

SPECIALIZED TELEVISION ENGINEERING

TELEVISION TECHNICAL ASSIGNMENT

SERIES-PARALLEL CIRCUITS



Copyright 1947 by Capitol Radio Engineering Institute Washington, D. C.



www.americanradiohistorv.com

FOREWORD

This assignment gives you an opportunity to test your overall knowledge of applied mathematics and a.c. circuit theory in applications to radio frequency networks. Almost all r.f. circuits can be broken down into series or parallel circuits or combinations of the two. In the preliminary design state almost any group of specifications may be given with the requirement that the remainder be calculated.

For example, the tuned r.f. network between a transmitter power amplifier and an antenna may include a tuned tank circuit, a harmonic suppression circuit, a tuned antenna circuit and provision for coupling the first to the In addition, in a television transmitter there also last. will be a vestigial side-band filter to remove a portion of one of the side-bands. This is usually-but not alwaysdone between the power amplifier and the antenna. Normally most of the L, C and R values must be calculated on the basis of given frequency, required power output, known amplifier tube characteristics, and measured antenna resistance. If a directional antenna array is to be used, the factor of phase angle becomes important. Practically all television and FM broadcasting radiating systems employ multi-element units which require the use of phasing networks.

This assignment will give you an overall picture of such r.f. circuits and will prepare you to handle specific applications of such circuits when they are taken up in later assignments.

Note particularly in this assignment how easily the most complex circuit is broken down for analysis into simple component circuits. This is extremely important. Just as the expert mathematician looks at a complex algebraic equation as a series of simple algebraic processes

and attacks the problem in a systematic manner, so should the solution of a complex electrical circuit be viewed and attacked. Problems such as those studied in this assignment will help you to acquire the habit of orderly thinking. Nothing else will simplify the analysis of complex circuit operation so much as the orderly approach.

The fact that you have reached this assignment indicates that you have the necessary background to handle it without difficulty. You should enjoy it immensely. When you have completed the exam, check your work carefully and then go on to the next assignment which takes up practical radio applications of r.f. tuned circuits.

> E. H. Rietzke, President.

- TABLE OF CONTENTS -

TECHNICAL ASSIGNMENT

SERIES-PARALLEL CIRCUITS

		Page
	SIMPLE PARALLEL CIRCUIT	1
	PARALLEL COMBINATION OF SERIES LCR	
	CIRCUITS	8
	ADDITION OF SERIES CIRCUIT TO	
	PARALLEL COMBINATION	11
	COMPLEX NETWORK	12
	SUMMARIZING THE VOLTAGE, CURRENT AND	
	IMPEDANCE RELATIONS IN SERIES AND	
	IN PARALLEL CIRCUITS	17
THE	LC TABLE	18

www.americanradiohistory

SIMPLE PARALLEL CIRCUIT: THE CHARACTERISTICS OF PARALLEL CIRCUITS HAVE BEEN COMPARED WITH THE CHARACTERISTICS OF SERIES CIRCUITS UNDER SIMILAR CONDI-TIONS. MOST OF THE CHARACTERISTICS ARE EXACTLY OPPOSITE IN THE TWO TYPES OF CIRCUITS. CONDITIONS HAVE BEEN ASSUMED AND BOTH SERIES AND PARALLEL CIRCUITS HAVE BEEN DISCUSSED BOTH THEORETICALLY AND MATHEMATICALLY. IT IS NECESSARY TO THOROUGHLY UNDERSTAND THE PARALLEL CIRCUIT MATHEMATICALLY IN ORDER THAT THE OPERATION OF SUCH A CIRCUIT MAY BE PREDICTED FROM ITS KNOWN VALUES. A MATHE-MATICAL UNDERSTANDING OF A CIRCUIT WILL GREATLY SIMPLIFY THE STUDY OF THE THEORY AND OPERATION OF THAT CIRCUIT. BEFORE PROCEEDING TO A MORE COMPLEX CIRCUIT ANOTHER SIMPLE PARALLEL CIRCUIT PROBLEM WILL BE SOLVED. PARALLEL LCR CIRCUITS AND COMBINATIONS INVOLVING SUCH CIRCUITS ARE VERY EXTENSIVELY USED IN RADIO, BOTH TRANSMITTERS AND RECEIVERS.

CONSIDER THE CIRCUIT IN FIGURE 1. THIS REPRESENTS A PARALLEL TUNED CIR-

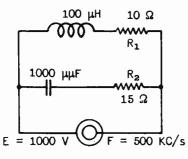


FIG. 1.

10 Ω CUIT IN WHICH THE RESISTANCE SHOWN IN SERIES WITH THE INDUCTANCE IS THE RESISTANCE DISTRIB-UTED THROUGHOUT THE INDUCTANCE COIL, AND THE RESISTANCE IN SERIES WITH THE CAPACITY REPRE-SENTS THE RESISTANCE LOSSES IN THE CONDENSER, F = 500 KC/s PARTLY IN THE DI-ELECTRIC AND PARTLY IN THE

PLATES AND CONNECTING LEADS. AS WAS SHOWN IN

THE STUDY OF SERIES CIRCUITS, THE EFFECTS OF THESE RESISTANCES ARE AS IF THE RESISTANCES WERE LUMPED IN SERIES WITH PURE INDUCTANCE AND PURE CAPACITY. IT IS EVIDENT THAT EACH BRANCH OF THE PARALLEL CIRCUIT MUST BE TREATED FIRST AS A SEPARATE SERIES CIRCUIT.

IN ORDER TO FIND THE TOTAL IMPEDANCE OF A PARALLEL CIRCUIT IT IS FIRST

www.americanradiobiston

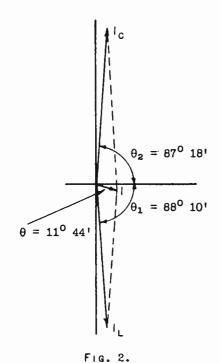
NECESSARY TO DETERMINE THE TOTAL CURRENT IN THE EXTERNAL CIRCUIT, THAT TOTAL CURRENT DIVIDED INTO THE APPLIED VOLTAGE GIVING THE TOTAL IMPEDANCE OF THE CIR-CUIT AT THAT FREQUENCY. THE EXTERNAL CIRCUIT CURRENT EQUALS THE VECTOR SUM OF THE CURRENTS THROUGH THE INDIVIDUAL BRANCHES. THE CURRENT THROUGH EACH BRANCH CIRCUIT EQUALS THE APPLIED VOLTAGE DIVIDED BY THE IMPEDANCE OF THAT BRANCH. THE IMPEDANCE OF EACH BRANCH CIRCUIT IS A FUNCTION OF THE RESISTANCE AND REACT-ANCE OF THAT INDIVIDUAL BRANCH. IN THE SOLUTION OF THIS PROBLEM IT IS CLEAR THAT EACH BRANCH CIRCUIT MUST FIRST BE SOLVED INDIVIDUALLY. FIRST, SOLVE FOR ALL THE VALUES OF THE UPPER CIRCUIT CONSISTING OF AN INDUCTANCE OF 100 MICRO-HENRIES AND RESISTANCE OF 10 OHMS. STATED IN EQUATION FORM:

 $R_{1} = 10 \text{ ohms}$ $X_{L} = 2 \text{ fr} L = 628 \times 5 \times 10^{-1} = 314 \text{ ohms}$ $Z_{1} = \sqrt{R^{2} + X_{L}^{2}} = \sqrt{10^{2} + 314^{2}} = 314.1 \text{ ohms}$ $I_{L} = E/Z = 1000/314.1 = 3.18 \text{ amperes}$ $T_{AN} \theta_{1} = X/R = 314/10 = 31.4$ $\theta_{1} = 88^{0}10^{1} \text{ Lag}$

IN THE INDUCTIVE BRANCH THE IMPEDANCE IS 314.1 OHMS, THE CURRENT IS 3.18 AMPERES, AND THE CURRENT LAGS 88 DEGREES 10 MINUTES BEHIND THE APPLIED VOLTAGE. THIS CURRENT IS SHOWN IN FIGURE 2 AS IL. IN THE CAPACITY BRANCH!

 $R_{2} = 15 \text{ ohms}$ $X_{C} = 1/2\pi FC = \frac{10^{6}}{628 \times 5} = 318.4 \text{ ohms}$ $Z_{2} = \sqrt{R^{2} + X_{C}^{2}} = \sqrt{15^{2} + 318.4^{2}} = 318.7 \text{ ohms}$ $I_{C} = E/Z = 1000/318.7 = 3.13 \text{ amperes}$ $T_{AN} \theta_{2} = X/R = 318.4/15 = 21.2$ $\theta_{3} = 87^{0}18' \text{ Lead}$

IN THE CAPACITY BRANCH THE IMPEDANCE IS 318.7 OHMS, THE CURRENT 3.13 AM-PERES, AND THE CURRENT LEADS THE APPLIED VOLTAGE BY 87 DEGREES AND 18 MINUTES-



THIS CURRENT IS SHOWN ON FIGURE 2 AS I ...

THERE ARE TWO CURRENTS, IL AND IC, BOTH CAUSED TO FLOW THROUGH THEIR RESPECTIVE CIR-CUITS BY THE SAME APPLIED VOLTAGE, E. THEY MAY THEREFORE BE CONSIDERED AS TWO FORCES ACT-ING UPON A COMMON POINT, AND THE TOTAL RE-SULTING CURRENT IN THE EXTERNAL CIRCUIT IS EQUAL TO THE VECTOR SUM OF THE INDIVIDUAL CUR-RENTS.

A REFERENCE TO THE ASSIGNMENTS ON VECTOR ANALYSIS AND PARALLEL CIRCUITS WILL SHOW THE METHOD OF HANDLING THIS PROBLEM. SINCE THE ANGLE BETWEEN THE TWO CURRENTS IS NEITHER ZERO DEGREES NOR 180 DEGREES, THAT IS.

They are neither exactly in phase nor exactly opposite, it is necessary to find the two components of each current in order to add them vectorally. Each current has two forces acting on it, reactance and resistance. Each current therefore has a resistance component and a reactive component. The resistance component of current tends to flow in phase with E and is equal to 1 Cos θ . The reactive component tends to lead or lag by 90⁰ and is equal to 1 Sin θ . A study of Figure 2 will show that the resistance components of both are to the right of the vertical bisector therefore adding. The total resistance component is equal to 1_cCos 87⁰18¹ + 1_iCos 88⁰10¹.

The reactive components, I Sin θ , are in exact opposition, the smaller therefore subtracting from the larger and the total reactive component equalling I Sin 88°10' - I Sin 87°18'.

IN THIS CASE WHERE THE DIFFERENCE BETWEEN THE REACTIVE COMPONENTS IS VERY SMALL AND A CASUAL INSPECTION DOES NOT INDICATE WHICH IS THE LARGER, THE SINE

value of I_{L} or the Sine value of I_{C} , the total may be written as above and if the results show $I_{C}Sin \theta$ as greater than $I_{L}Sin \theta$ it may be reversed and written $I_{C}Sin \theta - I_{L}Sin \theta$, or the result may simply be considered as a negative value.

AGAIN REFERRING TO VECTOR ANALYSIS, THE RESULTANT IS EQUAL TO THE HYPOTENUSE OF A RIGHT TRIANGLE FORMED BY THE TOTAL RESISTANCE COMPONENT AND THE TOTAL REACT-IVE COMPONENT. THE TOTAL CURRENT IS THEN,

	$I = \sqrt{I_{R}^{2} + I_{X}^{2}}$
But,	$I_{R} = I_{C} \cos 87^{\circ} 18' + I_{L} \cos 88^{\circ} 10'$
AND	$I_X = I_L SIN 88^{0}10' - I_C SIN 87^{0}18'$
THEREFORE,	$I = \sqrt{ I_{c}Cos 87^{0}18' + I_{c}Cos 88^{0}10' ^{2} + (I_{c}SIN 88^{0}10' - I_{c}SIN 87^{0}18' ^{2})^{2}}$
	IL = 3.18 AMPERES IC = 3.13 AMPERES
	$C_{0S} 87^{0}18^{1} = .04711$ $S_{IN} 87^{0}18^{1} = .99889$
	$C_{0S} 88^{0}10' = .03199$ $S_{IN} 88^{0}10' = .99949$
THEREFORE,	l_{c} Cos 87 ⁰ 18' = 3.13 x .04711 = .1475
	$I_{\rm L} \cos 88^{\rm O} 10^{\rm I} = 3.18 \times .03199 = .1017$
AND	$I_{\text{LSIN}} 88^{\circ}10' = 3.18 \times .99949 = 3.1783$
	$l_{c}S_{1N} 87^{0}18^{1} = 3.13 \times .99889 = 3.1265$

SUBSTITUTING THESE VALUES FOR THE SYMBOLS IN THE ABOVE EQUATION, IT BECOMES

$$| = \sqrt{(.1475 + .1017)^2 + (3.1783 + 3.1265)^2}$$

ADDING, $I = \sqrt{.2492^2 + .0518^2}$

SQUARING THE VALUES AND ADDING,

 $| = \sqrt{.0621 + .00268} = \sqrt{.06478}$

EXTRACTING THE SQUARE ROOT, I = .254 AMPERE.

IT WILL BE SEEN THAT THERE IS A CURRENT OF 3.18 AMPERES IN THE INDUCTIVE BRANCH AND 3.13 AMPERES IN THE CAPACITY BRANCH BUT ONLY .254 AMPERE IN THE EX-TERNAL CIRCUIT OR LINE THROUGH THE ALTERNATOR. THE TOTAL IMPEDANCE IS EQUAL TO

E/I = 1000/.254 = 3936 OHMS.

THIS CIRCUIT, EVEN THOUGH NOT OPERATED EXACTLY AT RESONANCE, HAS AN IMPED-

ANCE OF **3956** OHMS WHILE THE INDIVIDUAL IMPEDANCES OF THE TWO BRANCHES IN PARAL-LEL ARE ONLY **314** AND **318** OHMS RESPECTIVELY. THE SMALL DIFFERENCE BETWEEN THE TWO IMPEDANCES SHOWS THAT THE FREQUENCY OF **500** KC/s is not far from the reso-NANT FREQUENCY OF THE CIRCUIT AT WHICH FREQUENCY THE RESULTING IMPEDANCE WOULD HAVE BEEN SOMEWHAT HIGHER.

Since the reactive value of I_L is slightly greater than the reactive value of I_C , the resulting current will tend to LAG behind the applied voltage. The tangent of the angle of LAG of the line current behind the applied voltage is equal to the total reactive component of current divided by the total resistive component. The equation then becomes

$$T_{AN} \theta = I_X / I_R = \frac{I_L S_{IN} \theta - I_C S_{IN} \theta}{I_C Cos \theta + I_L Cos \theta}$$
$$T_{AN} \theta = \frac{.0518}{.2492} = .20786$$

$$\theta = 11^{0}44' \text{ LAG.}$$

TABULATING THE RESULTS OF THIS PROBLEM IT IS FOUND THAT AT A FREQUENCY OF 500 KC/s with an applied voltage of 1000 volts,

TOTAL IMPEDANCE = 3936 OHMS TOTAL CURRENT = .254 AMPERE TOTAL CURRENT LAGS THE VOLTAGE BY 11⁰44'

AT THE RESONANT FREQUENCY OF THE CIRCUIT THE RESULTING CURRENT WOULD HAVE BEEN STILL LESS AND THE TOTAL IMPEDANCE HIGHER. THE RESONANT FREQUENCY IS FOUND BY THE USE OF THE EQUATION

 $F = \frac{1}{2\pi \sqrt{LC}}$ L in Henries, C in Farads, F in cycles. $2\pi = 628 \times 10^{-2}$ C = 1000 x 10⁻¹² = 10⁻⁹ F L = 100 x 10⁻⁶ = 10⁻⁴ H THEREFORE, F = $\frac{1}{628 \cdot 10^{-2} \cdot \sqrt{10^{-9} \times 10^{-4}}}$

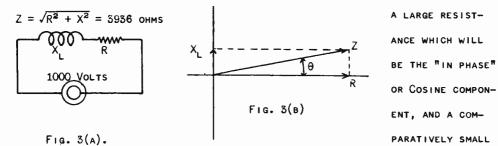
$$F = \frac{1}{628 \cdot 10^{-2} \sqrt{10 \times 10^{-14}}} = \frac{1}{628 \cdot 10^{-2} \sqrt{10} \cdot 10^{-7}}$$

$$F = \frac{1}{628 \cdot \sqrt{10} \cdot 10^{-9}} = \frac{10^9}{628 \sqrt{10}} = \frac{10^9}{628 \times 5.16}$$

$$F = \frac{10^{11}}{198448} = 503,910 \text{ cycles or } 503.91 \text{ KC/s.}$$

IT HAS BEEN SHOWN IN A PRECEDING LESSON THAT AT A FREQUENCY LOWER THAN ITS RESONANT FREQUENCY, THE CIRCUIT WILL ACT AS AN INDUCTANCE. AT RESONANCE THE CIRCUIT ACTS AS A RESISTANCE. IN THIS PROBLEM THE CIRCUIT HAS A RESONANT FRE-QUENCY OF 503.91 KC/s. It is operated at a lower frequency of 500 KC/s and a CURRENT LAG OF SLIGHTLY MORE THAN 11 DEGREES RESULTS. A LAGGING CURRENT IS THE RESULT OF AN INDUCTIVE CIRCUIT, THIS CIRCUIT THEREFORE HAS THE EFFECT OF A HIGH RESISTANCE AND A COMPARATIVELY SMALL INDUCTANCE IN SERIES.

IN CIRCUIT CALCULATIONS IT IS OFTEN HELPFUL TO REPLACE A COMPLEX CIRCUIT WITH AN EQUIVALENT MORE SIMPLE CIRCUIT. IT IS INTERESTING AT THIS POINT TO SEE JUST WHAT COMPONENTS AN EQUIVALENT SERIES CIRCUIT MUST HAVE TO REPLACE THE PAR-ALLEL COMBINATION OF FIGURE 1 AT 500 KC/S WITHOUT CHANGING IN ANY WAY THE LOAD ON THE GENERATOR. THE EQUIVALENT SERIES CIRCUIT MUST BE SUCH THAT THE SAME GENERATOR CURRENT, .254 AMPERE, WILL FLOW UNDER THE PRESSURE OF 1000 VOLTS WITH A CURRENT LAG OF $11^{0}44^{1}$. THE EQUIVALENT CIRCUIT WILL OBVIOUSLY CONSIST OF IN-DUCTANCE AND RESISTANCE IN SERIES. SEE FIGURE 3(A). Z = 3936 OHMS, MADE UP OF



INDUCTIVE REACTANCE WHICH WILL BE THE "QUADMATURE" OR SINE COMPONENT. THIS IS SHOWN IN FIGURE 3(B).

7.

FROM FIGURE 3(B) IT IS EVIDENT THAT R = Z COS θ and X_L = Z SIN θ .

R = $Z \cos \theta$ = 3936 Cos 11⁰44¹ = 3936 x .97910 = 3854 ohms X_L = $Z \sin \theta$ = 3936 Sin 11⁰44¹ = 5956 x .20336 = 800 ohms

$$L = \frac{X_L}{2\pi F} = \frac{800}{6.28 \times 5 \times 10^5} = 254.7 \,\mu\text{H}.$$

Thus <u>at this frequency</u>, 500 KC/s, so far as the load effect on the generator is concerned, the parallel circuit of Figure 1 could be replaced with the equivalent series circuit of Figure 3(a) having the following constants: R =3854 ohms, L = 254.7 μ H.

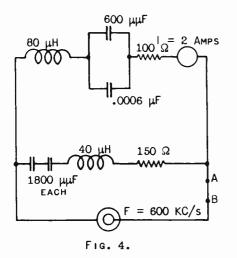
A MENTAL CONCEPT OF THE OPERATION OF A COMPLEX CIRCUIT MAY OFTEN BE SIM-

IN THE PARALLEL CIRCUIT PROBLEM DISCUSSED ABOVE, WHEN SOLVING FOR THE IM-PEDANCE OF THE ENTIRE PARALLEL CIRCUIT IT WAS NECESSARY FIRST TO FIND THE CUR-RENT THROUGH EACH OF THE BRANCHES, AND AFTER ADDING THOSE CURRENTS VECTORALLY TO DIVIDE THE RESULTING CURRENT INTO THE APPLIED VOLTAGE TO OBTAIN THE IMPED-ANCE; Z = E/I. TUNED CIRCUITS OF THIS TYPE WILL ORDINARILY BE USED WITH VARY-ING VOLTAGES, AND WHEN IT IS REQUIRED TO FIND THE IMPEDANCE NO VOLTAGE WILL BE SPECIFIED. IN THAT CASE IT IS MERELY NECESSARY TO ASSUME A VOLTAGE. THIS IS SEEN TO BE CORRECT WHEN IT IS REMEMBERED THAT THE CURRENT THROUGH ANY CIRCUIT VARIES DIRECTLY AS THE VOLTAGE. IF THE VOLTAGE IS DOUBLED THE CURRENT THROUGH THE CIRCUIT IS ALSO DOUBLED. IF THE VOLTAGE IS DECREASED TO ONE-HALF ITS NOR-MAL VALUE THE CURRENT IS ALSO DECREASED TO ONE-HALF.

Since Z = E/I, if the applied voltage is 100 volts and the current is one ampere, the impedance is equal to 100/1 or 100 ohms. If the voltage is increased by ten times, the current will also be increased ten times. And the equation Z = E/I becomes Z = 1000/10 = 100 ohms, the same as before. It is evident that the impedance of the circuit does NOT change with a variation of the applied voltage and therefore ANY voltage may be assumed when solving for the impedance.

IT MUST BE REMEMBERED HOWEVER THAT THE SAME ASSUMED VOLTAGE MUST BE USED THROUGHOUT THE ENTIRE PROBLEM. WHEN WORKING WITH FAIRLY HIGH VALUES OF IMPED-ANCE IT IS USUALLY SOMEWHAT MORE SIMPLE IF A COMPARATIVELY HIGH ASSUMED VOLTAGE IS USED BECAUSE UNDER THAT CONDITION COMPARATIVELY LARGE CURRENTS WILL BE OB-TAINED IN THE PARALLEL BRANCH CIRCUITS. THE HIGHER VALUES OF CURRENT ARE USUALLY CONSIDERED EASIER TO HANDLE THAN VERY SMALL DECIMAL AMOUNTS. OF COURSE, IF A CERTAIN VOLTAGE IS SPECIFIED IN THE PROBLEM, IT IS UNNECESSARY TO USE AN ASSUMED VOLTAGE.

PARALLEL COMBINATION OF SERIES LCR CIRCUITS: IT MAY BE NECESSARY TO DE-TERMINE THE ACTUAL VOLTAGE ACROSS A PARALLEL CIRCUIT AND THE TOTAL CURRENT IN THE EXTERNAL OR GENERATOR CIRCUIT WHEN ONLY THE CIRCUIT CONSTANTS AND THE CUR-RENT THROUGH ONE BRANCH OF THE PARALLEL CIRCUIT ARE GIVEN. THE FREQUENCY MUST BE SPECIFIED IN ALL A. C. WORK INVOLVING INDUCTANCE AND CAPACITY. SEE FIGURE 4.



cuits, each composed of L, C and R, in parallel.

IN THIS DIAGRAM THERE ARE TWO SERIES CIR-

IT IS FIRST NECESSARY TO SOLVE FOR THE IMPEDANCE, ANGLE OF LEAD OR LAG, AND THE CURRENT IN EACH BRANCH.

NO APPLIED VOLTAGE IS SPECIFIED BUT THE CURRENT THROUGH THE UPPER BRANCH IS GIVEN AS 2 AMPERES. SINCE E = 1Z, if the impedance OF THE UPPER BRANCH IS FIRST DETERMINED, THE

VOLTAGE ACROSS THE CIRCUIT MAY BE FOUND BY MULTIPLYING THAT IMPEDANCE BY THE CURRENT OF 2 AMPERES. THIS BEING A PARALLEL CIRCUIT, THE SAME VOLTAGE IS ALSO APPLIED ACROSS THE LOWER BRANCH, AND BY DIVIDING THIS COMMON VOLTAGE BY THE IM-PEDANCE OF THE LOWER BRANCH, THE CURRENT THROUGH THE LOWER BRANCH MAY ALSO BE CALCULATED. THIS. PROBLEM IS STARTED IN THE SAME WAY AS WAS THE PRECEDING

PROBLEM, I. E., THE INDIVIDUAL BRANCHES ARE TREATED AS SIMPLE SERIES CIRCUITS AND SOLVED FOR IMPEDANCES AND CURRENTS.

UPPER BRANCH:

R = 100 ohms. $X_{L} = 2\pi FL = 628 \times 6 \times 8 \times 10^{-2} = 301 \text{ ohms.}$ $C = 600 \ \mu\mu F + .0006 \ \mu F = 600 + 600 = 1200 \ \mu\mu F. \qquad \text{ing in parallel, add.})$ $X_{C} = \frac{1}{628 \times 6 \times 12 \times 10^{-7}} = \frac{10^{7}}{45216} = 221 \text{ ohms.}$ $Z_{1} = \sqrt{R^{2} + (X_{L} - X_{C})^{2}} = \sqrt{100^{2} + (301 - 221)^{2}} = \sqrt{100^{2} + 80^{2}} = 128 \text{ ohms.}$ $Tan \ \theta_{1} = X/R = 80/100 = .8$ $\theta_{1} = 39^{0} \text{ Lag.} \qquad (X_{L} \text{ is greater than } X_{C})$ $E = 1Z = 2 \times 128 = 256 \text{ volts.}$

As was previously explained, this is also the voltage across the lower branch; in other words, the total applied voltage. Solving for the lower branch in a similar manner:

$$X_{L} = 628 \times 6 \times 4 \times 10^{-2} = 151 \text{ ohms.}$$

$$C = \frac{1}{-\frac{1}{1} + \frac{1}{-\frac{1}{12}}} = \frac{1}{-\frac{2}{2}} = \frac{1800}{2} = 900 \text{ }\mu\mu\text{F} \qquad (\text{The two capacities are in series.})$$

$$X_{C} = \frac{1}{628 \times 6 \times 9 \times 10^{-7}} = \frac{10^{7}}{33912} = 295 \text{ ohms.}$$

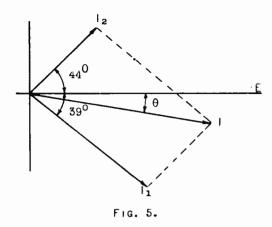
$$Z_{2} = \sqrt{R^{2} + (X_{C} - X_{L})^{2}} = \sqrt{150^{2} + (295 - 151)^{2}} = \sqrt{150^{2} + 144^{2}} = 208 \text{ ohms}$$

$$\text{Tan } \theta_{2} = X/R = 144/150 = .960$$

$$\theta_{2} = 44^{0} \text{ Lead.} \qquad (X_{C} \text{ is greater than } X_{L}.)$$

$$I_{2} = E/Z = 256/208 = 1.23 \text{ amperes.}$$

IT HAS BEEN DETERMINED THAT THE APPLIED VOLTAGE OF 256 VOLTS CAUSES A CUR-RENT FLOW OF 2 AMPERES, l_1 , in the upper branch, that current lagging 39^0 be-HIND THE APPLIED VOLTAGE; ALSO THAT THE SAME VOLTAGE CAUSES A CURRENT FLOW OF 1.23 AMPERES, l_2 , THROUGH THE LOWER BRANCH, THIS CURRENT LEADING THE SAME VOLT-



BY 44⁰. This is shown vectorally in Figure 5.

THE RESULTANT CURRENT, THAT IS, THE LINE CURRENT IN THE EXTERNAL CIR-CUIT, IS ALSO SHOWN AS THE VECTOR SUM OF THE CURRENTS THROUGH THE INDIVIDUAL BRANCHES. THE VECTOR SUM OF THE CUR-RENTS MAY BE EXPRESSED BY THE FOLLOWING EQUATION:

$$I = \sqrt{(I_1 \cos 39^\circ + I_2 \cos 44^\circ)^2 + (I_1 \sin 39^\circ + I_2 \sin 44^\circ)^2}$$

$$I_1 \cos 39^\circ = 2 \times .777 = 1.554 \text{ amperes}$$

$$I_2 \cos 44^\circ = 1.23 \times .719 = .884 \text{ amperes}$$

$$I_1 \sin 39^\circ = 2 \times .629 = 1.258 \text{ amperes}$$

$$I_2 \sin 44^\circ = 1.23 \times .695 = .855 \text{ amperes}$$

REPLACING THESE FIGURES IN THE ABOVE EQUATION, IT BECOMES

$$I = \sqrt{(1.554 + .884)^2 + (1.258 + .855)^2}$$

$$I = \sqrt{(2.438^2 + .403^2)} = \sqrt{6.1062} = 2.47 \text{ amperes}.$$

$$Tan \theta = I_X / I_R = \frac{1.403}{2.438} = .165$$

$$\theta = 9.5^0 \text{ Lag.}$$

SINCE THE TOTAL IMPEDANCE OF A PARALLEL CIRCUIT IS EQUAL TO THE APPLIED VOLTAGE DIVIDED BY THE TOTAL CURRENT, THE IMPEDANCE OF THIS PARALLEL COMBINA-TION IS EQUAL TO:

Z = E/I = 256/2.47 = 103 ohms.

TABULATING THE RESULTS:

I (UPPER BRANCH) = 2 AMPERES

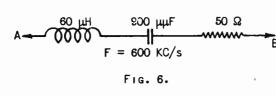
 θ " = 39⁰ LAG

| (LOWER BRANCH) = 1.23 AMPERES

 θ " " = 44^O LEAD

TOTAL CURRENT = 2.47 AMPERES θ (TOTAL CURRENT) = 9.5° Lag Applied voltage = 256 volts. Z(Total circuit) = 103 ohms.

ADDITION OF SERIES CIRCUIT TO PARALLEL COMBINATION: SUPPOSE IT IS DESIRED TO ADD ANOTHER IMPEDANCE, COMPOSED OF L, C AND R IN SERIES, TO THE PARALLEL COMBINATION AND IN SERIES WITH IT. IN FIGURE 6 IS SHOWN THE SERIES CIRCUIT,



THE FREQUENCY IS THE SAME AS IN FIGURE 4, I. E., 600 KC/s.

$$R = 50 \text{ ohms}$$

$$X_{L} = 2\text{mFL} = 628 \times 6 \times 6 \times 10^{-2} = 226 \text{ ohms.}$$

$$X_{C} = \frac{1}{628 \times 9 \times 6 \times 10^{-7}} = \frac{10^{7}}{628 \times 9 \times 6} = 295 \text{ ohms.}$$

$$Z = \sqrt{R^{2} + (X_{C} - X_{L})^{2}} = \sqrt{50^{2} + (295 - 226)^{2}} = \sqrt{50^{2} + 69^{2}} = 85 \text{ ohms}$$

$$\tilde{T}_{AN} \theta = X/R = 69/50 = 1.38$$

$$\theta = 54^{0} \text{ Lead } (X_{C} \text{ is greater than } X_{L}.)$$

The impedance of the parallel circuit Z_p was found to be 103 ohms and to be such as to cause a current Lag of 9.5 degrees. The impedance of the series circuit Z_s is 85 ohms and its effect is to cause a current lead of 54 degrees. The total impedance of the two circuits, <u>in series</u>, will be the vector sum of Z_p and Z_s and is shown in Figure 7. Both will have an effect on the current flow through the circuit according to their individual magnitudes and angles.

THE TOTAL IMPEDANCE FOR THE ENTIRE CIRCUIT CONSISTING OF THE PARALLEL COM-BINATION SHOWN IN FIGURE 4 IN SERIES WITH THE SERIES COMBINATION SHOWN IN FIG-URE 6 IS EXPRESSED IN THE FOLLOWING EQUATION:

$$Z = \sqrt{(Z_{p}Cos \ 9.5^{\circ} + Z_{s}Cos \ 54^{\circ})^{2} + (Z_{s}SIN \ 54^{\circ} - Z_{p}SIN \ 9.5^{\circ})^{2}}$$

$$Z_{p}Cos \ 9.5^{\circ} = 103 \ x \ .987 = 101.6$$

$$Z_{s}Cos \ 54^{\circ} = 85 \ x \ .588 = 49.9$$

$$Z_{s}SIN \ 54^{\circ} = 85 \ x \ .809 = 68.7$$

$$Z_{p}SIN \ 9.5^{\circ} = 103 \ x \ .165 = 16.9$$

REPLACING SYMBOLS WITH NUMERICAL VALUES, THE EQUATION BECOMES:

$$Z = \sqrt{(101.6 + 49.9)^2 + (68.7 - 16.9)^2} = \sqrt{151.5^2 + 51.8^2} = 160 \text{ ohms.}$$

Tan $\theta = X/R = 51.8/151.5 = .341$
 $\theta = 19^0 \text{ Lead}$

THE TOTAL IMPEDANCE OF THE TWO CIRCUITS CONNECTED IN SERIES IS 160 OHMS AT

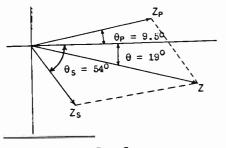


FIG. 7.

A FREQUENCY OF 600 KC/s. The reactive component of Z_s is larger than the reactive component of Z_p , therefore since the effect of Z_s is to cause a leading current the effect of the total impedance is to cause a current lead of 19^0 .

COMPLEX NETWORK: IN ORDER TO

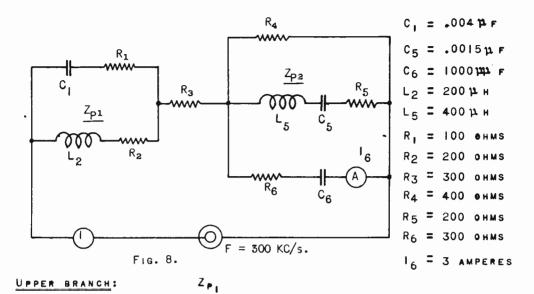
DEMONSTRATE THE EASE WITH WHICH ALL THE VOLTAGE, CURRENT, AND IMPEDANCE RELA-TIONS IN A SOMEWHAT COMPLEX NETWORK MAY BE DETERMINED FROM WHAT MAY AT FIRST APPEAR TO BE RELATIVELY LITTLE INFORMATION, CONSIDER THE SOLUTION OF THE CIRCUIT OF FIGURE 8.

Fundamentally this is a series circuit consisting of Z_{P1} , R_3 and Z_{P2} in series. It is first necessary to solve for Z_{P1} and Z_{P2} in terms of ohms and θ . In a problem of this type the most essential factor is an orderly and method-ical attack.

To find the Line current i_T , the impedance of the total load across the generator, the generator voltage, and the phase angle of the load.

FIRST, SOLVE Zp,.

www.americanradiohistorv.com



 $\begin{array}{c} R_{1} = 100 \text{ ohms} \\ X_{C_{1}} = \frac{10^{3}}{6.28 \times .3 \times 10^{6} \times 4 \times 10^{-9}} = \frac{10^{3}}{7.536} = 133 \text{ ohms} \\ Z_{1} = \sqrt[7]{R^{2} + X_{C_{1}}^{2}} = \sqrt{100^{2} + 133^{2}} = 166 \text{ ohms} \\ T_{AN} \theta_{1} = \frac{X}{R} = \frac{133}{100} = 1.33 \quad \theta_{1} = 53^{\circ} \frac{166}{166} = 1 \text{ amp.} \\ Assume \text{ voltage drop is 166 volts, then } I_{1} = \frac{166}{166} = 1 \text{ amp.} \end{array}$

LOWER BRANCH:

 $R_{2} = 200 \text{ ohms}$ $X_{L_{2}} = 6.28 \times 3 \times 10^{5} \times 2 \times 10^{-4} \pm 37.68 \times 10 = 376.8 \text{ ohms}$ $Z_{2} = \sqrt{200^{2} + 376.8^{2}} = 427 \text{ ohms}$ $TAN \theta_{2} = \frac{X}{R} = \frac{376.8}{200} = 1.88 \quad \theta_{2} = 62^{\circ} \text{ LAG}$ $I_{2}(ASSUMED) = \frac{166}{427} = .39 \text{ Ampere}$ The Assumed currents are shown in the vector of Fig. 9.

TO FIND THE TOTAL LINE CURRENT WE MUST COMBINE THE TWO BRANCH CURRENTS VECTORIALLY AS FOLLOWS:

$$I_{T} = \sqrt{(1_{1}\cos 53^{\circ} + 1_{2}\cos 62^{\circ})^{2} + (1_{1}\sin 53^{\circ} - 1_{2}\sin 62^{\circ})^{2}}$$

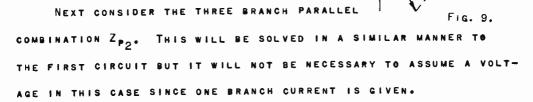
= $\sqrt{(.602_{+}.183)^{2} + (.799 - .344)^{2}} = \sqrt{.785^{2} + .455^{2}}$
= .907 AMPERE

11

$$Z_{P_{1}} = \frac{E_{ASSUMED}}{I_{ASSUMED}} = \frac{166}{.907} = \frac{183 \text{ ohms}}{.907}$$

TAN $\theta_{P_{1}} = \frac{.455}{.785} = .58$ $\theta_{P_{1}} = \frac{.30^{\circ} \text{ LEAD}}{.907}$

THIS IS THE TRUE IMPEDANCE OF Z_{P1} EVEN THOUGH AN ASSUMED VOLTAGE WAS USED IN ITS DETERMINATION. THE ASSUMED VOLTAGE AND THE BRANCH CURRENTS <u>WILL NOT BE USEFUL</u> FOR ANY OTHER WORK IN THIS PROBLEM.



7

LOWER BRANCH:

$$R_6 = 300 \text{ ohms}$$
 $C_6 = 1000 \text{ mus} = 10^{-9}\text{Farad}$ $I_6 = 3 \text{ amp.}$
 $x_{c_6} = \frac{1}{6.28 \times 3 \times 10^5 \times 10^{-9}} = \frac{10^4}{18.84} = 531 \text{ ohms}$
 $Z_6 = \sqrt{300^2 + 531^2} = 609 \text{ ohms}$ $\text{Tan} \theta_6 = \frac{531}{300} = 1.77$
 $= 60.5^{\circ} \text{ Lead}$
 $E_{Z_{P_2}} = I_6Z_6 = 3 \times 609 = 1827 \text{ volts}$ (This is the actual
voltage across the circuit, no assumed voltage necessary)

$$R_{5} = 200 \text{ ohms}$$

$$X_{L_{5}} = 6.28 \times 3 \times 10^{5} \times 4 \times 10^{-4} = 75.36 \times 10 = 753.6 \text{ ohms}$$

$$X_{c_{5}} = \frac{1}{6.28 \times 3 \times 10^{5} \times 15 \times 10^{-8}} = \frac{10^{3}}{282.6} = 354 \text{ ohms}$$

$$Z_{5} = \sqrt{200^{2} + (753.6 - 354)^{2}} = \sqrt{200^{2} + 399.4^{2}} = 447 \text{ ohms}$$

$$T_{AN} \theta_{5} = \frac{399.4}{200^{-2}} = 1.997 \qquad \theta_{5} = 63.5^{\circ} \text{ LAG}$$

$$I_5 = \frac{1827}{447} = 4.09$$
 AMPERE

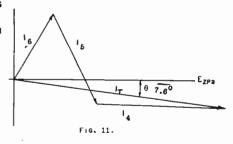
UPPER BRANCH:

 $Z = R_4 = 400 \text{ eHMS}$ $\theta_4 = 0^0$ $I_4 = \frac{1827}{400} = 4.57 \text{ AMPERES}$

THE CONDITIONS IN EACH PART OF THE CIRCUIT ARE CLEARLY SHOWN IN FIG. 10. 1₆ LEADS THE APPLIED VOLTAGE OR VOLTAGE DROP ACROSS Z_{P2}, 1₅ LAGS AND

¹4 IS IN PHASE WITH E. FIG. 11 MORE CLEARLY SHOWS THE OPERATION OF THE CIRCUIT AS A WHOLE. IT IS SEEN THAT THE TOTAL CURRENT THROUGH Z_{P2} is quite large and the current lags the circuit volt-Age by a few degrees. The total current I will next be found.

FIRST NOTE THAT IN FIG. I'I THE CURRENTS ARE THOSE FOUND FROM ACTUAL CALCULATION WITHOUT ASSUMING ANY VOLTAGE AND THE REFERENCE VECTOR IS THE VOLTAGE DROP ACROSS THE PARALLEL COMBINATION. THE TOTAL CURRENT LAGS THIS VOLTAGE DROP



AS SHOWN IN THE FIGURE. THE TOTAL APPLIED VOLTAGE TO THE ENTIRE CIRCUIT IS THE VECTOR SUM OF THE VOLTAGE DROPS ACROSS THE SEPARATE CIRCUITS Z_{P1} , R_3 , and Z_{P2} . This can be used as a check on the voltage as found later and on the solution as a whole.

$$I_{T} = \sqrt{(I_{6}\cos 60.5^{\circ} + I_{5}\cos 63.5^{\circ} + I_{4})^{2} + (I_{5}\sin 63.5^{\circ} - J_{6}\sin 60.5^{\circ} + I_{4}\sin 0^{\circ})^{2}}$$

= $\sqrt{(3\cos 60.5 + 4.09\cos 63.5 + 4.57)^{2} + (4.09\sin 63.5 - 3\sin 60.5 + 0)^{2}}$
= $\sqrt{(1.48 + 1.82 + 4.57)^{2} + (3.66 - 2.61)^{2}}$
= $\sqrt{7.87^{2} + 1.05^{2}} = 7.94$ AMPERE

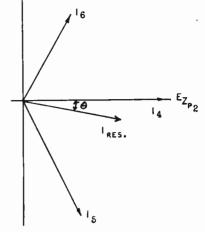
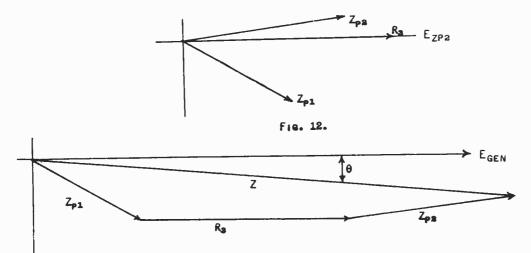


Fig. 10.

 $T_{AH} \theta = \frac{1.05}{7.87} = .133$ $\theta_{P_2} = \frac{7.6^{\circ} LAG}{7.6^{\circ} LAG}$ $Z_{P_2} = \frac{1827}{7.94} = \frac{230 \text{ ohms}}{7.6^{\circ} LAG}$

NOTE THAT THIS CURRENT IT IS THE TRUE CURRENT OF THE ENTIRE CIRCUIT WHICH IS SUPPLIED BY THE GENERATOR.

Now that the three parts of the original circuit of Fig. 8 have been solved they can be considered as an equivalent series circuit as illustrated by the vectors in Fig. 12 which can then be combined into a resultant total impedance as shown in Fig. 13.



THE SOLUTION OF TOTAL Z CAN BE DONE GRAPHICALLY BY PLOTTING THE VECTORS TO SCALE OR BY TRIGONOMETRY AS SHOWN BELOW:

Z	= $\sqrt{(Z_{P_1}\cos 30^\circ + Z_{P_2}\cos 7.6^\circ + R_3)^2} + (Z_{P_1}\sin 30^\circ - Z_{P_2}\sin 7.6^\circ)^2}$
	$= \sqrt{(183\cos 30^{\circ} + 230\cos 7.6^{\circ} + 300)^{2} + (183\sin 30^{\circ} - 230\sin 7.6^{\circ})^{2}}$
	$= \sqrt{(158 + 228 + 300)^2 + (91.5 - 30.4)^2} = \sqrt{686^2 + 61.1^2} = 689 \text{ ohms}$
I _T	= 7.94 AMP. $E_{gen} = 1Z = 7.94 \times 689 = 5471 \text{ volts}$
As	A CHECK WE CAN SOLVE FOR THE VECTORIAL SUM OF THE VOLTAGES.
	$E_{Z_{P_1}} = 183 \times 7.94 = 1453 \text{ volts } 30^{\circ} \text{ Lead}$ $E_3 = 300 \times 7.94 = 2382 \text{ v}.$
E _G	= $\sqrt{(1827\cos 7.6 + 1453\cos 30 + 2382)^2 + (1827\sin 7.6 - 1453\sin 30)^2}$
	$\frac{1827 \times .9908 + 1453 \times .866 + 2382}{2} + (1827 \times .11667 - 1453 \times .5)^2}$
	$= \sqrt{(5451)^2} + (513.5)^2 = 5480$ volts (a close check)

16.

 $T_{AN} \theta = 61.1/686 = .08906$

 $\theta = 5.1^{\circ}$ LEAD.

This is the entire solution. This complex circuit has a total impedance at 300 KC/s of <u>689 ohms</u>. To cause 3 amperes to flow in the branch circuit $^{C}6R_{6}$ requires a generator voltage of <u>5471 volts</u>, and this voltage causes a total line current of <u>7.94 amperes</u> which <u>leads</u> the generator voltage by <u>5.1⁰</u>. So far as the generator load is concerned, this complex circuit could be replaced by a simple series circuit consisting of resistance and capacity, the total impedance and phase angle at this frequency being equal to the calculated value, that is, Z = 689 ohms, $\theta = 5.1^{\circ}$ Lead. <u>This would require a resistance of 686</u> <u>OHMS AND CAPACITY SUCH THAT X_C = 61.1 ohms</u>, as is apparent from the final equation for Z above.

IT SHOULD BE NOTED THAT, EVEN THOUGH AT FIRST GLANCE THE SOLUTION OF THIS CIRCUIT SEEMS QUITE COMPLEX, ACTUALLY WHEN HANDLED IN A SYSTEMATIC MANNER IT IS NOT AT ALL QIFFICULT. ANY SERIES-PARALLEL COMBINATION CAN BE SOLVED BY THE GENERAL METHODS USED IN THIS PROBLEM. IT SHOULD BE EMPHASIZED THAT THE SOLU-TION IS CORRECT ONLY FOR THE SINGLE FREQUENCY AT WHICH THE CALCULATIONS ARE MADE.

VARIOUS PRACTICAL APPLICATIONS OF SERIES AND PARALLEL CIRCUITS WILL BE DISCUSSED IN CONSIDERABLE DETAIL ALONG WITH THEIR CONSTRUCTION, IN FOLLOWING LESSONS.

SUMMARIZING THE VOLTAGE, CURRENT AND IMPEDANCE RELATIONS IN SERIES AND IN PARALLEL CIRCUITS:

1. SERIES.

VOLTAGE ACROSS CIRCUIT EQUALS VECTOR SUM OF INDIVIDUAL VOLTAGES. CURRENT IS THE SAME THROUGH ALL PARTS OF A SERIES CIRCUIT.

IMPEDANCE IS EQUAL TO THE VECTOR SUM OF THE INDIVIDUAL IMPEDANCES.

2. PARALLEL.

VOLTAGE IS THE SAME ACROSS EACH BRANCH OF THE CIRCUIT. CURRENT EQUALS THE VECTOR SUM OF THE INDIVIDUAL CURRENTS. IMPEDANCE EQUALS THE TRUE OR ASSUMED VOLTAGE DIVIDED BY THE TRUE OR ASSUM-ED CURRENT WHICH THE TRUE OR ASSUMED VOLTAGE FORCES THROUGH THE CIRCUIT.

THE LC TABLE.

The LC TABLE, A COPY OF WHICH IS INCLUDED IN THE BACK OF THIS LESSON, IS ONE OF THE MOST CONVENIENT TABLES USED IN RADIO WORK. IT HAS BEEN SHOWN THAT FOR ANY ONE FREQUENCY OR ANY ONE WAVELENGTH, THERE IS ONLY ONE VALUE OF THE LC PRODUCT. THAT IS, THE INDIVIDUAL L AND C VALUES FOR TWO CIRCUITS MAY BE EN-TIRELY DISSIMILAR BUT SO LONG AS THE PRODUCTS OF L TIMES C IN THE TWO CIRCUITS ARE THE SAME, BOTH CIRCUITS WILL HAVE THE SAME WAVELENGTH OR RESONANT FREQUEN-CY.

WHY IS THIS LC TABLE SO USEFUL? BECAUSE IN ALMOST ANY TYPE OF RADIO PROB-LEM DEALING WITH WAVELENGTH OR FREQUENCY THE LC PRODUCT IS ENCOUNTERED. FOR EXAMPLE,

 $F_{REQUENCY} = \frac{1}{2\pi \sqrt{LC}}$ $F_{IN CYCLES, L AND C IN UNITS.$ $F_{REQUENCY} = \frac{159}{\sqrt{LC}}$ $F_{IN KILOCYCLES, L AND C IN MICRO-UNITS.$

Wavelength = $1884 \sqrt{LC}$. Wavelength in meters, L and C in micro-units. Given a circuit consisting of known values of inductance and capacity, it is possible of course to find either the wavelength or the resonant frequency from one of the equations shown above. This involves several steps in either case. By the use of the tables, however, it is only necessary to multiply the values of L and C, (both in micro-units), and refer to the table for the corresponding value of either wavelength of frequency. It should be observed that in the enclosed tables, frequency is designated by "N" and is expressed in <u>cycles</u>. To convert to kilocycles it is only necessary to divide the frequency in cycles by 1000.

ANOTHER EXAMPLE WHERE THE LC TABLE CAN BE USED TO ADVANTAGE. SUPPOSE THERE IS GIVEN A CIRCUIT OF KNOWN WAVELENGTH OF RESONANT FREQUENCY, AND THE VALUE OF C; TO FIND L. IT IS ONLY NECESSARY TO REFER TO THE LC PRODUCT FOR THE KNOWN FREQUENCY OR WAVELENGTH AND DIVIDE THE LC PRODUCT BY THE KNOWN VALUE OF C IN MICROFARADS. THE ANSWER WILL BE THE INDUCTANCE IN MICROHENRIES. IF L IS KNOWN AND C MUST BE DETERMINED, THE LC PRODUCT IS DIVIDED BY THE VALUE OF L IN MICROHENRIES, THE ANSWER BEING THE CAPACITY IN MICROFARADS.

THE ENCLOSED TABLES ARE COMPUTED FOR A MINIMUM WAVELENGTH OF 100 METERS OR 3000 KILOCYCLES. IN MODERN RADIO WORK IT IS OFTEN NECESSARY TO CALCULATE VAL-UES IN WAVELENGTHS MUCH SHORTER THAN 100 METERS. IT IS A VERY SIMPLE MATTER TO USE THE LC TABLE FOR SUCH VALUES.

For example, the LC product for a wavelength of 1000 meters is .282; for 100 meters the LC product is .00282; for 10 meters the LC product will be .0000282.

THE LC PRODUCT FOR 3000 METERS IS 2.53; FOR 300 METERS, THE LC PRODUCT IS .0253; FOR 30 METERS THE LC PRODUCT WILL BE .000253; FOR 3 METERS, .00000253.

IN OTHER WORDS, FOR EACH PLACE IN THE WAVELENGTH COLUMN THE DECIMAL POINT IS MOVED TO THE LEFT, THE DECIMAL POINT MUST BE MOVED <u>TWO</u> PLACES TO THE LEFT IN THE COLUMN OF LC PRODUCTS. THIS IS DUE TO THE FACT THAT THE WAVELENGTH VARIES AS THE <u>SQUARE ROOT</u> OF THE LC PRODUCT; WAVELENGTH = $1884 \sqrt{LC}$. IN EXTRACTING THE SQUARE ROOT OF THE PRODUCT OF LC, THE PRODUCT IS POINTED OFF <u>IN GROUFS OF TWO</u> PLACES FROM THE DECIMAL POINT.

FREQUENCY VALUES MAY BE HANDLED IN A SIMILAR MANNER EXCEPT THAT AS THE DECIMAL POINT IN THE NUMERICAL VALUE OF FREQUENCY IS MOVED <u>ONE PLACE TO THE</u> LEFT, THE DECIMAL POINT IN THE LC PRODUCT IS MOVED <u>TWO PLACES TO THE RIGHT</u>. THIS IS DUE TO THE FACT THAT THE FREQUENCY VARIES <u>INVERSELY</u> AS THE SQUARE ROOT OF THE LC PRODUCT: $F = \frac{1}{2\pi \sqrt{10}}$.

REGULAR USE OF THE LC TABLES WILL GREATLY DECREASE THE TIME REQUIRED IN THE SOLUTION OF PROBLEMS INVOLVING L, C, WAVELENGTH AND FREQUENCY, AND WILL ALSO DECREASE THE CHANCES OF ERROR.

A problem involving these values will show clearly the use of this table. Given a variable condenser of 500 micro-microfarads maximum capacity and 40 micro-microfarads minimum. It is desired to use the condenser with an inductance that will enable the circuit to be tuned to 550 meters with all the capacity in the circuit. What must be the effective value of inductance?

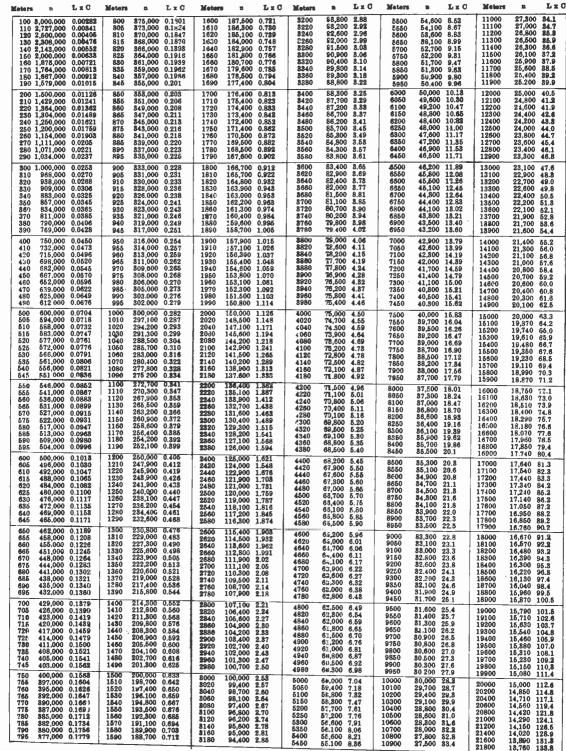
Referring to the LC table, it is found that the LC product at 550 meters is .0852; 500 $\mu\mu F$ = .0005 μF . L = LC/C = .0852/.0005 = 170.4 μH .

WITH THIS VALUE OF INDUCTANCE WHAT WILL BE THE SHORTEST WAVELENGTH TO which this circuit can be tuned? Minimum capacity is 40 $\mu\mu$ F or .00004 μ F. L is 170.4 μ H. LC = 170.4 x .00004 = .006816.

THE LC TABLE STATES THAT .00633 REPRESENTS A WAVELENGTH OF 150 METERS AND THAT .00721 REPRESENTS 160 METERS. .006816 IS ABOUT HALF-WAY BETWEEN THESE TWO VALUES SO THAT THIS CIRCUIT WOULD TUNE DOWN TO APPROXIMATELY 155 METERS. AN ACCURATE FIGURE COULD BE DETERMINED BY INTERPOLATION, BUT GREATER ACCURACY IS USUALLY NOT NECESSARY IN A PROBLEM OF THIS TYPE.

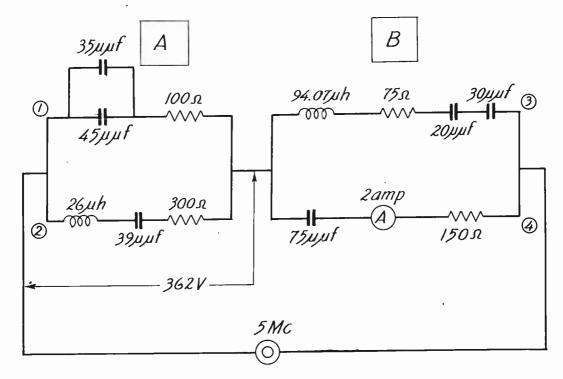
WHEN WORKING WITH QUITE HIGH FREQUENCIES, THAT 'S, FREQUENCIES IN EXCESS OF 1000 KC;s, the frequency is commonly expressed in megacycles. Thus 15,000 KC/s = 15 MC, 8700 KC/s = 8.7 MC, 2300 KC/s = 2.3 MC, 120,000 KC/s = 120 MC, ETC. ONE MEGACYCLE EQUALS ONE THOUSAND KILOCYCLES. TO CONVERT MC TO KC, MUL-TIPLY BY 10^3 . To convert KC to MC, MULTIPLY BY 10^{-3} .





EXERCISE PROBLEMS

SERIES AND PARALLEL CIRCUITS



1). Impedance, and current of branch 1

2). Impedance, and current of branch 2

3). Impedance, and current of circuit \blacksquare By two methods

4). Impedance, and current of branch 3

5). Impedance, and voltage of branch 4

6). Impedance, and current of circuit B by two methods

7). Total Impedance of the circuit by two methods

8). Total line current

9). Total voltage drop

10). Total Power dissipated in the circuit by two methods. (Note: Vector diagrams for each circuit will be helpful).

www.americanradiohistorv.com

ANSWERS

1.	$Z_1 = 100 - j398.0 = 410.4/-75^{\circ}54'$ ohms
	I ₁ = .882 amps.
2.	$Z_2 = 300 + j0 = 300/0^{\circ}$ ohms
	$I_{2} = 1.206 \text{ amps.}$
3.	$Z_a = 186.93 - j112.54 = 218.2/-31°4'$ ohms
	I = 1.66 amps.
4.	$Z_3 = 75 + j300 = 309.8/75^38'$ ohms
·	I ₃ = 2.906 amps
5.	$Z_4 = 150 - j424.6 = 450.2/-70^{\circ}32'$
	$E_h = 900.4$ volts.
6.	$Z_{b} = 449.23 + j303.67 = 542.3/34^{\circ}4'$ ohms
	$I_b = 1.66$ amps.
7.	Z _t = 636.14 + j191.13 = 664.45/16 [°] 43' ohms
8.	$I_t = 1.66 \text{ amps}$.
9.	$E_t = 1102 \text{ volts.}$
10.	$P_t = 1752$ watts.