# Popular Electronics 

# Build an IC Voltage Regulator for Your Car Tune In Africa for Exciting English Broadcasts Ultrasonic Detector Reveals Sounds of Insects How to Get Hi-Fi Sound in Any Auto! NEW SPEAKERS, TAPE DECKS, FM RADIOS, POWER BOOSTERS, EQUALIZERS + INSTALLATION TIPS 



Tes


New Quartz-Lock System In This Infinity Speaker - New Induction Tweeter Issue

# COBRA RE-INVENTS THE REMOTE CB. 

The first remote CBs were nothing more than a CB transceiver that you locked in your trunk and an oversized mike that could barely fit in your hand.

Such was the state of the art when the first remotes were introduced. Which was why Cobra spent till now improving the state of the art.

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The receiver has automatic gain control, switchable noise limiting, plus Dual-Gate Mosfet and Monolithic Crystal Filter to keep interference to a minimum. So the voice you hear always comes through loud and clear.

The streamlined mike puts all the controls at your fingertips. Speaker, channel selector and squelch are built right in. So there's no fiddling around while you're driving around.

And with Cobra's reputation for building them right and our nationwide network of Service Centers making sure they stay that way, you can be pretty sure that nobody's ever going to improve on the 62XLR.

Cobra may not have been the first to make a remote. But we were the first to do it right.


Punches through loud and clear. Cobra Communications Products DYNASCAN CORPORATION 6460 W. Cortland St., Chicago, Ilinois 60635 Write for color brochure


## fact:

 the IV does more... much more!

The creation of the new V15 Type IV is a tour de force in innovative engineering. The challenge was to design a cartridge that would transcend all existing cartridges in musical transparency, technical excellence, and uniformity.
The unprecedented research and design disciplines that were brought to bear on this challenge over a period of several years have resulted in an altogether new pickup system that exceeds previous performance levels by a significant degree-not merely in one parameter, but in totality.

In fact, this pickup system has prevailed simultaneously over several extremely difficult music re-creation problems which, until now, have defied practical solutions. Most of all, this is an eminently musical cartridge which is a delight to the critical ear, regardless of program material or the rigorous demands of today's most technically advanced recordings.

## THE V15 TYPE IV OFFERS:

- Demonstrably improved trackability across the entire audible spectrum-especially in the critical mid- and high-frequency areas.

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*Cartridge-tone arm system trackability as mounted in SME 3009 lone arm at I gram tracking force.

- Dynamically stabilized tracking overcomes record-warp caused problems, such as fluctuating tracking force, varying tracking angle and wow.
- Electrostatic neutralization of the record surface minimizes three separate problems: static discharge; electrostatic attraction of the cartridge to the record; and attraction of dust to the record.
- An effective dust and lint removal system.
- A Hyperelliptical stylus tip configuration dramatically reduces both harmonic and intermod ulation distortion.
- Ultra-flat response -individually tested to within $\pm 1 \mathrm{~dB}$.
- Lowered effective mass of moving system results in reduced dynamic mechanical impedance for superb performance at ultra-light tracking forces.

For more information on this remarkable new cartridge, write for the V15 Type IV Product Brochure (ask for AL569), and read for yourself how far Shure research and development has advanced the state of the art.


Shure Brothers Inc., 222 Hartrey Ave., Evanston, IL 60204 In Canada: A. C. Simmonds \& Sons Limited Manufacturers of high fidelity components, microphones, sound systems and related circuitry.

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MAX- 100 is a cinch to use. 1 g ves pou continuous readings from 2 Hz -o a guaranteed 100 WHz , with 8 digit accuracy. Fast readings with $1 / 6$-sec. update and 1 -sec. sampling -ate. Precise readings, Jerived from a crystal-controlled time base with 3ppm accuracy. High-sensiziv ty eadings from signals as law as 30 $\mathrm{m} V$ with diode overload protection Ip to 200 V peaks.

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and other digital circuits. Reparr of depth sounders and fisr spotters. Troubleshooting ultrasonic rerrote con:role. For these, and hundrexts of other a pplications, you'll find $t$ ndispensab.e.

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Editorial

## THE ELECTRONIC CURSE!

An editor from New York magazine telephoned me today and said, "l'm calling you because I'm doing a feature story on laziness." Allowing me a few seconds to recover from that opening line, she asked if I know anyone locally whose home is equipped to do most anything at the push of a button or on voice command.

Have our electronic servants made us lazy, as some people seem to believe? । think not, in the true sense of the word. But others may view this differently.

Many Japanese, for example, feel that some electronic aids can make a person sluggish, physically or mentally. That's one reason why the use of the abacus or soroban is so widespread in Japan (some 25 million Japanese use this device today). Interestingly, the abacus is considered to be a valuable aid for increasing one's mental alertness. Of course, the Japanese haven't forsaken electronic calculators. They've simply added a mechanical device to their ownership of an electronic product, much as many Americans use bicycles as well as automobiles.

The importance that such mental-aleriness training holds for the Japanese is underscored by the abacus instruction giver, every year to some two million elernentary school students. Furthermore, abacus instruction is mandatory for business administration students (about 300,000), and more than 300,000 Japanese are enrolled in special abacus schools. Every year, $11 / 2$ million Japanese take exams in the use of the abacus to get a degree in proficiency level. Sort of a "black belt" in abacus. Local abacus contests are held in Japan, with winners competing in an annual soroban contest.

Primitive product or not (the abacus pre-dates Christ), tests prove that experienced abacus users can match the speed of electronic calculator users for addition, subtraction, division and multiplication. More important, though, Japanese company spokesmen claim that the mental exercise garnered by using an abacus develops a cerebral agility that is beneficial in business. Abacus users are said to be able to figure out customers' change in their heads with amazing speed and accuracy, for example; at Hitachi, Ltd., computer key punch operators are chosen on the basis of abacus proficiency, etc. To see how you fare against a skilled abacus user, time yourself with an electronic calculator for answers to 30 problems at the level of $8546 \times 38$. (A recent 18 -year-old soroban champion required only $11 / 2 \mathrm{~min}$ utes on an abacus to solve 30 such calculations when he entered high school.)

I'm not suggesting that one forget about using an electronic calculator, of course. But in line with electronics being a possible curse in some respects, let's not forget the basics of creating and assembling these and other electronic devices. How you approach this makes the difference between a kit builder and an electronics experimenter. There's nothing wrong with the former, but to really learn about electronics, don't forget the latter! And in response to those people who think that electronics makes one lazy, electronics enthusiasts all know that it takes considerable work (and knowledge) to assemble or design electronic equipment so that we can take it easy.



# Personal Computers \& Microprocessing 

> Here are two inexpensive programmed learning courses designed to keep you up-to-date in digital electronics.

Design of Digital Systems - six volumes

The products of digital electronics technology will play an important role in your future. Calculators, digital watches and TV games are already commonplace. Now, microprocessors are generating a whole new range of products. Personal computers will be in widespread use very soon. Your TV, telephone and computer will combine to change your children's education, your jobyour entire way of life.

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These courses were written by experts in electronics and learning systems so that you could teach yourself the theory and application of digital logic. Learning by selfinstruction has the advantages of being faster and more thorough than classroom learning. You work at your own pace and respond by answering questions on each new piece of information before proceeding.
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## CONTENTS

The contents of Design of Digital Systems include:
Book 1: Octal, hexadecimal and binary number systems; representation of negative numbers; complementary systems; binary mulitplication and division.
Book 2: OR and AND functions; logic gates; NOT, exclusive-OR, NAND,NOR and exclusive - NOR functions; multiple input gates; truth tables; DeMorgan's Laws; canonical forms; logic conventions; Karnaugh mapping; three-state and wired logic.

[^0]Book 3: Half adders and full adders; subtractors; serial and parallel adders; processors and arithmetic logic units (ALUs); multiplication and division systems.

Book 4: Flip-flops; shift registers; asynchronous counters; ring, Johnson and exclusive -OR feedback counter; random access memories (RAMs); read-only memories (ROMs).

Book 5: Structure of calculators; keyboard encoding; decoding display data; register systems; control unit; program ROM; address decoding; instruction sets; instruction decoding; control program structure.

Book 6: Central processing unit (CPU); memory organization; character representation; program storage; adaress modes; input/output systems; program interrupts; interrupt priorities; programming; assemblers; executive programs, operating systems, and time-sharing.

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## CONTENTS

Digital Computer Logic and Electronics is designed for the beginner. No mathmetical knowledge other than simple arithmetic is assumed, though you should have an aptitude for logical thought. It consists of 4 volumeseach $111 / 2^{\prime \prime} \times 814^{\prime \prime}$-and serves as an introduction to the subject of digital electronics.

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## N <br> Letters

## "ROAOMATE" REVISITED

With reference to the "Roadmate" CB converter featured in the October 1976 issue of Popular Electronics, an r-f signal can leak from the antenna through C1, R6, and LED1 to the power system of the vehicle. Moreover, $r-f$ noise from the ignition system can enter the converter via the opposite path. Addition of a $10-\mathrm{mH}$ miniature choke in series with R6 will overcome this problem and result in a noticeably higher $S / N$ for the converter. - Mark W. Fleming, Watertown, MA.

## getting the name straight

Although the article "How to Upgrade a Basic ELF Microcomputer" (February 1978) appears to be technically sound and proficiently written, the contention that "Feelin" Groovy" was penned by Neil Simon looks very odd The author of "Feelin" Groovy" is Paul Simon. -Donald L. Wallace, Dayton, OH

## UFO RESEARCHER

Having researched the UFO (unidentified flying object) controversy for the past 21 years, I read with interest "Investigating UFO's and Other Magnetic Phenomena' (May 1978). I am very pleased to see some fresh material on the subject. Author George Lawrence's UFO article was most informative -Hayden C. Hewes, Director, International UFO Bureau, Inc., Edmond, OK.

## GREAT TUTORIAL

Thanks for the great article on "How to Design Power Supplies" in the April 1978 issue. It was one of the best tutorial articles on electronics I have ever seen. It combined clearly presented theoretical material, good practical examples (including diagrams and pictures). and practical suggestions for determining values for a given component. -Charles Ragland, San Francisco, CA

## ELF USER'S GROUP

I am certain that many Popular ElecTRONICS readers will be interested to know that we are forming a COSMAC-1802 User's Group for hobbyists who have any of the Elf versions available, including VIP, Infinite, and other 1802-based microcomputers. We will be corresponding, exchanging software and ideas, etc. Membership is free. Interested readers, please write directly to: Patrick Kelly, Box 7162, Los Angeles, CA 90022.

## A COMPONENT CHANGE

With reference to the "Real-Time $1 / 2$ Octave Analyzer" that appeared in the September and October 1977 issues of POPuLAR Electronics, please be advised that there has been a component substitution. An SN76502 was originally specified for IC36. This particular IC has been discontinued. The new part number is TL4441CN.

Also, Table I in the article had two errors in it. The filter capacitors should be labelied C 8 and C28 (not C9 and C29) and should be $0.068 \mu \mathrm{~F}$ (not $0.047-\mu \mathrm{F}$ ). -Richard Marsh

## Out of Tune

In "Microprocessor Microcourse, Part 2" (April 1978) the $Q$ and $\bar{Q}$ outputs of the flipflop in Fig. 13 should be transposed.
in "Elf II ROM Monitor" (March 1978) the connection to pin 17 of a 2102 in Fig. 1 should be to a 2101

In "How to Design and Build Power Supplies, Part I' (April 1978), on page 43 under "Filters," the second paragraph should read the higher frequency of the full-wave rectifier's output is easier to filter." Also the last sentence under the "Filters" heading should read $8000 \mu \mathrm{~F}-$ not 8000 pF .

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40894* President Model "Washington" 40-Ch. AM/SSB CB Base Station
40895 Yaesu Model FRG-7 AM/SSB Communications Receiver
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## New Products

Additional information on new prodacts covered in this section is available from the manufacturers. Either circle the item's code number on the Free Information Card or write to the manufacturer at the address given.

## B\&K Precision Portable Oscilloscope

The B\&K Precision Model 1432 portable, triggered-sweep, dual-trace oscilloscope has a rated bandwidth of 15 MHz and vertical sensitivity of 2 mV /division. Operation can be from 117 or 234 V ac or 12 V dc. Features include: channel $\mathrm{A} \pm$ channel B , 19 calibrated sweep ranges from $0.5 \mu$ s to $1.5 \mathrm{~s}, 5 \times$ magnifier, automatic selection of chopped or alternate display modes, and automatic TV line and frame sync selection. The TTL-compatible $Z$ axis permits intensity modulation of the $3^{\prime \prime}$ CRT over a 1MHz range. Battery overcharge is prevented by an automatic charge-limiting circuit. The Model 1432 is supplied with two 10:1/ direct probes, four accessory tips with carrying pouch, ac and dc power cables, and a viewing hood. Options include a rechargeable battery pack. \$750.
circle no 93 on free information caro

## Hidden Radar Detector Converter

Bel Sales' "Shadow" converter allows concealment of any radar detector under car hoods, says the manufacturer. When $X$ band or $K$ band radar is sensed by the radar detector, the "Shadow" flashes a red

light and sounds a warning beeper. The new device is powered from the car's cigarette lighter. Wiring and a waterproofing kit are included. $\$ 49.95$
clrcle no. 89 on free information caro

## Hustler Quick Mount CB Antenna

The Hustler "Speedy Seizer" mobile antenna is designed for fast, semipermanent mounting on such car-panel edges as doors and trunk lids. A built-in ball joint allows its use on vertical or horizontal edges requiring only a $1 / 8^{\prime \prime}$ gap with a $5 / s^{\prime \prime}$ minimum return behind the panel for clamping. The mount grounds directly to the vehicle body, yet can be readily removed for car washes or concealment. The $46^{\prime \prime}$ long antenna is a

stainless steel center-loaded whip with screwdriver-adjustable tip rod for SWR adjustments. A 17' RG-58 cable with factoryinstalled connectors is provided. \$19.95. circle no. ss on free information caro

## Automatic Garage Door Electronic Lock

Here is an electronic combination lock to activate automatic garage-door openers without using radio control or a key. The three-digit combination has a total of 390

possible combinations; wrong digits disable the system for 30 to 60 seconds. The lock is guaranteed to operate from $-50^{\circ}$ to $150^{\circ} \mathrm{F}$., and requires only two low-voltage wires and two screws for its installation. \$19.95. Address: Domino Engineering Corp., Box 376, Taylorville, IL 62568.

## Five-Volt

Powered
Breadboard
Designed primarily for TTL logic circuits, the Continental Specialties PB-203 ProtoBoard includes a $1 \%$-regulated, 5 -volt dc power supply. The breadboard area includes enough tie points to support fourteen, 14 -pin DIP IC's. Four binding posts

provide additional power and signal connections. The power supply is rated at 1 A , $5 \pm 0.25 \mathrm{~V}$, with 10 mV combined ripple and noise at $1 / 2$ A out, and is protected against short circuits. Dimensions are $61 / 2$ $\times 93 / 4 \times 31 / 4 \mathrm{in} .(16.5 \times 24.8 \times 8.3 \mathrm{~cm}) . \$ 80$.

CIRCLE NO. 91 on free information card

## Teac Open-Reel Tape Deck

Teac's new quarter-track, two-channel A-6600 is an open-reel tape deck with separate erase, record, playback, and re-verse-playback heads. Full IC logic pushbutton control is provided, along with automatic reverse (with sensing foil) and automatic repeat play. A two-capstan servotension system moves tape at speeds of

$3.3 / 4$ and $71 / 2 \mathrm{ips}$. The A-6600 also includes a $20-\mathrm{dB}$ microphone attenuator pad, cue selection, auto space, large and small reel tension switch, individual microphone and line controls for each channel, a master control, independent two-position bias and
(Continued on page 10)

## (i) 0

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equalization controls, large VU meters which can be switched to read up to $\pm 6 \mathrm{VU}$, independent input level controls for mike/ line mixing capability, and a separate output level control. Optional accessories include the RC-80 remote control. $\$ 1300$.

Circle no 92 on free informaton caro

## Portable Radio Direction Finder

The Apelco Marine Electronics DF-10 radio direction finder receives marine and aviation radio beacons, the marine band,

and standard AM broadcasts. The DF-10 features a switchable noise limiter and a bfo. Another tuning aid provided is a sensitivity meter, which doubles as a power cell strength indicator. The solid-state radio direction finder is powered by eight " $D$ " cells. Dimensions are $7.5^{\prime \prime} \mathrm{H} \times 11^{\prime \prime} \mathrm{W} \times 9^{\prime \prime} \mathrm{D}(19.1 \times$ $27.9 \times 22.9 \mathrm{~cm})$. Weight is $8.5 \mathrm{lb}(3.9 \mathrm{~kg})$. $\$ 249$.
clrcle mo bo on free information card

## Rabco Straight-LineTracking Turntable

The Rabco Model ST-8 straight-line-tracking 33-1/3-45-rpm turntable was recently unveiled by Harman-Kardon. The ST-8 plays records the same way the master is recorded; the company claims zero tracking error, zero skating force, and zero sty-

lus overhang. A "Hall-effect" servo-controlled dc motor drives the $2.4-\mathrm{lb}(1.1-\mathrm{kg})$ turntable via a belt. The platter is brought to selected speed from a dead stop in only one revolution, according to the manufacturer. Wow and flutter is said to be $0.05 \%$;
and rumble is rated at -65 dB . Turntable speed accuracy is maintained automatically but the user can also adjust it by $\pm 5.5 \%$. The 6-g low-mass tonearm has automatic lift-off and damped cueing. Other ST-8 highlights include a built-in bubble level with adjustable levelers and touch-sensitive resistance-type switches. Dimensions are $6.8^{\prime \prime} \mathrm{H} \times 16.5^{\prime \prime} \mathrm{W} \times 16.3^{\prime \prime} \mathrm{D}(15.7 \times 41.9 \times$ $41.3 \mathrm{~cm}) . \$ 499$.

CIRCLE NO 85 ON fREE INFORMATION CARO

## Satin Moving-Coil Cartridge

The Satin Model M-117G is claimed to be the only moving-coil stereo cartridge that offers an output great enough to drive amplifiers and receivers directly without the need for a transformer or pre-preamplifier. In addition, the Satin is also said to be the only moving-coil cartridge with userreplaceable stylus assembly. (It is held in place by magnetic force.) The key to the high $3.0-\mathrm{mV}$ output of the cartridge is the use of aluminum ribbon coils and an advanced magnetic structure to concentrate a high magnetic force in a 250-micron gap. The cartridge comes with a $0.2 \times 0.8$ mil elliptical diamond stylus. Frequency range is rated at 20 Hz to $25,000 \mathrm{~Hz}$. Compliance is rated at $12 \times 10^{-6} \mathrm{~cm} /$ dyne. Recommended tracking force is 0.5 to $1.5 \mathrm{~g} . \$ 115$.

CRCLE NO 87 on free information caro

## Chemtronics Solder/ Desolder System

The Chemtronics SD5 Modular Solder/ Desolder System consists of a pound or half-pound spool of MIL-spec solder and a D5 Desolder Wick Dispenser Tool

snapped into the core of the solder spool. The D5 features a $21 / 2^{\prime \prime}$ heat-resistant TefIon probe which is said to allow precise application of the flux-treated wick. The probe tip also aids "webbing" of the wick, which is available in $0.06^{\prime \prime}$ and $0.10^{\prime \prime}$ gauges. The entire SD5 system is refillable and available as separate components. Solder comes in 16-, 18-, and 21 -gauge sizes with $63 / 37,60 / 40,50 / 50$, and $40 / 60$ formulas.

[^1]
## Space Byte Modular Business Computer

Space Byte's new Modular Business Computer is a business-oriented system that includes the Space Byte SB85-16 Terminal Mounted Mainframe with 8085-based CPU and 16 K of read/write memory, a Hazeltine 1500 video display terminal, and an

iCOM 3712 dual floppy disk drive with iCOM FDOS III operating system. All hardware is mounted on an adjustable, rollaway floor stand. The BIZPAK businessapplication software provided is a reportgenerating system written in assembly language, and features interactive program modules for accounts payable and receivable, payroll, and general ledger. There is extensive operator prompting and transparent file maintenance for inexperienced computer operators. Other software available includes Disk Extended BASIC CP/M, and Fortran-80. Hardware and software maintenance contracts are available.

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CIRCLE NO 95 ON FREE INFORMATION CARO
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## Onkyo Digital Synthesized Tuner

Onkyo's Model T-909 FM tuner uses digital processing techniques throughout its circuitry, according to the manufacturer. The T-909 employs a quartz crystal controlled oscillator said to provide a tuning frequency accuracy of $\pm 30 \mathrm{ppm}$. Tuning is accom-

plished in $200-\mathrm{kHz}$ steps, and the T-909 displays the frequency to which it is tuned on front-panel, seven-segment LED's. Among the features of the T-909 is its sev-en-channel, user-programmable memory This allows station selection at the push of a button. Other features include Dolby NR adaptor plug-in provisions and a deemphasis switch, front-end dual-gate MOSFET's, multipath detector terminals, and a "birdie"
(Continued on page 12)

# m-pak 

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Housed handily in the same type of roll-up, plastic-coated, canvas case, the 27-piece 99SMW adds a Weller WP25 professional, pencil-style soldering iron with an extra, wider tip, and a No. 100 wire stripper/cutter. These plus the traditional 99SM tools that thousands of servicemen and technicians have liked so much so long: 20 Xcelite Series 99 quick-change, interchangeable blade tools - popular size nutdrivers, slotted and Phillips type screwdrivers, extension, reamer, regular and stubby handles: diagonal and long nose pliers; thinpattern, adjustable wrench. The handiest handiul of service tools you've ever laid your hands on!
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filter. Controls include SCAN UP, SCAN DOWN, MEMORY, and MUTING. The tuner's specifications are: usable sensitivity of 9.8 $\mathrm{dBf}(1.7 \mu \mathrm{~V})$ mono, $17.2 \mathrm{dBf}(4 \mu \mathrm{~V})$ stereo; capture ratio, 1.5 dB ; alternate channel attenuation, 80 dB ; and a f́requency response of $30-16,000 \mathrm{~Hz}+0.5,-2 \mathrm{~dB}$.

CIRCLE NO 96 ON FREE INFORMATION CARO

## TRS AM/SSB CB Base Station

The TRS Challenger Model 1400 is its new top-of-the-line 40 -channel AM/SSB base station CB transceiver. It features LED numeric channel display and a built-in digital

clock. AM and SSB sensitivity is reported as 0.7 and $0.3 \mu \mathrm{~V}$ at $10 \mathrm{~dB}(\mathrm{~S}+\mathrm{N}) / \mathrm{N}$, respectively. Selectivity is rated at -50 dB for $\pm 10 \mathrm{kHz}$ (AM) and $\pm 2.5 \mathrm{kHz}$ (SSB). PA audio output is 4 W . Front-panel controls and indicators include switchable anl and noise blanker, high and low tone-cut controls, panel-meter dimmer, r-f gain, squelch, and fine-tune controls, plus S/r-i and SWR meters. A front-panel headphone jack and dual speakers are also provided. The ac-powered transceiver measures $5.9^{\prime \prime} \mathrm{H} \times 15.8^{\prime \prime} \mathrm{W} \times 11.4^{\prime \prime} \mathrm{D}(14.9 \times$ $40.0 \times 28.9 \mathrm{~cm}$ ). $\$ 549.95$.
circle no 97 on free information card

## Jensen Spectrum Series Loudspeakers

The Spectrum Series is a new line of home stereo speakers from Jensen Sound Laboratories, headed by the top-of-the-line 550 . The 550 features a $15^{\prime \prime}$ woofer with polyurethane foam suspension, two $31 / 2^{\prime \prime}$ directradiating mid-range drivers, and a $11 / 2^{\prime \prime} \mathrm{My}$ lar dome tweeter. Frequency response is $45-20.000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$, and recommended

power input is $10-90 \mathrm{~W}$ continuous. The Comptrac crossover network provides uniform energy transfer between drivers with minimum phase shift. High- and mid-range frequency level controls are hidden behind the cocoa-colored, free-floating grille. Cabinetry is of hardwood. The 550 measures 31 " $\mathrm{H} \times 19.5^{\prime \prime} \mathrm{W} \times 15.5^{\prime \prime} \mathrm{D}(78.7 \times 49.5 \times 39.4$ cm). $\$ 299.95$.

CIRCLE NO 98 ON fREE, information caro

## Amateur TV Converter

Science Workshop's Model ATVC-10 amateur TV converter enables one to monitor transmissions in the $420-450-\mathrm{MHz}$ band on TV channels 2 through 6 . It connects to the vhf terminals of a standard TV receiver. The ac-powered converter features Varactor-tuned circuits and adjustable r-f gain. Its walnut and beige aluminum cabinet measures $1.8^{\prime \prime} \times 4.3^{\prime \prime} \times 4.1^{\prime \prime}$ $(4.6 \times 10.9 \times 10.4 \mathrm{~cm}) . \$ 49.95$ factory wired; $\$ 39.95$ semi-kit (critical circuits prewired and aligned). Address: Science Workshop, Box 393, Bethpage, NY 11714.

## 3M Instant-Record Cassettes

3M's new instant-record cassettes (IRC) eliminate the possibility of program material loss due to recording on cassette tape leaders. The new tapes feature leaders of low-noise recording tape, allowing recording over the entire length of the tape. The leaders are heavy-duty 1.5 -mil tape, with 1 -mil tape being used on the IRC-30 cassette. This new tape format should prove to be handy in all cassette tape applications, and especially valuable to blind students who tape notes. 3M IRC cassettes are available in C 30 ( $\$ 1.50$ ), C60 ( $\$ 1.80$ ), and $\mathrm{C} 90(\$ 2.50)$ versions.

> CIRCLE NO g9 on free information caro

## Shakespeare Digital Depth Indicator

Shakespeare Marine Electronics' new 400-D digital depth indicator, packaged in a water-tight case, offers interesting functions. One is an audio alarm when a preset depth has been reached. Another provides a sound warning if depth varies from a user-determined range or "window," as might happen when a ship slips anchor. "Fail-safe" programming of the $400-\mathrm{D}$ is said to prevent false readings from being displayed if positive bottom contact is lost. The 400 -D display consists of a $31 / 2$-digit LCD readout, back-lighted for night operation. \$349.95.

[^2]
# Radio Shack's personal computer system? This ad just might make you a believer. 

You can't beat the 4 K system at \$599
... or the step-up
16K system at \$899
... or the fast 4K/printer system at \$1198
... or the Level-II 16K/printer/disk system at \$2385


TRS-80 "Breakthru"
-TRS-80 microcomputer - $12^{\prime \prime}$ video display

- Proiessional keyboard
- Power supply
- Cassette tape recorder - 4K RAM, Level-I BASIC
- 232-page manual
- 2 game cassettes


TRS-80 "Sweet 16"

- Above, except includes 16K RAM


TRS-80 "Educator"

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## So how are you gonna beat the system that does this much for this little? No way!



TRS-80 "Business"

- Above, except
includes 32K RAM, line printer,
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Get details and order now at Radio Shack stores and dealers in the USA, Canada, UK, Australia, Belgium, Holland, France, Japan. Write Radio Shack, Division of Tandy Corporation, Dept. C-05\%, 1400 One Tandy Center, Fort Worth, Texas 76102. Ask for Catalog TRS-80.

#  <br> New Literature 

## B\& K FREQUENCY COUNTER GUIDE

A six-page brochure from B\&K-Precision provides application information and complete
specifications on four B\&K frequency counters. They include models that count up to 520 MHz , universal and autoranging counters, and portable instruments. A selection of frequency counter accessories are also described, including probes, power adaptors, carrying cases and an RF signal tap. Address: B\&K Precision, Sales Dept., 6460 W Cortland Ave., Chicago, IL 60635.

## EDMUND HOBBYIST CATALOG

Hundreds of new products are listed in Edmund Scientific's 165-page spring catalog for experimenters, do-it-yourselfers, hobbyists, and students. Some of these products in-
clude the Astroscan 2001, the 6 -in. f/6 and the 8 -in. $\mathrm{f} / 5$ telescopes; a Solar Water and Beverage Heater; a manual on "How to Reduce Heating Costs Without Alternative Energy Sources;" and a Power Miser that makes use of trapped air in a/c units which is usually wasted. Other new products featured are: a low-cost TV projection lens; UFO slide sets; electronic computers ready to be built, electronic experiments for youngsters; and a dictionary of scientific and technical terms. Address: Edmund Scientific, Co., 7782 Edscorp Bldg., Barrington, NJ 08007.

## SYSTRON-DONNER PRODUCT CATALOG

Systron-Donner Corp. has issued a 20-page catalog on its test instruments. Frequency counters from 10 MHz to 24 GHz ; universal counter-timers; digital volt/multimeters for portable, bench, and systems applications; dc laboratory and systems power supplies; pulse/function/data signal and sweep generators: time code readers/generators/displays; spectrum analyzers; microwave components and a new instrumentation controller for IEEE bus applications are included. Address: Systron-Donner Corporation, 10 Systron Dr., Concord, CA 94518.

## CONTINENTAL SPECIALTIES CATALOG

Continental Specialties Corp. has released a 12-page catalog describing its line of electronic prototyping, development and testing hardware. New products such as hand-held logic probes, and a matching digital pulserpart of CSC's test equipment family called The Logical Force-the MAX-100 compact frequency counter are introduced. Other products included are test sockets, solderless breadboards, test instruments and matching blank cases, and IC test clips. Address: Continental Specialties Corp., 70 Fulton Terrace, New Haven, CT 06509.

## HEATH INSTRUMENT CATALOG

Heath/Schlumberger has announced availability of its latest Assembled Instruments Catalog. It contains 32 pages of descriptions and specifications for Heath oscilloscopes, labo-ratory-grade Strip and $X-Y$ recorders, power supplies, signal and function generators, counters, multimeters (analog or digital), and a selection of accessories such as probes and interconnecting cables. A listing of selfinstruction courses in ac and dc electronics, semiconductor devices, digital techniques, microprocessors and others is included. Address: Heath/Schlumberger, Dept. 57-020, Benton Harbor, MI 49022

## WIDL VIDEO CATALOG

"The Catalog of Video Supplies and Accessories," now available from WIDL, Video, Chicago, contains over 500 items such as video tape, audio cassettes, tape labels, microphones, video and audio cables, connectors and adapters, printed video forms and other video accessories. Address: WIDL Video, 5325 N. Lincoln, Chicago, IL 60625.

## "A sight for sore ears

Hard to believe there's one simple solution to the noise, range and privacy problems of conventional $\mathrm{AM} / \mathrm{CB}$. But there is.

It's Midland Power Single-Sideband CB.

TUNE OUT TROUBLE. Instead of 40 overcrowded channels, Midland SSB splits each channel into three. And you get more output power than your ordinary AM/CB.

As for features, Midland makes SSB as simple or as advanced as you please. Take our Model 79-900, shown below.

It's the top of the Midland SSB line, with all the important controls-our 2button "Touch and Run" channel selector that runs through all 40 channels and stops where you want, SSB clarifier and volume control - -on the mike, as well as on the panel.

MOST ADVANCED. Electronically, it's the most advanced SSB in Midland's 17 years of electronics experience. A result of the attention Midland pays

> Midland SSB Run with Number 1.
to the needs and wants of serious CBers.

In the words of C. W. McCall, America's Number 1 CBer: "Do your ears a favor. See your Midland CB dealer now."

And while you're there, ask about the new Midland Mobile Audio AM/FM car stereo line, with MicroPrecision ${ }^{\text {TM }}$ Tuning.


# Learnto service Communications/EB with NRI's Complete 

Learn design, installation and maintenance of commercial, amateur or CB communications equipment.
There are more than 25 million CB sets out there, millions more two-way radios, walkie-talkies, and other communications apparatus in use by business, industry, government, police and fire departments, and individuals. That means a lot of service and maintenance jobs ... and NRI can train you at home to fill one of these openings. NRI's Complete Communications Course covers all types of two-way radio equipment... AM and FM transmission and reception, television broadcasting, microwave systems, radar principles, marine electronics, mobile communications, and aircraft electronics. And NRI guarantees you will pass the exam for the commercial FCC Radiotelephone License you need to perform most servicing work, or your tuition will be refunded in full. This money-back


Some designed-for-learning equipment you get.

agreement is good for six months after completion of your course.

## Learn on your own 2-meter, digitally synthesized VHF transceiver.

You'll learn to service all types of communications equipment as you assemble your own VHF transceiver. NRI engineers have designed it, not only as a commercial-quality, highperformance unit, but as a unique "power-on" training tool to give you actual bench experience with the principles needed to service CB , commercial, and amateur equipment.

## Then we help you

 get your FCC Amateur License so you can go on the air.The complete course includes 48 lessons, 9 special reference texts, and 10 training kits. Included are your own electronics Discovery Lab, ${ }^{\text {TM }}$
antenna applications lab, CMOS digital frequency counter, and an optical transmission system. You'll learn at home, at your own convenience, earning your FCC license and the preparation you need for the communications field of your choice. Mail the postage-paid card today. TMTrademark McGraw Hill.

## CB specialist course also offered.

If you prefer, you can concentrate on the booming field of CB radio with NRI's special course in CB servicing. You get 37 lessons, 8 reference texts and plenty of hands-on training with your own 40 -channel $\mathrm{CB}, \mathrm{AC}$ power supply, and multi-meter. Also included are 14 coaching units to make it easy to get your commercial Radiotelephone FCC License...required for you to test and service communications equipment.

# equipmentat home... Communications Course. 

## Or get into TV and audio servicing

NRI can train you at home to service TV equipment and audio systems. Choose from five courses that go up to our 48-lesson Master Color TV/Audio Course. With it you get 14 kits for practical bench training and demonstrations, including NRI's exclusive, designed-for-learning, 25 " diagonal solid state color TV, 4-channel audio system complete with speakers, and professional instruments you build and use for learning and earning. It's proven, effective training that's helped thousands of pros already. And it's the best value offered in the field. NRI's bite-size lessons speed learning, exclusive

"Power-0n" training makes it real. Send card for free catalog.

# Learn computer electronics 

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Stereo Scene

By Ralph Hodges

## POWER SUPPLIES AND OTHER REVELATIONS

SUDDENLY an enormous amount of attention is being paid to power supplies in audio amplifiers. To quote from a technical manual of a major Japanese manufacturer: "It may fairly be said that the most important thing in audio amplifier designing is the power supply circuit as a source of energy, because the fundamental function required from amplifiers is to drive speakers."

The same manual goes on to describe the "ideal" characteristics of a power supply: "(1) Low output impedance; (2) Good transient response; (3) Huge power capacity and stable energy supply." This appears to make good sense.

Some time ago, I wrote in this space about a modification to the Dynaco Stereo 400 that would make it a "super" amplifier. Much of the modification involved augmenting the power supply (to the tune of several hundred thousands of microfarads) to improve the amplifier's "audible" characteristics. At that time I was enthusiastic about the difference (created by Frank Van Alstine of the company of the same name) that I heard between the modified and unmodified versions of what was basically the same amplifier. I still am. So is Dynaco, to judge by the emergence of the new 416 amplifier with its add-on option of zillions of microfarads that can be purchased in a very attractive package, plugged into the supply rails of the 416 , and set on top of the amplifier to create a convenient and reasonably compact duo. And I truly believe it does.

How Does It Work? To observe the difference between an augmented and unaugmented power supply you merely connect a voltmeter across the supply rails of your amplifier and see (and perhaps hear) what happens as it's playing a record or tape. Does the voltage go up and down? Often it will, and presumably that's not good. Several amplifier manu-
facturers have begun using the term "dynamic crosstalk" to describe a condition (in an amplifier utilizing a single power supply) in which the demands made upon one channel will effectively modulate the output of the other channel, because the power supply feeding both is pumping up and down. You can eliminate dynamic crosstalk-if its possibility concerns you-by building separate and well-shielded power supplies for each channel, or by designing a power supply that refuses to quit under any conditions of current drain.

Separate power supplies have been cropping up in astonishing numbers recently. But anyone who has ever hefted a power transformer for a truly big audio amplifier knows that duplicating it (for the second, third, or even fourth channel) is not especially cost effective. Reverting to a single supply and adding umpteen filter capacitors can evidently help to an extent. But note that these capacitors, although they weigh very little, occupy a great deal of space. It takes a lot of capacitors to make any significant difference in the performance of a power supply, and this probably accounts for various manufacturers' policies of offer-


Fig. 1. Top trace is voltage variation of switching supply while playing music. Bottom is regular supply with same music.
ing additional filter capacitors as add-on rather than as built-in facilities.

Enter the presumed great hope of the future: the switching power supply. This concept reportedly comes to us fresh from the world of the computer, and if it's a little surprising that it's barely reached into the realm of audio applications before now, it seems the possible deluge is finally at hand.

The switching power supply is a high-frequency device, operating typically at a frequency between 20,000 and $35,000 \mathrm{~Hz}$. For such frequencies a transformer can be quite efficient and compact. Furthermore, it would stand to reason that the storage capacitors, being charged at such a high rate, would not need to incorporate the "extra" capacitance generally required by a highpower amplifier with a $60-\mathrm{Hz}$ supply. If the above suggests the possibility of an unusually small and lightweight assembly, you're on the right track. Recently 1 was shown a supply for a power amplifier worth 120 watts per channel into 8 ohms (and 240 watts per channel-an honest doubling!-into 4 ohms) that a group of schoolchildren could have effortlessly played catch with. And yet this supply exhibited almost no variation under music-playback conditions that had a conventional supply's voltages pumping up and down vigorously. (Fig. 1.)

The power amplifier was the M-7070, a new product from JVC; and by the time you read this. Sony will have announced two integrated amplifiers with switching supplies to the U.S. market. Block diagrams for the two designs are shown in Figs. 2 and 3. Note that in both cases some form of feedback is used to regulate voltages and-in JVC's case-to lower output impedance.

From all indications American manufacturers are hot on the heels of the overseas companies, although they are not necessarily in agreement as to the proper goals of power-supply design. At least one is unconcerned about maintaining tightly regulated supply voltages as long as the necessary voltages are there when called for. And the speed (transient response?) of a switching supply should enable extremes of the musical waveform to be "tracked" easily as they occur. At other times the supply will "stand down" in a much more efficient mode of operation.

A Head That's Finally Together? The latest evolution in the Nakamichi 600 Series of two-head cassette decks has a feature that's a little breathtaking:


## The XSV/3000 is the source of perfection in stereo sound!

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1976: Stereohedron ${ }^{®}$ This patented Stylus tip assures super traceAbility ${ }^{\top M}$, and its larger bearing radius offers the least record wear and longest stylus life so far achievable.

## 1975: High Energy Rare Earth Magnet

Another Pickering innovation, enabling complete


1. Technical drawing of the Stereohedron shape
 miniaturization of the stylus assembly and tip mass through utilization of this type of magnet.

## 1968: Dustamatic ${ }^{\circledR}$ Brush

This Pickering patented
invention dynamically stabilizes the cartridge-arm system by damping low frequency resonance. It improves low frequency tracking while playing irregular or warped records. Best of all, it provides record protection by cleaning in front of the stylus.

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Fig. 2. JVCcalls its switching supply D.P.S. or Digital Power Supply.

a record/playback head with a 0.9 mi cron gap. Now that should-and in fact does-make a lovely playback-only head for a three-head cassette deck, but the common understanding has been that you just can't use such a narrow gap for recording; flux saturation in the magnetic gap kills you almost immediately, driving distortion way up.

Well, perhaps not. According to Nakamichi, analysis of the flux pattern propagated by such a head (if it is properly designed) shows that the pattern-and particularly the so-called "critical zone" where recording actually takes placecan in fact remain well-defined and controlled, if a little distorted from its familiar shape. Experts in tape-recording theory contacted immediately after the Nakamichi announcement allowed that the whole idea was plausible, although they could not of course vouch for this particular execution of it sight unseen.

I won't exhaustively detail the specifications of the new Nakamichi machine here, except to note that its usable frequency response is claimed to extend comfortably beyond $20,000 \mathrm{~Hz}$, and that performance appears to be as much tape as machine limited. The "distortion compensating" circuits of the previous Model 600 have disappeared from the present machine, but there are phasecorrecting circuits.

Dialing Vertical Angle. Of late l've had very little time to spend on the phono vertical-tracking-angle issue, with which this column has dealt on several past occasions. (If you recall, the propo-
sition is that vertical tracking angle, which can be altered by raising or lowering the tonearm or shimming the phono cartridge in the headshell, is critical to within a degree or even less for proper record-player performance. It should also logically require at least occasional alteration for different records.) Fortunately, reports continue to come in from the field; reports that are positive, negative, and just plain frustrated.

It is extremely difficult to give adequate instructions on zeroing-in on approximately correct vertical tracking angle (VTA) to someone who lives at a distance. The situation is only made worse by certain cartridge/tonearm/ turntable combinations that assume somewhat improbable-looking geometries as they approach the optimum setting (although a careful analysis of the various angles involved usually reveals. that they are not so improbable at all). But there may be a little help in the offing for those inclined to pursue the matter.

The story is this: many have found playing a left-minus-right signal from a stereo record to be extremely helpful in establishing at least a ballpark setting for VTA. The idea makes perfect sense, and I'm abashed that I didn't think of it myself. In listening to the L-R, all you do is try to minimize familiar old distortion and mistracking effects, which will no doubt be plentifully evident on vocal sibilants and other demanding high-frequency signals. It is recommended that several records be used, since in many cases the vocalist you're trying to focus on will almost completely disappear in
the L-R mode. (In fact, if on such records he/she does completely disappear, that's a good sign.) What you really want is a record in which a centered vocalist acquires a distinct distant and reverberant quality. Once you've established an initial "optimum" setting for this one record, you can rest assured that optimum settings for others won't be far off.

Getting an L-R signal can be as simple or complex as you'd care to make it. The simplest way is to reverse the leads on one channel of the phono cartridge and then switch the amplifier to mono. Some fiddling with the balance control is then advisable (the outputs of a phono cartridge's two channels are rarely matched perfectly) to get as complete a drop-out of the central performers in the stereo panorama as possible. An alternate route to an L-R signal is, of course, the use of a phase inverter somewhere after the phono preamp to add the two channels in anti-phase. Whatever takes your fancy.

It's also been suggested that you use a mono recording in pristine conditionif you can find one-for the L-R test. In this case you should hear nothing but distortion, and the less distortion the better in terms of VTA adjustment.

Again, l'd like to emphasize that I'm not convinced this adjustment can be carried out successfully on every record player. It seems logical that a certain amount of rock-solid stability must be there to begin with before such subtleties can be heard. But l'd be glad to be proved wrong and look forward to further reports from the field.


Fig. 3. Block diagram of Sony supply. Note feedback derived from primary of transformer.

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circuitry, Panasonic's RF-4800 gives you all these sophisticated controls. Like an ell-gear-drive tuning control to prevent "backlash:" Separate wide/narrow bandwidth selectors for optir um reception even in crowded conditions. Adjustable calibration fo easy tuning to exact frequencies. A BFO pitch control. RF gat contral ior increased selectivity in busy signal areas. An ANL swifch. Even separate bass and treb e controls. And if all ihat short wave isn't enough. There's mare. Like SSB (single sideband) amateúr radio All 40 CE channels. Ship to shore Even Morse communications: ABADC operation. And with Panasonic's $4^{\prime \prime}$ full-range speaker, the tig sound of $A M$ and $F M$ will really sound b.g.

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# Julian Hirsch Audio Report 

## CARTRIDGE LOADING AND PREAMPLIFIER INTERACTION

Interaction [between cartridge and preamplifier] can modify response by several decibels.

FOR MANY YEARS, it was assumed that a magnetic cartridge (of the moving-magnet or moving-iron type) would perform correctly if it were simply terminated by a $47,000-\mathrm{ohm}$ resistance. (The adoption of this load value was one of the earliest instances of standardization in the hi-fi industry.) More recently, we have become aware that matters are not quite so simple. For instance, the input resistance is shunted to ground by a capacitance. Part of this is in the tonearm wiring that connects the record player to the preamplifier and part in the preamplifier circuit itself.

The reactance of the shunting capacitance decreases with increasing frequency, reducing the total load impedance presented to the cartridge. It might seem that this would reduce the cartridge's high-frequency output, but this is true only when the frequency is very high. At lower frequencies, the added shunt capacitance actually boosts the cartridge's output, and most cartridges are designed to be terminated in a specific capacitance as well as resistance for flattest overall frequency response.

To see why this is so, one must realize that the cartridge has a high-frequency mechanical resonance between its effective moving stylus mass (referred to the tip) and the compliance of the vinyl record material. This may occur at a frequency as low as $10,000 \mathrm{~Hz}$ in lower-priced cartridges, whose styli are more massive, and as high as 30,000 Hz or more in CD-4 cartridges. Most often, it falls in the $15,000-\mathrm{to}-25,000-\mathrm{Hz}$ range. The resonance takes the form of a peak in the cartridge's output, which may be more or less damped by mechanical means in the structure of the stylus system.

There is also an electrical resonance,
between the inductance of the cartridge coil and the total shunting capacitance of the load circuit. Here, the damping is supplied by the 47,000 -ohm load resistance and, to some extent, by the winding resistance of the cartridge. This resonance produces a response peak whose frequency is determined by the L and C values and whose amplitude is controlled by the " Q " of the system established by the circuit resistance. However the mechanical resonance and its frequency response characteristics are not affected by any electrical circuits within the cartridge or external to it.

By proper proportioning of the electrical and mechanical resonances of the cartridge, the combined response can be made very flat throughout the audiofrequency range. By placing the electrical resonance somewhat above the mechanical resonance and if both have the correct " Q " values, the mechanicalresonance peak will be attenuated and the rise in frequency will improve the overall flatness.

This is why a cartridge manufacturer will usually specify a range of load capacitance values (such as 250 to 300 pF or 400 to 500 pF ) into which his cartridge will deliver its rated frequency response. These values take into account the normal wiring capacitance of the record player's tonearm and its connecting cables (typically 100 to 500 pF in modern units) and assume about 150 pF of input capacitance in the preamplifier. The actual preamplifier capacitance, however, may vary widely, from nearly zero to many hundreds of picofarads. Sometimes, as in the Hafler Model DH-101 reviewed this month, the naturally low-input capacitance of the amplifier has been padded to 250 pF , making the load broadly optimum for a wide variety of cartridges.

The effect of too little capacitance is to produce a peak in the cartridge response, often at a frequency of 12,000 to $15,000 \mathrm{~Hz}$. Too much capacitance will often boost the high frequency output of the cartridge [in the 10,000-to-$15,000-\mathrm{Hz}$ range] but will attenuate it faster at higher frequencies. A greatly excessive capacitance will roll off the output pronouncedly above $10,000 \mathrm{~Hz}$ or so.

There is a second factor to consider that is entirely unrelated to the cartridge load but which also
influences the overall frequency response. In some phono preamplifier stages, the feedback network that provides the RIAA playback equalization is not well isolated from the cartridge input. When the stage is driven from the resistive source impedance of a signal generator, the frequency response may appear to be an accurate RIAA curve. However, the presence of the inductance of the phono cartridge across the input terminals may modify the amplifier feedback sufficiently to alter its
response at very high frequencies. In a good amplifier this effect is moderate, usually less than $\pm 1 \mathrm{~dB}$ of variation up to $20,000 \mathrm{~Hz}$. (This may be a boost or a cut or a combination of the two.) In a few cases, fortunately becoming rarer these days, the interaction can modify the response by several decibels, which is plainly audible and obviously undesirable. In the finest preamplifier designs, there is absolutely no interaction between the cartridge and the preamplifier equalization.

# Audio Test Reports/ 

HIRSCH/HOUCK LABORATORIES

## HAFLER MODEL DH-101 STEREO PREAMPLIFIER

## "Plain Jane" preamp produces exceptionally fine performance.




The Hafler Model DH-101 stereo preamplifier is almost starkly simple, but its performance is literally "state of the art." Its distortion is virtually unmeasurable, and every effort appears to have been made to eliminate the various flaws, both major and minor, that plague many preamplifier designs.

The preamplifier is available in both factory-wired and kit forms. The kit assembly process is principally one of mechanical assembly and the soldering of wires from the circuit boards to the switches and controls. The circuit board assemblies themselves come completely wired and tested. The Model DH-101 measures $133 / 4^{\prime \prime} \mathrm{W} \times 81 / 2^{\prime \prime} \mathrm{D} \times 31 / 4^{\prime \prime} \mathrm{H}$ $(34.9 \times 21.6 \times 8.3 \mathrm{~cm})$ and weighs 8 lb
( 3.6 kg ). Suggested selling price about $\$ 300$ factory-wired, $\$ 200$ kit.

General Description. The frontpanel controls of the preamplifier consist of three knobs and 10 pushbutton switches. The faces of the rectangular pushbutton switches appear in black when the switches are not engaged (out positions). Pressing in any of these switches causes the face color to change to white (yellow in the case of the TAPE monitor buttons), clearly indicating the status of the controls without requiring LED indicators or extra electrical switching.

In addition to the various signal input and output jacks, the rear apron contains two pairs of phono jacks labeled EXT PATCH, which are normally connected together by heavy jumpers. They are
in the signal path, after the input selector but ahead of any of the preamplifier's active circuitry, except for the phono preamplifier stages. They make it possible to connect an equalizer or other signal processor into the system without sacrificing any of the tape-recorder versatility of the preamplifier.
All the amplifier circuits are on a single circuit board, onto which the lugs of the control potentiometers are soldered. When they are installed on the front panel, the potentiometers are the sole mounting support for the circuit board, but they appear to be perfectly adequate for that purpose. A smaller board contains the power supply circuits (with IC regulators for the $\pm 18$-volt supplies) and another board carries the pushbutton switches. There is a large empty space in the preamplifier that can accommodate a moving-coil phono cartridge preamplifier (to be made available at a later date) that will connect to one of the phono inputs.
The performance specifications of the Model DH-101 can be summed up quite simply, although they are extensive. All noise levels are inaudible and, with " $A$ " weighting, are very difficult to measure. All frequency-response characteristics are within 0.5 dB of flat or the specified equalization response. All distortions are less than the residual levels of any standard laboratory instruments, up to the rated output of 3 volts. There is no


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PERFORMANCE SPECIFICATIONS

| Specification | Rating | Measured |
| :---: | :---: | :---: |
| Rated output | 3 volts, $10-100,000 \mathrm{~Hz}$ | As rated |
| Maximum output | 7 volts, $20-20,000 \mathrm{~Hz}$ | 12 volts at 1000 Hz |
| THD | Less than $0.001 \%$ at rated output | Less than 0.0025\% (instrument residual) |
| IM distortion | Below instrument residual | Less than 0.002\% (instrument residual) |
| Rise time | $2 \mu \mathrm{~S}$ | $3 \mu \mathrm{~s}$ |
| Slew rate | $12 \mathrm{~V} / \mu \mathrm{s}$ | 7 to $12 \mathrm{~V} / \mu \mathrm{s}$, depending on measurement method. |
| Frequency response | $\begin{aligned} & 20-20,000 \mathrm{~Hz}, \\ & +0 /-0.25 \mathrm{~dB} \end{aligned}$ | $\begin{aligned} & 20-20,000 \mathrm{~Hz}, \\ & +0 / 0.5 \mathrm{~dB} \end{aligned}$ |
| Hum \& noise | -90 dBV ("A" weighted) (high level) | Less than -80 dBV (measurement limit) |
|  | -86 dBV ("A" weighted; phono) | -68 dBV (unweighted) |
| Phono Frequency response | Within $\pm 0.5 \mathrm{~dB}$ of RIAA, $40-15,000 \mathrm{~Hz}$. | As rated |
| Phono overload ( 1000 Hz ) | 180 mV | 180 mV |
| Phono cartridge interaction at $20,000 \mathrm{~Hz}$ | Unmeasurable | Unmeasurable |
| Phono gain ( 1000 Hz ) | 34 dB | 35.4 dB |
| High level gain (1000 Hz) | $20 \mathrm{~dB} \pm 1 \mathrm{~dB}$ | 20.0 dB |
| Bass tone control range $(50 \mathrm{~Hz})$ | $\pm 12 \mathrm{~dB}$ | +14.5, - 16 dB |
| Treble tone control range ( $20,000 \mathrm{~Hz}$ ) | $\pm 10 \mathrm{~dB}$ | +11.5, -17 dB |

detectable interaction between the phono preamplifier response and the inductance of a phone cartridge.

User Comment. The basic appearance of the Model DH-101 is so plain and devoid of gadgetry and styling features that it is difficult to credit it with being quite possibly the most highly refined preamplifier one can buy (in terms of sheer performance). Our measurements convinced us of the accuracy of the claims for this preamplifier, and we looked to our use tests to either confirm or amend the conclusions of our measurements.

Connecting the preamp to a hi-fi system revealed what is probably its only
weak point. The phono connectors on the rear apron, at least on our early sample, seemed to be slightly oversize in their outer diameter, requiring considerable force to insert some of the mating plugs. This would not have been so bad, but the rear apron itself is a thin, flexible metal surface, liberally covered with holes and cutouts. It bent inward, sometimes to an alarming degree, when the necessary pressure was applied to the phono plugs as they were inserted. This did not result in any permanent deformation of the metalwork, but it did detract from the overall image of quality which is conveyed by every other aspect of the performance and operation of the preamp itself. (Continued on page 30)

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## Product Focus

The Hafler Model $\mathrm{OH}-101$ is the first preamplifier we've tested whose phono equalization has been designed to conform to the IEC characteristic, expected to be adopted by the RIAA at some future time. The major effect of the change is to incorporate a very effective infrasonic (rumble) filter in the phono equalization. Instead of the preamplifier gain continuing at a nearly constant value to some indeterminate frequency below 50 Hz (as in the present RIAA characteristic), the response is rolled off at low frequencies. The difference between the two curves is 3 dB at 20 Hz (and only 0.6 dB at 50 Hz ). Hence, audible differences can be expected to be negligible. However, the IEC response curve is down by about 18 $d B$ at 2 Hz , compared to the extended RIAA curve. Hafler has tailored the equalization of the Model $\mathrm{DH}-101$ to be a
compromise between the two, falling within 0.3 dB of the existing RIAA curve down to 30 Hz , yet matching the IEC curve within 1 to $2 d B$ throughout. A second difference in the IEC curve is merely one of definition; its high-frequency rolloff, at 6 dB /octave, is extended to 20,000 Hz instead of stopping at $15,000 \mathrm{~Hz}$ as before.

Both the phono preamplifier and the tone control (output) sections of the Model DH-101 employ differential amplifiers with complementary-symmetry output stages. All the preamplifier circuits use discrete components, the only IC's being the power supply regulators. The phono preamplifier has a very high output voltage capability (actually equal to that of the main tone control amplifier). Its very low output impedance enables it to drive an adequate signal into the low imped-
ance of the equalizing network in the feedback loop, even at very high frequencies, without distortion.

In the main amplifier section, the tone controls are located at the output of the amplifier and supply a feedback signal to the side of the differential input that is not driven by the program signal. In their configuration, the active circuits of the Model DH-101 resemble a typical power amplifier output stage (except for their power ratings). Hafler credits this circuit design for much of the exceptional performance of the preamplifier, including its extremely low distortion, excellent pulse handling and transient characteristics, and (especially in the case of the phono input) a constant input impedance at all frequencies of interest, regardless of the reactive nature of the phono cartridge or other source connected to the input.
(Continued from page 26)
There were no switching transients when any of the pushbuttons were operated. Although there is no obvious source of a time delay in the power supply circuits, there was no sign of a "thump" when the preamplifier was switched on, even when connected to an already energized power amplifier. The controls operated smoothly and with a quality "feel." We were unable to measure the preamplifier's noise level, which was below the minimum range of our test instruments. However, even with the volume set near maximum, which produced ear-splitting levels from records, lifting the pickup left a deafening silence, the hiss being barely audible
with one's ear against the speaker. Through the high-level inputs, there was absolutely no audible hiss or hum, at any setting of the volume control.

The preamp was designed to terminate most phono cartridges correctly and to be free of interaction with them. To this end, each phono input has a 220-pF capacitor wired across it to ground. With the normal amplifier input capacitance of about 30 pF and typical phono-cable and tonearm wiring capacitance of about 150 pF , this loads the cartridge with about 400 pF (in parallel with 47,000 ohms). This is an optimum condition for many cartridges, including those from Ortofon and many Shure models, among others. If the cartridge is
ciacle no 101 on free information card
meant to be terminated in a lower capacitance, such as 250 pF , or the connecting cables have a higher than usual capacitance, the manufacturer suggests that the 200-pF capacitors be removed or replaced with other capacitors of a lower value. This is not critical in most cases, but anyone who is convinced that he can hear the difference may wish to trim the amplifier input capacitance to an optimum value for his cartridge.

Although we tested a factory-wired Model DH-101, we have been informed by those who have built the kit that it is very simple, and that even a neophyte should be capable of assembling a properly performing preamplifier in a few hours or so.

## TECHNICS MODEL SL-1500MK2 DIRECT-DRIVE TURNTABLE SYSTEM

Quartz-control speed system also operates at each pitch control stop.



The Technics Model SL-1500MK2 is a complete direct-drive system for playing records. It consists of a turntable that is directly driven at either $331 / 3$ or 45 rpm by a dc motor whose speed (including pitch adjustment) is precisely controlled by a quartz oscillator; a precision tonearm with damped cueing, automatic muting and automatic return; and an acoustically isolated base. A seethrough plastic dust cover is hinged to the rear of the base.

The player measures approximately $1711 / 16^{\prime \prime} \mathrm{W} \times 151 / 8^{\prime \prime} \mathrm{D} \times 511 / 16^{\prime \prime} \mathrm{H}$

## PERFORMANCE SPECIFICATIONS

| Specification | Rating | Measured |
| :---: | :---: | :---: |
| Wow and Flutter | 0.025\% wrms (JIS) | 0.05\% rms (IRE) |
| Rumble | $\begin{aligned} & -50 \mathrm{~dB} \text { (DIN 45539A) } \\ & -73 \mathrm{~dB} \text { (DIN 45539B) } \end{aligned}$ | 34 dB unweighted (NAB) <br> -58 dB (ARLL weighted) |
| Build-up characteristic | $90^{\circ}$ or $1 / 4$ rotation at $33-1 / 3 \mathrm{rpm}$ | Less than 1 second to full speed or stop |
| Tracking error angle | $+3^{\circ}$ (outer groove of $12^{\prime \prime}$ disc); <br> $+1^{\circ}$ (inner groove) | As specified (less than $0.4^{\circ}$ in. throughout) |
| Effective tonearm mass | 22 g with $6.5-\mathrm{g}$ cartridge at $1.25-\mathrm{g}$ stylus pressure | 19 g net, less cartridge but with shell; measured with 6-g cartridge at l-g force |
| Tonearm/cable capacitance | Not specified. | 100 pF to ground; 6.5 pF between channels |
| Antiskating calibration | Not specified. | Requires approximately $\mathrm{l}-\mathrm{g}$ greater setting than tracking force |
| Arm cueing | Not specified. | Very slow; no drift |
| Base isolation | Not specified. | Much better than average for direct-drive turntables; exceptionally resistant to jarring. |

$(14.5 \times 45.3 \times 38.4 \mathrm{~cm})$ and weighs 26 lb ( 11.8 kg ). The manufacturer's suggested price, less cartridge, is $\$ 369.95$.

Technical Description. As exemplified in the Model SL-1500MK2, the di-rect-drive turntable motors used by Technics are dc motors with "heteropolar" construction. According to Technics, dc motors are more efficient than are ac motors in turntable applications, requiring less power and generating less heat The heteropolar motor used in the company's latest series of quartz-controlled turntables has a 12-pole stator, around which revolves a 16 -pole permanentmagnet rotor that is a part of the platter itself. This $3: 4$ pole ratio is claimed to be optimum for a high-torque, high-efficiency direct-drive motor.
Mounted concentrically with the di-rect-drive rotor and stator is a frequency generator (tachometer) that supplies a feedback signal to the control circuits that drive the motor. The generator consists of two 91-tooth gears that are smaller than the motor itself. One gear is on the platter and the other is fixed and JULY 1978
has a magnetic coil structure built into it. The gear teeth do not touch. As the teeth move past each other, a voltage is induced in the coils of the fixed gear. This supplies 91 pulses per revolution of the platter to the control system.

Current is supplied to the windings of the motor with precise timing from a bidirectional drive circuit. The drive circuit is effectively a three-phase, full-wave system that produces smaller torque pulsations and, consequently, less wow and flutter than the half-wave systems used to drive some other turntable motors. The timing of the signals sent to the stator windings is controlled by three sets of fixed-position sensing windings and a separate position detector rotor that turns with the motor. These elements perform the commutation function that is necessary for the operation of a dc motor. Since all commutation is accomplished by means of magnetic induction. there are no brushes to wear out or become noisy.

The speed of the turntable is locked to the frequency of a crystal-controlled oscillator and is maintained to an accuracy
of $\pm 0.002 \%$. Similar accuracy has been achieved in a number of other quartzlocked turntables in recent months, but only at the 33 1/3- and $45-\mathrm{rpm}$ nominal operating speeds. For vernier speed control, the quartz locks of these turntables are disabled and their speed references are then adjustable dc voltages, as they are in conventional direct-drive turntables.

In the Technics turntable, however, the full accuracy and stability of a quartz-lock system are maintained, while the speeds can be adjusted over a $\pm 9.9 \%$ range in discrete $0.1 \%$ steps. The exact deviation is decimally displayed by a digitally driven numeric readout, as are the basic 33 1/3- and 45-rpm operating speeds. The synthesized quartz frequency generator that makes this display possible would not have been practical before the development of largescale integrated (LSI) circuits.

At the heart of the system is the synthesizer/counter chip. In it, the frequency of an external voltage-controlled oscillator (vco) is divided by 1000 and compared with the frequency of a crystal oscillator, which is also divided by 1000. The filtered output of the phase comparator is fed back to the vco, locking the latter in frequency and phase to the crystal reference signal. The divided frequency from the vco is also counted by circuits within the IC and converted into digital display drive signals. A scanning counter, controlled by + and - pushbuttons on the control panel of the turntable, can be used to change the frequency division ratio of the vco from 901 to 1099 in integral steps. This changes the frequency in $0.1 \%$ steps from $-9.9 \%$ to $+9.9 \%$, while maintaining phase lock with the crystal reference.

A starting signal is applied to the turntable's coils when the system is first turned on. As the turntable speeds up, the frequency fed back from the frequency generator's gears is converted to a dc voltage in a frequency-to-voltage ( $F / \mathrm{V}$ ) converter whose output drives the motor's torque-control circuit. Simultaneously, the frequency feedback signal is compared with the vco signal, which is already locked to the divided crystal-oscillator frequency, in a phase comparator whose output is summed with the F/V converter's output. When the correct speed is attained, control is automatically transferred to the output of the phase comparator. The strobe markings under the turntable platter are illuminated by LED's and can be seen from above by a mirror system. The

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(Continued from page 31)
markings remain stationary, no matter what speed the platter is set for, as long as the phase-lock system is in control.

The motor drive is inherently bidirectional, allowing the speed of the platter to stabilize rapidly after it has been changed in either direction. A useful byproduct of this system is the rapid braking that accompanies the shutdown of the motor.

The tonearm furnished as part of the player is a mildly S -shaped aluminum tube with a lightweight diecast head shell. It is fitted with the four-pin bayonet locked plug that is now used almost universally with Japanese tonearms. The precision ball-bearing pivots are claimed to have less than 7 mg of friction in the horizontal or vertical planes.
(Continued on page 40)

The quartz synthesizer pitch control permits pitch variation by $\pm 9.9 \%$ in steps of $0.1 \%$ by dividing the oscillator frequency by 1000.

The principal circuit components of the phase control system are incorporated in four IC's. Shown here also are the direct-drive motor and control/display circuits.

Block diagram shows how motor rotational speed is 'locked" to a reference frequency originating in a quartz oscillator.


## THE N3W Stereo Review SRT14

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the trost compichensive, accurate ana easy-liouse test record ever developed.

Here, at last, is a record that contains everything you need to get the fullest, most realistic reproduction from your stereo equipment. Whether you've spent thousands on your stereo system or have a more modest setup, the SRT14 is an indispensable tool fcr helping you realize the full potential of your equipment.

Best of all, you don't have to be an electronics engineer to use it. You can actually perform a complete stereo-system checkup by ear slone.

## A test lab in a record jacket

Employing the most edvanced recording, mastering, and pressing techniques, the Stereo Review SRT14 is produced to strict laboratory standards. Engraved in its grooves are a series of precisely recorded test tones, frequency sweeps, and random-noise signals that enable you to accurately analyze and check your stereo system for:

- Frequency response.
- Stereo separation.
- Cartridge tracking.
- Channel balance.
- Hum and noise, including turntable rumble
- Wow and flutter

Optimum speaker placement, and more . . . much more:
And you can do It all without any instruments ... by ear alone.

## Step-by-step instructions

Included with the SRT14 is a detailed instruction manual complete with charts, tables, and diagrams. This takes you step by step through the testing process. It explains the significance of each test. It tells you what to listen for. It clearly describes any abberations in system response. And it details corrective procedures.

## For professionals too

The usefulness of the SRT14 is not contined to the nontecnnical listener. Included on the record are a series of tests that call for the use of sophisticated measuring instruments, such as oscilloscopes, chart recorders, and distortion analyzers. These tests permit the advanced audiophile and professional to make precise measurements of transient response, recorded signal velocity, anti-skating compensation, IM distortion, and a host of other performance characteristics.

## SRT14 record contents

FREQUENCY-RESPONSE TEST, LEFT AND RIGHT CHANNEES. CONsists of half-octave warble tones that permit testing and adjustment of frequency response of a sound system over the full audio range. STEREO SEPARATION. Indicates the amount of signal leakage from one channel into another using warble tones from 400 to $12,800 \mathrm{~Hz}$. PHONO-CARTRIDGE TRACKING, HiGH FREQUENCY. Consists of a two-tone test signal ( 16,000 and $16,300 \mathrm{~Hz}$ ) that repeatedly swoops to a high level and returns to a fixed low level. The level and quality of an audible "difference tone" indicates mistracking.
PHONO-CARTRIDGE TRACKING, LOW FREQUENCY. A single 300Hz tone recorded with similar swoops indicates mistracking as an increase in harmonic distortion.
CHANNEL BALANCE. Separate random-phase noise sources for the two channels permit balancing not only of overall channel levels, but also of the individual tweeters and mid-range drivers in the speaker systems.
PHASE TEST, SPEAKERS AND PHONO-CARTRIDGE, A low-frequency warble tone is recorded alternately in and out of phase several times to establish correct interchannel phasing.

NOISE TEST. A very low-level recording of a piano provides a reference playback level by which the low-irequency noise of a sound system playing an unmcdulated groove can be judged.
RECORD-PLAYER FLUTTER TEST. A passage of piano music is recorded three times with increasing amounts of flutter. The degree to which the record-player's flutter "masks" the recorded flutter indicates ts relative severity.
FREQUENCY-RESPONSE SWEEP, 10,000 TO $40,000 \cdot \mathrm{HZ}$. The frequency response and channel separation of a phono cartridge at ultrasonic frequencies can be measured with a voltmeter or other appropria:e instrument to give an indication of the cartridge's suitability for CD-4 reproduction.
FREQUENCY-RESPONSE SWEEP, 500 TO $20,000 \mathrm{HZ}$. Similarly, the cartridge's response and separation over the range of important audible frequencies can be measured.
SQUARE-WAVE TEST. The high-frequency response, phase shift, and resor ant characteristics of a phono cartridge can be evaluated quickly bw viewing the reproduced waveform of an RIAA-equalized $500-\mathrm{Hz}$ square wave on an osci loscope.
TONE-BU7ST TEST. Tone bursts sweeping from 500 to $20,000 \mathrm{~Hz}$ give an indication of a phono cartridge's transient response when the reproduced signal is viewed on the oscilloscope.
INTERMOOULATION-DISTORTION TEST. A phono cartridge's intermodulation distortion can be measured directly úsing a standard IM meter designed to analyze an SMPTE signal.
ANTISKATING TEST. A speecially designed test signal permits adjustmert of skating compensation for best reproduction of critical high-level recorded passages.
$1,000-\mathrm{HZ}$ ЭEFERENCE TONES. Four tones whose recorded velocities increase by 3-dB steps can be used to determine (by the comparison method) the recorded signal velocity on a disc recording.
FLUTTER AND SPEED TEST. A $3,150-\mathrm{Hz}$ tone recorded with great speed-accuracy and stability provides a signal for use with a flutter meter or frequency counter.
STEREO-SPREAD TEST. A series of recorded gun shots provide a guide to optimum speaker placement for the most subjective satisfying stereó image.

## The final step

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B


Bi-directional drive circuit of AN640 IC supplies signals to two rows of position detectors (A). Switching passes current through two sets of drive coils ( $B$ ) to provide smooth full-wave drive.

## (Continued from page 34)

The head shell has a calibrated scale and index marker that can be used to set the correct stylus overhang for any cartridge whose stylus-to-center spacing is known. The damped cueing system has a built-in muting switch that silences the audio outputs as soon as the tonearm is lifted and does not unmute until after the pickup had reached the surface of the record. Another useful feature of the tonearm is its adjustable height. By loosening a screw on the side of the tonearm's pivot support, the entire arm can be moved vertically over a $6-\mathrm{mm}$ range with $1-\mathrm{mm}$ calibrated accuracy. This makes it easy to set any cartridge parallel with the surface of a record for correct vertical tracking angle and to avoid mechanical interference between the body of the cartridge and the record.

The turntable's mounting base serves more than just a cosmetic function. The entire record player is supported on feltdamped feet, and the combined turntable/tonearm system is floated from the base on separate resilient mounts. The goal here was to isolate the turntable from external vibration and minimize acoustic feedback, often a problem with direct-drive record players.

User Comment. We tested the record player with an AKG Model P8ES phono cartridge installed in its tonearm and performed our listening tests with an Ortofon Model M20FL Super cartridge installed. Cartridge installation and setup were simple and straightforward.

The record player conveys a sense of
precision in both appearance and "feel" that is also evident in its performance. Having the operating controls out front and accessible even with the cover closed is a real advantage. The one slipup in this respect is that the cueing control is not accessible until the dust cover is lifted. We also found the dust cover itself awkward to lift with one hand. Its front surface slopes to the rear and is difficult to grasp and lift except by its side or with two hands.

The turntable cannot be faulted. It came swiftly up to speed and changed speed so rapidly and in such small increments that we could hardly believe that anything was really changing. When we played the $1000-\mathrm{Hz}$ band of a test record and displayed the output on a frequency counter, each touch of a vernier button changed the displayed frequency by exactly 1 Hz . The almost instantaneous stopping of the platter when the sTOP button was touched was a nicety we appreciated.

The cueing/muting system is ingenious, but was too slow for our taste. Some 10 seconds are required to unmute after the cueing lever is lowered, though the descent time of the tonearm can be adjusted over wide limits, depending on the height to which the lift is set. If the descent is too fast, a portion of the record will not be heard before the muting is disabled. If the descent is too slow, the unmuting will occur before it should and the thump of the stylus contacting the record will be heard. Optimally, the descent should be timed, by appropriate setting of the lift height, for


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Multi-user BASIC, suitable for program instruction and simple business applications, is included with the multi-user
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about nine seconds. This may tax the user's patience, but will result in ideally quiet cueing action.
The strobe pattern was always rock steady during our tests, except for brief transients as we were making large speed changes. The only way to make
the pattern move was to place so much drag on the turntable that it dropped out of synchronism with the quartz oscillator. Needless to say, under a condition like this, we did not have to look at the strobe pattern to know that the turntable was running slow.

The Model SL-1500MK2 is certainly one of the nicest looking pieces of record playing machinery we have seen. It offers a combination of mechanical and electronic sophistication that is nothing less than remarkable for a record-playing system of its modest price.

## INFINITY MODEL Qb SPEAKER SYSTEM

Three-way system features electromagnetic induction tweeter.



The three-way Infinity Model Qb speaker system features a new tweeter of unconventional design, called an EMIT (electromagnetic induction tweeter). This is backed up by a $10^{\prime \prime}(25.4-\mathrm{cm})$ woofer whose cone mass is progressively decoupled with increasing frequency to improve its performance in the upper-bass range. The $600-\mathrm{to}-4000-\mathrm{Hz}$ midrange is handled by a $4^{\prime \prime}(10.2-\mathrm{cm})$ cone driver. The system's impedance is rated at 4 ohms, and recommended driving power is 15 to 150 watts/channel.

The fully sealed enclosure is finished in birch-grain vinyl. It measures $25^{\prime \prime} \mathrm{H} \times$ $141 / 2 \mathrm{~W} \times 12^{\prime \prime} \mathrm{D}(63.5 \times 36.8 \times 30.5 \mathrm{~cm})$ and weighs nearly $40 \mathrm{lb}(18.2 \mathrm{~kg})$. Although it is no larger than many socalled "bookshelf" speaker systems, the Model Qb is meant to be placed on the floor, preferably on an optional steel pedestal available from Infinity. The pedestal raises the speaker about $12^{\prime \prime}$ from the floor and tilts it slightly backward, preferably no closer than a couple of feet from any room wall.

Two small knobs set into the rear of the cabinet are provided for varying the outputs of the midrange and treble drivers over a limited range. The frequency response of the speaker system is rated at 42 to $32,000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$, under unspecified test conditions. The horizontal dispersion is rated at $\pm 60^{\circ}$ at $20,000 \mathrm{~Hz}$ for an output-level decrease of 2 dB .

The nationally advertised value of the Model Qb is $\$ 192$ and the optional steel stands are $\$ 40$ per pair.

Laboratory Measurements. The averaged frequency response of the speaker system, measured in the reverberant field of our test room, was exceptionally flat from several hundred hertz to the $15,000-\mathrm{Hz}$ upper limit of our mi-
crophone's calibration. The overall variation of $\pm 2 \mathrm{~dB}$ from 450 to $15,000 \mathrm{~Hz}$ places the Model Qb in a very select group of speaker systems. (Our measurements are made at normal listening distance in a normally furnished room, rather than in the unnatural environment of an anechoic chamber.) The polar dispersion was good, although in our test room, there was about a 5-dB difference in high-frequency output measured onaxis and $30^{\circ}$ off-axis.
Locating the microphone close to the woofer, the output of the latter exhibited a downward-sloping characteristic at frequencies beyond 65 Hz . It amounted to about an $8-\mathrm{dB}$ decrease at 600 Hz , where the steeper attenuation of the crossover network began. Below 45 Hz , the output fell at the $12 \mathrm{~dB} /$ octave rate that would be expected from a sealed speaker system.

Splicing the woofer and midrange/ high-frequency curves together was not as unambiguous as we would have liked, but our best approximation of a combined curve revealed a considerable low-frequency rise below 200 Hz . This did not coincide with what we heard from the speaker system. It is probable that by installing the speaker on its tilt stand and locating it well away from any wall, the bass response would tend to be minimized, which is in all likelihood why Infinity recommends the use of the stand. In any event, it was clear that the woofer's output down to below 35 Hz , was at or above the midrange and high-frequency levels.
The midrange and tweeter level controls had very limited adjustment ranges, making it impossible to seriously degrade the performance of the speaker


Tone-burst responses at 100,500 , and 5000 Hz .

## Product Focus

The most unique feature of the Infinity Model Qb, visually and audibly, is its EMIT (Electromagnetic induction tweeter) speaker. The EMIT is used singly in the Model Qb (and its lower priced twoway version, the Qa) and in arrays in the company's higher-priced speaker systems. It is meant to provide the essential advantages of electrostatic speakers, such as extremely smooth and extended high-frequency response and excellent transient response. Unlike the electrostatic speaker, it is rugged, efficient, can handle considerable power input, and has superior dispersion.

From the front, the EMIT does not look at all like a conventional speaker. It is a Hat plate with four narrow slits, behind which there appears to be a plastic diaphragm. This diaphragm is a thin, lowmass plastic, resembling that of an electrostatic tweeter. Deposited on it is an etched conductor "winding" that consists of a number of turns in the shape of a highly elongated rectangle. The long sides of the winding are behind the slots in the front plate, and the conductors are in the field of two powerful samariumcobalt magnets.

The signal current passing through the conductors on the tweeter diaphragm produces a deflection of the thin plastic. The diaphragm is driven uniformly over its effective radiating surface, like that of an electrostatic speaker. But the efficiency and ruggedness of the EMIT give it a considerable advantage over the electrostatic type. In addition, the EMIT does not require a power source for a polarizing voltage. The vertical orientation of the slots in the front plate gives the driver excellent horizontal dispersion.
system by careless setting of the controls. The tweeter level could be varied over about $\mathrm{a} \pm 1-\mathrm{dB}$ range at frequencies beyond 3500 Hz , and the midrange control had a range of about $\pm 1 \mathrm{~dB}$ from 600 to 3500 Hz .

The impedance of the system attained its minimum of 4 ohms at 20 Hz . A second minimum impedance of 4 to 5 ohms occurred at 120 Hz . Over most of the audio range, the impedance measured between 5 and 15 ohms. At the $54-\mathrm{Hz}$ bass resonance point, the impedance was almost 20 ohms.

For a fully sealed speaker system, the Model $Q b$ is moderately efficient. When driven with 2.8 volts of random noise in the $1000-\mathrm{Hz}$ octave ( 2 watts into the rated 4 -ohm impedance), it produced an $89-\mathrm{dB}$ sound-pressure level (SPL) at a distance of 1 meter. The bass distortion
at a 1-watt level (2 volts) was less than $1 \%$ down to 65 Hz . It rose slowly to $4 \%$ at 40 Hz and to $7 \%$ at 30 Hz . With a $10-$ watt input, the distortion rose more rapidly. It reached $5 \%$ at 56 Hz and $10 \%$ at 40 Hz . The tone-burst response was good at all frequencies.

User Comment. The absence of peaks or dips in the response of the Model Qb is immediately apparent from its sound. The speaker system sounds smooth and uncolored. The EMIT tweeter has a crystalline clarity, with no trace of stridency. In general, we listened to the speaker system with its level controls set to their centers of rotation, which gave us the flattest response in our lab measurements. In our opinion, however, these controls could have been omitted altogether, since their effect on the sound is so subtle.

The bass performance of the system is not as easy to evaluate as that of the EMIT, since the bass is influenced to a greater degree by the listening room. Following Infinity's instructions for setup, we felt that the balance between lows, middles, and highs was just right. The bass was deep and solid when required, and there was little tendency to exaggerate the upper bass, which is a common fault with many speaker systems. On powerful organ-pedal notes, the woofer on one system tended to "flutter" at its extreme excursion. Investigating this, we felt an air leak around the tweeter when the woofer was driven hard. This was obviously a manufacturing defect because it did not occur with the other speaker system in our stereo setup.

The brown grille cloth is attached to a wooden frame that snaps onto the speaker enclosure with two plastic fasteners, one of which is located at the center of the top edge and the other at the center of the bottom edge of the frame. These fasteners support the grille about $1 / \mathrm{r}^{\prime \prime}(3.2 \mathrm{~mm})$ from the front of the enclosure. When the systems were driven hard, the grille assemblies rocked from side to side. This did not produce any audible buzzing or resonance effects in our tests, however.

The Infinity Model Qb is competitively priced with a number of other fine speaker systems. It is highly listenable, with a smooth, uncolored sound and no detectable audible weaknesses in its performance. Also, playing the speaker system for an extended period of time, we concluded that we would be happy to "live" with it.

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> The latest mobile sound components and how to match them to an automobile environment.

BY PETER SUTHEIM

ABOUNTIFUL crop of high-quality audio components for mobile use has been introduced this year. Consequently, it's now easier to obtain "hi-fi" sound in an automobile. Here is an indepth look at some of the new car speakers, power boosters, equalizers, tape decks and FM receivers and the special challenges you will face if you are to take full advantage of the better sound they offer.

Speakers. The primary problem involved in an audio system for a car concerns the speakers and their placement. As you know, there is little space available for normal-size speaker enclosures in an automobile. Thus, you must use tiny enclosures or mount "raw" loudspeakers in already available cavities. With stereo and quadraphonic sound, the problem is compounded.

If you have the space, excellent speaker performance can be had from several makes and models of small inte-
gral speaker systems. The ADS Models 2002 and 2001, which include biamplification systems, are prime examples of this type. The enclosures themselves are about the size of squat milk cartons. They are made of cast aluminum, and each contains a $4^{\prime \prime}(10.2-\mathrm{cm})$ woofer and a $1^{\prime \prime}(2.54-\mathrm{cm})$ dome tweeter. The individual speakers in each enclosure are driven by separate amplifiers, with frequency division for crossover occurring before the amplifiers.

ADS produces virtually the same speaker system without the amplifier as the Model 200. It is a 4 -ohm system specifically designed for a car stereo setup. The Model 200's thrive on hefty amounts of power, so they will not sound their best with the typical 3-watts/channel EIA-rated output of most low-cost indash car stereo units.

Similar to the ADS Model 200 are the Visonik "Little David" 50, the Braun "Output C," the ACR system, Tamon's Model LB-1030, and the Roadstar Mod-



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## SPARINMMTA

## For the Travelin' Man

Car Sound/CB Equipment/Auto Clocks/Shifters/Creepers
el RS-6040. The foregoing are merely the smallest high-quality speaker systems. The ADS Model 300, for example, is akin to them but it has a $6^{\prime \prime}(15.2-\mathrm{cm})$ woofer and offers correspondingly better bass and higher power-handling ability. Advent's Model 400 is a larger design.

If integral speaker systems, with their properly designed enclosures, appear to be impractical (for space or safety reasons) or unattractive, you will have to mount raw speakers into the body of your car. Unfortunately, the most popular speaker-mounting locations are not truly ideal acoustically. These include in the doors, under the seat, in the shelf under the rear window, etc. Such locations might excite low- and middle-frequency resonances in the car's interior air space or mechanical resonances in the body structure. Such problems, however, are usually minimized by ambient noise and other factors. Also, upper frequencies produced by speakers in such locations do not get a straight shot at the listener. However, these locations are the most likely available ones in a car, and do serve rather well in this strange acoustic environment.

Installing Raw Speakers. If you can accomplish it mechanically, the best place for a couple of small midrange drivers is either directly in front of you in the dashboard or just under the dash. Here, one driver would be placed at each end of the dash and aimed slightly upward. However, because space is limited and because the open-bottom dash makes a poor woofer baffle, they should be supplemented by a woofer system located elsewhere. A passive crossover or an active arrangement with two sets of amplifiers should be used in this case.

The speakers should be rigidly mounted as nearly flush with the outer surface as possible. (Shallow cavities created by rear mounting tend to color the sound.) Some protection, in the form of a suitably cut slab of soft open-cell foam (available from electronics and carstereo stores), is required for the speakers. If you wish, back up the foam with aluminum window screening or lightweight hardware cloth for protection.
With some midrange drivers, especially smaller ones, you may not need tweeters. If you do-and you will if you use $5^{\prime \prime}(13-\mathrm{cm})$ drivers-use only one per channel. Locate the tweeters as close as possible to the midrange drivers, preferably right next to them. The tweeters can use their own special amplifier or the same amplifier as the mid-.
range drivers. In either case, a dividing network is required to protect the tweeters from potentially damaging low-frequency energy. Use a minimum value of $4 \mu \mathrm{~F}$ (nonelectrolytic) capacitance in series with the tweeter and amplifier for an 8 -ohm tweeter or $8 \mu \mathrm{~F}$ for a 4 -ohm tweeter. This provides a $6-\mathrm{dB}$ /octave rolloff below about 5000 Hz . You can improve on this by adding a low-resistance, $0.25-\mathrm{mH}$ air-core inductor in series with an 8 -ohm midrange driver or a $0.12-\mathrm{mH}$ choke in series with a 4 -ohm driver. Connections are shown in Fig. 1.

Door Mounting. If you cannot find a suitable location for the speaker drivers up front, you will have to put them in the doors or somewhere to the rear. Doors are not the very best choice for mechanical and acoustic reasons. For example, some speakers are too deep to fit into such shallow locations. Moreover, without adequate damping material, the outer sheet metal tends to reflect short-er-wavelength energy back through the cone. But given limited auto interior space, doors are deservedly popular speaker-mounting locations.

There are some precautions to observe when installing speakers in doors. For example, damping material is difficult to install unless you can prevent it from interfering with the window and latch mechanisms. For the same reason, it is risky to cut into the interior door panel just below the window, which is the best acoustical location. Doormounted speakers are, therefore, usually installed far down, where some of the upper frequencies are radiated into upholstery and carpet.

The proper way to install a speaker in a door is to remove the entire interior trim panel, which means removing at least the window crank as well. Then examine the exposed mechanical setup to determine if there is clearance for a small $3^{\prime \prime}$ to $5^{\prime \prime}(7.6$ to 12.7 cm ) driver as high and as far forward as possible.

Once this has been determined, cement as large a square of $1 / 2^{\prime \prime}$-thick felt or sound-deadening board as possible to the inside surface of the outer sheetmetal panel at that location; use roofing cement or silicone-rubber adhesive. Crank the window up and down to be sure the damping material does not interfere with any mechanism. Do not cut any holes in the interior panel until you have determined what you plan to do.

Rear-shelf Mounting. Another possible location for mounting raw, unenclosed speakers is the shelf under the rear window. Though there are several drawbacks to this location, too, most can be overcome. The most serious drawback is that the speaker backs "look" into a trunk that's a large, rather reflective resonant cavity common to both speakers. This will tend to reduce stereo separation and promote boomy bass. In small cars that are nearly airtight, you may even run the risk of rupturing the trunk-speaker cones if you slam the door with all windows rolled up. Some manufacturers such as Acoustic Fiber Sound have made a bid to solve these problems with enclosed speaker systems.

If the speakers face upward, sound will be reflected and dispersed by the rear window of the car. This is an effeclive way of throwing the sound into the car, but it further reduces stereo separation and stereo imaging, which is already somewhat peculiar because the speakers are behind the listeners.

Finally, the sun tends to roast anything on the rear shelf. Therefore you must protect the speakers with a foam grille. The grille itself will likely have to be replaced after a year or so when it crumbles. For mobile stereophiles who like robust sound with lots of bass and "hot" highs, the rear shelf is the place to put multiple-driver assemblies with $6^{\prime \prime} \times$ $9^{\prime \prime}$ woofers and separate tweeters (and even separate midrange drivers in some

Fig. 1. Simple passive crossover for use with separate woofer and tweeter rolls off 6 dB/octave below 5000 Hz .



Advent EQ-1 self-amplified $6^{\prime \prime} \times 9^{\prime \prime}$ speaker is equalized for rear-deck mounting.



Unorthodox door mounting of Radio Shack "Minimus 0.5" aims sound at listener's ears, but could use separate wooler for deeper bass.


Jensen Triaxial ${ }^{10}$ for front door panels has low-mounted $51 / 4 /$ " woofer, and surface-mounting mid-high module.
models). Since the midrange drivers and tweeters in such systems often have closed backs, channel-to-channel coupling via the trunk at middle frequencies is less of a problem. At low frequencies, most of the material on commercial pop recordings is essentially monophonic anyway, so there is no loss.

Advent has announced its Model EQ-1 twin $6^{\prime \prime} \times 9^{\prime \prime}(15.2 \times 22.9 \mathrm{~cm})$ powered speaker system that is equalized especially for rear-deck mounting. Also worth noting is the Polk Mini-Monitor, which may be too large for most cars, but not for some vans.

Coaxial speakers or preassembled systems with a woofer and a tweeter make sense if you keep in mind the general principles stated here. Jensen Sound Laboratories and KLH, among others, even have 3-way coaxial types.

Your vehicle may offer unique mounting possibilities. Some van owners, for example, have been happy witi speakers mounted into the seats themselves
or into the roof. A few general hints may help. First, low frequencies are essentially nondirectional, and the ear is easily fooled into placing the origin of the bass frequencies at the source from which the middle frequencies are emanating. Therefore, you are free to work with small drivers, leaving the bass from as high as 200 Hz and down to be reproduced by other speakers located under a seat or on the rear shelf. Also, small drivers are far easier to install in nearoptimum locations.

The second hint is to aim the midrange drivers and accompanying tweeters, if any, so they have a clear path to your ears. The larger the cone, the lower the frequency at which uneven dispersion begins: for a $3^{\prime \prime}$ driver, it is at about 4000 Hz ; for a $4^{\prime \prime}$ driver, at about 3000 Hz ; and for a 5" driver, at about 2500 Hz . This means that for any place you sit, off-axis output from the driver will be erratic and not easily predictable beyond those frequencies. In a high-quality sys-
tem, this is the chief reason for crossing over to a tweeter at those frequencies.
The third hint is to mount the midrange driver, where possible, in a rigid airtight box filled (but not packed tightly) with fiberglass or other insulation. An internal volume of 180 cu in. $\left(9^{\prime \prime} \times 5^{\prime \prime} \times\right.$ $4^{\prime \prime}$ ) for example, will allow smooth response down to about 120 Hz with a suitable $3^{\prime \prime}$ or $4^{\prime \prime}$ high-compliance driver. The box need not be rectangular; in fact, an irregular shape will reduce cavityresonance problems.

Finally, protect all drivers with an acoustically transparent foam or screen grille.

Speaker Connections. Phasing multiple drivers can be a challenge. With loudspeakers close together and wired in series or parallel, they are either in- or out-of-phase. In-phase hookups are preferred, for better bass response and clearer stereo imaging. Therefore, with identical parallel-connected speakers,
wire "like" terminals together (one terminal is usually identified with a spot of paint or a " + "); with identical speakers in series, wire opposite terminals together as shown in Fig. 2.

If you mix dissimilar speakers, you must first establish the correct phasing. You can do this with a single flashlight cell by temporarily connecting it across the speaker's terminals and observing the direction of cone travel. If the cone moves outward, mark the speaker terminal to which the positive terminal of the battery is connected with nail polish. If the cone moves inward, mark the speaker terminal to which the battery's negative terminal is connected. Do this with every driver you plan to use, whether or not they are already marked.

It's not as easy when crossover capacitors and inductors are involved. With a simple 6-dB/octave circuit like that shown in Fig. 1, there is a $90^{\circ}$ phase difference between the low (midrange) output and the high (tweeter) output. The usual practice is to connect the tweeter in a way that would be out-of-phase if there were no crossover network. Try making connections both ways. If you hear a difference, connect it the way that sounds best to you. With a properly phased set of drivers, the sound is more coherent than with improperly phased speakers.

A separate woofer system calls for a more sophisticated crossover. Because of the low frequencies involved, this is best handled with a separate amplifier and an active crossover network. Alternatively, you could use a passive network, of course. However, inductive values become large and the coils are therefore costlier. Additionally, there is some loss of power. If you want a passive woofer network, however, the circuit in Fig. 3 works well. It requires two $12-\mathrm{mH}$ iron-core chokes of less than 1ohm resistance (available from TSR Engineering, 5146 W. Imperial Hwy., Los Angeles, CA 90045 for $\$ 7.50$ each).

An active-filter circuit for one channel is illustrated in Fig. 4. The "low" output is intended to feed a separate power booster. More economically, a single crossover can be used for both channels, feeding a common bass amplifier/ speaker channel as shown. This requires only one bass amplifier (Setton makes a single-channel one), and one woofer. The main speakers then each receive a full-range signal.
A cluster of four $5^{\prime \prime}$ or $6^{\prime \prime}$ woofers makes sense at low frequencies if you can find the space for them. Use two 8-


Fig. 2. For correct phasing in multiple-speaker installations, wire "unlike" terminals in series as at (A), "like" ones in parallel as at (B).
ohm units in parallel across each power booster output. Unlike typical power amplifiers, boosters generally "like" 4-ohm loads. Be sure to check the manufacturer's instructions on this, though.

If you listen mostly to classical music, you can "fake it," if it is more convenient, by taking the bass from only one channel with a below-100-Hz dividing network (Fig. 3).

Finally, you may wish to consider a neat trick to give a sense of expanded space without expensive reverberation or delay devices. Simply connect two small ( $3^{\prime \prime}$ is fine), inexpensive drivers, wired in series and out-of-phase, from the "hot" side of one channel to the "hot" side of the other channel. (See Fig. 5.) Locate the drivers someplace to the rear of the car, on either side of the rear window, for example. The type of enclosure used is unimportant.

A 25-ohm speaker level control potentiometer wired in series with the drivers will allow you to experiment with the level of the rear ambience until it satisfies you. Also, a nonpolarized 8- or $10-\mu \mathrm{F}$ capacitor connected across the speaker pair attenuates the upper couple of octaves to make the "recovered ambience" sound more like actual reflected sound. The effect of this hookup is very similar to some matrix quadraphonic schemes at nowhere near the expense. (For true 4-channel sound, an eighttrack tape deck would be required.) It also does wonders for stereo; rear-seat passengers will enjoy it, too.

Amplifiers. A single transformerless class-B output stage working with a 12 volt supply (actually about 13.8 volts) cannot produce more than about 2 watts of continuous power into an 8 -ohm load or 4 watts into a 4 -ohm load without being overdriven. Since this is what most low-cost car stereos use, you can see that the output power claims for some of them are wildly optimistic. Hence the market for power boosters.

There are several ways to obtain more power from a 12 -volt source. The most common and least expensive is limited to a theoretical maximum of four times the "simple" power output. How this is done is shown in Fig. 6. Two identical amplifiers are fed the same input, both out-of-phase. The speaker load is connected between the "hot" outputs, neither side of which can be grounded. One amplifier "pushes" as the other "pulls." The net effect is to double the supply voltage, which produces four times the power in the same resistive load. This works out to about 8 watts into 8 ohms and 16 watts into 4 ohms.



Unscrewing carpet-retaining plate provides handy channel for wires feeding signals to amplifiers or to the rear speakers.


Reinstalling the door panel, a snap-in grille mounting plate is secured. With some makes, a grille snaps onto the speaker or is held by speaker-mount screws.


Cut speaker hole in rear deck with a sharp blade or sabre saw. Many trunks have oval stampings underthe deck that can be used as guides when cutting $6^{\prime \prime} \times 9^{\prime \prime}$ holes.


Removing front door panel shows where
this "Kriket" speaker can mount.
(Here, there's a pre-cut hole.)
Attach wires before securing speaker.


Sound-deadening board, secured with roofing cement, suppresses rattles and resonances which may occur with sheet-metal door panels.


Add speaker grilles to protect cones and to enhance appearance.
Use bolts, not screws, to fasten grille.
One of two speakers on back deck is shown.

| DESIRED <br> CROSSOVER <br> FREO (HI) | $C$ <br> $(\mu \mathrm{~F})$ |
| :---: | :---: |
| 70 | 0.28 |
| 80 | 0.25 |
| 100 | 0.20 |
| 120 | 0.16 |
| 150 | 0.13 |
| 200 | 0.10 |



Fig. 4. Electronic crossover feeds highs and lows to separate amplifiers. Crossover frequency depends on value of capacitor. Summing network
(dashed box) lets you use one crossover and one bass amplifier to feed a common woofer. Otherwise, 2 crossovers and 4 amplifiers are required.

A variation of this technique uses an autotransformer to step up the voltage to the speaker. Since almost unlimited current is available from the car's electrical system, it is possible to use high-current output transistors to swing as much as 3 or 4 amperes into a transformer winding. In effect, the speaker's impedance is stepped down to 2 ohms or less by the transformer. Since output power is inversely proportional to load impedance, it goes up accordingly.

A third way is to use a dc-to-dc converter to raise the supply voltage. The output power available with this technique is limited only by the current that can be drawn from the car's electrical system without taxing other functions it must also serve. This approach makes it possible to adapt the power-amplifier circuits from home audio components to mobile service. All that is necessary is to design a switching-type dc converter with sufficient voltage, current, and regulation to feed the desired amplifier.

A fourth method you can use is biamplification. By dividing the audio spectrum into two or more parts, with corresponding numbers of amplifiers in each channel, each feeding a suitable speaker, you effectively parallel the amplifiers across the 12 -volt supply. The combined output from this scheme is several times greater than you would expect, thanks to the uneven distribution of energy in mu-
sic. (This is the technique employed in the ADS Model 2001 and 2002 systems and some Sanyo stereos.)

Unless a specification sheet explicitly mentions "FTC ratings," car-amplifier output power figures should be taken with a large grain of salt. Copy writers can come up with amazing power figures by ignoring clipping, assuming a 14 - or 15 -volt supply and a 4 -ohm or less impedance, using "peak" power, etc. Read before you buy, and be prepared to pay reasonably high prices for good amplifiers. The $\$ 20$ power boosters are likely to be disappointing.

Most boosters are designed to operate from the speaker outputs of a complete in-dash or under-dash radio receiver and/or tape player. Since the outputs of such a unit generally include a large series dc-blocking electrolytic capacitor, the boosters usually have a lowvalue resistor ( 10 to 100 ohms) across their inputs. This reduces the ill effects of dc leakage through the capacitor. Unfortunately, the low input shunt resistance also makes it impossible to drive the boosters like normal power amplifiers, directly from the volume control of the main unit. It is usually possible to snip one lead of this input shunt to remove it. Bear in mind, however, that by opening the booster's case you will in all likelihood void its warranty if tinkering is noticeable.

Once you remove the shunt, you have a two-channel power amplifier that can be used directly from the volume or tone controls of your main unit or with the active crossover described in this article.

Some power boosters have a transformer input and no shunt resistor. These can definitely not be used from high-impedance internal feeds.
One manufacturer hopes to standardize the industry on a 1000 -ohm source impedance and 10,000-ohm input, with typically $250-\mathrm{mV}$ input levels. But until that occurs, you are on your own. Impedance matching is not the issue here. However, a 50,000 -ohm volume control will not take kindly to a load of 100 ohms from its wiper to ground. The output level will be severely reduced, and there may be distortion at the top end of the volume control. If you mate components from different manufacturers, you will have to obtain schematic diagrams from each manufacturer and examine them to determine if the units you plan to use are compatible. However, most units are compatible, with minor modifications.
You can solve some mating problems by using a pair of emitter-follower circuits, as shown in Fig. 7. (The maximum source impedance at a volume-control wiper is one-fourth the value of the control if the control is fed from a low impedance. It occurs at the setting where the
output level is 6 dB down from maximum.) Assuming a 50,000 -ohm control, the highest source impedance will typically be between 10,000 and 15,000 ohms, depending on what precedes the control. It is unwise to drive more than $10^{\prime}$ to $15^{\prime}$ ( 3 to 4.6 m ) of shielded cable at this impedance. If you do, high-frequency losses may become audible. However, you may need greater length than this in a large car if the power amplifier is located in the trunk. This is where an emitter-follower circuit can prove useful!
The outputs of mobile booster amplifiers usually float with respect to ground. Hence, use separate cunductor pairs for connections to all speakers. If either side of a speaker output is grounded or a common lead between outputs is used, the booster amplifier can be damaged. Use nothing smaller than No. 18 wire for dc supply leads and speaker wiring. In fact, No. 16 is better, especially with 4 -ohm speakers and speakers wired in parallel.

Booster amplifiers can usually be located anywhere. Some have power switches, but you can avoid the need for accessing them by using the circuit in Fig. 8. This allows all accessories to be turned on and off from the switch on the main dashboard receiver or player. If the dash receiver you choose has a connection for a power-driven antenna, you can use that to operate a relay like the one in Fig. 8. The contacts of the relay can then switch the boosters and other devices. Be sure, however, that the relay can handle at least half of the maximum current drawn by the booster amplifier at full output.

Avoid mounting boosters (or any electronics, for that matter) near heater outlets or where they will be exposed to direct sunlight or engine heat. Some boosters are quite heavy, which necessitates the need for secure mounting. The heavier ones are best laid flat on a sturdy surface and then bolted down with No. 8 or No. 10 machine screws, lockwashers, and nuts.

You are on your own when it comes to connectors. Only a small number of mobile electronics gear uses the familiar phono jack and plug. With the manufacturer's instructions and a schematic diagram, which is often but not always supplied, you will usually be able to identify the various leads and attach suitable connectors if needed. Do not expect one manufacturer's booster amplifier to interface with another's dashboard unit without connector modification.

Tape Units. Most serious audiophiles have a strong preference for cassettes over the eight-track tape format. Cassette tapes are small and improving constantly. Furthermore, there is a greater variety of home-recording cassette hardware enabling one to use the same tapes in an auto cassette machine. Also, tape winding and rewinding are much faster, and Dolby noise reduction is available. Moreover, some manufacturers produce mobile cassette machines with recording provisions. At least two of them (Sanyo and Blaupunkt) make indash systems with AM and stereo FM receivers and cassette decks that allow recording of cassettes in stereo from the receiver or in mono via a microphone. But if you want eight-track cartridge capability, several manufacturers will oblige you with in-dash and under-dash models.

The top-of-the-line cassette tape units at this writing are from Nakamichi and Uher, both of which are quite expensive but can also be used as self-powered portables. The Nakamichi Model 250 is a player only, while the 350 also records. Both have built-in Dolby NR circuits. The Uher Model 210 is a recorder/ player sans Dolby. Some better-thanaverage players have begun to appear in in-dash combination systems with stereo receivers. Examples of these include the Concord Models HP-100 and HP-350, Motorola TC876AX, Sanyo

Model FT1490A with Dolby, Kraco KID-588 with auto reverse, and Pioneer Model KPH-9000, among others. Dynascan Corp.'s Cobra 50XLR in-dash combo also incorporates a CB transceiver.

When you are out shopping, read the specifications carefully and check out missing entries. And don't always believe that a low price indicates a bargain. Sorre run-of-the-mill cassette machines exhibit inordinately high flutter, especially with thinner C-90 and C-120 tapes. A special nuisance is "warble" on a bumpy road, caused by a poorly designed or adjusted tape transport.

FM Receivers. Until recently, the best mobile FM tuners were poor performers. Then Pioneer spearheaded a change in this image a while back with its "Supertuner" line. More recently, Sanyo with its "Audio Spec" line, Concord, and others have joined the ranks.

If you want absolutely everything in your car stereo, now you can almost get it. The latest in car-stereo units incorporate stereo cassette with auto-reverse, plus AM/FM-stereo radios with digital tuning and readouts (that double as clocks) plus three tuning modes: autoscan, preselect, and manual.

The first three of what promises to be a new wave of high-end, do-all stereo units are the Fujitsu Ten ETX-41B (\$275), the J.I.L. 634E (\$350), and the



The Pioneer "Supertuner" underdash TP-900 has easy-to-read circular FM dial, 8 -track tape.

Separate components-cassette deck, tuner and amplifier-make up this Panasonic stereo system.



Sanyo underdash cassette player comes with slide-in/out bracket.


Cassette, CB and AM/stereo FM are combined in this $5^{\prime \prime}$-deep Cobra 50 XLR.

Panasonic Q8520 (\$399.95). Their tuning facilities differ somewhat: the J.I.L. has only 4 station-select buttons (which bring in 8 stations- 4 AM and 4 FM ), but auto-scans in two modes, either stopping at the first station found, or pausing there unless instructed to stop. The Fujitsu unit can preselect $5+5$ stations (with a rotary switch instead of pushbuttons), and only has the stop-until-restarted auto-tuning mode. The Panasonic has similar auto-tuning, but preselects $6+6$ stations and has an unconventional "manual" tuning mode: a pair of buttons which scan the dial up or down until released.
As to tape features, the Panasonic offers Repeatrack, which automatically restarts the tape after rewinding, the Fujitsu has Dolby, and the J.I.L. is also available in an 8 -track version, the Model 874E.

On the presumption that you'll use amplifiers or boosters with them. all three have low power-about 15 watts per channel at $10 \%$ distortion for the J.I.L. and Panasonic, and zero watts for
the Fujitsu (it has no power amplifier stage at all).

If the other features don't impress you that much, but the digital tuning does, less feature-laden digital stereo/clock units are available from Audiovox, Boman, Craig and Sanyo.

Though still not on a par with fine
home component FM tuners, the new breed of high-quality car radios packs a lot of performance capability into a small package. When you are shopping around, read the manufacturers' specifications for their receivers very carefully. The published specifications for the better receivers are quite comprehensive.

Fig. 7. Pair of emitter-follower circuits permits using lengthy speaker lines without highfrequency loss.




Spark'Dmatic GE-500 40-W rms equalizer/ booster displays illuminated response curve, features front-rear fader control.

Trurk-mounted 7-inch Isophon woofer "speaks" into rear-seat back betow 120 Hz . Driven by one channel of Jandy booster.


Motorola equalizer/booster has $30-\mathrm{W}$ rms power output, LED power indicators.


Fosgate 200-W amp has preamp with source selector, power LED's, 3 -way equalization.

Fujitsu Ten preamp has 5 inputs, including mike mixing.

They give figures for sensitivity in stereo as well as mono, capture ratio, alter-nate-channel selectivity, etc. Always look for a complete statement for a specification. For example, a " $2-\mu \mathrm{V}$ " sensitivity rating is meaningless if the background noise suppression is not
stated. It might take $50 \mu \mathrm{~V}$ or more to give you a listenable signal in stereo.

FM reception in a moving vehicle is difficult at best, of course. An unaided vertical whip antenna does not capture much signal, especially from stations that transmit only a hosizontally polar-


Fig. 8. Relay circuit for remote control of booster amplifier.
ized signal. (More and more FM stations are going to circular polarization, however, which includes a vertical field component that maintains its strength near the ground.) The single omnidirectional whip antenna is also vulnerable to multipath interference (delayed reflections combining with the direct signal), which severely distorts the signal and causes dropouts and noise bursts in the sound.

Horizontally oriented windshield strip antennas eliminate the polarization problem, but tend to favor the fore and aft directions. The best antenna at present is still a single front-fender-mounted $31^{n}(78.7-\mathrm{cm})$ stainless-steel whip, connected by a short length of coaxial cable directly to the antenna input of the receiver. The coax shield should be well

Amplifiers, Boosters, Equalizers-<br>Who Makes What?

If you're searching for an amplifier or booster for your car stereo, it helps to know who has products in your desired power range. Here's a quick run-down of the power ranges available from each company, in watts per channel for 4 ohms , at less than $1 \%$ distortion. Power figures with higher or unspecified distortion levels are marked with asterisks. Figures marked " $E$ " are for booster/equalizers.

| Company | Watts/Channel |
| :--- | :--- |
| ADS | 25 |
| Audiomobile | $20-75$, E20 |
| Audiovox | E16 $^{\circ}$ |
| Boman Astrosonix | $25^{\circ}$ |
| Clarion | E10-20, 12.5 |
| Concord | 18 |
| Craig | $12-25$, E36 |
| Fosgate | $20-50$ |
| Fujitsu Ten | 20 |
| Jandy Car-Fi | $30-80$, E20-30 |
| Kraco | $20^{*}$, E25-30 |
| Kustom Kreations | $37.5^{\circ}$ |
| Laser Acoustics | $65-175$, E65-175 |
| Linear Power |  |
| (Shmegg) | $15-125$ |
| Mega | 25 |
| Motorola | $12-20$ |
| MetroSound | $20-35$, E20 |
| Muntz | $15-20^{*}$ |
| Panasonic | 10 |
| Pioneer | $12-20$ |
| Prime | E10 |
| Pyramid | $20-40$, E22 |
| Royal Sound | $10-30^{\circ}$, E25 |
| Sanyo | $15-25$ |
| Sanyo biamp | $23+5 \mathrm{~W}$ (14 3\% |
| Setton | 40 |
| Sonic Boom | $12.5-45$ |
| SparkOmatic | $15 *^{*}$, E20 |
|  |  |

grounded to the vehicle's chassis at the antenna end. Telescoping designs that can be extended to several feet may be somewhat better for AM reception. But they may be a source of noisy or erratic reception as they age and contact between the sections deteriorates. They are also more easily broken.

FM boosters are of questionable usefulness. With a low-grade FM receiver in a weak-signal area, they can be helpful. But in urban areas where signals are strong and multipath reception is a problem, such a preamplifier can make things worse by increasing crossmodulation distortion. The best mobile tuners are the least likely to benefit from an
auxiliary preamp, and their performance may actually be degraded by one unless its noise figure, almost never published for these devices, is better than that of the $r-f$ stage in the receiver itself.

Accessories. A variation of the power booster amplifier that includes tone controls or a "graphic" equalizer is a growing and popular addition to a mobile sound system. It can be very useful, but bear in mind that even a relatively small audible boost in some parts of the frequency spectrum can demand significantly more power from your amplifiers. Also, don't apply excessive boost at very low frequencies since small speakers may reach the limits of their cone excursions, resulting in distortion and possible damage.

Though a graphic equalizer may help even out a vehicle's peculiar acoustics, the typical five-band unit will not provide enough compensation to wholly flatten out the overall system response. But it may alter the sound's tonal balance to reduce some of the masking caused by noises in a moving vehicle. There is at least one seven-band car equalizer. from Jandy Car-Fi. Moreover, there are nongraphic equalizers available from Laser Acoustics. These have provisions for screwdriver-adjustable frequency settings which can't be misadjusted by curious passengers.

At this writing, the power amplifiers built into most graphic-equalizer accessories are not particularly distinguished, some being rated at as much as $10 \%$ distortion. This picture will undoubtedly change. Again, remember to read and interpret all published specifications.

Finishing Up. Ignition noise is not the problem it once was. Most recent cars are equipped by the factory with interfer-ence-reducing ignition wiring. Older cars and cars ordered without factory-
installed radio receivers may have to be fitted with resistor-type spark plugs or add-on resistors or be rewired with resistive ignition cable.

Alternators are not usually prime causes of FM interference. If yours is, you can get suppression kits from, say, Radio Shack, Lafayette Radio and other sources to eliminate the interference. Too, if your car has the older type of dc generator, you might have to install a suppressor. The kit should include coaxial capacitors rather than the more common axial- or radial-lead capacitors.

Coaxial capacitors can also aid in reducing noise from fan and wiper motors and electric fuel pumps. A single $0.25-\mu \mathrm{F}$ coaxial capacitor connected very near the offending device, and solidly grounded nearby through its mounting lug, will often render the interference inaudible. Sometimes you will have to add inductance in series with the supply lead to the motor. A value of 1 mH or thereabouts should do nicely. Be sure that the inductor and coaxial capacitor are rated to handle the current drawn by the motor when it is operating at maximum speed.

If you run into persistent electrical interference in your car, the Mobile Manual for Radio Amateurs contains much information that is useful for combatting the problem. It is published by the American Radio Relay League (ARRL) and is available from most electronics parts and equipment stores.

In Conclusion. With careful selection of mobile audio components, installation, and interfacing, you can easily obtain good high-fidelity sound in your car, van, or RV. The hints presented here should help you tailor your system to the peculiar acoustical and electrical environment of your vehicle. Then you and your passengers can enjoy more realistic music reproduction.
$\diamond$


Solid-state system uses new IC to achieve high reliability at low cost.

UNTIL recently, all automotive voltage regulators have been electromechanical devices. As a result, they exhibit all the disadvantages associated with relays and breaker points-wear, chatter, changes in spacing, and pitted contacts due to arcing. The latest model solid-state regulators are inherently more reliable. Now, with the introduction of a new IC from Motorola, it's possible to build one at low cost to replace mechanical or costly discrete component solid-state regulators.

In addition to the new IC, the full-feature, solid-state voltage regulator presented here employs only a handful of discrete components. It is easy to construct and install, and offers selectable temperature coefficient, overvoltage protection, as well as automatic shutdown in case of loss of battery voltage.

About the Circuit. A typical automotive charging system is shown in Fig. 1. The alternator differs from a generator with a fixed field in that the magnetic field is derived from dc flowing in the ro-
tating field winding. Field current, usually adout 3 amperes, is supplied via the voltage regulator. The alternator stator windings are usually connected in a three-phase wye ( $Y$ ) configuration producing an ac output. Six silicon diodes form a three-phase, full-wave rectifier which converts the ac to dc.

The output of the alternator is a function of both shaft speed and magnetic field strength. Variations in load resistance and shaft speed can be compensated for by changing the amplitude of the field winding current and hence the strength of the magnetic field. That's exactly what the voltage regulator does.

An electromechanical regulator accomplishes this task by "chopping" the field current or inserting into and removing a fixed resistor from the field current loop. The contacts of such a regulator can be opening and closing more than 200 times each second. Solid-state voltage regulators, including this project, govern the amplitude of field current electronically. A voltage-dependent current source is used to drive a power

transistor whose output current excites the field winding. Stock regulators typically employ a zener diode, several fixed resistors, a thermistor which sets the regulator's temperature coefficient, a driver transistor and a power output transistor.

The internal structure of the Motorola MC3325 monolithic silicon IC, upon which this project is based, is shown in Fig. 2. The integrated circuit can be viewed as composed of four separate sections-a temperature coefficient circuit, a battery-voltage loss detector, an overvoltage detector and an output drive amplifier. Each of these sections can be biased independently by the addition of a few passive components. The IC is designed to drive an npn Darlington transistor which in turn controls current through the field winding.

Shown in Fig. 3 are the three basic alternator types commonly used in automobiles. The regulator presented in this artide can be used with the floating field (A) and pulled-up field (B) alternators. However, the regulator cannot be used with a grounded field alternator (C) unless the output stage of the regulator is modified. Alternatively, the alternator can be modified to make it a floating field type. The project has not been tested with this type of alternator, but provisions for the required regulator modifications have been included on the pcboard foil pattern.

Some automobiles employ alternators with integral voltage regulators. That is, the voltage regulator is enclosed in the alternator housing. The project can be used with such an alternator if the housing is removed, the voltage regulator by-


Fig. 2. Schematic of the internal structure of the Motorola MC3325 automotive voltage regulator IC.

## PARTSLIST

C1. C2 $-0.01-\mu \mathrm{F}$ disc ceramic
DI- 1 N 4003 rectifier
ICI-MC3325 automotive voltage regulator (Motorola)
Q1-2N6059 npn Darlington transistor
The following are $1 / 4$-watt, $5 \%$ carbon-composition resistors:
RI- 13,000 ohms
R2- 1000 ohms
R3- 1600 ohms
R4-2200 ohms
R5-2000 ohms
R6- 1500 ohms
R7- 3000 ohms
Misc.-Printed circuit board, suitable metallic enclosure, heat sink, power transistor socket, silicone thermal compound, terminal strip. IC socket or Molex Soldercons, pc board standoff, machine hardware, etc.
Note: The following are available from Questar Engineering Co., 50 South McDonald Street, Mesa, Arizona 85202: Etched and drilled glass-epoxy printed circuit board for \$5.25; 2N6059 npn Darlington transistor for $\$ 5.85$; MC3325 IC for \$1.75. Arizona residents add sales tax.

passed, and the alternator rewired so that the field winding and rectifier output are connected to the external terminal block. The exact modifications required vary with each charging system.

The schematic diagram of the IC voltage regulator is shown in Fig. 4. Very few external components are needed. Darlington transistor Q1 drives the alternator field coil. The fixed resistors are

Fig. 3. Three types of alternators found in automobiles: floating field (A); pulled-up field (B); grounded field (C).
used to bias the various sections of IC1. Resistor R3 functions as a current limiting resistor in case an overvoltage condition occurs at the output diodes of the alternator. The magnitude of the regulated voltage is determined by R5. Resistor R1 establishes the current in the IC's diode string. The magnitude of this current determines the regulator's temperature coefficient. The maximum

(B)
overvoltage of the charging system is limited by $R 6$.

Construction. Printed circuit construction techniques are recommended for the assembly of the voltage regulator. Etching and drilling and parts placement guides for a suitable board are shown in Fig. 5. All components except Q1 are mounted on the printed circuit board. Use a minimum amount of heat consistent with the formation of good solder joints and pay close attention to polarities and pin basing of semiconductors. An IC socket or Molex Soldercons will facilitate the installation of IC1.

The Darlington transistor should be mounted on a heat sink attached to the metallic project enclosure. A power transistor socket, a mica insulating washer, silicone thermal compound, machine hardware and shoulder washers should be employed. A suitable terminal (pre-


For use with grounded-field alternator, use dashed lines.
ferably of the barrier block type) should be mounted on the enclosure to provide convenient points for the ground, battery sense, field and alternator output connections. Interconnect the pc board, transistor socket and terminal strip using No. 16 or heavier insulated, stranded copper wire. Mount the pc board in the
enclosure using metallic standoffs and machine hardware.

Adjustments. The key circuit parameters can be optimized for use in a particular charging system by changing the values of certain fixed resistors, namely, R3, R5, and R6. Before changing any of


Fig. 5. Etching and drilling guide for pc board is at left, with parts placement guide shown below.

the values specified in the parts list, consult your automobile's specifications. Then use the following information to determine the resistor values.
Resistor R3 limits current through the zener diode internally connected to pin 3 of IC1. In operation, the voltage between pin 3 and ground (the sum of the zener voltage and the voltage drop across the transistor's base-emitter diode) will be about +7.5 volts. The value of $R 3$ should be chosen so that the current flowing through the resistor, internal zener, and base-emitter diode at maximum overvoltage is between 2.0 and 6.0 milliamperes.

The following equation is used to compute the value of $R 3$ : $R 3=\left(V_{\text {overmax }}\right.$ $\left.-V_{2}\right) / I_{3}$, where $V_{2}$ is the zener voltage, +7.5 volts, $I_{3}$ is the current flowing into pin 3, between 2.0 and 6.0 mA , and $V_{\text {overmax }}$ is the maximum output voltage of the alternator. The specified value of $R 3$ is 1600 ohms. If the alternator output varies between 13.0 and 16.0 volts, $I_{3}$ varies between 3.4 and 5.3 mA . Thus, over the full range of voltages anticipated, the zener current is less than 6.0 and greater than 2.0 mA .
The magnitude of the regulated voltage is determined by the value of $R 5$, which is defined by the following equation: $V_{\text {reg }}=(1+R 5 / R 1)(8.4)+(n+$ R5/5000) (0.7), where $n$ is 3 for the circuit shown in Figs. 4 and 5 and R1 is chosen so that the current in the battery sensing diode string is between 0.5 and 1.0 mA . This current in part establishes the regulator's temperature coefficient, which can be varied from approximately $-9.0 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ to $-13.0 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ by changing the number of diodes in the string.

The approximate temperature coefficient of the zener diode is $+3.0 \mathrm{mV} /{ }^{\circ} \mathrm{C}$, and that of the string diodes and transistor base-emitter diodes is $-2.0 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ each. Starting at pin 1 (refer back to Fig. 2), and counting upward, we add -2.0 $\mathrm{mV} /{ }^{\circ} \mathrm{C}$ for the first base-emitter diode, $-2.0 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ for the second, $+3.0 \mathrm{mV} /$ ${ }^{\circ} \mathrm{C}$ for the zener, and $-10 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ for the five diodes between pins 8 and 6 . This results in a total temperature coefficient of $-11 \mathrm{mV} /{ }^{\circ} \mathrm{C}$.

The voltage between pin 8 and ground should range from +7.9 volts minimum to 8.8 volts maximum. The resistance of R1 should be greater than 7800 ohms and less than 16,600 ohms, where $R_{\text {max }}=V_{\text {max }} / I_{\text {min }}$ and $R_{\text {min }}=V_{\text {min }}$ $/ t_{\text {max }}$. If a typical voltage of 8.35 volts and a current through the diode string of 0.64 mA are assumed, the resistance of Ri should be 13,000 ohms. This results


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in a minimum current of 0.61 mA and a maximum of 0.68 mA for the specified range of voltages between pin 8 and ground. In most charging systems, the normal charging voltage will be between 13.3 and 15.3 volts. Accordingly, 14.0 volts is a good choice for $V_{\text {reg }}$. Solving the equation previously given for a $V_{\text {reg }}$ of 14.0 volts results in an $R 5$ resistance of 2333.3 ohms. A 2000 -ohm resistor will cause $V_{\text {reg }}$ to be 14.3 volts.
The value of $R 6$ is more easily determined. This resistor limits the maximum overvoltage, which can be expressed as: $V_{\text {overmax }}=[(R 3+R 6) / R 6](7.5)$. We have previously calculated the resistance of $R 3$ to be 1600 ohms, so we can simplify the equation: $V_{\text {overmax }}=[(1600+$ $R 6$ )/R6 ] (7.5). If the maximum overvoltage is chosen to be 16.0 volts, then R6 should have a resistance of 1411.8 ohms. The closest "standard" value, 1500 ohms, will limit $V_{\text {overmax }}$ to 15.5 volts. Because this 88.2 -ohm change in resistance varies the maximum overvoltage by 0.5 volt, a $5 \%$ (or closer) tolerance component should be used for R6. In fact, it's a good idea to use resistors with such tolerances for all three biasing resistors (R3, R5, and R6).

If a Darlington transistor other than the type specified in the schematic and Parts List is used as Q1, it might be necessary to change the value of $R 2$. Check the data sheet of the Darlington transistor for the beta of the device. Then select a value of R2 that will provide enough drive when the alternator output voltage is at its minimum level. An expression you can use to determine the required resistance is: $I_{\text {drive }}=\left(V_{\min }\right.$ $2.8) /(R 2+50)$.

Installation. The project should be mounted at or near the location of the voltage regulator that is being re-placed-except if the existing regulator is an integral one. In any event, the regulator should be positioned near the battery for good thermal tracking. All interconnections between the regulator and the rest of the charging system should be made with flame-retardant, oil-resistant insulated stranded wire. Use No. 14 wire for the battery, ground, and field connections and No. 10 wire for the connection to the alternator output terminal. Most, if not all, of the wires needed should be already present. If not, add a required length of suitable wire, keeping the gauge requirements just mentioned in mind. Finally, be sure the regulator is firmly secured in place to prevent possible damage caused by vibration.

POPULAR ELECTRONICS

# Low-cost loop antenna EXTENDS AM RADIO RECEPTION 

> Easy-to-build air-core loop helps pull in distant stations on inexpensive radios.

EVEN IF you're vacationing too far from home for normal AM reception, you can still pick up home-town broadcasts with an ordinary AM radio. Alternatively, if you're staying at home, you can receive out-of-town sports broadcasts to keep tabs on your favorite team. Using an inexpensive external loop antenna will do the trick. Here's why it works and how to build one.

Because portable and desk-top AM receivers employ relatively small, internally mounted ferrite-core loop antennas, they can deliver only enough signal for good reception of local stations. However, if an external loop with a larger effective cross-sectional area is substituted for the internal one, or used in tandem with it by mutual coupling, the working sensitivity of the receiver is increased in direct proportion to the ratio of the loop areas.

If the loops are used in tandem, no connections or modifications to the receiver are necessary. Signals will be coupled to the small internal loop induc-

tively when the two loops are placed in proximity to each other. If your home is of wood-frame construction and the walls do not have metal lath, you can mount the large loop on a wall or even conceal the loop behind it. The loop can then be a source of fun as a mystery spot where your neighbor's $\$ 5$ transistor radio will work better than ever before!

Constructing a Loop. A typical large loop antenna is shown in the figure. It is made simply by winding a series of turns of wire on some supporting structure. The loop is tuned by a variable capacitor connected across it. The antenna can be supported by wooden pegs inserted into the wall or by a free-standing wooden cross frame. Insulated copper wire, No. 20 or larger, should be used. Bell wire or even No. 14 house wire will yield excellent results. Such a loop can be concealed if other members of the family consider it unsightly.

Plan to make your loop square, or at most, slightly rectangular. This makes it easy to compute the area inside the loop. Construct your loop so that it is as large as possible. A $7-\mathrm{ft} \times 9-\mathrm{ft}(2.1-\times$ $2.7-\mathrm{m}$ ) loop, for example, is suitable if you have 8 - $\mathrm{ft}(2.4-\mathrm{m})$ walls. If possible, mount the loop on a wall which is in line with the distant radio stations you want to receive. The antenna is most sensitive to signals parallel to the plane of the loop, and least sensitive to signals propagating in directions perpendicular to it (striking the antenna broadsides).

To calculate how many turns of wire are needed, compute the area of the proposed loop and use the following formula: $N=242.3 / \sqrt{ } A$ where $N$ is the number of turns and $A$ is the area in square inches ( 1 square inch $=6.45$ $\mathrm{cm}^{2}$ ). For example, suppose the planned loop will measure $6^{\prime} 9^{\prime \prime}(2.1 \mathrm{~m})$
on each side. Its area will be 6,561 square inches ( $4.2 \mathrm{~m}^{2}$ or $42,330 \mathrm{~cm}^{2}$ ) and the number of turns required will be three. For your convenience, here are the loop sizes corresponding to an integral number of turns:

| N | Length of each side |
| :---: | :--- |
| 3 | $80-11 / 16^{\prime \prime}(2.05 \mathrm{~m})$ |
| 4 | $60-5 / 8^{\prime \prime}(1.60 \mathrm{~m})$ |
| 5 | $48^{\prime \prime}(1.23 \mathrm{~m})$ |
| 6 | $40-3 / 8^{\prime \prime}(1.03 \mathrm{~m})$ |
| 7 | $34-5 / 8^{\prime \prime}(88 \mathrm{~cm})$ |
| 8 | $30-5 / 16^{\prime \prime}(77 \mathrm{~cm})$ |
| 9 | $26-7 / 8^{\prime \prime}(68.3 \mathrm{~cm})$ |
| 10 | $24-3 / 16^{\prime \prime}(61.4 \mathrm{~cm})$ |

Incidentally, you can make a small, portable loop on a wooden frame to take along on picnics, or on a boat. A loop two feet square ( $0.3716 \mathrm{~m}^{2}$ or $3716 \mathrm{~cm}^{2}$ ) will provide good results with a "pocket portable" receiver.

Connect the loop ends to each side of an ordinary air dielectric variable tuning capacitor (one loop end to the rotor plates and the other to the stator plates). The capacitor, which can be removed from a junked AM receiver or purchased new (or surplus), should have a maximum capacitance of at least 360 pF . Multisection capacitors can be wired in parallel to extend the loop's tuning range. Be sure to solder all connections using rosin core solder.

Using the Loop. A loop antenna will provide some improvement in reception of all stations, not just the one at the frequency to which it is tuned. However, for best results the loop should be resonated. Tune the receiver to the desired station's frequency and place it in the vicinity of the large loop. Orient the receiver so that its internal ferrite bar is perpendicular to the plane of the loop. Then rotate the shaft of the antenna's tuning capacitor until the signal peaks.

Enhanced reception will be experienced when the receiver is placed up to approximately one side dimension of the loop in front of, or behind the wall on which the loop is mounted. Experiment with the placement of the receiver to determine the location that gives best results. The closer the receiver is to the loop, the more signal coupled to the internal ferrite antenna. For casual listening, as opposed to chasing weak DX signals, the degree of coupling between the loop and the receiver will not be critical, thanks to the large measure of improvement the loop provides.

# Shortwave DX "CATCHES" FROM AFRICA 

## Broadcasts from a changing continent can be informative and interesting.

sOME people tune the shortwave bands to keep closer tabs on what's happening in the rest of the world. Others seek distant, rarely heard DX stations. It's not often that one section of the world can offer exceptional opportunities for both categories of SWL's, but Africa certainly can. There are numerous stations to challenge your skills as a dial-spinner, while at the same time offering the politically-aware news-oriented SWL some of the most potentially explosive listening excitement to be found anywhere.

Major DX'ing fireworks are provided by the rapidly escalating radio war between the white minority regimes in South Africa and Rhodesia, and the black nations that surround them in southern Africa. More interesting listening comes from the numerous former colonies struggling to achieve stable nationhood, undergoing various internal upheavals in the process. Such instability makes extracting a QSL from many African outlets an even greater chal-
lenge than hearing the station! Yet such situations can provide supreme DX satisfaction when overcome, resulting in the arrival of a highly prized verification.

South Africa's Big Voice. There's no question as to which nation has the biggest signal in Africa: The Republic of South Africa puts a thumping signal into all parts of the world with its external service, Radio RSA. In addition, it has an extensive network of domestic shortwave services which can be easily heard throughout North America.

Radio RSA currently broadcasts in English to North America at 2230-2320 GMT on 9585,11800 , and 11900 kHz . Programming is typical of most govern-ment-operated broadcasters: news, editorials, features, music, and mailbag programs. For the SWL interested in African affairs and politics, Radio RSA can provide insights into South African foreign policy that cannot be obtained through conventional American news media.
(SSL card sent by the Ghana Broadcasting Corporation verifying reception.

## GHANA BROADCASTING CORPORATION

P.O. BOX 1633

ACCRA, GHANA

Dear Sir/Madam,
Thank you for your reception report of our transmission(s)
on $\quad 11850$
ke/s heard at 2015 -
2100
G.M.T. on

1: $5: 68$
We have pleasure in verifying your report which is much appreciated.


The governmental body responsible for domestic broadcasting services is the South African Broadcasting Corporation (SABC), which provides programming in English, the Dutch dialect Afrikaans, and various tribal vernaculars. SABC's English service is widely heard on 4875 kHz from 0300 GMT sign-on until fade out at sunrise in South Africa. Programming is pop music and a generally subdued announcing style. Sometimes telephone call-in shows are featured, and these are among the most fascinating programs you're likely to find on shortwave.

Private broadcasting stations are not allowed in South Africa, but SABC provides its own commercial service known as Springbok Radio. This service operates all through the South African night and offers the same pop music and mellow style of SABC's noncommercial service. You will recognize many of the brands mentioned in the commercials, since many American firms market their goods in South Africa, while some brands, such as "Jungle Oats" cereal, are clearly unique to South Africa!

In the summer of 1977 SABC opened yet another domestic service, this time featuring rock and roll music in a style similar to that of American "Top Forty" stations, although with fewer suggestive lyrics and songs than heard on American radio. This service is known simply as Radio Five and best reception is on 3388 kHz from 0300 GMT onward.
The Rhodesia Broadcasting Corporation can be heard on 3396 kHz from 0355 GMT sign-on in English. Programming is mainly news, pop music, and advertising, delivered in a sedate style that seems oblivious to the explosive situation facing the nation. But SWL's have heard items warning Rhodesians to prepare for the "war sacrifices" that might lie ahead. If the guerrilla campaign against the lan Smith regime isn't settled in some peaceful manner, Rhodesia might provide some of the most dramatic
listening available to the SWL. This station is a good verifier, by the way, and the wise SWL should secure a QSL now before any future political changes.

Voices of Opposition. None of the black-ruled nations of southern Africa opposed to the regimes in South Africa and Rhodesia have the commanding shortwave signals possessed by South Africa. This makes them better DX, yet none are extraordinarily difficult. Most are audible, at varying levels, throughout North America on average receiving equipment.

Perhaps the most virulent opposition voice belongs to Radio Tanzania, which can be heard in English from 1730 to 1915 GMT on 15435 kHz . A regular program at 1800 GMT is "Liberation for South Africa and Liberia," consisting of stinging, barbed commentary directed against those two nations. Also occasionally heard at 1830 GMT is the "Voice of Namibia" program, calling for the liberation of the territory of South West Africa, which is controlled by South Africa. Radio Tanzania's newscasts are similarly peppery. Radio Tanzania allows the American SWL to directly hear the views and opinions of the more militant nationalist groups. Moreover, the American news media often fail to capture the impassioned nature of the nationalist movements which comes across so clearly over Radio Tanzania.

Zambia is a nation which is also steadfastly opposed to the regimes of South Africa and Rhodesia, yet it maintains a more moderate radio voice. Radio Zambia transmits its general service, consisting of English and local vernacular languages, on 3346 kHz from 0350 to 0530 GMT and on 7250 kHz from 0630 GMT until fade-out due to sunrise in Zambia. Reception of either frequency cannot be termed easy, but Radio Zambia does manage to put a readable signal into the United States most days. Newscasts feature extensive coverage of African nationalist activities but without the stridency of Radio Tanzania. Much African music is played, making it a listening treat even for those uninterested in politics. One distinctive characteristic of Radio Zambia is the practice of referring to areas under white rule by their nationalist or native names. For example, Rhodesia is always called "Zimbabwe" by Radio Zambia. Zambia can also often be heard sending greetings to listeners in white-ruled areas.

Amin Speaks to the World. Few


Radio Senegal's QSL card gives the station's frequencies and power outputs. Note that it is in French.
leaders in world history can compare in notoriety and controversy with Idi "Big Daddy" Amin of Uganda. Listeners who wish to keep up with his latest escapades can do so by tuning in to the external services of the Uganda Broadcasting Corporation. It is on the air in English from 1880-1830 GMT on 15325 kHz (although with interference from Radio Canada International) and from 2030-2100 GMT on 9730 kHz . Transmissions are to Africa, but Uganda's powerful 250-kilowatt transmitters often deliver strong signals to North America.

Uganda's broadcasts are not as exciting as one might expect, given Amin's pyrotechnics. Programming is a varied hodge-podge, seemingly spontaneous,
with musical selections running the gamut from traditional African music to Judy Garland and American country and western. Yet interspersed are various features and commentaries, giving firsthand glimpses into Uganda and its policies. Uganda's newscasts are often the means by which Amin springs his surprises on the world. The first news of Amin's "post-operative coma" came from UBC's external services, for example. Thus, Uganda always has the potential for exciting and dramatic listening as long as Amin remains in power.

Other English Voices. Not all African broadcasters are as heavily politicized in their English programs as the

Verification card from the Nigerian Broadcasting Corporation.

## NIGERIAN BROADCASTING CORPORATION

broadcasting house.lagos.nigeria

stations we have discussed so far. Some offer relatively balanced news coverage and much interesting African music. For those seeking a more moderate perspective on African affairs, the Voice of Nigeria is a listening must. Best North American reception is on 7275 kHz at 0600 GMT sign-on and on 11770 kHz from 1800-1930 GMT. These are external services intended for foreign audiences. Also easily heard is the domestic service, Radio Nigeria, on 4990 kHz . Listeners in the East will find this frequency audible from fade-in around 2130 GMT until sign-off at 2305 GMT. Listeners throughout North America can try for their sign-on at 0430 GMT. Programming consists of authentic African music, news, and interviews. Both the Voice of Nigeria and Radio Nigeria are very friendly verifiers.

Ghana is an African nation which has of late curtailed much of its international broadcasting activity. Once a station which rivaled Radio RSA as Africa's leading radio voice, Radio Ghana is now restricted to one frequency, 6130 kHz , where it is heard until its 2300 GMT sign-off in English. It now uses the slogan, "The Voice of the Revolution," and bears watching as this nation attempts to stabilize its internal situation. SWL's in the West will find Ghana easier to catch via their domestic shortwave serv-
ice, the Ghana Broadcasting Corporation, which signs on at 0600 GMT on 4915 kHz . Plenty of African music and commentaries are featured. If you send a reception report to either station, don't be surprised to receive letters from Ghanaians wanting to be your pen-palnumerous SWL's have received them for several years. Apparently many letters from the United States are opened in Ghana by local postal employees!

A station completely different from the norm in African broadcasting is the Sierra Leone Broadcasting Service, operating from a nation established by the British for much the same purpose as Liberia: as a home for former slaves. Listeners in the East can hear it on 3316 kHz at 2335 GMT sign-off, usually with religious programming, while listeners further west can listen for their 0600 GMT sign on. Programming then is rock and pop music, commercials, and a DJ-all sounding rather like the radio voice of another former British colony, Radio Belize in Central America! Yet even this relatively tame station had many interesting news items during a coup attempt in early 1977.

Those who seek a true challenge should try to hear and verity the English service of Radiodiffusion Television Ivoirienne broadcasting from Ivory Coast. English is scheduled around 2000 GMT

| Thank you for your report on the reception of our Domestic English Service/Afrikaans Service/ |  |
| :---: | :---: |
| H.F. Verwoerd Shortwave Station |  |
| DATE : .......19....actober... 19.7.5.............. |  |
|  |  |
|  |  |
|  | MR, He.e. HELMS JR |
| SOUTH AFRICAN BROADCASTING CORPORATION sUID-AFRIKAANSE UITSAAIKORPORASIE |  |
|  |  |
|  | U.S.A. |
| P.O. Box 4559, Johannesburg, South Africa | for: DIRECTOR: TECHNICAL SERV. |
| Supreme |  |

South African Broadcasting Corporation QSL card verifying reception on 3980 kHz .
on 11920 kHz , and programming features American soul and African pop music. You're likely to hear a variety of station identifications used, but the two most common are Radio Abidjan (after the name of the national capital) and Radio Ivory Coast International. Signals are uften good throughout North America, and the real challenge starts after you've heard the station. Their QSL policy is sporadic and totally unpredictable. Some listeners receive a verification with their first report. Others cannot get a reply despite numerous reports. Your author falls into the latter category, having sent several reports since 1969 with nothing to show for the effort!

Non-English Stations. For those more interested in QSL's than program content, or for linguists, don't ignore non-English broadcasts, as follows.

Since many African nations were formerly French colonies, it is hardly surprising that French is a very common language over the African airwaves. Thus, you'll have to resurrect your high school or college French to follow the news programs and commentaries on these stations. But if you don't know French, you'll find that you can follow the programming well enough to prepare a reception report, and these stations will usually verity a reception report written in English.

Benin is a new name for the former nation of Dahomey, and its La Voix de la Revolution Beninaise is a most unusual experience for North American listeners. Programming is in French, with local music, frequent excited political speeches and slogans, and repeated mentions of Cuban troops in Africa. It operates on 4870 kHz , and is scheduled to sign on at 0415 GMT although at the time of this writing it is sometimes operating all night. English has been reported in the past around 2030 GMT by European SWL's.

Another all-night operation is Radiodiffusion Nationale in Guinea, operating on 7125 and 4910 kHz . Although it identifies as La Voix de la Revolution on occasion, its programming has toned down over the years. French is used along with local vernacular languages, and much African music is featured.

The Congo is well-heard in North America via Radiodiffusion Television Congolaise on 4765 kHz at 0400 GMT sign on. Tribal rhythms and drumming are often featured, and are a delight if you haven't heard them before. Listeners in the East might want to try for their

English programs scheduled for 2130 GMT, although these are presented on a somewhat irregular basis.

Chad is heard well throughout North America on 4905 kHz via the programs of Radiodiffusion Nationale Tchadienne.

Sign-on is at 0430 GMT in French with pop and rock music mixed with more traditional native flute and drum music. If you are proficient in French you will also find the numerous political commentaries to be of interest.

## ENGLISH BROADCASTS FROM AFRICA

Times in GMT

Frequency
Station
3250 South Africa, Springbok Radio, Johannesburg. Approximately 2215-0300 sign-off. Pop music and commercials.
3316 Sierra Leone, Freetown. 0600 sign-on with pops and rock music, advertisements.
3346 Zambia, Lusaka. General service. Sign on at 0350-0530 in English and local languages African music and news.
3388 South Africa, Radio Five, Johannesburg. Audible 0300 until sunrise fade-out. Rock music.
3396 Rhodesia, Salisbury. Sign on 0355 in English. Pop music and ads. Low-key generally but increasingly reminding listeners to prepare for possible wars with neighboring states.
4875 South Africa, SABC, Johannesbuirg. Sign-on 0300 with pop music. Identifies as Radio South Africa.
4915 Ghana, Accra Domestic service. Sign-on 0600 . News, commentary, African music.
4990 Nigeria, Lagos. Fade approximately 2130 until 2305 sign-off, also at 0430 sign-on. Authentic African music, news and features.
6130 Ghana, Accra. International service, until 2300 sign-off.
7250 Zambia, Lusaka. General service. Sign-on 0630. Similar programming as 3346 kHz outlet.
7275 Nigeria, Lagos. Voice of Nigeria international service. News, commentaries, and music from 0600 sign-on.
9585 South Africa, Radio RSA, Johannesburg. International service beamed to North America 2230-2320.
9730 Uganda, Kampala. External service. Varied programming and unpredictable news 2030-2100.
11770 Nigeria, Lagos. Voice of Nigeria, similar to $7275 \mathrm{kHz}, 1800-1930$.
11800 South Africa, Radio RSA, Johannesburg. Same schedule as 9585.

11900 South Africa, Radio RSA, Jo-
hannesburg. Same schedule as 9585.

11920 Ivory Coast, Abidjan. African and U.S. soul music around 2000.

15325 Uganda, Kampala. External service, 1800-1830.
15435 Tanzania, Dar es Salaam. AntiRhodesian and South African programming 1730 to 1915 signoff. Often stinging in its criticism of white minority regimes.
3210 Mozambique, Maputo. Portugese talks and music from 0315.

4765 Congo, Peoples Republic of, Brazzaville. Sign-on 0400 French with much native music. Also before 0000 sign aff along East Coast.
4820 Angola, Luanda. All night. Best after Latin American stations clear the frequency after 0500. Portugese talk and local and pop music.
4870 Benin, Cotonou. French language revolutionary speeches and slogans. Many mentions of Cuba and local music. 0415 sign on, occasionally all night.
4890 Senegal, Dakar. 0600 sign-on French and native languages with much exotic local music.
4895 Mozambique, Maputo 0400 sign-on. Portugese with talks and instrumental music, some English identifications.
4905 Chad, N'djamena. 0430 sign-on. French with African music. Also audible in East at 2200 sign-off.
5038 Central African Empire, Bangui. French talks with native and pop music until 2300 sign-off, also at 0430 sign-on.
5047 Togo, Lome. French talk and music like Bangui-5038-be careful not to confuse the two! In East until 2300 sign-off. Also at 0530 sign-on.
7125 Guinea, Conakry. French and vernacular languages with African music all night.
9535 Angola, Luanda. Sign-on 0500. Portugese with chanting, African and pop music, and talks.

Senegal's Radiodiffusion du Senegal puts good signals into North America on 4890 kHz at its 0600 GMT signon. They sign on with a melody played on a native instrument known as the "khora" and feature its distinctive sound on many of their musical selections. Much exotic chanting can be heard as well, along with some native languages. French announcements are likely to be the only things familiar to most listeners.

Two stations often confused by new SWLs are Radiodiffusion du Togo on 5047 kHz and La Voix de L'Empire Centrafricane (from the former Central African Republic, now called Empire) on 5038 kHz . Both stations program in French with African and pop music. Both sign off at 2300 GMT and put powerful signals into the East Coast at that time. Togo signs on at 0530 GMT , however, while the Central African Empire starts operations at 0430 GMT.

Portugal's Former Colonies. The shortwave stations of Angola and Mozambique were exciting listening a few years ago as they gained independence from Portugal and were racked by internal conflicts. SWL's could tune in live political rallies with excited speakers and crowd shouts. Programming slants varied as different factions gained control of the broadcasting stations.

The situation is more stable now. Angola's Radio Nacional can be heard at 0500 GMT sign-on in Portugese with African and pop music, certainly a change from their former all-night political harangues.

Mozambique is also heard in Portugese with much guitar music on 4895 kHz at 0400 GMT sign-on. Fortunately for SWL's, Radio Mozambique also has some English station identifications. DXers seeking a challenge can shoot for Radio Mozambique's 3210 kHz outlet, which is sometimes heard past 0315 GMT with similar programming. Both Angola and Mozambique are quiet now, but alert SWL's should keep abreast of African events which could offer excellent listening possibilities.

The above advice could be applied equally well to any station in Africa. Not only does shortwave allow you to hear news direct from the source, but it also sometimes permits you to hear news in progress. There have been cases where SWL's have heard live gunfire in the background of African shortwave broadcasts! And given the highly explosive nature of certain sections of the continent, that sound might be heard again.

## Listen to a NEW WORLD OF SOUNDS WITH ULTRASONIC DETECTOR

# Inexpensive detector converts ultrasonic sounds from insects, compressed gas leaks, etc., to an audio output. 

EXPLORING the world of ultrasonic sound-which lies above approximately 20 kHz -can be exciting and educational. Here is a frequency spectrum beyond human hearing where many insects and rodents communicate with each other, where sounds from leaks in pressurized gas lines occur, etc.

The inexpensive circuits presented here convert these ultrasonic sounds to audio frequencies, enabling anyone to hear them. Also included is a simple ultrasonic transmitter circuit that will enhance your ability to probe this interesting electronics area.

An Ultrasonic Receiver. The schematic diagram of a heterodyne-type ultrasonic receiver is shown in Fig. 1. This receiver hyterodynes ultrasonic signals with those from an internal oscillator, converting them to audible frequencies for reproduction by a dynamic speaker. Thus, it allows you to "hear" any signals it detects.

Piezoelectric transducer TR1 converts ultrasonic waves impinging upon it into ac waveforms which are applied to the noninverting input of operational amplifier IC1A. Because a single-ended power supply is used, resistors R1 and R2 bias the noninverting input to onehalf the supply voltage. Resistor R3, effectively connected across TR1 by electrolytic capacitor C1, damps the transducer's resonant response and broadens its bandwidth. At dc, R5 provides 100\% negative feedback to stabilize the operating point. At signal frequencies of interest, the gain of IC1A is 60 dB for the values given in Fig. 1.

The output of IC1A is directly coupled to op amp IC1B, a similar amplifier stage. The voltage gain of IC1B, about 43.5 dB with the component values specified, is somewhat lower than that of the preceding stage. Signals at the output of IC1B are capacitively coupled by C5 to diodes D1 and D2.

Also applied to the diodes is the output of an ultrasonic oscillator comprising IC3 and its related components. The frequency of this oscillator is determined by the setting of potentiometer R12 and the capacitance of C 9 , which is chosen so that the oscillator output corresponds to the resonant frequency of the transducer. (Transducers are readily available from surplus dealers with resonant frequencies ranging from 22 to 44 kHz .)

The two diodes form a nonlinear net-


Fig. 1. An ultrasonic receiver, where incoming signals heterodyne with those from the local oscillator to produce an audible output.

## PARTS LIST FOR FIG. 1

Cl.C8-10- $\mu \mathrm{F}, 25-\mathrm{V}$ tantalum C2.C4.C6.C7-0.1- $\mu \mathrm{F}$ disc ceramic C. $3-0.01-\mu \mathrm{F}$ dise ceramic C5-0.005- $\mu \mathrm{F}$ dise ceramic
C9-180-pF (or $330-\mathrm{pF}$ ) dise ceramic, polystyrene glass or silver-mica (see text)
C10-0.003- $\mu$ F dise ceramic
CII- $0.05-\mu \mathrm{F}$ dise ceramic
CI2-50- $\mu \mathrm{F}$, 25-V electrolytic
D1.D2-IN914 signal diode
|C 1-TBA231 dual op amp (see note)

IC2-LM386 audio amplifier
IC3-CD4()01 quad 2-input NOR gate
The following fixed resistors are $1 / 4-W$. 10\% carbon composition:
R1.R2.R5- 150.000 ohms
R.3- 1000 ohms

R4.R6- 150 ohms
R7.RII-22.000 ohms
R8- $\mathbf{3 3 0 . 0 0 0}$ ohms
RIO- -470.000 ohms
R13- 10 ohms

R9- 10.000 -ohm linear-taper potentiometer R12-20.000-ohm linear-taper potentiometer SPKR-8-ohm dynamic speaker
TRI-Piezoclectric ultrasonic transducer
Misc.-Printed circuit or perforated board: suitable enclosure: hook-up wire: dc power scurce: machine hardware: etc.
Note-The TBA231 dual op amp is imported from the U.K. by SG-ATES Semiconductor
Corp., 435 Newtonville, MA 02160 (Tel:
617-969-1610).
work. Hence, when signals from the oscillator and the op amp are applied, they heterodyne with each other. If IC3 oscillates at a frequency fairly close to that of an ultrasonic wave detected by TR1, an audible beat signal will appear at the cathode of $D 2$ at a frequency equal to the difference between the two ultrasonic frequencies. The process is similar to that performed in a conventional superheterodyne $r$ - $f$ receiver. The beat note, which can be tuned by adjusting R12, is amplified by IC2, an audio IC, to a level sufficient to drive the dynamic speaker. Potentiometer R9 serves as an audio gain control.

An Ultrasonic Transmitter will help you explore the ultrasonic region more fully. A suitable design is shown schematically in Fig. 2. The circuit is
similar to the local oscillator stage in the receiver, but the previously unused fourth gate in the 4001 is employed to provide push-pull drive for transducer TR2. The output frequency is variable by means of R3. The capacitance of C1 should be chosen so that the nominal oscillating frequency corresponds to the resonance of the transducer. As was the case with $C 9$ in the receiver, $C 1$ should be 180 pF if $44-\mathrm{kHz}$ transducers are used, or it should be 330 pF for use with $22-\mathrm{kHz}$ transducers.

Construction. Either printed circuit or perforated board can be used to duplicate the transmitter and receiver circuits. Parts placement is not especially critical. The use of sockets or Molex Soldercons is recommended when mounting the IC's on the boards. Be sure to
observe normal precautions when handling the CMOS devices. Install polor ized capacitors and semiconductrers with due regard for polarity and pin hasing. Batteries are well suited to power the transmitter and receiver circuits. Note that, when transmitter switch S1 is in the OFF position. the output states of IC1's gates are frozen. The quiescent current drain of the circuit is so small that no power switch is necessary. If a battery supply is used with the receiver, however, an spst power switch should be used to disconnect the circuit from the supply when it is not being operated.

Use. Receiver potentiometer $R 12$ tunes the circuit across a limited portion of the ultrasonic frequency range. Apply power and adjust audio gain control R9 until some noise is heard through the speak.-


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er．Then rub the palms of your hands in front of TR1．The receiver will detect the ultrasonic energy from the rubbing．

You will notice that TR1 has a very di－ rectional response．This is due to the fact that ultrasonics have very short wavelengths（compared to those at au－ dio frequencies）and are thus subject to less diffraction at the edges of large ob－ jects．Also，ultrasonic waves behave like light waves in that they tend to travel in straight lines．

It＇s interesting to note that if coupling capacitor C10 in the receiver is discon－ nected from the diode mixer，the receiv－ er will still detect ultrasonic signals if more than one frequency is present．The frequencies present at the input will beat against each other to produce an audi－ ble output．This can be verified by re－ peating the palm－rubbing experiment described earlier after the coupling ca－ pacitor has been disconnected．The speaker will still generate an audio out－ put even though no local oscillator signal is being injected into the diode mixer．
If an ultrasonic wave generated by transmitter transducer TR2 now im－ pinges upon TR1，the random noise re－ produced by the speaker will drop to a low level．No tone will be heard because only one frequency is applied to the mix－ er．Stray coupling that allows a portion of the local oscillator output to reach the mixer will create an audible beat．

When the receiver and transmitter are operating in the same room，a signal will be heard as R12 tunes the receiver
across its range．The two transducers do not have to be directly facing each other if enough hard surfaces in the room reflect the ultrasonic waves，and the room is not so large that it introduces excessive signal attenuation．

The circuits presented have been suc－ cessfully used with ultrasonic transduc－ ers from many different sources，includ－ ing those used in television receiver re－ mote control accessories．Of course，if you want to tune in several ultrasonic ＂bands，＂you can use a multiple－pole ro－ tary switch to select the appropriate transducer and its corresponding oscil－ lator capacitance．Experimentation indi－ cates that the receiver can＂hear＂the transmitter at distances up to 125 feet if the transducers are aimed at each other． The use of a suitable parabolic reflector in tandem with TR1 and／or multiple driv－ en transmitter transducers should result in even greater useful range．

Other Suggestions．We have al－ ready mentioned the possibility of using these circuits for signalling purposes． Many other practical applications exist． For example，leaks in the rubber sealing of car doors and windows or in the seal－ ing of a freezer door．The transmitter is placed in the car or freezer and fills the interior with ultrasonic waves．The walls of the interior reflect the waves to create a wide dispersion of ultrasonic energy．If the receiver＇s transducer is moved over the exterior，a tone will be heard when－ ever it passes any leaks．


Fig．2．This ultrasonic transmitter employs four NOR gates．

## PARTS LIST FORFIG． 2

CI－180）－pF（or $330-\mathrm{pF}$ ）disc ceramic，poly－ styrene，glass or silver mica capacitor ICI－CD4001 quad dual－input NOR gate R1－470．000）－ohm $10 \%$ ． $1 / 4$－W resistor R2－22．000－ohm $10 \%$ ． $1 / 4-\mathrm{W}$ resistor R．3－20．000－ohm linear－taper potentiometer

## SI－Spdt switch

TR2－Piezoclectric ultrasonic transducer Misc．－Printed circuit or perforated board； suitable enclosure；hook－up wire；dc power source：machine hardware，etc．


PART 5: THE CONTROL SECTION OF PIP-2.

THUS FAR in this series, we have covered the basics of number systems, digital logic and microprocessor organization. We have also introduced PIP-2, a simple 4-bit educational microprocessor, and learned how it's organized and programmed.
Now let's take a detailed look at the control section of PIP-2. We will see how instructions are fetched from the program memory, decoded and executed. We will also learn how to revise PIP-2's instruction set by modifying the microinstructions stored in control's ROM.

PIP-2's Control Section. The most important and complex section of a microprocessor is its control circuitry. This is the element that fetches instructions from the microprocessor's memory in the proper sequence, then decodes and executes the instructions.

The overall operation of the control section is a perfectly synchronized sequence of individual operations that fetch instructions, transfer data, advance counters and perform arithmetic operations.

The control section responds to a load instruction, for example, by simultaneously connecting the memory address containing the data word to be loaded (the source) and the input of the appropriate register (the destination) to the microprocessor's bidirectional bus. The control then sends a clock pulse to the register to complete the load operation and proceeds to fetch the next instruction.

While all this might seem extremely complicated to the unititiated, it's really quite simple since the program instruc-
tion is merely a binary bit pattern that can be interpreted by the control section to perform a specific task. In simplest terms, the control section is no more complicated (at least in principle) than the decoder circuit that lights up the proper segments of a seven-segment display in response to a binary-coded decimal (BCD) input nibble.
The heart of the control section of some microprocessors is a complex combinational network of gates that decodes program instructions and activates the appropriate control inputs of the various sections of the processor. More advanced microprocessors employ a special ROM that contains the sequences of microinstructions necessary to accomplish each program instruction. These so-called microprogrammable microprocessors are much more versatile since their instruction sets can be extensively revised by simply modifying the microinstructions stored in the ROM.

PIP-2, the educational microprocessor we've been studying, is microprogrammable and the block diagram shown in Fig. 1 illustrates the general organization of PIP-2's control section. You might want to refer back to Part 4 of this series to see how control interfaces with the remainder of PIP-2.
A detailed breakdown of PIP-2's control, including the organization of the microprogrammable ROM containing the microinstructions, the microinstruction decoders and the clock, is shown in Fig. 2. We will now discuss each part of the control section.

Clock. The clock is a relatively simple but vital part of CONTROL since it provides the synchronized train of pulses that cycle PIP-2 through a program. The clock's output is said to be two-phase since it supplies two streams of pulses having identical frequency but different phases from outputs $\phi 1$ and $\phi 2$. Fig-


Fig. 1. Organization of the PIP-2 control section.


Fig. 2. Internal details of the PIP-2 control section showing how the three decoders interface with the microprogram with in the ROM. Here, the ROM is shown as a matrix with 0's being unconnected junctions and 1 's having a connection.
ure 3 shows the timing diagram for these two clock signals.

The clock has two control inputs. A low at C/S applied by pressing the START switch starts the clock. A low at C/D applied by pressing stop, or by a signal from the microinstruction decoder (activated by a HLT instruction in the program), disables the clock.

Instruction Register and Microprogram Counter. This is a 4-bit counter that doubles as a 4-bit register. It receives the op-codes from the program memory, which are actually control ROM addresses, and feeds them into the control ROM address decoder.

Signals from clock phase $\phi 1$ increment the instruction register and cause it to step through a sequence of addresses in the control ROM, much like PIP-2's program counter steps through addresses in the program memory when executing a program. That's why the instruction register can also be called a microprogram counter.

The instruction register has a couple of other control inputs. When IR/W is low, a $\phi 2$ pulse from the clock writes the instruction on PIP-2's address/data bus into the instruction register. When IR/C is low, the instruction register is cleared to 0000.

## Control ROM Address Decoder.

This is simply a 1 -of-16 decoder that activates appropriate addresses in the control ROM in response to the data in the instruction register. When the nibble 0000, for example, is in the instruction register, the first address in the control ROM is selected.

Control ROM. This is a 128 -bit ROM organized as sixteen 8 -bit bytes. Each byte is assigned a unique address (0000 to 1111) and comprises a single microinstruction. As shown in Fig. 2, the control ROM is loaded with microroutines (sequences of microinstructions) for six separate program instructions. As we'll soon see, these microroutines can be
easily changed by simply reprogramming the ROM.

Microinstruction Decoders. Control has a pair of 1-of-8 decoders (Source and Destination), and a single 1-of-4 decoder (Operation). The selected output of each decoder goes low while the remaining outputs stay high.

These decoders convert the microinstructions encoded in the selected ROM address into the appropriate operations necessary to execute the microinstruction. As you can see in Fig. 2, the control ROM is divided into sixteen 8 -bit bytes. The first two bits of each byte are fed into the operation decoder. The next three bits go to the source decoder and the final three bits go to the destination decoder.

The outputs from the three decoders and from the clock form PIP-2's control bus. The outputs of the source decoder go to the read ( $R$ ) control inputs of the various sections of PIP-2. The outputs of the destination decoder go to the write
(W) control inputs of the various sections. And the outputs of the operation decoder go to the special operation control inputs, clock disable (C/D) and program counter increment (PC/I).

Note that several outputs of the source and destination decoders and two outputs from the operation decoder are not used. This means that additional circuits (maybe a C register, perhaps an arithmetic-logic unit) can be connected to PIP-2's address data bus. These lines may also be used to control external devices. In both cases, of course, new microinstructions would have to be added to the control ROM to activate the new circuits.

Note also how the bit pattern stored in the ROM activates the decoders. Address 0001, for example, contains the microinstruction 00001001. Let's divide this byte into each of the three bit fields applied to the decoders and see what happens:


The operation field (00) does nothing since it activates the unconnected 0 output of the operation decoder.

The source field (001) activates the 1 output of the source decoder. This applies a low to RAM/R.

The destination field (001) activates the 1 output of the destination decoder. This applies a low to IR/W.

The result? The output of the program memory (RAM) and the input of the instruction register (IR) are simultaneously connected to the address/data bus, and the arrival of the next $\phi 2$ pulse from the clock loads the instruction register with the selected instruction op-code in the program memory.

Now that we know something about each of the sections of PIP-2's control and how an individual microinstruction is executed, let's see how control fetches and executes an instruction from the program memory.

Fetching and Executing. Understanding how control fetches (retrieves) an instruction from the program memory and then executes it will take you a long way toward understanding how real microprocessors work. You might find it handy to have Part 4 of the Microcourse available since we'll be referring to PIP-2's instruction set mnemonics and op-codes.

Let's assume the first instruction in the program memory (address 0000) is LDA. This is a memory reference instruction that is followed by a 4-bit data nibble in the next program memory address. When executed, LDA will load the A register with the data nibble in program memory address 0001.

After the program containing the LDA instruction is loaded into the program memory, the initiate switch is pressed to return the program counter to program memory address 0000. The instruction register doubles as a microprogram counter and pressing inITIATE clears it to 0000 also.

The two microinstructions that comprise NOP occupy the first two bytes of the control ROM. When START is pressed, the first clock pulse advances the combination instruction register/ microprogram counter to the second NOP microinstruction (control ROM address 0001).

What's the byte stored in this address? Figure 2 shows that this microinstruction is 00001001 -which activates the RAM/R and IR/W control inputs discussed earlier. When the $\phi 2$ clock pulse arrives, the instruction register
advanced to the next address in the program memory (which contains the data nibble to be loaded into the A register). Signal $\phi 2$ is a do-nothing clock pulse since there is no data located on the address/databus.

The third $\phi 1$ clock pulse advances the instruction register to the second microinstruction in the LDA microroutine (control ROM address 0011). This microinstruction (00001010) applies lows to RAM/R and A/W. When clock pulse $\phi 2$ arrives, the $A$ register copies the contents of the data nibble following the LDA op-code in the program memory.

Now that the A register has been loaded with the specified data nibble, the most important part of the LDA instruction has been accomplished. The remaining two microinstructions fetch the next step from the program memory.

The fourth $\phi 1$ clock pulse advances the instruction register to LDA microinstruction 01000000. This increments the program counter to the next address in the program memory (0011). The next $\phi 2$ clock pulse is another do-nothing pulse. The fifth $\phi 1$ clock pulse advances the instruction register to the final LDA microinstruction, 00001001.


Fig. 3. Timing diagram of the PIP-2 two-phase clock.
copies the op-code of the instruction in program memory address 0000 . The opcode for LDA is 0001, so in this case the instruction register doesn't change states. (What would happen if the opcode was 1011 or 0101?)

Thus far, all of control's operations have been preprogrammed and completely automatic with the specific goal of fetching the first instruction from the program memory. What happens next?

Recall that the op-code for each instruction is a binary number that is 0001 less than the starting address of the microroutine in the control ROM that executes the instruction. When the next $\phi 1$ clock pulse arrives, the instruction register advances to the first microinstruction in the LDA microroutine and things start to happen. Let's follow the various steps in the execution of the LDA microroutine to see how.

The first LDA microinstruction (Fig. 2) is 01000000 . Only the PC/I control input is activated; the program counter is

This loads the op-code of the next instruction in the program memory into the ins:ruction register.

All the steps necessary to execute LDA appear rather complicated at first. But if you'll browse back through the preceeding paragraphs again you'll see that LDA, like all of PIP-2's instructions, is merely a collection of very simple operations neatly strung together by the $\phi 1$ and $\phi 2$ pulses from the clock. Figure 4 is a sequence of diagrams that shows exactly what happens.

Summing Up Control. Now that you've seen how PIP-2 fetches, decodes and executes an instruction, you can better appreciate the sophistication of control. You can even think of control as a simple microprocessor inside PIP-2. The control ROM contains the program, the instruction register serves as the program counter and the microinstruction decoders implement the various instructions.


The Table summarizes the microroutines necessary to execute each PIP-2 instruction. In addition to the mnemonics and their op-codes, the table contains the entire truth table of the control ROM. it also shows the operations that take place for each microinstruction.

Microprogramming PIP.2. Look back at the table of microroutines for a moment. Notice how often the fetch operations PC/I and RAM/R $\rightarrow$ IR/W occur? Remove these microinstructions from the table and we're left with only five additional microinstructions.

Obviously there are more possible microinstructions than just these seven. All that's necessary to arrive at a new microinstruction is to place one source and one or more destinations on the address data bus. Here are some possibilities:

$$
\begin{aligned}
& \begin{array}{l}
A / R \longrightarrow I R / W \\
A / R \longrightarrow P C / W \\
B / R \longrightarrow I R / W \\
B / R \longrightarrow P C / W \\
B / R \longrightarrow A / W
\end{array} \\
& R A M / R \longrightarrow B / W \\
& A D D / R \longrightarrow B / W \\
& A D D / R \longrightarrow P C / W \\
& A D D / R \longrightarrow I R / W
\end{aligned}
$$

Of course these are only some of the additional microinstructions that are possible. All sorts of possibilities open up if we activate more than one destination device. For example, RAM/R $\longrightarrow$ A/W;B/W;PC/W.
If we assume that you have assembled a working version of PIP-2, it's quite possible the original instruction set will not fill your requirements. If that's the case, you can substitute new microinstructions to devise your own special instruction set.

Suppose you want to replace LDA with LDB (load the $B$ register). All you have to do is find the LDA microroutine in the control ROM and reprogram the byte that loads the A register (address 0011) so that the B register is loaded instead. The original byte is 00001010 . The new byte is 00001011 . The remaining bytes are unchanged. The op-code for LDA becomes the op-code for LDB since we haven't changed the location of the microroutine in the control ROM.

You can use this same procedure to microprogram other new instructions into PIP-2. Just remember these points:

1. Be sure to assign the correct opcode to each new instruction. Remem-
ber, the op-code is a binary number that is 0001 less than the first address of the microroutine in the control ROM.
2. If necessary, be sure to include the appropriate fetch microinstructions in each new microroutine so the next instruction in the program memory will be retrieved.
3. Be sure the microinstruction at the 0001 address in the control ROM is always 00001001 . This is necessary since this microinstruction plays a key role in fetching the first instruction from the program memory during PIP-2's automatic start sequence.
4. Plan ahead! Are you eliminating an existing instruction(s) you might need later? Does the control ROM have room for the new instruction(s)? Are there any possible programming shortcuts you can use to implement instructions not in the control ROM?
5. Document your work so you'll know what you've done.

Don't let these simple precautions stop you from having a go at microprogramming PIP-2! Some of the possibilities are very interesting.

For example, an instruction that loads the program counter with the contents of

PIP-2'S MICROROUTINES

| Program MemoryMnemonic OP-Code | Control ROM |  |  |  | Operation |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Address | Microroutine |  |  |  |
|  |  | OP | 5 | D |  |
| NOP 1111 | 0000 | 01 | 000 | 000 | PC/I |
|  | 0001 | 00 | 001 | 001 | RAM/R $\rightarrow$ IR/W |
| LDA 0001 | 0010 | 01 | 000 | 000 | PC/I |
|  | 0011 | 00 | 001 | 010 | $\mathrm{RAM} / \mathrm{R} \rightarrow \mathrm{A} / \mathrm{W}$ |
|  | 0100 | 01 | 000 | 000 | PC/I |
|  | 0101 | 00 | 001 | 001 | RAM $/$ R $\rightarrow$ IR/W |
| ADD 0101 | 0110 | 00 | 100 | 010 | $A D D / R \rightarrow A / W$ |
|  | 0111 | 01 | 000 | 000 | PC/I |
|  | 1000 | 00 | 001 | 001 | $R A M / R \rightarrow I R / W$ |
| JMP 1000 | 1001 | 01 | 000 | 000 | PC/I |
|  | 1010 | 00 | 001 | 100 | RAM $/ R \rightarrow$ PC/W |
|  | 1011 | 00 | 001 | 001 | $R A M / R \rightarrow I R / W$ |
| MOV 1011 | 1100 | 00 | 010 | 011 | $A / R \rightarrow B / W$ |
|  | 1101 | 01 | 000 | 000 | PC/I |
|  | 1110 | 00 | 001 | 001 | RAM/R $\rightarrow$ IR/W |
| HLT 1110 | 1111 | 11 | 000 | 000 | C/D |

the A register permits the program to branch to an address specified by the result of an addition. This procedure is called indirect addressing. It gives a microprocessor the ability to branch to one of several possible addresses in its program memory depending upon the results of an earlier operation.

Here's one possible microroutine that
performs the indirect addressing operation we've been discussing:

| Microinstruction | Operation |  |  |
| :--- | :---: | :---: | :---: |
| $\mathbf{O P}$ | $\mathbf{S}$ | $\mathbf{D}^{*}$ |  |
| 00 | 100 | 100 | $A D D / R \rightarrow P C / W$ |
| 00 | 001 | 001 | $R A M / R \rightarrow I R / W$ |

*OP $=$ operation; $S=$ source; $D=$ destination decoders.

Since the second microinstruction of this microroutine is the same as the second microinstruction of NOP in PIP-2's original instruction set, we can easily substitute it for NOP. We just reprogram the first address in the control ROM with 00100100 and assign NOP's op-code to the new instruction.

For convenience, it's nice to assign a mnemonic to the new instruction. Since the instruction is an indirect jump, one possibility is JMI. You might want to be more specific since other indirect jump instructions are possible. Since this is a "JUMP INDIRECT TO ADDRESS IN A," a betfer mnemonic might be JIA.

Now that you know how PIP-2 is microprogrammed, how about adding a new instruction or two on your own? With a little care you just might come up with an instruction set that's better.

Summing Up. If you've stayed with the course, you should have a fairly respectable knowledge of some of the fundamental basics of microprocessors. To be sure, real microprocessors are far more sophisticated than PIP-2. But PIP-2 has prepared you to move up to real microprocessors.


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Solid State

By Lou Garner

## CROSSING THE BRIDGE

MODERN solid-state equipment designs are often an interesting blend of the very old with the very new. To achieve their goals, engineers frequently employ state-of-theart semiconductor devices in networks which have been in use for many decades. Back in 1847, well over a century ago, Sir Charles Wheatstone adapted a basic circuit developed by S. H. Christie (in 1833) for use in making precision resistance measurements. His adaptation of the Christie design has been known ever since as the Wheatstone Bridge, and is still used today in its original or modified form in equipment and instrument designs.


Fig. 1. Developing the bridge circuit: (A) Basic series-parallel network; (B) Meter between common point; (C) Wheatstone bridge; (D) Bridged -T network.

Let's examine the evolution of the bridge circuit. Consider the simple network shown in Fig. 1A: just two parallel branches, each consisting of two resistors in series. Resistors R1 and R2 make up one branch, R3 and R4 the parallel branch. Neither a spectacular nor unusual arrangement, but consider the potentials at points " $X$ " and " $Y$ " if a voltage is applied across the two branches. If R1 and R2 have the same resistance ratio as R3 and R4, the relative potentials at " $X$ " and " $Y$ " will be the same, regardless of the actual resistance values. As an example, if $R 1$ is a 1000 -ohm unit, $R 2$ is 9000 ohms, then the voltage at point " $X$ " will equal that at point " $Y$ " if R3 is, say, 10,000 ohms and $R 4$ is 90,000 ohms, for both branches have the same ratio of 1:9. In practice, this can be demonstrated quite easily by connecting a high-impedance voltmeter or sensitive galvanometer between the two points, as illustrated in Fig. 1B.

This simple fact provided the scientist and engineer with an
extremely powerful tool for measuring the value of unknown resistances, for if an accurately known standard resistance were used for R3 and means were provided for determining the ratio of R1: R2, then R4's value could be calculated easily and accurately once the voltage at " $X$ " equalled (or balanced) that at " $Y$," and the balance could be established by adjusting the R1:R2 ratio.

So far, so good... but what, you may wonder, does all of this have to do with bridges? Simple-the meter ( $M$ ) has "bridged" corresponding points in two parts of the network, and if the basic circuit is redrawn as in Fig. 1C, you'll recognize the standard form of the Wheatstone bridge. The basic design may be used for other than resistance measurements if an ac rather than a dc voltage source is used, and if reactive elements (inductances and capacitances) are used in the arms of the bridge, with a sensitive ac detector substituted for the meter. If a series RC network is used in place of R3 and a parallel RC network for R4, the arrangement becomes a Wien bridge that will "null" (zero voltage between points " X " and " $Y$ ") at a specific frequency dependent upon the relative RC values. The Wien bridge can be used for measuring capacitances and in other applications.

Going a step further, a "bridge" circuit need not always be a measurement device. It can be another network element, as in the bridged- $T$ arrangement shown in Fig. 1D. Here, R4 is the bridging element. Used in attenuators and for impedance matching, bridged-T networks also can serve as special-purpose low- and high-frequency filters if reactive components (inductors and/or capacitors) are used in the arms.


Fig. 2. Rectifier circuits: (A) Full-wave; (B) Modified full-wave; (C) Full-wave bridge.

In a broad sense, the conventional full-wave rectifier is a type of "bridge" circuit, for the dc load serves as a bridge between the transformer center tap (CT) and the corresponding junction between two rectifiers (D1 and D2), as illustrated in


Fig. 4. Using a bridge rectifier as an automatic polarity switch.


Fig. 2A. If the transformer secondary has no center tap, an artificial center may be set up by using two resistors of equal value (R1 and R2) as in Fig. 2B. Unfortunately, while this arrangement will provide full-wave rectification, it is rather inefficient. Resistors R1 and R2 place a constant drain on the transformer, whether or not current is required by the load and, in addition, extra power is dissipated in the resistors as load current increases. These disadvantages may be offset by replacing R1 and R2 with additional rectifier diodes D3 and D4, as shown in Fig 2C, and, of course, we now have the widely used bridge rectifier. Note the configuration's overall similarity to the original Wheatstone bridge.
Although the bridge rectifier's primary application is as an ac-to-dc converter, it can be used in other ways. The amount of ac drawn from the source, whether a transformer secondary or the power line, is directly proportional to the dc load; the greater the load current, the greater the ac, and vice-versa. This characteristic can be utilized to control the current through an ac load, such as a motor, solenoid, or incandescent lamp. Simply connect the ac load in series with a suitable bridge rectifier and provide a variable dc load, such as a power transistor or SCR, as illustrated in Fig. 3A and 3B, respectively. Both are useful techniques to remember if you don't have a Triac handy.

Another interesting application for the bridge rectifier is given in Fig. 4. Here, the bridge rectifier is connected between a dc source and a dc load (such as a CB transceiver, battery operated amplifier, or similar unit), insuring that the correct polarity will be applied to the equipment regardless of the original source connections. This is an inexpensive precaution for equipment which may be installed incorrectly.

A useful and versatile arrangement, the bridge rectifier is by no means the only application for the bridge concept in modern solid-state circuit designs. Another useful application is shown in Fig. 5. Here, the inputs of a conventional op amp are connected across the two branches of an RC Wien bridge,

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Fig. 5. In this circuit, the inputs of an op amp are connected across the branches of an RC Wein bridge with the output furnishing ac drive.
with the output furnishing ac drive to the network. The result is a simple and reliable audio oscillator. The circuit's frequency of operation is established by the component values in the series and parallel RC arms according to the equation $f=1 /(2 \pi$ RC). For operation at, say, 1 kHz , the two " $R$ " resistors might have values of 16,000 ohms each, while the two " C " capacitors could be $0.01-\mu \mathrm{F}$ units. Other RC combinations could be used for the same frequency of operation, of course, as long as the basic equation is satisfied, but, as a general rule, it is best to stick with standard off-the-shelf values. The Wien bridge oscillator can be duplicated quite easily and inexpensively in the home laboratory. It can be assembled as a standalone circuit for experimental tests or, if preferred, incorporated as a circuit element in more complex designs, such as musical instruments or test equipment. The two diodes are gen-eral-purpose types while the specified op amp is one section 74
of a type LM148. A quad device, the LM148 is essentially four standard type 741 op amps in a single 14-pin DIP and, if desired, a single type 741 unit can be used in the circuit without changing component values. Except for the 5,000 -ohm adjustment potentiometer, the resistors may be either $1 / 4$-or $1 / 2$ watt types, while the tuning capacitors should be high quality, low-voltage ceramic or plastic film types. The dc source voltage ( $V_{C C}$ ) may range from 4.5 to as high as 16.0 volts, but the input reference voltage ( $V_{\text {ref }}$ ) should be set at one-half the source voltage for optimum performance.
Another type of bridge circuit suitable for experimenter and hobbyist solid-state projects is illustrated in Fig. 6. Here, a


Fig. 6. Power arnplifier bridge configuration.

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loudspeaker load (RL) is "bridged" between the outputs of two integrated circuit power amplifiers to achieve a power output nearly twice that available from a simple amplifier. The input signal is coupled simultaneously to the noninverting ( + ) input terminal (pin 1) of one amplifier through a $10-\mu \mathrm{F}$ dc blocking capacitor and to the inverting ( - ) input terminal (pin 2) of the second amplifier through a 1 -ohm series isolation resistor and a $220-\mu \mathrm{F}$ dc blocking capacitor. Used alone in a single-ended arrangement, each amplifier is capable of delivering approximately 8 watts to a low-impedance load when operated on a 14.4 -volt dc power source. In the bridge configuration, the pair of amplifiers can deliver 15 watts to a comparable load when operated on a similar dc source.

With neither layout nor lead dress overly critical, the bridge audio amplifier circuit can be assembled using chassis construction or either perf or pc board assembly techniques, as preferred. Good audio wiring practice should be followed, of course, with the signal-carrying leads kept short and direct and ample spacing between the input and output leads. Except for the 100,000 -ohm balance control, all resistors are half-watt types. Capacitors marked with polarity symbols are electrolytics, others are low-voltage ceramic or plastic film types. The semiconductor amplifiers should be attached to an adequate common ground heat sink by their metal tabs.

Assembled in small plastic packages with extended metal mounting tab/heat sinks, the (Motorola) TDA2002 devices are 8 -watt monolithic silicon class-B power amplifiers designed primarily for automotive and general purpose applications. Each device features internal thermal overload and supply over-voltage protection as well as short-circuit current limiting.

The TDA2002 may be used on dc supply voltages from 8 to 18 volts and deliver intermittent peak currents of up to 4.5A. As a general rule, the higher the supply voltage in a given circuit, the larger the maximum output power up to the device maximum ratings.

Reader's Circuit. Guy Isabel (1725, Henri-Bourassa East Blvd., Apt. 25 Montreal, Quebec, H2C 1J5, Canada) uses the simple burglar alarm circuit shown in Fig. 7 in his own home. It includes a lantern battery, a spst power lock switch, an alarm device such as a heavy-duty bell, a medium-current silicon contralled rectifier, a capacitor and a group of normally open (NO) spst magnetic switches, S1 through SN. The required capacitor value will vary somewhat with the gate sensitivity of the SCR used but, generally, will range from $0.05 \mu \mathrm{~F}$ to about $0.5 \mu \mathrm{~F}$ although a small electrolytic of 1 to $5 \mu \mathrm{~F}$ may be needed in some installations. In practice, the magnetic switches are mounted to protect doors, windows and other access openings, arranged to close when the door or window is opened, and wired in parallel.


Fig. 7. Burglar alarm uses sensors in parallel.

The circuit draws virtually no current from the battery when in a "stand-by" condition with the on/off switch closed. If any door or window is opened, closing the corresponding reed switch, however, C1 will charge and this current surge will fire SCR1, sounding the alarm. Once SCR1 switches to a conducting state, the control switches have no effect and the alarm will continue to sound until the battery runs down or until the system is reset by opening the power/reset switch.

Guy's design, while simple, fairly reliable, and offering the user a virtually zero stand-by current drain, does suffer from two disadvantages. First, the control switches are wired in parallel, requiring two wires to be run to every switch. Second, the circuit is not "fail safe." That is, if an accidental (or deliberate) break should occur in the wiring to the control switches, the system will offer reduced or zero protection. Most commercial burglar alarm systems employ a "supervised" alarm line . . . i.e., one in which a small current flows at all times, so that any break (as by a burglar cutting wires) will result in an alarm.


Fig. 8. Burglar alarm with supervised line for "fail safe" operation.

A "fail safe" design using virtually the same number of components is illustrated in Fig. 8. Here, normally closed magnetic switches are used to protect the access openings and these (S1 to SN) are wired in series, requiring a single lead alarm line. A current-limiting resistor (R1) is needed, but, generally, C1 is no longer required and may be eliminated (as shown by the dotted line connection). A half-watt resistor, R1's value will depend on the gate sensitivity of SCR1. If the device requires, say, only 1 mA to fire, a 10,000 -ohm resistor may by used, assuming a 12 -volt battery. The stand-by current drain can be determined by dividing the battery voltage by R1's value.
In operation, a small current flows through R1 and the control line circuit whenever the system is switched on but the control line effectively shorts out the SCR's gate voltage and this device remains in a nonconducting state. If there is a break in the line, however, whether caused by an open switch or someone cutting the line, gate current is applied to the SCR through R1, firing this device and sounding the alarm. As before, once the SCR has fired, the alarm will continue to sound until the battery is exhausted or until the system is reset by opening the on/off switch.
With either circuit, care must be taken when choosing the alarm device, whether a bell, buzzer, siren or gong. If this is an "interrupter" type electromechanical unit, it may be possible for an intruder to silence the alarm, once activated, simply by restoring the control line circuit to its normal standby condition (i.e., either opening or closing the appropriate switch, depending on which circuit is used). This problem may be avoided by shunting the alarm device with a small resistor of adequate value to maintain the SCR's sustaining current even when the alarm device interrupts the normal current flow.

Another DVM. The ink was barely dry on our March column describing digital meter circuits when the National Semicon-
ductor Corporation (2900 Semiconductor Drive, Santa Clara, CA 95051) announced a new 31/2-digit, 0.5 inch high LED display designed especially for instrumentation applications, including power supply readouts, multimeters and digital panel meters. Designated type NSB5388, the new common-cathode multiplexed GaAsP display is compatible with National's own ADD3501 DVM chip as well as with comparable chips offered by other manufacturers, and car, be connected easily by PCB type terminals on the edge of the device. Featuring separate access to its plus/minus sign and decimal points, the display offers a digit light intensity rated, typically, at 1.6 mcd at 10 mA per segment peak current.

Device/Product News. Motorola Semiconductor Products, Inc. (Box 20912, Phoenix, AZ 85036) has announced its entry into the "BIFET" operational amplifier market with a line of twelve devices, all based on the generic LF155A type. The LF155/255/355 series offer low supply current requirements; the LF156/256/356 group, a 5-MHz gain-bandwidth at a higher current; and the decompensated LF157/257/357, a 20 MHz gain-bandwidth.

The Intel Corporation (3065 Bowers Ave., Santa Clara, CA 95051) has introduced a new family of HMOS $4096 \times 1$-bit fullystatic random-access memories. Identified as the 2141 series, the family includes seven types providing four speed versions and three low-power selections. Requiring only a fraction of the power of conventional MOS static RAM's, the new devices offer maximum access times ranging from 120 to 250 ns, with minimum cycle times equalling the maximum access times. Assembled in 18-pin DIP's, all seven units use a single $+5 \mathrm{~V},-10 \%$ dc power source and are directly compatible with TTL on all inputs and outputs.

Raytheon's Semiconductor Division (350 Ellis Street, Mountain View, CA 94040) has announced a new high-performance dual operational amplifier. Designated type 4559, the new unit is specified for use in audio systems, data modems, telecommunications equipment, function generators, and similar equipment. Guaranteed to be unity gain stable, the 4559 has a minimum unity gain bandwidth of 3.0 MHz , a slew rate of $1.5 \mathrm{~V} / \mu \mathrm{s}$ equalization, a noise voltage of only $2.0 \mu \mathrm{~V} \mathrm{rms}$ maximum and a full power bandwidth of 25 kHz .

Fairchild's Semiconductor Products Group (464 Ellis Street, Mountain View, CA 94042) is now offering a new series of inexpensive 3-terminal, half-amp voltage regulators. Identified as the $\mu \mathrm{A} 78 \mathrm{C}$ family, the devices are offered in the special packages with heat sink tabs which may be used as direct replacements for units assembled in standard TO-202 packages. Nine voltage options of the regulator are available as stock items: $8,10,12,15,17,18,20,22$ and 24 volts.

RCA's Solid State Division (Box 3200, Somerville, NJ 08876) is now producing the first multiple-technology dualvoltage comparators available from the semiconductor industry. The CA3290 series of BiMOS dual comparators feature two independent single- or dual-supply voltage comparators on a monolithic chip and a high common-mode input voltage range, making then well suited for applications in long-timedelay circuits, square-wave generators, A/D converters, and high-source-impedance voltage comparators. Gate-protected MOS/FET transistors in the input circuit provide very high imput impedances ( 1.7 terohm typical), extremely low input currents ( 3.5 pA typical at +5 V ), and high-speed performance. With a dc supply voltage range of from 4 to 36 V , the devices are compatible with TTL, DTL, ECL, MOS, and CMOS logic systems. Different versions of the CA3290 family are available in TO-5 cans, 14 -lead DIP's and 8-lead MiniDIP's. $\diamond$

#  

## DERIVING 60 Hz

a. How do clock chip manufacturers obtain an accurate $60-\mathrm{Hz}$ timing signal from a $3.579545-\mathrm{MHz}$ crystal? The nearest frequency that permits the use of a decent dividing network seems to be 3.6000 MHz . -George Rogers, Waynesboro, VA.
A. The key word in your question is "decent." I assume you mean that the nearest frequency which can be divided by a nice round number is 3.6 MHz . You are correct-applying a signal at that frequency to four successive decade counters and one $\div 6$ counter (or a $\div 3$ stage and a simple flip-flop which divides by two) will result in an output at exactly 60

By John McVeigh
Hz . However, semiconductor manufacturers have chosen a different route.

Shown in Fig. A is the block diagram of National Semiconductor's MM5369 oscillator/17-stage programmable divider IC. The programmable modulus of the counter can vary from 10,000 to 98,000 . If a $3.579545-\mathrm{MHz}$ quartz crystal is connected to the IC and tuned (via a small trimmer capacitor) to oscillate at exactly that frequency, the MM5369 will produce a $60.0000838-\mathrm{Hz}$ output-according to my calculator-when the counter's modulus is 59,659 . The output waveform for this combination of oscillator frequency and counter modulus is shown in Fig. B.

Now, why use a $3.579545-\mathrm{MHz}$ crystal and a modulus of 59,659 ? I suspect that the reason is that high-quality crystals at that frequency have been massproduced for years. Every color television contains one because that is the frequency of the chroma subcarrier. The chroma oscillator is locked to that at the transmitter by PLL techniques.
The use of feedback via gates permits a designer to obtain nonstandard moduli such as $3,5,7,9$, etc. The 7490 bi-quinary counter with a $\div 5$ stage as well as a $\div 2$ flip-flop is a good example of this. Combining such feedback with presettable counters via LSI technology enables manufacturers to produce such sophisticated counters as the MM5369 at reasonable cost. Apparently, producing such counters and combining them with mass-produced color TV crystals was found to be more economical than using IC counters having standard moduli teamed up with quartz crystals that oscillate at:"nice, round". frequencies but that were not being mass-produced.


## BUZZING DIODES

Q. My stereo system includes a separate tuner and amplifier. I am experiencing heavy interference when the tuner is in the AM mode. There's an awful buzz clear across the band which only the strongest signals can overcome. Investigating the problem with a small transistor radio, I discovered that the amplifier is the source of the noise. Also, the noise signal is very strong near the tuner's ferrite rod antenna-whether the tuner is plugged into the ac outlet or not. Both components' chassis are grounded to a ground rod. What causes this, and what can be done to eliminate it? -David Shoulders, Eugene, OR.
A. I suspect that the interference is being caused by transients in the amplifier's power supply. When an ac voltage is applied to a silicon diode, the diode

does not conduct exactly for one halfcycle and then shut off for the other halfcycle. Rather, a silicon diode will not begin to conduct until the barrier potential at the diode junction (about 0.6 volt) is overcome. Also, the diode is a very nonlinear device, especially at the knee (the region in which it starts to conduct) of its characteristic curve.

As a result, turn-on and turn-off transients are generated near the $0^{\circ}$ and $180^{\circ}$ points in the sinusoidal cycle. These transients are rich in interferencecausing harmonics of the line frequency. I suspect that they are being radiated by wiring in the amplifier and perhaps by the line cord. The reason the signals are stronger near your tuner's ferrite rod AM antenna is transformer action. Mutual coupling between the two ferrite antennas causes signals picked up by the tuner's coil to be passed to that in the transistor radio.

The diode transients can be dealt with by installing $0.1-\mu \mathrm{F}$ disc ceramic capacitors on both sides of each diode as shown in the figure. Here, a full-wave center-tapped power supply is shown. If a dual polarity supply is used, repeat the procedure for the negative supply. If a
full-wave bridge rectifier is employed, install bypass capacitors at each corner of the bridge. The capacitive reactance of these components is too high to interfere with the rectifying action of the power supply.

All chassis in the system should be well grounded to a good earth ground. You mentioned that both components are grounded. That's good-but beware of hum-producing ground loops. If separate grounding wires are attached from each chassis to earth ground, no conductors should run from one chassis to the next. The shields of signal cables, if connected to a chassis at each end of the cable, will cause a loop to occur between the already grounded chassis. Such a ground loop can cause hum problems, but not the "buzz" you have described. Look to the amplifier's power supply for the source of that signal.

[^3]
# Advanced Electronic Career 

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# 불ํํIㅇํ Experimenter's Corner 

By Forrest M. Mims

## DIGITAL TO ANALOG CONVERTERS, PART 1

A
LMOST any electronic circuit can e classified as either analog or digital. Analog circuits are those in which the signal voltages present may be at any level between low and high extremes determined by the power supply. Many analog circuits are known as linear circuits since they produce an output directly proportional to an input signal over a limited range of amplitude and frequency. Digital circuits are, on the other hand, those in which signals can

D/A Conversion. Let's suppose that you've built a simple digital controller circuit that will turn individual lamps in an array on and off in any pattern you specify. The brain of the controller is a semiconductor memory that you can program with the desired information. How would you use your controller to adjust the brightness of a single lamp without modifying the controller circuitry?
The solution to this problem is a digi-tal-to-analog (D/A) converter. The D/A


Fig. 1. Connecting digital controller to D/A converter.
assume only one of two distinct levels. Typically, one is at or near ground potential and the other near the powersupply voltage. In TTL digital integrated circuits, the two voltage levels are a low of a few tenths of a volt and a high of about 3.3 to 5.0 volts.

Although an amazing variety of circuit functions can be performed using only analog or digital techniques, some applications can only be accomplished by combining the two methods. Some examples of combining analog and digital techniques are the digital voltmeter, speech recognition circuitry, sophisticated motor-speed controllers, digital data transmission and many kinds of computer output circuits for controlling electromechanical devices like solenoids.


Fig. 2. Simple 4-bit
D/A converter using
resistor ladder.
converter is connected directly to the controller's output and adjusted to produce an output voltage proportional to the controller's binary output. Figure 1 shows how the controller is connected to the D/A converter.

There are several ways to design a D/A converter circuit, but the most common uses a resistor network followed by one or more operational amplifiers. Figure 2 shows a simple, 4-bit D/A converter that uses a ladder-like network of parallel input resistors. The values of the resistors are determined by their binary weighting factors. A 4 -bit input has binary weighting factors of $2^{3}\left(8_{10}\right), 2^{2}$ $\left(4_{10}\right), 2^{1}\left(2_{10}\right)$ and $2^{0}\left(1_{10}\right)$. If the lowest order (20) resistance is R , then the values are R, R/2, R/4 and R/8.
Though the circuit shown in Fig. 2 is very simple, it has two major drawbacks. First, it's difficult (at the hobbyist level) to find resistors having the precise resistances that are required. Second, the resistance values become spread over a very wide range for a relatively small number of input bits. Thus, for a 10 -bit D/A converter, the input resistors must range from $R$ to $R / 1024$. The digital circuit connected to the D/A converter, often a chain of flip-flops or gates, must be
able to supply a wide range of currents (high currents for low resistances and low currents for high resistances).

The problems of the D/A converter in Fig. 2 can be solved by increasing the number of resistors in the ladder network. The result is the R-2R ladder network shown in Fig. 3. As you can see, the ladder resistors have values of $R$


Fig. 3. $R-2 R$ resistor
ladder network
for D/A converter.
and $2 R$. This means only two readily available resistance values are required. It's possible to use a single value if you're willing to connect two $R$ resistors in series to obtain the $2 R$ values.

D/A Conversion Demonstrator. If you're serious about electronics experimentation and want to stay abreast of the latest developments, you should assemble a D/A demonstration circuit like the one shown in Fig. 4. This circuit is the basis for the practical D/A converter we'll discuss later.

The demonstrator circuit uses four spdt switches to achieve a 4-bit input. There's nothing improper about a mechanically switched binary output. Many real-world circuits use them. Most D/A converters, however, are connected directly to a digital circuit that provides a binary output.

You can test the operation of the D/A converter by connecting a voltmeter across its output while switching in various binary outputs. Since we're using a 9 -volt battery as a reference voltage and since there are sixteen possible input combinations, the output voltage should range from 0 volts to slightly under 9 volts in increments of $9 / 16$ volt.

Here are the actual voltages measured with the demonstrator circuit:

| Binary In | Voltage Out |
| :---: | :---: |
| 0000 | .00 |
| 0001 | .57 |
| 0010 | 1.12 |
| 0011 | 1.69 |
| 0100 | 2.19 |
| 0101 | 2.76 |
| 0110 | 3.32 |
| 0111 | 3.89 |
| 1000 | 4.50 |
| 1001 | 5.06 |
| 1010 | 5.60 |
| 1011 | 6.15 |
| 1100 | 6.69 |
| 1101 | 7.78 |
| 1110 | 7.82 |
| 1111 | 8.40 |

The output of the circuit is plotted on a graph in Fig. 5. As you can see, the re-


Fig. 4. D/A demonstrator circuit.
sponse of the circuit is reasc nably linear, even though I used $10 \%$ tolerance resistors. Commercial D/A converters are made with resistors having tolerances of $1 \%$ or better. When very close tolerances are necessary for superaccurate D/A converters, a plised laser is used to vaporize minute fortions of the carbon or metal-film resistive elements until the exact values required are obtained.

It's handy to be able to predict in advance the analog voltage output for a specific input bit pattern. The weighting factors for our 4-bit network are:

Most Significant Bit-

$$
\begin{array}{lll}
2^{3}=1 / 2 & \text { Reference Voltage } \\
2^{2}=1 / 4 & " & " \\
2^{1}=1 / 8 & " & "
\end{array}
$$

Least Significant Bit-
$2^{0}=1 / 16$ Reference Voltage

To calculate the analog oulput, simply multiply the reference voltage by the weighting factor for each bit portion with a 1 and sum the products. Thus 1100 is:

$$
\begin{array}{ll}
1-1 / 2 \times 9=4.50 \\
1-1 / 4 \times 9=2.25 \\
0-0 & 0 \\
0-0 & \frac{0}{6.75 \text { volts }}
\end{array}
$$

The calculated analog output, 6.75 volts, is only 0.06 volt higher than the value measured with the prototype circuit. That's an error of less than $1 \%$ !


Fig. 5. Voltage output vs. binary input for Fig. 4.

[0] Product

## SONY MODEL ICB-1020 PORTABLE CB TRANSCEIVER

## Hand-held AM, 40-channel transceiver has I-W r-f power and hot front end.



The hand-held Model ICB-1020 AM transceiver from Sony uses the latest in phase-locked-loop (PLL) synthesizers to provide coverage on all 40 CB channels. It is powered by eight AA cells that fit within its housing and has provisions for obtaining power from a 12 -volt mobile electrical system and, with an adapter, standard 117 -volt ac line power. Whichever power source is used, the rated $r$-f output power from the transmitter is 1 watt.

The transceiver provides: volume and squelch controls; separate power switch with power ON and $\mathrm{CH}-9$ positions (the latter bypasses all channels for instantaneous access of channel 9); earphone, external mic, and accessory power jacks; and a large S/r-i/battery condition meter.

The walkie-talkie type transceiver measures $105 / 8^{\prime \prime} \mathrm{L} \times 41 / 2^{\prime \prime} \mathrm{W} \times 35 / 8^{\prime \prime} \mathrm{D}(27$ $\times 11.4 \times 9.2 \mathrm{~cm})$ and weighs 2.75 lb $(1.25 \mathrm{~kg})$. Its suggested retail price is \$174.95.

General Description. We did not receive a schematic diagram for the transceiver, nor were we able to determine much of the circuitry by visual observation. From what we could observe, however, the receiver employs double-conversion to i-f's at 10,695 and 455 kHz , with selectivity obtained at the latter.

The PLL system is more or less standard. It engages a $10,240-\mathrm{kHz}$ crystal signal from which the standard reference is derived and which also is used for the second-conversion oscillator. The technique used is to have the volt-age-controlled oscillator (vco) used for the first conversion on the low side of the CB signal instead of the upper side. This ensures better image rejection and a reduced chance of adverse receiver-toantenna or case radiation on vhf.

The transmitter carrier is derived from the PLL system. It is amplified and raised to a high-level output by a power amplifier, where a multisection network matches the output to the telescoping antenna built into the transceiver. As usual, the driver and power-amplifier stages are collector modulated.
The case of the transceiver is designed along the lines of a handset. The speaker is at the top end and is equipped with a cushioned pad for comfortable listening. Even so, sufficient audio level is avaitable to seldom require ear-contact listening.

At the center of the case is a small hinged panel that can be depressed to activate the transmitter. Slightly recessed in the center of this panel is the channel selector with an easy-to-grip operating bar. The channel numerals around the periphery of the selector are small but easy to read.

The power switch is a three-position lever located on the right side of the transceiver's case. Pushing the lever one position forward powers the transceiver. Pushing the lever to its third posi-
tion immediately switches the system to channel 9 , at which time, a tiny LED comes on at the channel-9 position of the selector switch. Pulling the lever back one position cuts out channel 9 and reinstates operation on the channel to which the channel selector is set. This is a convenience for occasional monitoring of channel 9 without disturbing normal operation and for quickly accessing the channel in an emergency.
The large round-faced meter, calibrated from S1 to S9 for receiving, is exceptionally readable. Colored portions of the meter scale indicate battery condition and the proper modulating levels.

A "lip" type microphone is built into the bottom of the transceiver's case. This is backed up by an external-microphone jack that can be more conveniently used in fixed and base-station locations with an accessory mike.
At the bottom of the case are located the volume and squelch controls. The jacks for external facilities are on the right side of the case and are protected by a rubberized cover. The two halves of the case are also sealed with a weatherproof Neoprene gasket.

The antenna is on the left side of the case. It is a swivel telescoping whip that can lock in any orientation between vertical and horizontal. When not in use, the antenna stows along the side of the case, where a retaining clip holds it.
The batteries that power the transceiver install in a plastic holder that inserts and locks into a cavity at the lower end of the case. This arrangement eliminates the need to open the case to change batteries.

Laboratory Measurements. The transmitter developed a nominal 1-watt output on our test bench. It is presumed that to conform with FCC type-acceptance requirements, some form of automatic modulation control (amc) is used. However, we discovered that overmodulation at greater than $100 \%$ of the nega-tive-modulation peaks occurred when speaking into the microphone at very close range. Even so, adjacent-channel splatter with voice was a minimum of 55 dB down.

The $6-\mathrm{dB}$ down audio response on transmit was a nominal 300 to 2200 Hz . In on-the-air tests, the signal sounded exceptionally clean and crisp. The r-f output frequency's tolerance held to within $\pm 5 \mathrm{~Hz}$ on all channels, centered at +270 Hz .

Measurements on the receiver were quite difficult to perform because the
telescoping antenna is permanently attached to the transceiver, allowing interference from CB signals almost continuously. We did, however, manage to squeeze in most tests, but our results must be viewed as nominal values.
The receiver's sensitivity was $0.5 \mu \mathrm{~V}$ for $10 \mathrm{~dB}(\mathrm{~S}+\mathrm{N}) / \mathrm{N}$ at $30 \%$ modulation and 1000 Hz . Listening tests indicated that this figure would have been even better under ideal laboratory test conditions. The maximum squelch threshold sensitivity measured 0.5 microvolts, while the meter registered S9 with a 300-microvolt input.
Image and i-f rejection were an excellent 80 dB , while other unwanted-signal rejection (mostly due to overloading) was 55 to 60 dB . Adjacent-channel rejection and desensitization was 55 to 60 dB. Translated into signal strength, desensitization occured with a nominal $1000-\mu \mathrm{V}$ signal, which is also the level required to produce overload responses.

The 6-dB down audio response was 225 to 2300 Hz , and the audio output power with a $1000-\mathrm{Hz}$ tone into a $4-\mathrm{ohm}$ speaker was 250 mW with a sine wave at $3.5 \%$ THD and 340 mW at $10 \%$ THD with slight clipping.

We were unable to obtain an accurate determination of the agc characteristic because the permanently attached antenna picked up case or lead radiation from our signal generators. However, listening tests indicated about a $10-\mathrm{dB}$ audio output variation over a $60-\mathrm{dB}$ input range at 10 to $10,000 \mu \mathrm{~V}$.
The receiver drew about 70 milliamperes with no input signal. On transmit and without modulation, the drain was 250 milliamperes.

User Comment. This transceiver has a really high-quality professional look and feel about it. Its workmanship reminds one of a fine camera's.

We noted that the transceiver does not employ a noise-limiting system. However, the service for which the transceiver was designed does not generally require one.

To summarize, this is one of the finest transceivers we have ever used, both in appearance and in performance. It has a really "hot" receiving section and signal delivery that comes up to all our expectations for a 1-watt transmitter working into a limited antenna system. Reliable communication coverage can be maintained over a 1-to-5-mile range, depending on terrain and location.


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By PE Editorial Staff

## INSIDE A SIDEBAND CLUB

THE WHISPERS started as soon as I entered a meeting room where some 75 CB single sidebanders were gathered for a weekly meeting in an Eastern-US town: "Is he from the FCC?" Finally confronted openly, I assured attendees that I wasn't, displaying my PE business card and a copy of the latest issue to prove it.

Suspicions allayed, members candidly discussed their club and "hobby" with me. According to "call sign" numbers assigned, I was told, there are about 2700 sidebanders who belong to this club, one of many such organizations around the country. Attempting to ferret out club officers, I learned that there were none: no president, vice president, treasurer or what-have-you. The only title (unofficial) was the Master Holder of the Log (club member \#386)-a listing of club on-the-air numbers that started with \#1 and is said to now be past \#2700. If a member retires, his or her number is not issued again. In addition to the numbers, the log contains the members' first names and the towns in which they reside. Perhaps 150 members are considered to be very active.

In response to my question about the purpose of the club, members told me that the reasons are outlined in the club's bylaws. They added that there are no copies of the bylaws available because they are not written down, simply passed from one member to another by word-of-mouth. Essentially they cover how to break properly, use of $Q$ codes (which are employed extensively), and channels "reserved" for their local use ( 16 LSB, 18 USB, and 36 to 40 upper and lower sidebands). Moreover, there are no club fees.

Listening to these club members modulating prior to attending a meeting, it's clear that they are polite-no profane language whatsoever, a wait of about three or four seconds before keying a mike in the event a breaker wants to announce himself, and so on. It's equally
clear that FCC call signs are not used and that the five-minute talk limit prohibition is not followed.
"Does anyone here use a linear?" | innocently asked. After a spate of "What's a linear?" responses, I learned that everyone used either a linear or an overpowered basic rig. Power is the name of the game. But DX'ing is not! In fact, club members are so unhappy about purposeful incoming long-distance transmissions, that they expressed a wish that there were some channels set aside by the FCC just for this use so that those who like this form of CB communications would stay away from the local channels. Most skip in this eastern locale comes from Texas and Oklahoma, they told me.

Every area has its equipment favorites. In this club, it was certain Cobra, Courier and President CB models. "They use the same Signetics PLL chip and a fine pc board... we've got it down pat on how to adjust the rigs for higher power." No one, it seems, uses stock models here. Interestingly, many club members use amateur radio gear.

Virtually all of the club members said they own both a mobile and a base station. In many cases, the base was originally the first SSB mobile purchased, with addition of a power supply. Also, it was claimed that more than 95 percent of the club members had 40-channel rigs, even if they "rolled their own" in one way or another.

Many of the members own beam antennas so that they can really reach out. The most popular ones cited were: Avanti's "Moonraker IV" and "PDL-II," and Wilson's "Y Quad." Antenna Specialists' "Super Scanner" was very popular, too, though more for its omniswitched position than for its beam application. Shakespeare's "Big Stick" was a special favorite among omni's at the club. The K-40 mobile was held in high regard here.

One member (\#2584) is a long-time ham, claiming a General license. To prove it, he whipped out his amateur license, which drew some gasps from fellow members because it listed his surname, which is information rarely issued to even club members. He participates in training club members for their amateur radio license, mostly the Novice, he said. "About 70 percent of the sidebanders are trying to become hams."
The club members don't use call signs on the air because most of them operate contrary to some of the FCC rules and regulations. Only first names (no handles) and club numbers are employed. Also, when using AM they find it necessary to key the mike and get in quickly. If legal call signs were used, it would be impossible to "break" in the area, they say. The majority of members operate



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above channel 40 when the assigned channels are too noisey. "Sliders" are de rigueur.

There are national CB clubs that specialize in long-distance communications, it was pointed out. The Whiskey Club was cited as an example. The illegal call sign for this club starts with the number that represents the order in which the state that the CB'er resides in joined the Union, followed by a W, followed by an assigned number. Another illegal DX club is the Echo club from Germany. Both clubs are said to form networks directly above Channel 40's frequency.

Typical CB communication range with the rigs used in this group is said to be: mobile-to-mobile, 10 miles; mobile-tobase, 15 miles; base-to-base with omnidirectional antennas, 20 to 25 miles; with beams, 25 to 50 miles. All club members agreed that during traffic rush hour, it's impossible to communicate more than a mile or so from a mobile.

Asked how much cooperation they get from AM'ers on the channels they carved out for themselves, they said that most AM'ers cooperate. The few that don't are "wiped out" by using transmit sliders to put out signals on top of them.
"The people are really nice," was the most common reason given for working CB sideband. They come from all walks of life, are mostly adults ( 30 to 50 years of age, it seemed to me), and among the sidebanders present were lwo physicians, one school teacher, a hospital administrator, and an auto-service station manager. The members all said, too, that their hobby has motivated them to learn a lot about electronics. It was estimated that a typical sidebander spends from $\$ 2000$ to $\$ 3000$ on CB/amateur radio gear until the desired mobile and base systems are owned.

What do they want from the FCC? "Just for them to let us alone because we've got the best of all worlds right now for our hobby," was one response that drew agreement from everyone. If there was a change, they'd like channels assigned strictly for SSB use, where AM'- $^{\prime}$ ers couldn't use them unless a rig was doctored. SSB'ers and AM'ers are incompatible, they noted. Further, they would like some of their practices legalized, especially higher r-f power than presently permitted. Members added that they don't know of any problems caused by running higher power, stressing that they don't use illegally high power between 7 p.m. and midnight, the most popular TV viewing hours.

Shades of Prohibition days!

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By Hal Chamberlïn

## COMPUTER ARITHMETIC

ASK A LAYMAN what a computer does best and he will probably say that it is best at computing complicated mathematical formulas. However, ask the same question of a hobbyist who has obtained a computer for the purpose of mathematical computation and he will probably say that his machine handles text much better than numbers. The truth is that microcomputers have very littlè "number crunching" ability built-in. All that is normally available is addition and subtraction of 8 -bit numbers and some can't even subtract directly! Automatic handling of decimal numbers is frequently provided also, but proper use of decimal arithmetic is far more complex than the normal binary arithmetic. As a result of this limited arithmetic capability, all other mathematical operations must be broken down into addition and subtraction of 8 -bit numbers.
Arithmetic with integer numbers larger than the capacity of a computer word is termed "multiple precision arithmetic" with "double precision" used to signity the special case of two-word numbers. The words (bytes in an 8-bit processor) forming the number are simply strung end-to-end. A 24 -bit number for example would consist of 3 bytes. The leftmost byte is called the "most significant" or "high" byte, and the rightmost is called the least significant or "low" byte. The bytes in the middle, if any, have no special name.
The most important multiple precision operations for general computation are addition, subtraction, multiplication, division, comparison, negation, incrementing, decrementing, and left and right rotates including the carry flag. Usually the software necessary to do these operations is organized into a subroutine package. It is a common practice in such software packages to define "registers" for multiple precision numbers in main memory. At least two registers are typically needed; the left operand, and the right operand, quite analogous to the equivalent operation done on paper. 88

The package may be written for a specific number length such as 32 bits (4 bytes) or the length may be variable and passed to the subroutines as arguments. Besides the actual arithmetic subroutines, a "number move" routine is needed to conveniently move the multibyte data into and out of the pseudo registers. With a variable length arithmetic package, it becomes easy to do calculations to dozens or even hundreds of "decimal places" of accuracy in assembly language, which is much better than any BASIC language system.

Increment and Decrement. Incrementing and decrementing multiple precision numbers is probably the simplest of this type of operation. Assume for the moment that a 16 -bit number, which is stored as two bytes in memory, is to be incremented by one. Although some machines may have an instruction to do this to a pair of registers, let's try to do it directly in memory using a 6502 microprocessor. For incrementing, the first step is to increment the least significant byte directly in memory using the INC instruction of the 6502. Next it is necessary to determine if an "overflow" of that byte occurred. This could normally be determined by looking at the carry flag but on the 6502, INC does not change the carry. Close examination of the overflow situation, which only occurs if the byte was equal to $\mathrm{FF}_{16}$ before incrementing, will reveal that after the overflow the byte will always be zero! Therefore, the Z-flag can be tested instead. If an overflow did indeed occur, then the most significant byte should be incremented; otherwise the job is done. This procedure can be extended to numbers of any length by continuing to move left. byte-by-byte, incrementing as long as the previous byte overflowed.

Multiple precision decrement is nearly as easy. First the low byte is examined to determine if it is zero. If it is not, it is decremented and the job is finished. If it is zere, a decrement will cause it to un-
derflow. In that case, we decrement it anyway and then move left to the next byte and repeat the sequence. When the most significant byte is reached because of underflow of all previous bytes, it is simply decremented without any testing for zero. Note that these algorithms work equally well for signed two's complement multiple precision numbers.

Add and Subtract. Multiple precision addition and subtraction are more interesting. These operations require the left and right operand pseudo registers. The usual convention is to put the answer into the left operand register, much like arithmetic instructions themselves. The first step in double-precision addition is to clear the C flag and then, using the ADC instruction, add the low byte of the left operand to the low byte of the right operand and store the result back into the low byte of the left operand. Now, being careful not to disturb the carry flag, the ADC instruction is used to add the high bytes of the two operands together and store the result in the high byte of the left operand which completes the operation. For multiple precision one continues left adding pairs of bytes together with the ADC instruction until the most significant bytes are added.

Quadruple precision addition operation is shown below. The C-flag is used
Addition of two 4-byte numbers.

| 02791256 | 02 | 79 | 12 | 56 |
| :---: | :---: | :---: | :---: | :---: |
| $+0 \mathrm{FE} 534 \mathrm{B3} 3$ | 0 F | E 5 | 34 | B 3 |
| 125 E 4709 | $\frac{1}{12} ?_{1} \frac{0}{5 E} \sum_{0} \frac{1}{047} \eta_{1} \frac{0}{09} \leftarrow \mathrm{C}=0$ |  |  |  |

to transfer carry information from lesser significant bytes to more significant ones. One can actually think of it as adding multiple digit numbers together where each "digit" is a byte between 0 and 255. The carries are transferred from digit to digit just like decimal addition on paper. If care is taken on the return sequence from the add subroutine, except for $Z$, the status flags will correctly indicate the result of the operation just performed.

For subtraction, one could write a similar subroutine using the SBC instruction on each byte in the numbers. Before starting, however, it is necessary to set the C-flag for proper operation. Another way to do subtraction is to complement the right operand and then add it to the left operand using the add subroutine just described. Usually a complement routine is needed anyway, so this
popular electronics
scheme can also save some memory.
A two's-complement operation consists of merely inverting the bits of the number and then incrementing the result. Thus a multiple-precision complement routine would invert the bits of each byte of the right operand and then call the multiple precision increment routine described earlier. An exclusive-OR of a byte with all ones, using an "EOR \#\$FF" instruction, is all that is necessary to invert it.

Comparison. Not all needed multipleprecision functions are involved with computing answers, some comparison operations are necessary also. Probably the most important of these is a comparison with zero since the Z-flag is not meaningful after a multiple-precision add or subtract. Such a subroutine could be used to logic-OR all of the bytes in the number together using the accumulator. If the result of the OR'ing is zero, then each byte in the number must have been zero.

One way to do a signed comparison between two signed numbers is to subtract them and then see if the result is negative (right operand is larger), zero, or positive (left operand is bigger). Besides destroying one of the numbers being compared, this method suffers from a subtle pitfall. If the left operand is a large positive number and the right operand is a large negative number, the subtraction can overflow and the comparison result will be invalid. The converse case, right operand positive and left operand negative, creates the same problem. There is never any possibility of overflow when the numbers are of like sign however.

A dedicated comparison subroutine overcomes both problems. Unlike previous routines where the operation started at the right, comparison should start with the most significant bytes. The first step is to look at the sign bits. If they differ, the comparison result is clear already and any overflow problems are avoided. If the sign bits are the same, then the most significant bytes are subtracted but the result is not stored. If the result of the subtraction is nonzero, then the outcome of the comparison is known (negative means right operand larger, positive means left operand larger) and a return can be taken. If the result is zero, then lesser significant bytes must be subtracted until either a nonzero result is obtained or the entire number has been processed. In the latter case, equality between the two is the conclusion.

Rotation. Multiple precision rotation is much like addition in that the C-flag is used to transfer bits from one byte to the next. Although a more comprehensive set is easily written, subroutines to rotate left and right including $C$ are usually sufficient. The rotate is effectively changed to a shift if the calling program clears $C$ before calling multiple rotate. Rotates by several bit positions are accomplished by repeated calls to a single bit position rotate routine.

For a multiple rotate left, start at the rightmost byte. The byte is rotated left with carry by use of the ROL instruction which puts the old C-flag into bit-zero, shifts the whole byte left by one, and
puts bit-7 into the C-flag. Following this, the next byte to the left is rotated in a similar manner. The C-flag serves to transfer bit-7 of the low byte into bit-zero of the next higher byte. This process is repeated until the entire number has been done. Note that in the 6502, all of the manipulation can be performed directly in the pseudo register.

Rotate right is the exact opposite of rotate left. Start with the leftmost byte and use the ROR instruction on each lower byte in sequence. Unfortunately, some early production 6502's were manufactured without the ROR instruction. For these, an ROR can be simulated by doing 8 ROL's in a row instead. $\diamond$

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Lavole Laboratories, Inc. spectrum analyzer. Model LA-18A-82-includes indicator unit and tuner unit. Operation and service manuals. Yuma Two-Way Radio Services, Inc. Box 693, Yuma, AZ 85364

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# HLEmCHIRONTICS WTOIE3ILID News Highlights in inief 

## Double-Duty Digital

With cars getting smaller, finding space on the time-honored dashboard gets harder every year. Ford's solution for its 1978 Pinto and Bobcat cars is a digital clock that doubles as the AM radio's tuning display. The clock's display turns off when the ignition is off, preventing battery drain; to check the time when the engine's not running, you simply press a button. As with all digital displays, this one will make it easier to see at a glance exactly what the time is or exactly to what frequency the radio is tuned. One other advantage Ford doesn't mention: digital clocks, in our experience, seem to stand up better to the jolting and temperature extremes that have given older mechanical and electrical clocks the reputation of keeping time only until the payments stop.

## REACT Keeps Reacting

REACT International, Inc, an organization of volunteer emergency CB radio owners was 15 years old last year. Though the oldest (and largest) group of its kind, it happily responds to ideas originating in the field. For example, one innovation planned for this year is a full computerization of individual team and membership records. which will simplify direct mailings to individual members. (numbering about 100,000 ) rather than bulk mailing to REACT team leaders. That should get information into team members' hands sooner, and save the local teams a fair amount in remailing and clerical expenses. There will also be a bi-monthly publication on team management for team leaders. For youth members. there will be a junior REACT program, too. Obviously. REACT listens to its members-not just to channel 9. To join a REACT team, contact them at 111 E . Wacker Drive, Chicago, IL 60601. Dues are $\$ 5$.

## Databanks on Home TV

"Pages" of special "magazines" as well as computerstored information will be popping up on some British TV screens shortly. The "magazine" pages come from a service called "Teletext," developed jointly by the British Post Office (which is in charge of telecommunications in the U.K.), the British TV-set industry, and the semiconductor and information industries. Teletext sends its information in the intervals between frames of the broadcast TV signal. Teletext-compatible TV receivers should be available in Britain about now. By the middle of this year, a limited number of sets should appear with facilities for reception of still another TVbased information system. The new system, Viewdata, will provide viewers with thousands of pages of information, transmitted over regular telephone lines. Dial Viewdata. and the TV screen fills with a "menu"' page explaining the general subject areas available. Using the phone dial as a "computer terminal." the viewer works his way quickly through more and more detailed "menus" until he finds the exact topic he wants. More
than 6000 "pages" (each about equal to a doublespaced, typewritten page) of information are now available, with up to 60,000 planned for the systems' midyear trial introduction.

## "Smart" Pinball Machines

Now they're even building computers into pinball machines. The immediate advantage of replacing time-honored relay logic by microprocessors is the new features this allows: commercial pinball machines can now include such things as a memory to allow a player to compete against the high score of the week, the month or the year. (An alphanumeric display lets the year's high scorer see his name in lights-until some new player broke his record.) Bally's "Fireball" home machine (available from both Sears and JS\&A Sales, and in kit form from Heath) stores up to four players' scores, can be programmed for different degrees of player skill, and has a built-in synthesizer that plays seven different tunes when extra scores or bonuses are won. The main advantage may eventually prove to be lower cost. That could lead to more home pinball machines.

## Recording Studio Equipment Study

In a recent survey, US recording studios were polled to see what audio equipment was most popularly used. Interestingly, the 569 questionnaires returned to Billboard magazine, conductor of the survey, revealed that quite a few familiar, home-equipment brand names were in use in recording studio applications. The most favored monitor amplifiers were Crown and McIntosh, for example, and the tape-recorder (with fewer than 16 tracks) popularity list was headed by Ampex, Scully, Teac, and Sony. Nearly all brand names in the portable mixer category are well-known, too. Similarly, microphones in use, except for front-runner Neumann, could be found in any tape enthusiast's home: Electro-Voice, Shure, AKG, Sony, and Sennheiser. Preferred headphones included such familiar names as Koss, Sennheiser, and AKG. Shure, Stanton, and Pickering are the most widely used phono cartridges, and the overwhelming choices in audio tape are Ampex and Scotch. The speaker-system field was dominated by JBL and Altec, with Electro-Voice also being popular.

## Hand-Held, No-Battery Calculator

Tired of replacing calculator batteries? Then Photon, a new $\$ 39.95$ hand-held calculator developed by TEAL Industries, Inc., may be for you. Photon is powered by a 5.6 mm by 1.6 mm bank of solar cells which operate in all wavelengths of visible light, indoors or out. Among the more mundane features of Photon are an eight-digit liquid crystal display, computation in both mixed and chain modes, four-key memory, and live percent, square-root, and sign-change keys. But don't look for an on/off switch-there isn't one!

## Get out a whole lot forther with Super Sconner

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[^7]:    BATTERIES, NiCad new surplus. 10 pack - $\$ 3.50$. G. White. P.O. B. 19279, San Diego, CA 92119.

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