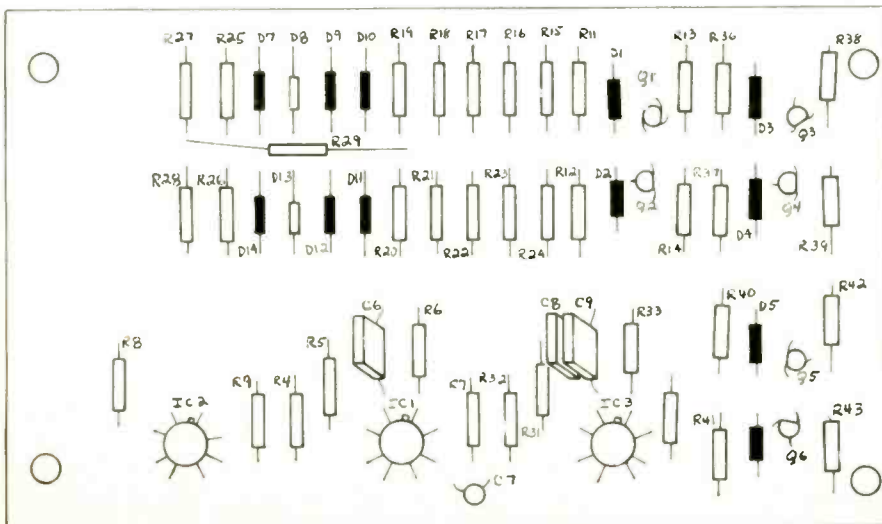
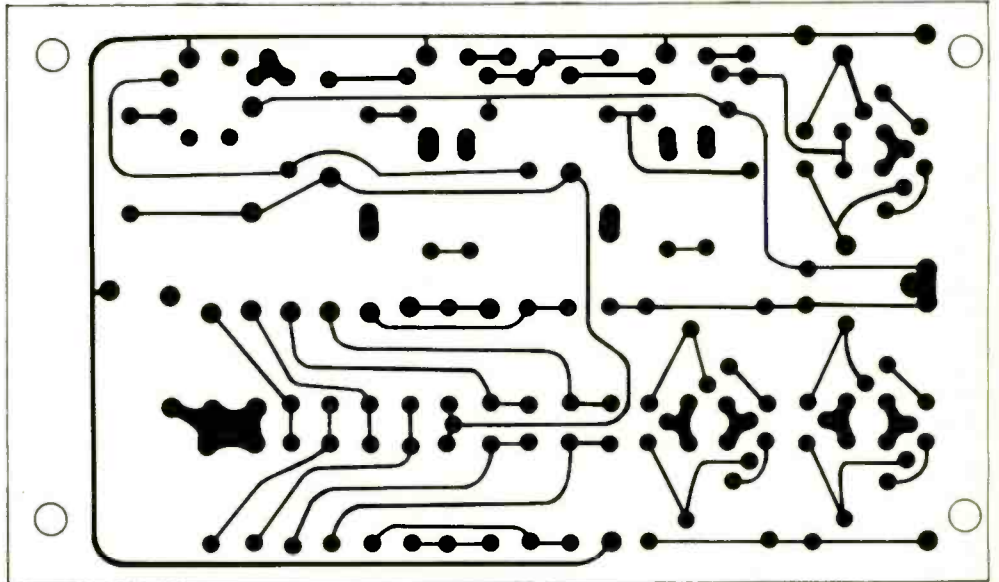
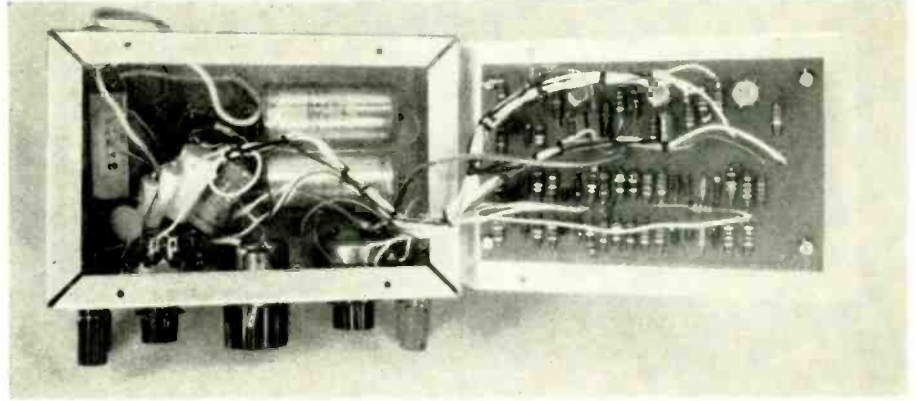
PRINTED-CIRCUIT PATTERN (right) is half actual size.

PARTS LAYOUT (below) shows component location on the circuit board.

There are three output drives, one for each waveform. They consist of complementary emitter followers. Diodes, such as D1 and D2, compensate for the base-to-emitter drop of the transistors. For positive inputs, the npn transistor supplies the current to the load, and for negative inputs the pnp supplies the current. The 470-ohm resistors limit the current to prevent damage to the transistors, as well as provide an output impedance of about 500 ohms.

Construction techniques

A 3 x 4 x 6-inch chassis, such as a Bud AC-430 with BPA-1505 bottom plate makes an economical, compact cabinet. Mount power supply components on the inside top of the chassis box using insulated standoff terminals (Grayhill type 18-1 or similar). Mount all controls and binding posts symmetrically on one 3 x 6-inch face. After all components have been mounted in the chassis and on the PC board, connect the two thru twisted



pairs of wires as per the wiring list.

Resistor R1 is connected from the top of R10 to the top of R2, and resistor R3 is connected from the bottom of R2 to the ground post.

The completed PC board is attached to the center of the bottom plate with 1/4-inch spacers.

The completed unit should be checked for proper operation. Notice there are no adjustments to be made. If all components are good, and the wiring is correct, the unit will function immediately.

After checkout is completed, fold the PC board up into the chassis in a manner similar to closing a book. Attach the bottom plate with four 1/4-inch No. 6 self-tapping screws. **R-E**

30 IC CIRCUITS

(continued from page 40)

The close parameter matching and tight thermal coupling of the CA3018's transistors result in exceptionally stable operation when they are used in the dc differential mode as a meter amplifier. The device outperforms conventional units in this type of application.

Fig. 24 shows how to connect the IC so it effectively increases the sensitivity of a 100- μ A meter to 1 μ A. Transistors Q3-Q4 and Q1-Q2 are wired in the differential mode as Super-Alpha emitter followers, with emitter loads R6 and R7 and with bias set by bridge network R1-R2-R3-R4. Resistors R5 and R9 enable the bridge bias network to be precisely balanced. Resistor R8 permits balancing the two emitter loads so the meter (connected between the emitters of the two emitter followers) reads zero when zero input is connected between the input terminals (pins 9 and 6) of the unit.

The emitter-follower actions of the circuit cause the design to deliver high current gain, but zero voltage gain, between the input terminals and the actual meter. The basic circuit gives a current gain of approximately 1000 times. This gain is subject to some variation with changes in ambient temperature.

In the interest of stability the actual gain of the circuit is reduced to only 100 by connecting SENSITIVITY control R10 across the input terminals.

To set up the circuit:

1. Set R10 (SENSITIVITY) to maximum, and R9 at mid-range. Short the input terminals, and adjust R8 for a zero reading on the meter.

2. Remove the short from the input terminals, and adjust R5 for a zero reading on the meter. Finalize the zero reading with R9.

3. Apply a 1 μ A current between

the circuit's input terminals. Then adjust R10 for full-scale deflection on the meter. The circuit is now ready for use.

Fig. 25 shows how to wire the IC as a dc millivoltmeter with a sensitivity of 200,000 ohms/volt, using a 100- μ A meter. Transistors Q1 and Q2 are used as a differential amplifier, with the meter connected between their collectors. Transistors Q3-Q4 are used as a constant-current generator in the differential amplifier's emitter line.

The circuit delivers both voltage and current gain between the input terminals (pins 6 and 3) and the meter. Resistor R_v acts as a range multiplier, and has a value of 200,000 ohms/volt—if the meter is required to read 10 volts full scale; R_v = 2 megohms.

To set up the circuit:

1. Short pins 3 and 6, and adjust R9 for zero meter reading.

2. Remove the short from pins 3 and 6, and with no input connected, adjust R5 for zero meter reading.

3. Connect the calculated value of R_v (depending on full-scale deflection voltage required), and apply the required voltage to the input terminals. Adjust R10 for full-scale meter deflection.

4. Recheck steps 1 and 2, and readjust if necessary.

Finally, Fig. 26 shows how to wire the IC as a dc millivoltmeter with a 2-megohm/volt sensitivity. The circuit is basically similar to Fig. 25, except that Q3-Q4 and Q1-Q2 are Super-Alpha-connected in the differential mode, and Q5 and D1 are used as the constant-current generator. Resistor R_v is given a value of 2 megohms/volt, depending on the full-scale deflection voltage required.

To set up the circuit:

1. Short pins 9 and 6, and adjust R7 for zero meter reading.

2. Remove the short from pins 9 and 6, and adjust R9 and R10 for zero meter reading.

3. Connect the calculated value of R_v, and apply the required full-scale deflection voltage to the input terminals. Adjust R8 for full-scale deflection.

4. Repeat steps 1, 2 and 3, and trim if necessary.

Since this circuit has very high sensitivity some care is needed in the construction to insure stable operation. The circuit must be well screened, to keep stray pickup away from the input network. Resistors R1-R3 and R2-R4 must be mounted close together, so they operate at similar temperatures. All must be 5% or better, high-stability types.

This circuit, and those of Figs. 24 and 25, can be used as multirange meters, by switch-selecting R_v values.

Schmitt trigger circuits

The high package density and gain-bandwidth products of the CA3018 make the IC ideal for compact, high-performance Schmitt trigger projects. Fig. 27 shows how to connect the IC as a simple Schmitt trigger with dc input and emitter-follower output. Transistors Q1-Q2 form the Schmitt, and Q4 acts as the emitter-follower output stage.

With zero input connected to the unit, the output is close to zero volts. When the input is raised to roughly $V_{\text{supply}} \times 0.22$, the Schmitt changes state and the output jumps up to a positive voltage. The trigger point of the circuit can be changed, if required, by using different R4 values. The circuit acts as a useful dc level sensor.

Fig. 28 shows how the circuit can be used as a simple sine/square converter. Here Q1 has a fixed bias applied via R1-R2 and R3, and the input signal is applied to the Schmitt via C1. The circuit needs a sine-wave input greater than 1 volt peak to peak for correct sine/square conversion, and gives a useful performance up to several hundred

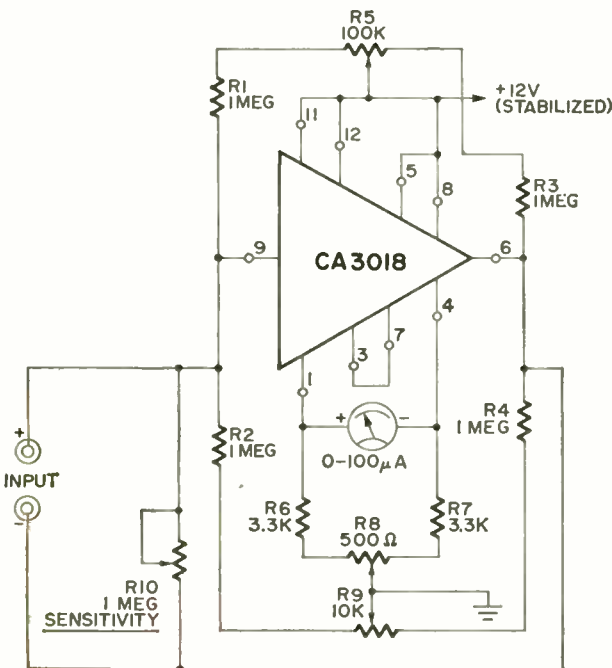


Fig. 24—ELECTRONIC MICROAMMETER uses 100- μ A meter and IC amplifier with 100X gain.

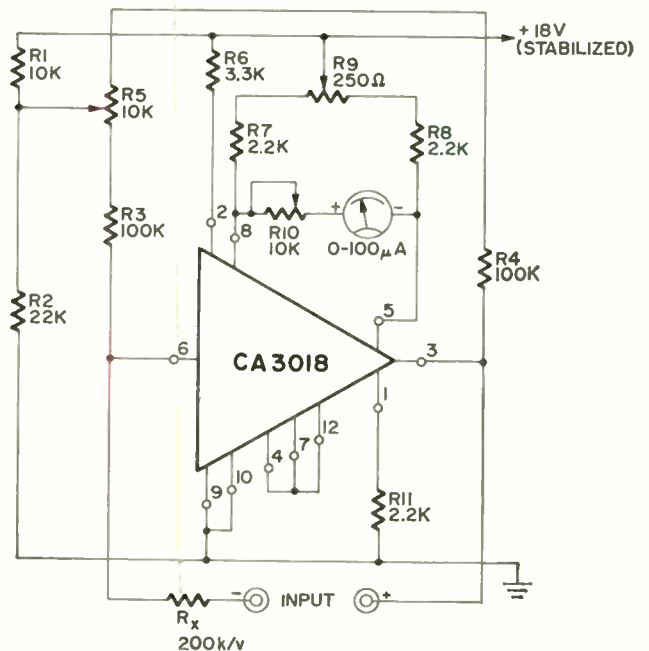


Fig. 25—ELECTRONIC MILLIVOLTMETER has IC amplifier. Make R_v switchable for multiple ranges.

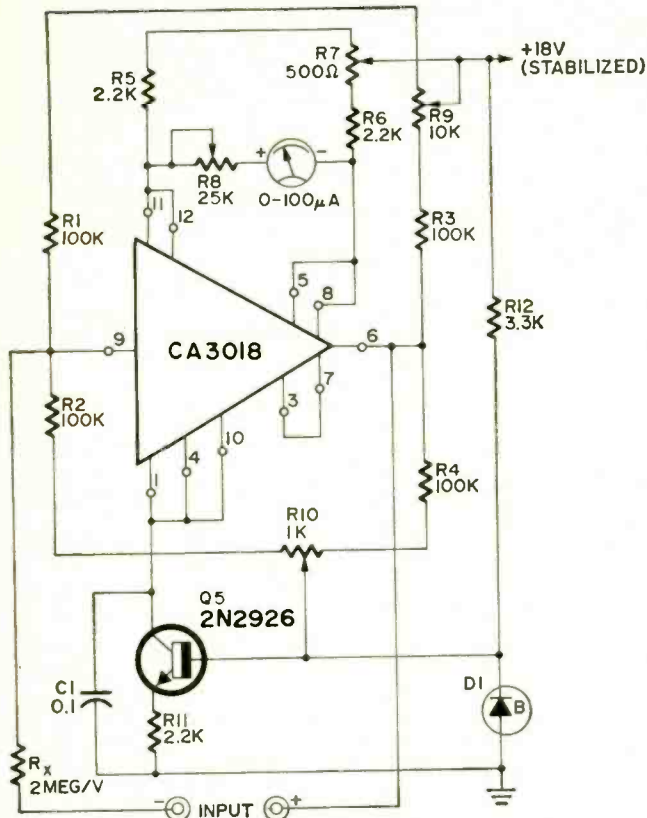


Fig. 26—TRANSISTOR AND DIODE raise this IC mV meter to 2 megs per volt.

kHz. Capacitor C2 is a speedup unit that reduces the output rise time.

Fig. 29 shows how to wire the IC so that it acts as a sensitive sine/square converter. Input impedance is 100,000 ohms and an input of only 50 mV peak to peak is needed for correct sine/square conversion.

Transistors Q3-Q4 are used as a Super-Alpha-connected common-emitter preamp with input connected to pin 9 and output direct-coupled to the input of the Q1-Q2 Schmitt trigger stage. The preamp is operated near saturation. The square-wave output of the circuit is taken from Q2's collector via C2, and the waveform is useful up to a couple of hundred kHz.

Multivibrators to try

The high package density and very high available current gain of the CA3018 also make the IC ideal for multivibrator projects. Fig. 30, for example, shows how the IC can be set up as a long-period stable multivibrator that operates at a frequency of 0.2 Hz—at 5 sec per cycle.

Transistors Q3-Q4 are wired in the Super-Alpha mode to form one half of the multivibrator. Transistors Q1-Q2 are wired in the Super-Alpha mode to form the other half. Because of the very high current gains of the Super-Alpha-connected transistors, the circuit can be given very high values of time-constant resistance (R1 and R2).

Two outputs are available from the circuit (one at pins 11 and 12, and the other at pins 5 and 9), and are in anti-phase. The operating period of the circuit can be increased, if required, by increasing the values of C1 and C2.

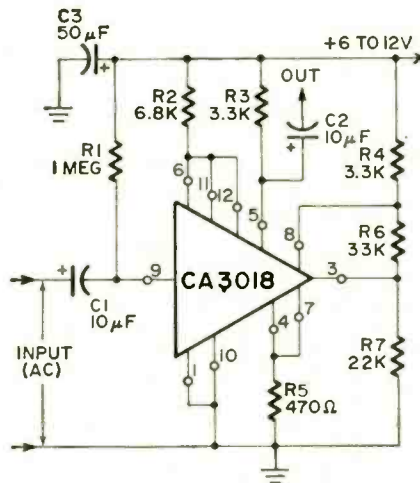


Fig. 31 shows how to connect the IC as a long period one-shot (or monostable) multivibrator, with emitter-follower output. Transistors Q3-Q4 are again wired in the Super-Alpha mode, and form one half of the multivibrator. Transistor Q2 forms the other half and Q1 is used as the emitter-follower output.

The output of the circuit (taken from R7's slider) is normally at zero volts, but rises to positive for a fixed period whenever the monostable is triggered. The period has a duration of approximately 2.5 sec/μF of C1's value—if C1 is 10 μF, the period is 25 seconds. The circuit can be triggered manually by momentarily operating S1, or electronically by a positive pulse on pin 3.

(continued on page 101)

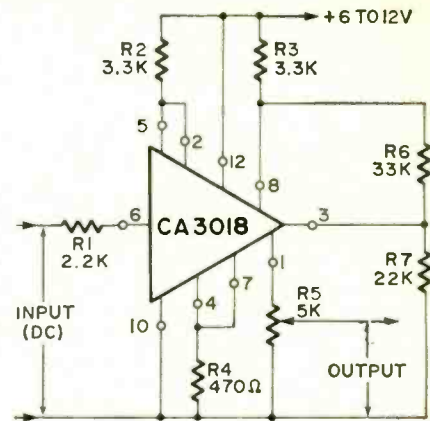


Fig. 27—SCHMITT trigger (right) is versatile pulse generator. Fig. 28 (below)—Schmitt trigger converts input sine wave to a square wave at the output.

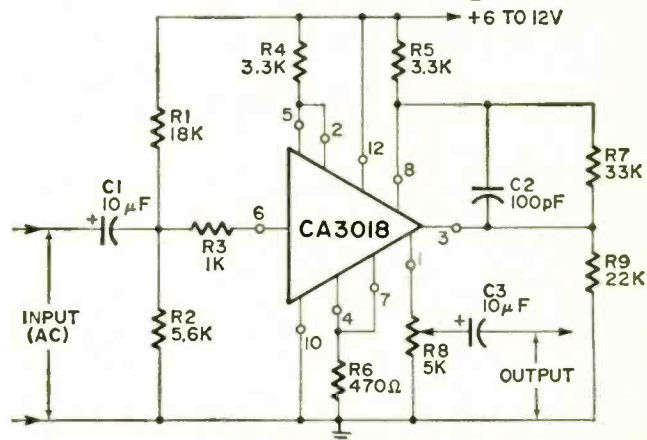
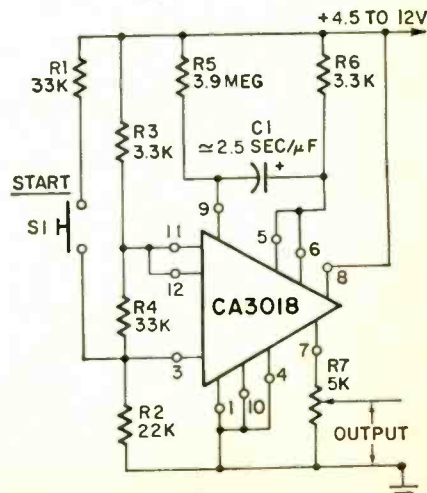
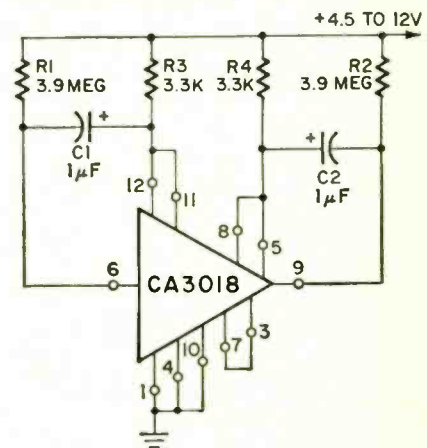
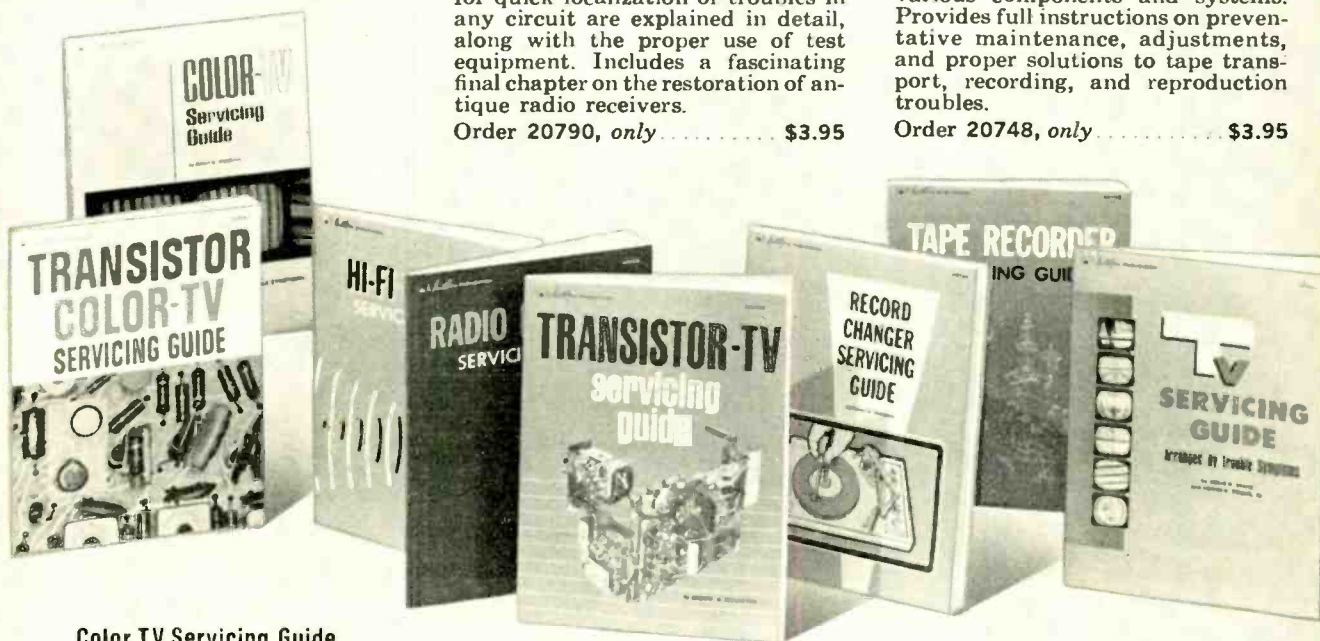


Fig. 29 (left), 30 (right) and 31 (below)—SENSITIVE SINE/SQUARE CONVERTER, low-frequency oscillator and a long-period one-shot multivibrator, respectively.



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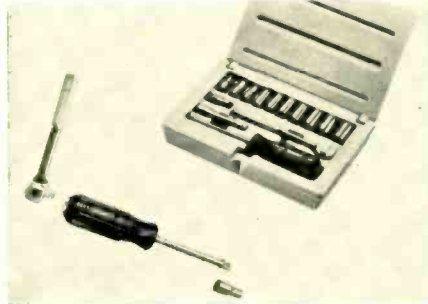
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More information on new products is available from the manufacturers of items identified by a Reader Service number. Use the Reader Service Card on page 103 and circle the numbers of the new products on which you would like further information. Detach and mail the postage-paid card.

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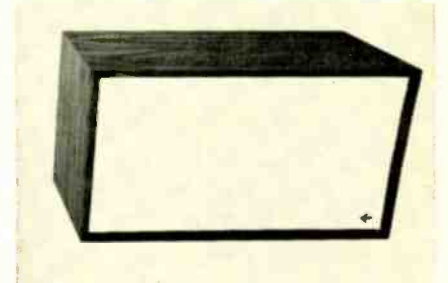
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tests entire assemblies for cold weather and high altitude operation. Troubleshoot for intermittents by heating chassis and spraying each suspected component for a few seconds. Healthy components will not be affected, but faulty ones will act up. To test entire units, place unit in cardboard box and flood with Blue Frost. Tight fitting metal parts that are difficult to mate can easily be pressed together if the inner part is first shrunk with the spray. Recommended for industrial or professional use only. Available in 8 oz. and 20 oz. cans.—Tech Spray, P.O. Box 949, Amarillo, Texas 79105.

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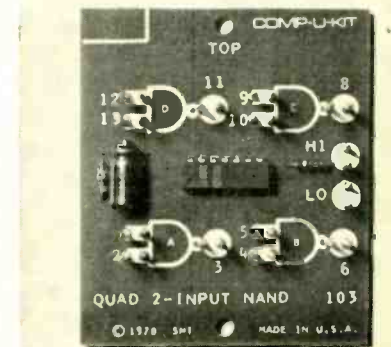
above 200 Hz to eliminate boxiness or



hints of a nasal effect. The free air cone resonance is just over 20 Hz. 8-ohm unit can handle 10-watts of audio. Enclosure measures $24 \times 14 \times 10\frac{1}{2}$ inches. . . . \$69.95, Audionics, 9701 S.E. Mill, Portland, Oregon 97216.

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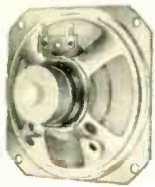
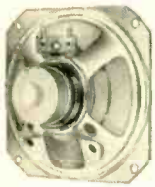
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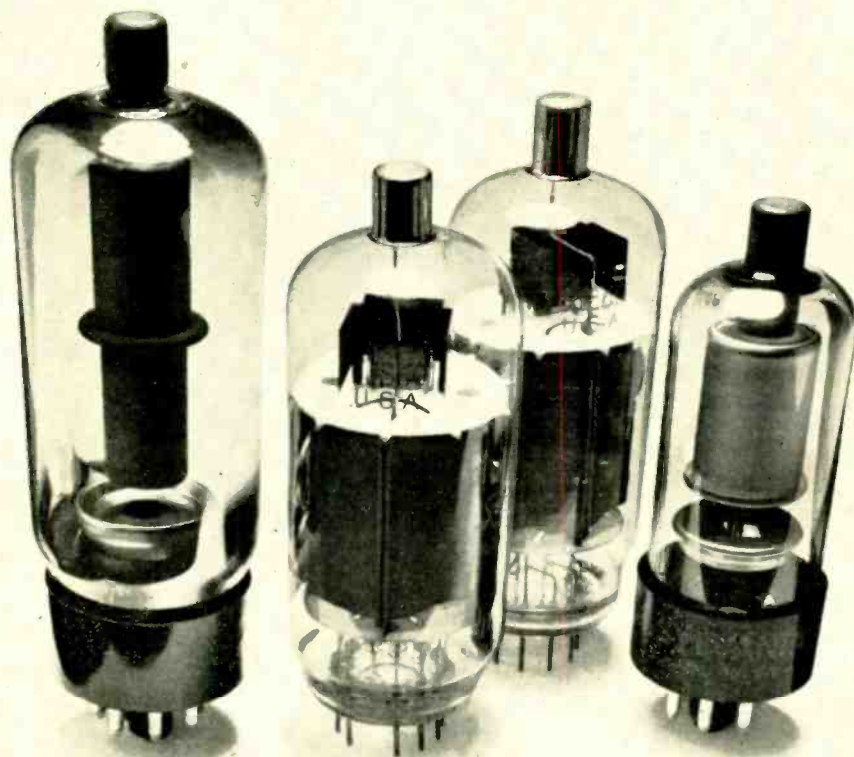
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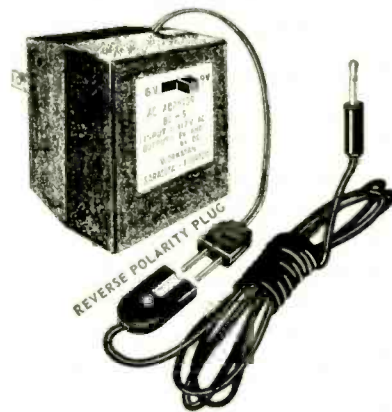


on a small meter, when the voltage exceeds levels set by the set manufacturer. Consists of a sensor unit which is in-

stalled in the set, and a meter placed in a bracket in the rear of the set. A needle on the meter enters a danger area when voltage exceeds limits.—Audio Equipment Co., Div. of Walter Kidde & Co., Clifton, New Jersey 07012

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(continued on page 85)



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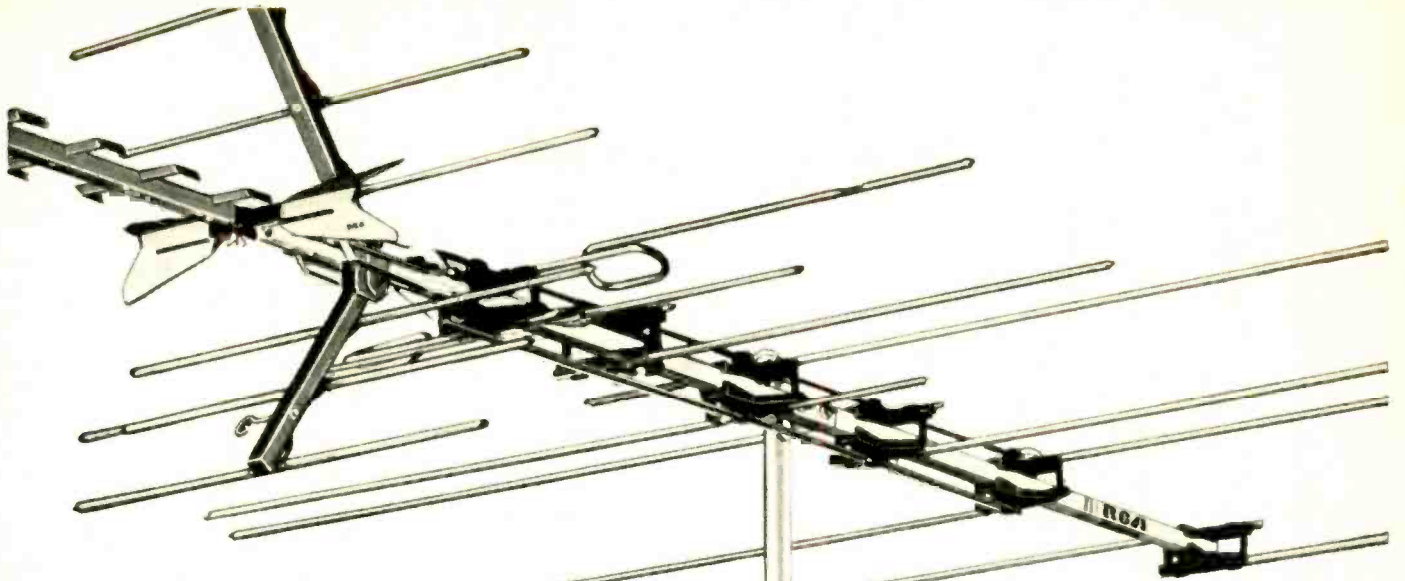


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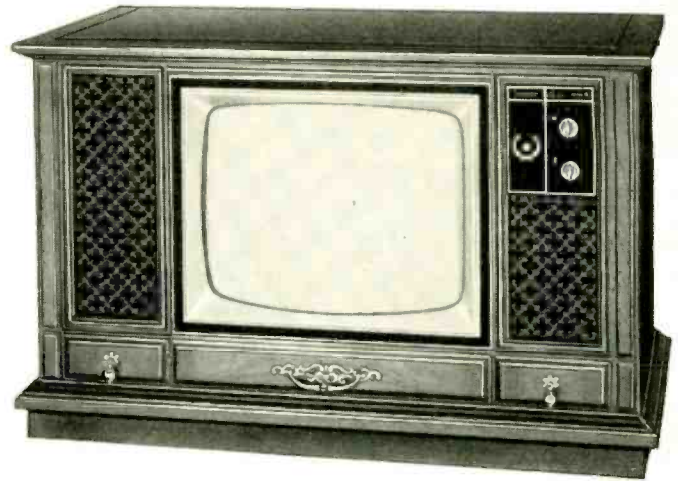
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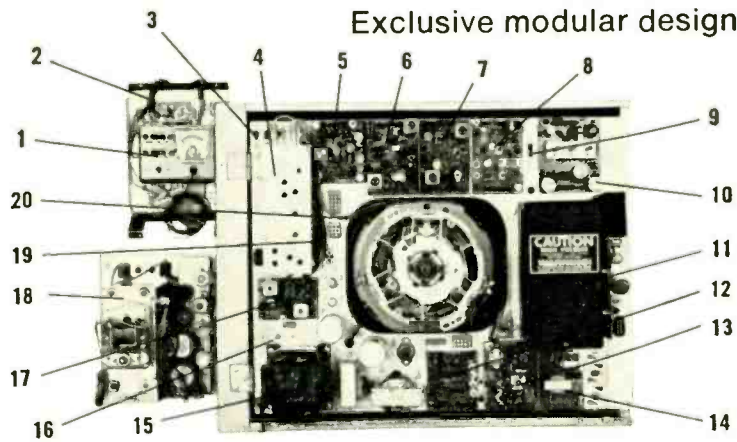
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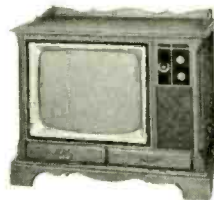
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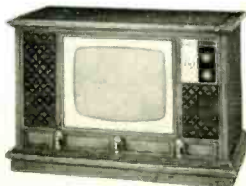


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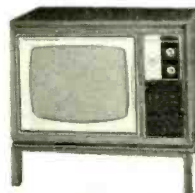


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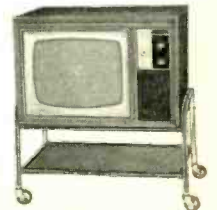
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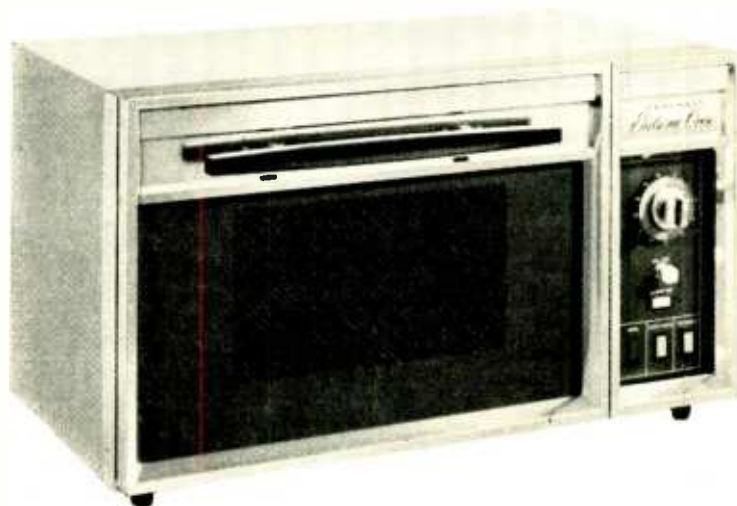
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- 102", 14 V picture tube

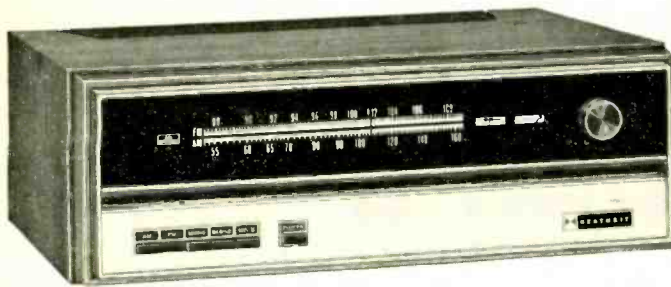


What do you do for an encore after you've created the solid-state GR-270 and GR-370 big-screen sets? Simple. Make them portable. That's virtually what's been done in the new Heathkit GR-169 solid-state portable color TV. Heath engineers took the same cool-running solid-state circuitry from the large screen chassis and packaged it in an easy-to-assemble compact chassis...with the same nine plug-in glass epoxy circuit board modules used in the big sets. In fact the only difference is the smaller preassembled horizontal deflection and high voltage power supply. The same MOSFET tuner and high gain 3-stage IF found in the big sets offer superlative color performance. And, as in the larger sets, complete owner-service features are provided by inclusion of built-in dot

generator and degaussing along with an exclusive volt-ohm check-out meter. 48-hour factory service facilities for modules are also provided with the GR-169. Other features include: built-in antennas and connections for external antennas; instant picture and sound; complete secondary controls available behind the hinged door on the front panel; high resolution circuitry for sharp, crisp pictures; adjustable noise limiting to keep external interference to a minimum. If you're looking for big set color fidelity and performance with portable convenience...put the new Heathkit GR-169 on your Christmas shopping list now!

Kit **GR-169**, 48 lbs. **\$349.95***

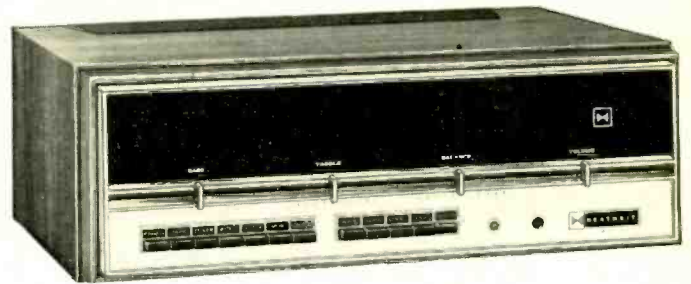
New Heath-gift ideas... for



New Heathkit® AJ-29 AM-FM-FM stereo tuner

This is the feature-packed tuner section of the famous Heathkit AR-29 stereo receiver . . . now available as a stereo "separate." The pre-assembled, factory-aligned FM tuner boasts 1.8 uV sensitivity for whopping station pulling power using FET design for superior overload characteristics. Three IC's in the IF section offer superior AM rejection, hard limiting, temperature stability, and outstanding reliability. Other features include a computer-designed 9-pole L-C filter for greater than 70 dB selectivity; new "blend" and "mute" functions; and a built-in AM rod antenna that swivels for best reception.

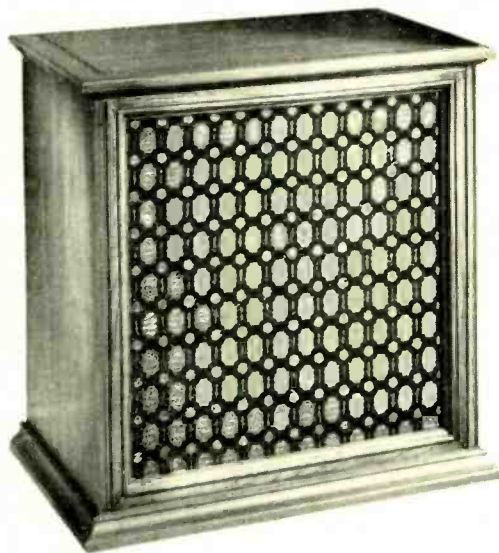
Kit **AJ-29**, 19 lbs., less cabinet **\$169.95***
Assembled **AE-19**, oiled pecan cab., 9 lbs. **\$19.95***



New Heathkit® AA-29 100-watt stereo amplifier

Power-packed amplifier section of the Heathkit AR-29, the AA-29 stereo "separate" marks another milestone in superior Heathkit amplifier design. Its 70-watts of continuous power is more than enough to drive even the most inefficient speaker systems. A massive, fully-regulated and filtered power supply, 4 conservatively heat sunk output transistors and the best IM and harmonic distortion specifications in the industry add up to sound fidelity you never expected to hear outside the theater. Modular plug-in circuit boards make assembly easier . . . snap out in seconds for future servicing.

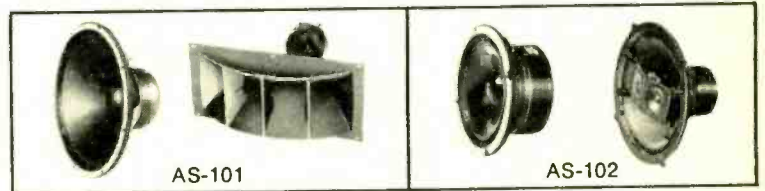
Kit **AA-29**, 27 lbs., less cabinet **\$149.95***
Assembled **AE-19**, oiled pecan cab., 9 lbs. **\$19.95***



New Heathkit® floor model speaker systems

In the new Heathkit AS-101 and AS-102 speaker systems, Heath engineers have combined the best of both worlds of sound and beauty. The AS-101 Heath/Altec-Lansing 2-way system features a 15" woofer and sectoral horn delivering from 35 to 22,000 Hz with uncompromising accuracy. The AS-102 Heath/Bozak 3-way system uses a 12" woofer, 6" mid-range, and two 2 1/2" tweeters in an infinite baffle design to produce clean natural reproduction from 40 to 20,000 Hz. Both systems are housed in assembled Mediterranean pecan cabinets, 29 5/8" H x 27 3/4" W x 19 7/8" D.

Kit **AS-101**, 53 lbs. **\$259.95***
Kit **AS-102**, 39 lbs. **\$259.95***



New Heath stereo equipment credenza



Romantic Mediterranean styling in wife-pleasing one-piece console design . . . yet with plenty of room for your favorite separate stereo components. Six-and-a-half feet of solid craftsmanship executed in North American Hickory veneers and solid oak trim, finished in oiled pecan. Completely assembled and finished, ready for installation of Heath or other components. Speaker enclosures

are ducted port reflex design, pre-cut for 12" speakers. An adjustable shelf has room for stereo receiver, cartridge or cassette tape player or separate tuner and amplifier. Below the shelf is room for your turntable and record storage. Accessory matching drawers on ball bearing slides are available for turntable and tape player.

Model **AE-101**, 90 lbs. **\$189.95***

home, shop and ham shack

New Heathkit® IC 15 MHz frequency counter...199.95*

A highly accurate, low cost frequency meter for anyone requiring accurate frequency measurements. Compare these features to counters selling for over twice this low price: accurate counting, 1 Hz to over 15 MHz; integrated circuitry; automatic trigger level for wide range input without adjustment; five digit readout with Hz/kHz ranges and overrange indicators for eight digit capability; high input impedance; storage circuitry for non-blinking, no-count-up readout; computer-type circuitry, no divider chain adjustment; temperature-compensated crystal time base oscillator; BNC input with cable; double-sided, plated-thru circuit board with sockets; three-wire, removable line cord; heavy-duty aluminum case handle/tilt stand and die cast zinc front panel; no special instruments required for accurate calibration.

Kit **IB-101**, 7 lbs. **\$199.95***



IB-101 SPECIFICATIONS: Frequency Range: 1 Hz to greater than 15 MHz. Accuracy: ± 1 count \pm time base stability. Gate Times: 1 millisecond or 1 second with automatic reset. Input Characteristics: Sensitivity: 1 Hz to 1 MHz, less than 100 mV rms. 1 MHz to 15 MHz, less than 250 mV rms. After 30 minutes warmup. Trigger Level: Automatic. Impedance: 1 Meg ohm shunted by less than 20 pf. Maximum Input: 200 V rms, DC-1 kHz. Derate at 48 V per frequency decade. TIME BASE: Frequency: 1 MHz, crystal controlled. Aging Rate: Less than 1 PPM/month after 30 days. Temperature: Less than ± 2 parts in 10^7 /degree C. 20 to 35 degrees C after 30 minutes warmup. $\pm .002\%$ from 0 to 50 degrees C. GENERAL: Readout: 5 digits plus overrange. Temperature Range: Storage; -55 to 80 degrees C. Operating; 0 to 50 degrees C. Power Requirements: 105-125 or 210-250 VAC, 50/60 Hz, 8 watts. Cabinet Dimensions: 8 1/4" W x 3 1/8" H x 9" D not including handle. Net Weight: 4 1/2 lbs.

See these and 300 other
Heath-gift suggestions at one of the
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165 Worcester St.

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3462-66 W. Devon Ave.

Chicago Area
Downers Grove, Ill. 60515
224 Ogden Avenue

Cleveland, Ohio 44129
5444 Pearl Road

Dallas, Texas 75201
2715 Ross Avenue

Denver, Colorado 80212
5940 W. 38th Ave.

Detroit, Michigan 48219
18645 W. 8 Mile Road

Fair Lawn, N. J. 07410
35-07 Broadway (Rt. 4)

Houston, Texas 77027
3705 Westheimer

Los Angeles, Calif. 90007
2309 S. Flower St.

Milwaukee, Wisc. 53216
5215 W. Fond du Lac

Minneapolis Area
Hopkins, Minn. 55343
101 Shady Oak Road

New York, N. Y. 10036
35 W. 45th Street

Philadelphia, Pa. 19149
6318 Roosevelt Blvd.

Pittsburgh, Pa. 15235
3482 William Penn Highway

St. Louis, Mo. 63123
9296 Gravois Ave.

San Diego Area
LaMesa, Calif. 92041
8363 Center Drive

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Redwood City, Calif. 94063
2001 Middlefield Road

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2221 Third Avenue

Washington, D. C. Area
Rockville, Md. 20852

5542 Nicholson Lane

All Heathkit Electronic Centers are units of Schlumberger Products Corporation.
Heathkit Electronic Center Prices Slightly Higher.

During 1971, consult Heathkit Catalog Supplements and local newspapers for announcements of new Heathkit Electronic Centers opening in these places:

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Cincinnati, Ohio

Los Angeles Area
Woodland Hills, California
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Prices listed are factory mail order.
Retail prices are slightly higher.

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New Heathkit® HW-101 SSB transceiver

The Hams at Heath have done it again... with an updated version of the Heathkit HW-100, one of the most popular pieces of ham gear on the market! The HW-101 features improved receiver circuitry resulting in better than 0.35 uV sensitivity for 10 dB S+N/N. Image and IF rejection are better than 50 dB. Other improvements are a new 36-to-1 ball-bearing dial drive; new selectable SSB or CW filters and attractive new front-panel styling.

Kit **HW-101**, 23 lbs. **\$249.95***

New Heathkit wattmeter/SWR bridge



Two switch-selected ranges allow measurement of RF output from 10-200 W and 100-2000 W. Built-in calibrator permits 10% accuracy throughout the 80-10 M ham bands.

Kit **HM-102**, 3 lbs. **\$29.95***

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Prices & specifications subject to change without notice. CL-394

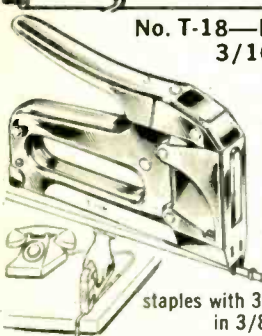
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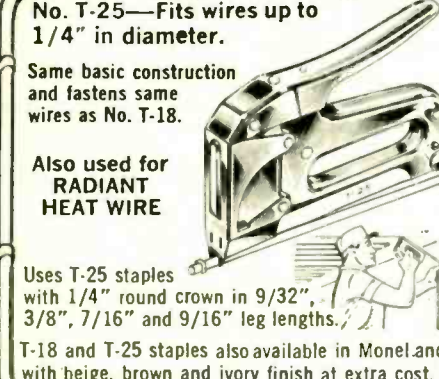
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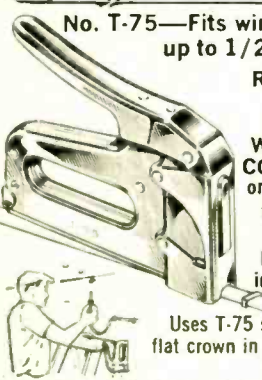
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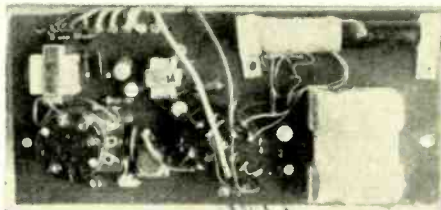
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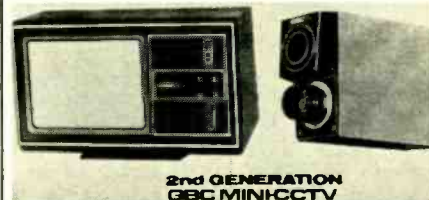
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EQUIPMENT REPORT

GBC CLOSED-CIRCUIT TV SYSTEM

For manufacturer's literature, circle No. 96 on Reader Service Card.



2nd GENERATION
GBC MINICCTV

A CLOSED CIRCUIT TELEVISION SYSTEM has many everyday applications. Some of them include seeing who's at the front door before opening it and keeping an eye on poolside activities from your kitchen. Combined with a video tape recorder, it makes a complete videotape system. Use the camera from the CCTV system to take the pictures you want to tape, the recorder to tape them, and the monitor to view what you taped.

CCTV systems come in many combinations. A basic system consists of a camera and monitor. A better system, like the GBC MC-15 described here, has a built-in audio channel.

The camera in the GBC system is light-weight and readily portable. It has only one cable dangling behind it. This links the video and audio signals from the camera to the monitor and carries operating power to the camera.

The camera has several interesting features. Its lens automatically adjusts to existing light levels to produce a proper picture. A microphone is built into the camera.

The only controls on the camera are focus and an on-off switch. This makes operation simple. An indicator lamp lets you know when the power is on. There is a mount for attaching the camera to a tripod.

The operating controls are on the monitor. There's the inevitable on-off switch, brightness and contrast controls, a volume control for the audio plus a push-to talk-button (the camera microphone becomes a speaker) and a focus adjustment that permits limited variation of focus from the monitor. The only other control is a standby switch.

After using the system I found picture quality excellent, the camera and monitor easy to use. The 2-way audio takes a little getting used to, but as soon as you remember not to talk while the other party does, you're in business. All in all, a system worth owning.—Warren Roy

NEW BOOKS

CBers' HOW-TO BOOK, by Leo G. Sands. Hayden Book Co., Inc. 116 W. 14 St., New York, N.Y. 10011. 5 1/2 x 8 1/4", 120 pp. Softcover, \$3.50.

Explains how to operate CB gear and includes tables for all CB codes and channels. Tells how to select and install the right unit, with analysis of mobile unit and base-station features. Walkie-talkies and phone patches also discussed. Covers test and repair techniques and modifications to improve operation.

1970 POPULAR TUBE/TRANSISTOR SUBSTITUTION GUIDE, by TAB editorial staff. Tab Books, Blue Ridge Summit, Pa. 17214. 224 pages, 5 1/2 x 8 1/2 in. Leatherette, \$4.95, paper, \$2.95.

First four sections cross reference popular American receiving tubes; commercial and industrial tubes are also covered. Third and fourth sections show substitutes for foreign tubes and base diagrams. Final four sections list transistors in a similar format.

COMPUTERS AND ARTIFICIAL INTELLIGENCE, by Robert G. Middleton. Howard W. Sams & Co. Inc., 4300 W. 62 St., Indianapolis, Ind., 46206. 126 pages, 5 1/2 x 8 1/2 in. \$3.95.

Early chapters describe analog and digital computers and basic logic circuits. Other topics discussed are: information theory, automation deductive logic, inductive logic; later chapters discuss possibilities that a computer might show true intelligence.

HANDBOOK OF OSCILLOSCOPES: THEORY AND APPLICATION, by John D. Lenk. Prentice-Hall, Inc., Englewood Cliffs, N.J. 07632. 212 pages, 6 1/4 x 9 1/4 in. Hardcover, \$10.95.

Supplements operating instructions of a scope by describing how to use them for everything from finding a diode's switching time to the hysteresis factor of a memory core. After scope-basics chapters, voltage, current, time, frequency and phase measurements are described. Other chapters explain scope use with sweep generators, checking components, amplifiers, communications gear, TV and industrial items.

COLOR TV—RECEPTION AND DECODING TECHNIQUE, by B. W. Osborne. Hart Publishing Co., 510 Sixth Ave., New York, N.Y. 9 1/4 x 6 1/4 in., 168 pp. Hardcover, \$12.50.

Presents circuits for color TV reception from antenna input circuit through decoder. Also introduces decoder design for color systems in use in different countries. Text will serve as an introduction to assessment of TV circuit performance in terms of waveform design. Covers signal-to-noise problems of vhf and uhf TV reception. Includes receiver input circuit, design for low distortion in the luminance and chrominance channels. NTSC decoder, PAL decoder, SECAM and NIR decoders, applications of waveform measurement techniques. R-E

NOVEMBER 1970

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(continued on page 94)

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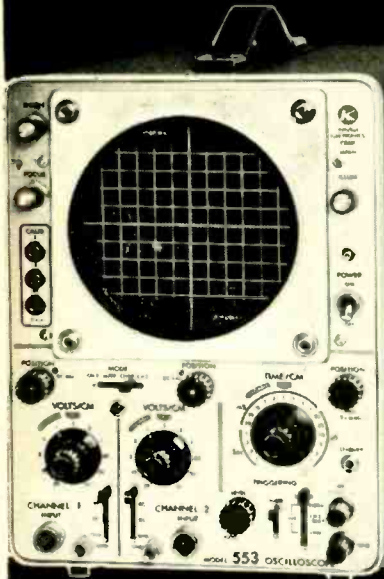
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- **Designing Semiconductor Hi-Fi Stereo Amplifiers**
First in a series of articles that shows how to custom-design the features you want into a solid-state amplifier
- **Darkroom Temperature Monitor**
Know what your solution temperatures are instantly and precisely
- **Building Solid-State Power Supplies**
Matt Mandl tells how to make the power supply you need for that special application

PLUS . . .

Jack Darr's Service Clinic
Your Career In Electronics
Kwik-Fix Troubleshooting Charts

NEW PRODUCTS

(continued from page 74)

CONTACT OVERHAUL KIT, for TV tuners, contains cleaning, polishing, and lubricating materials for switch contacts in all color and black & white TV tuners. Includes aerosol spray of degrea-



ser/cleaner for high pressure removal of "muck" from the surfaces with minimum of frost. An aerosol spray of lubricant completes the kit. Comes with instructions . . . \$5.50.—Castle TV Tuner Service, 5710 N. Western Ave., Chicago, Ill.

Circle 40 on reader service card

NEXT MONTH

R-E's Logic Laboratory is big news. Use it to discover the wonders of computer logic operations. Conduct experiments at home. Learn what makes computers tick.

NEW LITERATURE

All booklets, catalogs, charts, data sheets and other literature listed here with a Reader's Service number are free for the asking. Turn to the Reader's Service Card facing page 102 and circle the numbers of the items you want. Then detach and mail the card. No postage required!

PANEL METER CATALOG No D-70. Twenty pages packed with panel meter data. 1½", 2½", 3½", 4½", 5½", 6½", 7½", & 8½" models are described. Special units, including edgewise meters; wattmeters, and suspension meters are also included.—Triplett Corporation, Bluffton, Ohio 45817

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ELECTRONIC ENGINEERING CAREERS for experienced electronic technicians are spelled out in a special 22-page booklet. Prepared for the person "planning his future", this fact-filled booklet details opportunities in the field of electronics today. It tells how to get the additional education required to take advantage of "the need for highly qualified people who really understand electronics".—Grantham School of Engineering, 1550 N. Western Avenue, Hollywood, Calif. 90027

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ELECTRONIC KITS CATALOG, EICO 1970. Features do-it-yourself kits and factory-assembled electronic equipment. Lists color organs, strobes, translators, solid-state test instruments, electronic project kits, amateur shortwave equipment, Cortina stereo components, oscilloscopes, generators, power supplies, multimeters, probes and many others. Also includes an index of authorized dealers.—Eico, 283 Malta St., Bklyn, New York 11207

R-E

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AH6	6	4-5/8	2-5/8	3/4	8	1.79
AH6R	6	4-1/2	2-5/8	11/16	8	2.95
AH10	10	4-3/4	2-3/4	1-3/16	16	2.29
AH10	10	4-3/4	2-3/4	1-3/16	16	4.95
AH12	12	5-3/4	2-5/16	1-1/16	18	2.79
AH34	34	9-1/4	3-1/8	1-3/8	52	4.95
AH60	60	11-1/4	5	1-3/8	104	5.95

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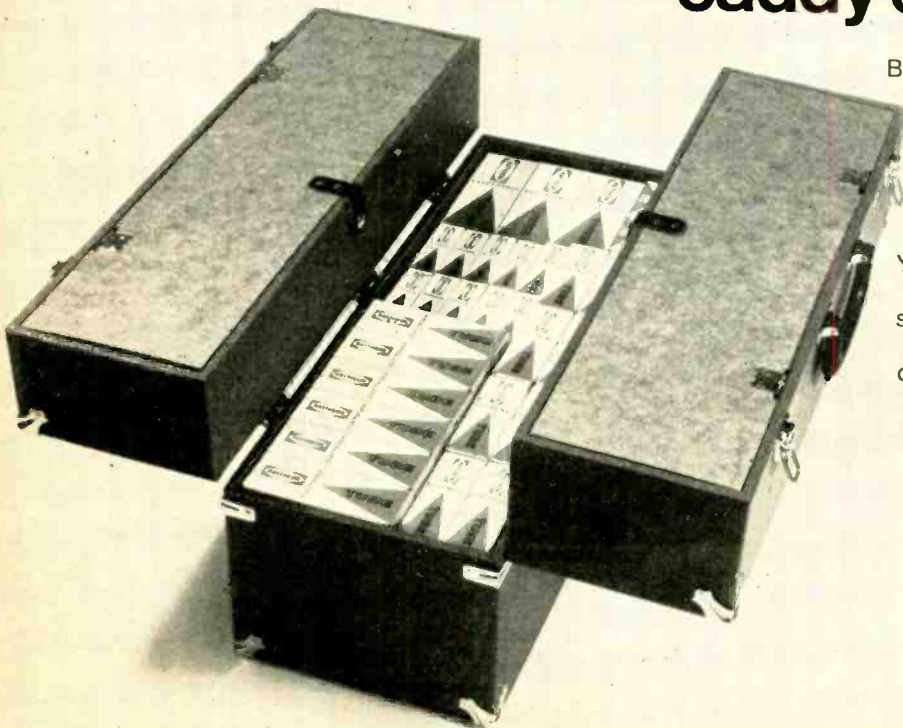
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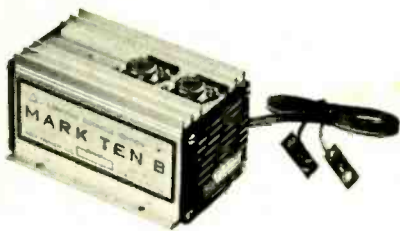
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Circle 73 on reader service card

IN THE SHOP

(continued from page 17)

times the normal current. Likely suspects, an internal leakage in the transistor or incorrect bias voltage on the base. So *that's* where we should start looking for the trouble.

We must learn to think in terms of "What is the *normal* reduction in this circuit?" In tube circuits they were all the same—more negative grid voltage, less plate current. In transistor circuits, a more-negative base voltage means less collector current *only* if the transistor is an npn! A negative voltage is a *reverse* bias (in the direction which *reduces* collector current) for npn's. For pnp's, a more-positive voltage reduces the collector current, since a positive voltage is reverse-bias for these.

I know that this is going to be hard to do, for some of us "set in our ways" old-timers. In fact, every now and then you'll probably find "up" or "down" sneaking into this space! The habits of a lifetime aren't easy to break; I write "Hertz" but I still say "cycles!" However, if you can discipline your thinking to use these terms, it will be a big help in dealing with circuit reactions in modern electronic equipment. And, frankly, the way things are now, we can use all the help we can get!

R-E

Service Clinic

By JACK DARR
SERVICE EDITOR

This column is for your service problems—TV, radio, audio or general and industrial electronics. We answer all questions individually by mail, free of charge, and the more interesting ones will be printed here.

If you're really stuck, write us. We'll do our best to help you. Don't forget to enclose a stamped, self-addressed envelope. Write: Service Editor, Radio-Electronics, 200 Park Ave. South, New York 10003.

Negative Pix, Old Conar TV

I've got a very weak negative picture on an old Conar 70 TV that I'm fixing up.—J. B., Dabneys, Va

Check the 6AS8 used as the 3rd video i.f. If the diode (video detector) has mount-leakage or is shorted that'll do it. Also see if the video output amplifier plate load resistor hasn't gone away up in value. This can cause the same symptoms too.

Intermittent color flashes

This color set works perfectly most of the time; every now and then the screen suddenly flashes a bright

blue. Sometimes, I can tap the chassis and make it flash. Changed tubes, checked picture tube, etc. Do you think this could be an intermittent heater-cathode short in the CRT?—R. G., Chickasha, Okla.

Possible, but let's exhaust all of the cheaper alternatives first! For one common cause of this symptom, check the plate connections to the B-Y color difference amplifier tube. If, as usual, there's a tiny rf choke in there, this can be intermittently opening. This lets the plate voltage jump away positive, and turns the blue gun hard on, causing the flash. Intermittent solder joints on the PC board in the same circuit can also cause identical symptoms.

This can affect any color, of course, not blue alone.

Fanout of R,G,B lines

The red, green and blue vertical lines fan out like a palm tree in this Silvertone color TV chassis. Controls don't seem to have a lot of effect on it. Top and bottom of the screen. What's going on?—A. M., Lake Charles, La.

This has been known to happen in these sets when the vertical centering control opens. This upsets the vertical convergence waveforms. Easy to check; just see if the vertical centering control will move the picture up and down. If it won't, there you are.

Yoke replacement

I have a Sharp TSP109A, with a bad yoke. Can't find a replacement listed for it, and can't get one from the company. Do you know of a U.S. manufacturer who makes one?—D. K., New York.

The Merit MDF-127 will replace this one. Your distributor may have to order it specially, but it'll fit.

Faint vertical bars

I've got faint vertical bars all the way across the screen, on a Sylvania D12 color chassis. These seem to show up mostly on black and white pictures. Checked tubes, etc. No help.—J. Q., Bangor, Maine.

Check the horizontal blanking amplifier transistor, Q604 (Q14 in photofacts). If you find the same pulse waveform on both base and emitter, there you are. This transistor is leaky. Replace it.

Hot horizontal output transistor

I've got a mysterious ailment in a little transistor TV. The horizontal output transistor gets very hot, I have no high voltage, and I get too much current. Everything else seems OK: no shorts in the flyback circuits, and the supply voltage is a little low. What's going on here? Tried a new

output transistor, and it gets hot too.—P. D., Tucson, Ariz.

Check the frequency of your horizontal oscillator. The drive pulses in transistor horizontal output stages must be of the right width, to get the correct ON time for the output transistor. They must also be at the right frequency. If the oscillator is running much too fast, you'll have pulses at the input, but there will be too many of them. This has the same effect as a too-wide pulse: it keeps the output transistor turned on too long, thus drawing too much current. Check with scope, comparing to video signal.

One line or three?

When I throw the switch to the

service position in color sets, I often get three separate lines. I thought you were supposed to get a single white line when you had the screens set correctly. How come?—R. G., Altoona, Penna.

When you set the switch to "service," you kill the vertical output stage. So, you lose your vertical dynamic convergence waveforms. Therefore, the three lines may or may not cover each other. Pay no attention to this; just set the three lines to equal intensity and turn it back to normal. Personally, I like to set screens so that the lines have just gone out. I think you get a more accurate adjustment by doing this. Others differ and have their own favorite adjustment. **R-E**

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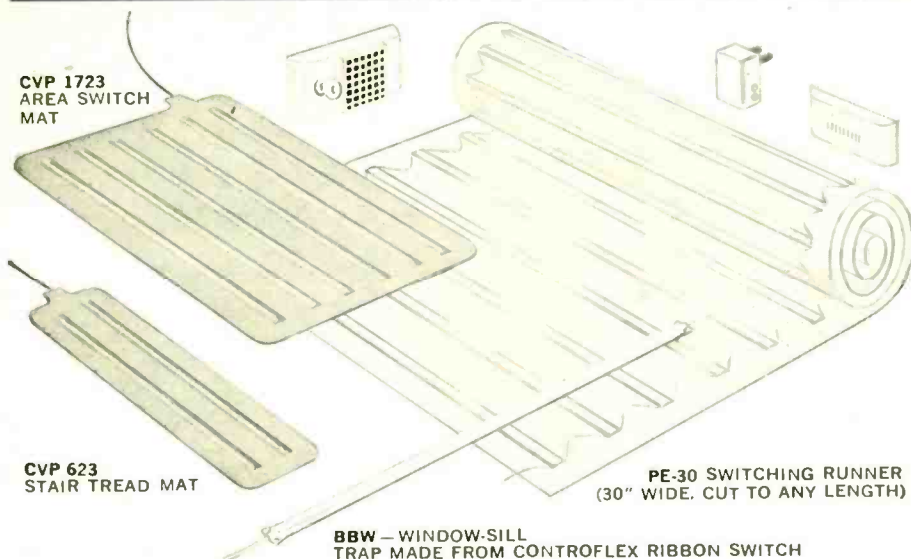
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LOGIC DESIGN COURSE

Recently introduced by the Center for Technical Development, a new Ohio-based company, the Logic Design Program DDP-1 consists of a 6 x 10-inch PC board, a 13-lesson loose-leaf notebook along with 13 programmed learning tests and 10 breadboard project instruction sheets. All components for the lessons are provided with the kit, and the parts may be removed from the push-in terminals for repeated use.

By lesson three, you've derived the alpha and beta formulas for transistors, covered the forward and reverse bias characteristics of diodes, and have been introduced to transistor load curves. Lesson four looks at design procedures for simple inverters, showing how to find I_C , I_B , R_B and R_{BB} . Five is a study of emitter follower operation. Starting with lesson four, separate breadboard lesson sheets tell you how to install components on the PC board to measure (a voltmeter is needed) and observe what you've been studying.

The remaining lessons (6-14) describe logic circuits, starting with diode gates and truth tables. Diode transistor logic (DTL) and resistor transistor logic (RTL) are covered in seven and eight—with corresponding breadboard projects. The ON or OFF states of the circuits is indicated by small lamps, operated with two switches premounted on the board.

Next comes NAND/NOR gate design and set/reset flip-flops. Here's how a typical design example goes: "Design an SR flip-flop, using NOR gates capable of driving 6-volt, 150-ohm lamps. Use germanium transistors type 2N404. Make $V_{BB} = +3V$ and $R_C = R_L$." The lesson shows how to complete this design. At this stage all four of the premounted transistors are used in the breadboard projects.

The final four lessons (11-14) are on bistable and monostable flip-flops and Schmitt trigger design. Also covered are differentiation networks and commutating capacitors. A scope is useful—but not necessary—with these lessons to observe waveforms.

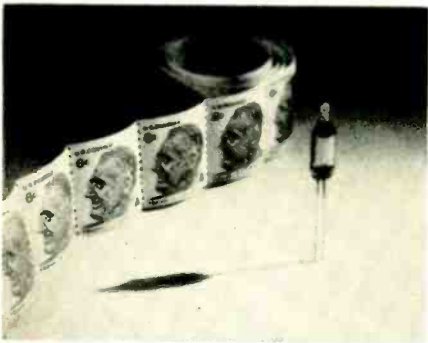
The course packs a lot of basic solid-state design theory into each of the brief lessons (each one is about 6 pages long). Examples and solutions for important design principles are provided in the lesson and tested in the sheets of the programmed course. The breadboard projects provide practice at wiring up components from simple schematics, and classroom instructors should find the non-permanent feature useful. Cost of the course is \$99.—John Williams

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NEW TUBES AND SEMICONDUCTORS

LONG-LIFE NEON LAMP

The A297 is a new neon lamp specifically designed for long-life use in a seven-bar numeric display. It is constructed with 7-mm electrodes in a glass envelope considerably longer than is necessary for an ordinary neon lamp. The longer glass envelope and shorter electrodes eliminates the



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The Signalite A297 has a breakdown voltage rating of 215 volts max and a maintaining voltage rating of 128 to 148 volts. Design current is 5 mA. The lamp is in a T2-size envelope that is 27/32 inch long and 0.244 inch in diameter.

For information, contact Signalite Inc., Neptune, N.J.

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The 2N5787 through 2N5792 are six new low-cost Fairchild SCR's with forward and reverse voltage ratings from 30 to 400 volts. They are electrical equivalents of the 2N5060 through 2N5063 and the Texas Instruments TIC-44 through TIC-47. Gate sensitivity is 10 μ A (typical) and on-state voltage drop is 1.0 volt (typical at 200 mA). All have a maximum current rating of 500 mA. Housing is Fairchild's TO-106 package, which has the same circular pin configuration as the TO-18 metal can. Guaranteed to operate at junction temperatures of -65° to $+125^{\circ}$ C. Ratings and prices (1 to 99) of these Fairchild devices are: 2N5787, 30 V, \$0.77; 2N5788, 60V, \$0.83; 2N5789, 100 V, \$0.96; 2N5790, 200 V, \$1.09; 2N5791, 300 V, \$1.61 and 2N5792, 400 V, \$2.00. **R-E**



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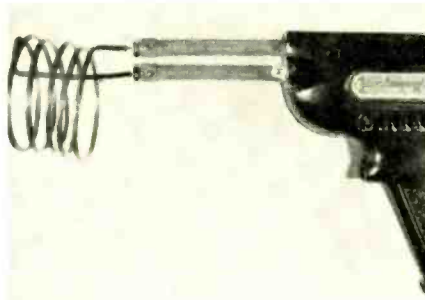


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DEMAGNETIZER

A simple tape head demagnetizer can quickly be made by attaching No. 14 household wire formed into a two-



or three-inch coil to your pistol grip soldering gun. A reminder for this as well as all demagnetizers, always withdraw the field from the head before turning power off.—John F. Agee

REMOVING PC COMPONENTS

When you have to remove multiple-terminal components like tube sockets; i.f. transformers; etc. from printed circuit boards use this trick. Rather than heating each terminal successively and hoping that the solder remains molten until all the terminals are heated, form a piece of No. 12 or

No. 14 wire to fit down on the terminals of the part to be removed. Tin the wire and make sure it touches all the terminals. Then apply heat to two or three places along this "heat sink" with your soldering gun. The part may then be removed with a minimum of heat—avoiding needless damage to circuit board or component.—H. Josephs

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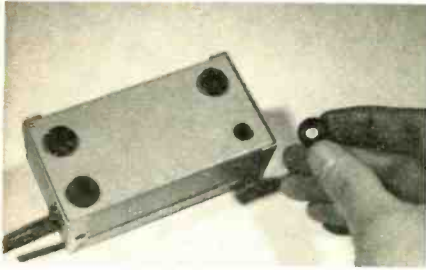
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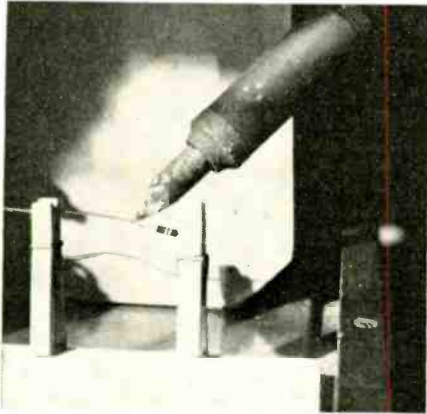
You can be saved by No. 12 or 14 bus bar wire. Cut a piece about 5" long, and shape into U form. Insert into the gun as usual and tighten. Then squeeze the tip in a vise so that the wires are only about 1/4" apart (near the U curve only). Then tin and use.

The substitute heats just like a ready-made tip and will give excellent results. It won't last nearly as long,

however. Bus-bar is sold at radio supply store, in 3-ft lengths and on 100-ft spools.—I. Queen

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clothespins at an angle, on a small wood block. With the wire and component held in the grasp of the clothespins, your hands are left free to handle soldering iron or gun and the solder.—S. Clark. R-E

DID YOU MISS?

Back on page 59 is a story about a new IC that may someday replace the i.f. strip in all radio receivers. Do you know how it works?



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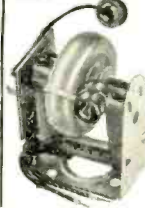
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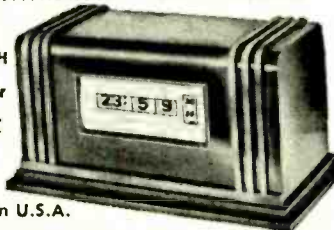
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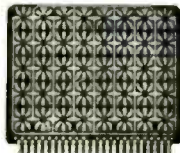
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(continued on page 97)

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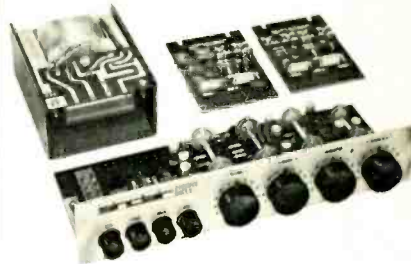
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sinclair PROJECT 60 Modular Stereo



Sinclair Project 60 modules may be used together or separately in home or professional applications requiring high performance at modest cost. Complete Project 60 systems come supplied with necessary hardware to enable you to easily install the modules in cabinet or enclosure of your choice. The Project 60 modules are fully guaranteed for TWO YEARS. A stereo FM tuner will be available shortly. If you have not found the amplifier which meets your installation or performance requirements, the PROJECT 60 SYSTEM by Sinclair may be your answer.

The Sinclair Z-30/50 audio amplifier modules are of advanced design, using silicon epitaxial planar transistors for ruggedness and reliability. The Z-50 is basically the same as the Z-30 but uses higher current output transistors. The Z-30/50 may be used in numerous audio applications such as electronic music and instruments, sound reinforcement and laboratory work in addition to home stereo.

POWER OUTPUT: Z-30, 15 watts RMS; Z-50, 30 watts RMS; Both into 8 ohm loads.

FREQUENCY RESPONSE: 20 Hz* to 300 KHz ±1 db at normal level. *adjustable to meet installation requirements.

DISTORTION: Less than .02% at 1 KHz at any level up to maximum rated output.

OUTPUT IMPEDANCE: 3 to over 16 ohms. . . . May be used with electrostatic speakers.

SENSITIVITY: 250 mV high impedance.

OPERATING VOLTAGE: As low as 8 Vdc with reduced output.

SIZE: 3 1/2 x 2 1/4 x 1/2 inches

Z-30 \$15.95 Z-50 \$18.95

The Stereo 60 module is a matching unit in performance for use with the Z-30/50 modules but may be used with other amplifiers. The Stereo 60 features low distortion (less than .05%), three inputs, tape output, RIAA equalization within 1 db on magnetic input and operates over a wide voltage range from 9 to 50 Vdc. Size: 8 1/4 x 1 1/2 x 4 inches. \$29.95

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HF FILTER: Variable 5 KHz to 28 KHz.

LF FILTER: Variable 25 Hz to 100 Hz.

SINCLAIR ACTIVE FILTER UNIT \$20.95

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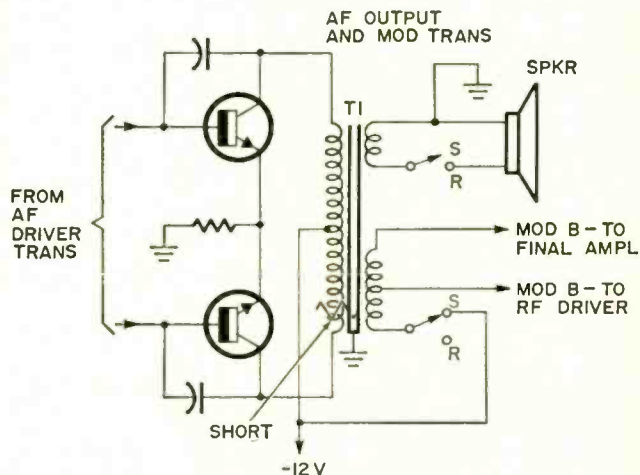
Circle 86 on reader service card

**CB Troubleshooter's
Casebook**

Compiled by
Andrew J. Mueller*

Case 1: Radio is dead. New batteries go dead when the switch is on for 30 seconds.

Common to: Midland 13-775B

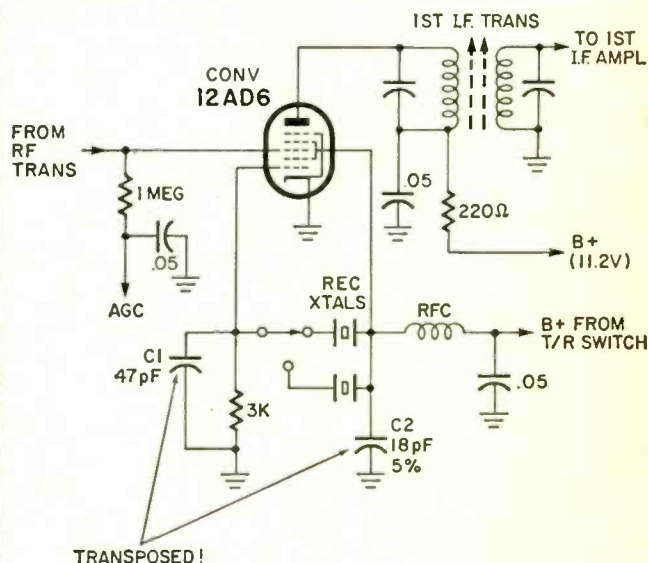


Remedy: Replace audio transformer, T1.

Reasoning: The audio output transformer primary has shorted to the core. This connects battery voltage directly to ground. Replacing T1 will restore normal operation.

Case 3: Radio receives poorly. Transmit is ok.

Common to: Pearce-Simpson Sentry



Remedy: Exchange C1 and C2

Reasoning: C1 and C2 influence the local oscillator injection voltage. From the factory these parts were reversed, reducing receiver sensitivity. When they are properly installed, sensitivity will return to normal. The schematic shows their proper locations. **R-E**

*Service Manager, Tel-Air Communications, Inc., Pewaukee, Wis.

ALARM SYSTEM DIRECTORY
(continued from page 95)

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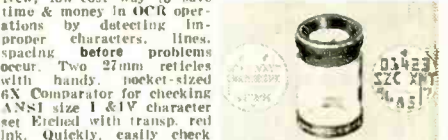
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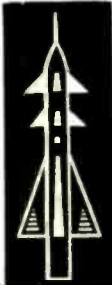
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400	1.40	2.20
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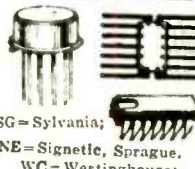
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<input type="checkbox"/> 903	Dual 3 input gate	5
<input type="checkbox"/> 903*	Dual 3 input gate	F
<input type="checkbox"/> 904*	Dual half adder	N
<input type="checkbox"/> 908	Full adder	5
<input type="checkbox"/> 909	Buffer	5
<input type="checkbox"/> 910	Dual two input gate	5
<input type="checkbox"/> 912	Half adder	5
<input type="checkbox"/> 913	Shift register	5
<input type="checkbox"/> 914/921	Dual 2 input gate	5
<input type="checkbox"/> 925	Dual 2 in. w/expand	N
<input type="checkbox"/> 926	Hi speed 923	F

Type	Description	Case
<input type="checkbox"/> 960	Dual buffer	S
<input type="checkbox"/> 923/940	J-K flip flop	S
<input type="checkbox"/> WC210**	2-3 in line driver	F
<input type="checkbox"/> WC215**	J-K flip flop	F
<input type="checkbox"/> WC226**	Triple 3 In. gate	F
<input type="checkbox"/> WC227**	10 In. Diode expand	F
<input type="checkbox"/> WC261**	Dual 4 in. gate	F
<input type="checkbox"/> WC266**	Quad 2 input gate	F
<input type="checkbox"/> SG-43	Dual 3 In. gate	N
<input type="checkbox"/> SG-123	Expandable 8 In. gate	N
<input type="checkbox"/> SG-163	Triple 2 in. bus driver	N
<input type="checkbox"/> NE-124**	R-S-T binary	F



Code: * 2 identical ICs in 1 pak, ** DTL, otherwise RTL. F=flat pak; N=dual inline; 5=TO-5 case.

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<input type="checkbox"/> SN7473N	Dual J-K master slave flip flop	1.49
<input type="checkbox"/> SN7474N	Dual 'D' type edge triggered flip flop	1.49
<input type="checkbox"/> SN7475N	Quad bi-stable latch	3.95
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<input type="checkbox"/> SN7302F*	Dual J-K flip flop with preset	1.29
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* DTL, others TTL. F=flat pak. N=dual inline

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<input type="checkbox"/> 2-UNIJUNCTION transistors 2N4891	\$1.19
<input type="checkbox"/> 1-IG MOS FET P channel 2N4353	\$1.00
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<input type="checkbox"/> 3-TRIGGER diodes for Sers, Triacs	\$1.00
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<input type="checkbox"/> 2-WATT G-E PA237 audio amp	\$3.98
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<input type="checkbox"/> 1-15-AMP Triac 200 PRV	\$1.49
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30 IC CIRCUITS

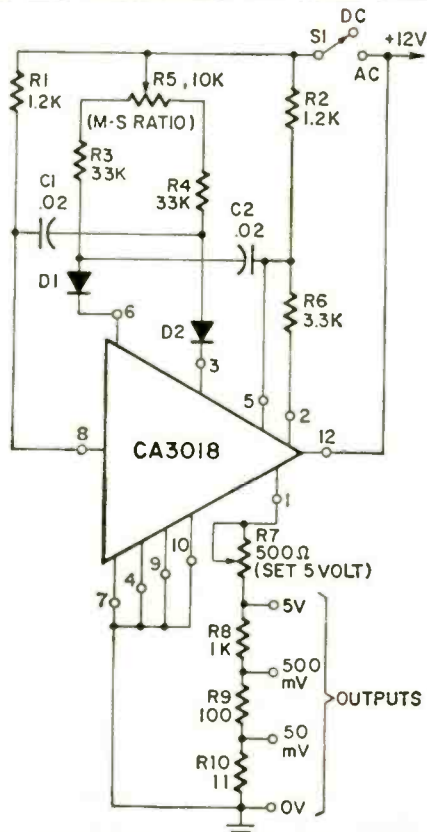
(continued from page 70)

Finally, Fig. 32 shows how the IC can be wired as an ac/dc oscilloscope calibrator. In this circuit, Q1 and Q2 are wired as a free-running (astable) multivibrator, operating at a frequency of about 1 kHz. Diodes D1 and D2 protect the transistors against reverse base-emitter breakdown, and R5 allows presetting the mark-space ratio.

Output is clipped at about 6.5 volts by Zener-connected Q3, and is passed on to the R7-R10 attenuator network via emitter follower Q4. In use, R7 is adjusted so that a signal of 5 volts peak appears at the R7-R8 junction. Calibration voltages of 50 mV, 500 mV and 5 volts are then available at the outputs.

When S1 is set to the DC position, the multivibrator is switched out of circuit, but Q3, Q4 and the attenuator network are still operating. Dc calibration voltages of 50 mV, 500 mV and 5 volts are now available at the attenuator outputs. Due to Zener stabilization by Q3, these calibration voltages are virtually independent of supply line variations.

The unit can be calibrated initially by turning S1 to the DC position and connecting a sensitive (20,000 ohms/volt or greater) meter between the 5-volt terminal and ground. Then adjust R7 for precisely 5 volts output, and the unit is ready to use as a scope calibrator. R-E



NOTE: D1, D2 = GENERAL PURPOSE SILICON DIODES
R8, R9, R10 = 5% TO 1% TYPES, DEPENDING ON CALIBRATION ACCURACY REQUIRED.

Fig. 32—OSCILLOSCOPE CALIBRATOR provides dc and peak-to-peak outputs.

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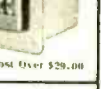


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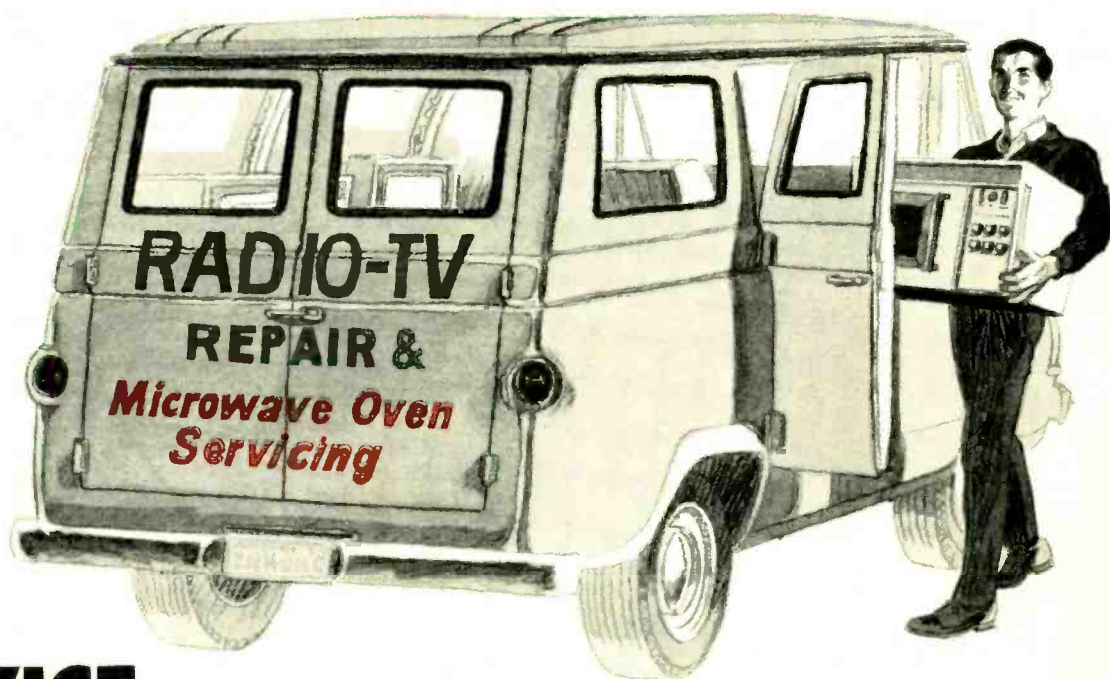
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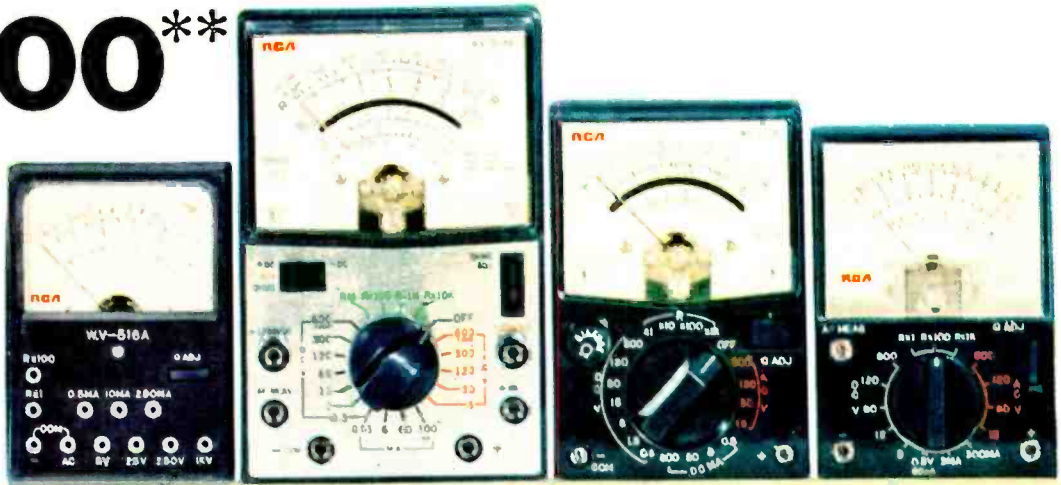
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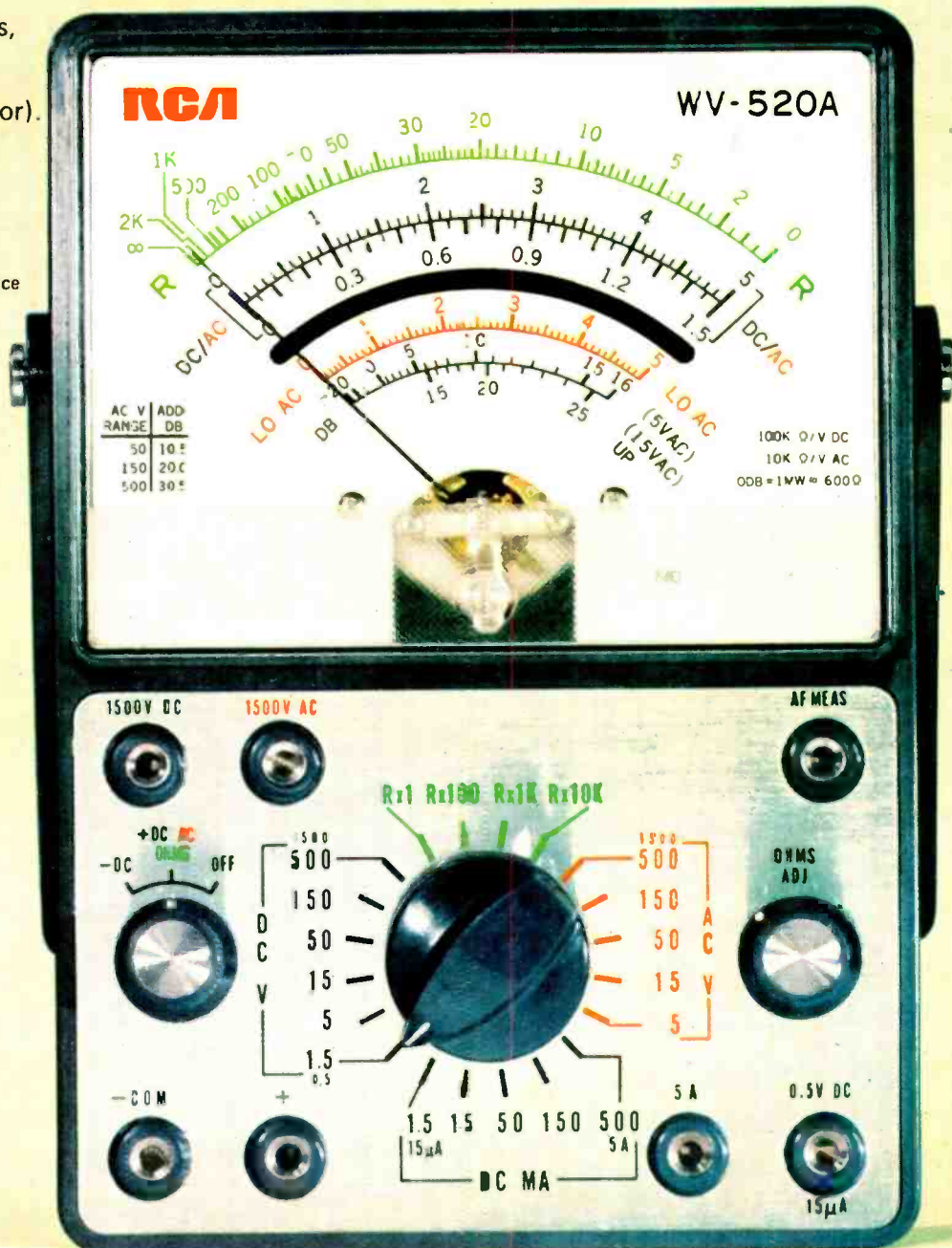
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